## Finite-Length Discrete Transforms

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**NOTE:** This document is a brief summary (a cheatsheet, actually) of Chapter 5 from the textbook *Digital Signal Processing*, by S. K. Mitra.

• The N-point DFT of a sequence x[n] is defined as:

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-j\frac{2\pi}{N}kn} = \sum_{n=0}^{N-1} x[n]W_N^{kn} \qquad 0 \le k \le N-1$$
 (1)

where  $W_N = e^{-j\frac{2\pi}{N}}$ 

• The N-point inverse DFT (IDFT) of a sequence X[k] is defined as:

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-kn} \qquad 0 \le n \le N-1$$
 (2)

• The N-point DFT of a sequence x[n] of length L < N can be obtained by zero-padding the sequence x[n] until it has the desired length of N samples. That is:

$$DFT_N\{x[n]\} = DFT_N\{x_{zp}[n]\}$$

where

$$x_{zp}[n] = \begin{cases} x[n] & \text{if} \quad 0 \le n \le L - 1\\ 0 & \text{if} \quad L \le n \le N \end{cases}$$

• Whenever computing DTFs or IDFTs it is useful to remember that:

$$\sum_{n=0}^{N-1} W_N^{An} = \begin{cases} N & \text{if } A = rN \text{ with } r \text{ an integer} \\ 0 & \text{otherwise} \end{cases}$$
 (3)

• The DFT can also be expressed in matrix form like:

$$\mathbf{X} = \mathbf{D}_N \mathbf{x} \tag{4}$$

where

$$\mathbf{X} = [X[0], X[1], ..., X[N-1]]^T$$
  
 $\mathbf{x} = [x[0], x[1], ..., x[N-1]]^T$ 

and  $\mathbf{D}_N$  is the  $N \times N$  DFT matrix given by:

$$\mathbf{D}_{N} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & W_{N}^{1} & W_{N}^{2} & \cdots & W_{N}^{(N-1)} \\ 1 & W_{N}^{2} & W_{N}^{4} & \cdots & W_{N}^{2(N-1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W_{N}^{(N-1)} & W_{N}^{2(N-1)} & \cdots & W_{N}^{(N-1)^{2}} \end{bmatrix}$$
(5)

• Equivalently, the inverse DFT (IDFT) can be expressed in matrix form:

$$\mathbf{x} = \mathbf{D}_N^{-1} \mathbf{X} \tag{6}$$

where:

$$\mathbf{D}_{N}^{-1} = \frac{1}{N} \mathbf{D}_{N}^{*} = \frac{1}{N} \begin{bmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & W_{N}^{-1} & W_{N}^{-2} & \cdots & W_{N}^{-(N-1)} \\ 1 & W_{N}^{-2} & W_{N}^{-4} & \cdots & W_{N}^{-2(N-1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W_{N}^{-(N-1)} & W_{N}^{-2(N-1)} & \cdots & W_{N}^{-(N-1)^{2}} \end{bmatrix}$$
(7)

• An important operator when working with DTFs is the *modulo* operator. This operator is denoted by  $\langle a \rangle_b = c$ , which reads as "c is equal to a modulo b". This operator works differently depending whether a is a positive or negative integer.

$$\langle a \rangle_b = \begin{cases} a - \lfloor \frac{a}{b} \rfloor b & \text{if } a > 0 \\ \lceil \frac{|a|}{b} \rceil b + a & \text{if } a < 0 \end{cases}$$

where  $|\cdot|$  and  $|\cdot|$  denote the floor and ceiling functions, respectively. For instance:

$$<10>_4 = 10 - \left\lfloor \frac{10}{4} \right\rfloor \cdot 4 = 10 - 2 \cdot 4 = 2$$
  
 $<-5>_3 = \left\lceil \frac{|-5|}{3} \right\rceil \cdot 3 + (-5) = 2 \cdot 3 + (-5) = 1$ 

• The *circular* convolution can be defined using the *modulo* operator:

$$g[n] \overset{N}{\otimes} h[n] = \sum_{m=0}^{N-1} g[m]h[\langle n-m \rangle_N] \qquad 0 \le n \le N-1$$
 (8)

• The DFT has several symmetry properties that can greatly simplify the computation of the DFT:

Length-N sequence	N-point DFT
x[n]	X[k]
$x^*[n]$	$X^* \left[ < -k >_N \right]$
$\operatorname{Re}\{x[n]\}$	$X_{ccs}[k] = \frac{1}{2} [X[k] + X^*[<-k>_N]]$
$j\cdot \operatorname{Im}\{x[n]\}$	$X_{cca}[k] = \frac{1}{2} [X[k] - X^*[<-k>_N]]$
$x_{ccs}[n] = \frac{1}{2} [x [n] + x^* [\langle -n \rangle_N]]$	$\operatorname{Re}\{X[k]\}$
$x_{cca}[n] = \frac{1}{2} [x [n] - x^* [<-n>_N]]$	$j \cdot \operatorname{Im}\{X[k]\}$

• Furthermore, if x[n] is a length-N real sequence then the corresponding N-point DFT has the additional symmetry properties:

$$\begin{array}{rcl} X[k] & = & X^*[<-k>_N] \\ \operatorname{Re}\{X[k]\} & = & \operatorname{Re}\{X[<-k>_N]\} \\ \operatorname{Im}\{X[k]\} & = & -\operatorname{Im}\{X[<-k>_N]\} \\ |X[k]| & = & |X[<-k>_N]| \\ \operatorname{arg}\{X[k]\} & = & -\operatorname{arg}\{X[<-k>_N]\} \end{array}$$

• The following theorems are also very useful when computing the DFT of a length-N sequence:

Length-N sequence	N-point DFT
g[n]	G[k]
h[n]	H[K]
$\frac{\alpha g[n] + \beta h[n]}{\alpha g[n] + \beta h[n]}$	$\alpha G[k] + \beta H[k]$
$g[\langle n - n_0 \rangle_N]$	$W_N^{kn_0}G[k]$
$W_N^{-k_0n}g[n]$	$G[\langle k - k_0 \rangle_N]$
$g[n] \overset{N}{\otimes} h[n]$	G[k]H[k]
g[n]h[n]	$\frac{1}{N}G[k] \stackrel{N}{\otimes} H[k]$

• The Parseval's relation can be also useful, specially when we want to compute the energy of a sequence g[n] that has a DFT G[k]:

$$\sum_{n=0}^{N-1} |g[n]|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |G[k]|^2$$

• The circular convolution of two sequences g[n] and h[n] can be computed using the DFT:

$$\left. \begin{array}{ccc} g[n] & \stackrel{DFT}{\rightarrow} & G[k] \\ h[n] & \stackrel{DFT}{\rightarrow} & H[k] \end{array} \right\} \Longrightarrow g[n] \stackrel{N}{\otimes} h[n] = \mathrm{IDFT} \left\{ G[k] H[k] \right\}$$

• The linear convolution of a length- $L_x$  sequence x[n] and a length- $L_y$  sequence y[n] can also be computed using the DFT:

$$x[n] \otimes y[n] = x[n] \overset{N}{\otimes} y[n] = \text{IDFT} \{ \text{DFT}_N \{ x[n] \} \text{DFT}_N \{ x[n] \} \}$$

where  $N = L_x + L_y - 1$ .

• The DFTs of two length-N real sequences g[n] and h[n] can be computed using the DFT of a single complex sequence x[n] = g[n] + jh[n]:

$$\begin{cases}
g[n] & \stackrel{DFT}{\rightarrow} & G[k] \\
h[n] & \stackrel{DFT}{\rightarrow} & H[k] \\
x[n] & \stackrel{DFT}{\rightarrow} & X[k]
\end{cases}
\Longrightarrow
\begin{cases}
G[k] = \frac{1}{2}[X[k] + X^*[\langle -k \rangle_N]] \\
H[k] = \frac{1}{2j}[X[k] - X^*[\langle -k \rangle_N]]
\end{cases}$$