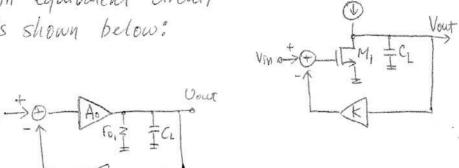
10. An equivalent circuit is shown below:



Open-loop transfer function (without feedback):
$$A(s) = \frac{A_0}{1 + \frac{s}{w_p}} = \frac{-g_{m_s} r_{o_s}}{1 + \frac{s}{r_{o_s} c_b}}$$

$$\Rightarrow Closed-loop -3aB bandwidth = B = (ro, C_L)(1+Aok) = ro, C_L(1+gm, ro, k)$$

3dB带宽指幅值等于最大值的二分之根号二倍时对应的频带宽度

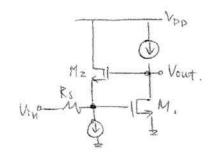
闭环传递函数
$$H(S) = -\frac{g_{m1}r_{01}/(1-g_{m1}r_{01}K)}{1+Sr_{01}c_L/(1-g_{m1}r_{01}K)}$$

$$\pm |\mathbf{H}(\mathbf{j}B)| = \frac{1}{\sqrt{2}} |\mathbf{H}(\mathbf{j}0)|$$

得到K=-
$$\frac{\mathbf{B}r_{01}c_L-1}{g_{m1}r_{01}}$$

23. First, recognize that

(a) both input & output are voltages.



* Vin primarily drives the Gate of Mi.

Sequence: Suppose Vin increases by Avin

- >> Vout drops by +gm, ADin × Fo, (Common-Source)
- Source of M2 decreases by same amount (Source follower)
- ... Vin 4 ⇒ VM,0 + ⇒ VM, q + ⇒ effective Vin driving M, q +
- > negative feedback

(b) Vin + > Vout + > Unz, 4 +

- > effective oin driving M, € \$
- =) positive feedback.

- (C) Din \$ => Dout \$ => Umi, \$ \$\div \$

 => effective Din driving Mi, \$\div \$\div \$

 => negative feedback.
- (d) Vin \$ => Voux \$ (common-base, M,)

 >> VMI, s \$

 >> effective vin driving MI, s \$

 -> regative feedback.

(Without feedback)

$$V_{DO}$$
 V_{DO}
 V_{D

$$\Rightarrow A_{C.L.} = \frac{V_{Out}}{V_{in}}\Big|_{C.L.} = \frac{A_{O.L}}{1 + A_{O.L} \cdot K} = \frac{g_m R_D}{1 + \frac{R_Z}{R_i + R_Z}} g_m R_D$$

$$R_{in,closed} = \frac{1}{g_{m_i}} \left(1 + \frac{R_z}{R_i + R_z} g_{in} R_b \right)$$

Rout, closed =
$$\frac{R_D}{1 + \frac{R_Z}{R_1 + R_2} g_M R_D}$$

$$A_{OL} = g_m R_D \| (R_1 + R_2) \| r_{01} \approx g_m R_D \| r_{01}$$

$$R_{out,Closed} = \frac{R_D || r_{01}}{1 + \frac{R_2}{R_1 + R_2}} g_m R_D$$

34.

$$R_{0L} = \frac{V_{out}}{v_{iN}}$$
 (no feedback)
 $= R_D$
 K (feedback factor)

$$K$$
 (feedback factor)
= $g_{m_2} \times \frac{R_2}{R_1 + R_2}$

$$\Rightarrow R_{C.L.} = \frac{Vout}{\overline{z_{IN}}} = \frac{R_D}{1 + R_D \times g_{M_2} \frac{R_Z}{R_I + R_Z}}$$

$$\Gamma_{\text{IN}}|_{\text{C.L.}} = \frac{|g_{m_1}|}{1 + R_D \times g_{m_2} \frac{R_2}{R_1 + R_2}}$$

$$\Gamma_{\text{Out}|C.L.} = \frac{R_D}{1 + R_D \times g_{M_2} \frac{R_2}{R_1 + R_2}}$$

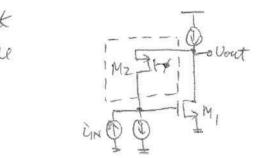
$$k: = \frac{1}{\sqrt{1}} = g_{m_2}$$

".
$$R_{CL} = \frac{R_{OL}}{1 + R_{OL}K} = \frac{(r_{oz} ll gm_z) gm_1 r_{o_1}}{1 + gm_1 gm_2 r_{o_1} (r_{oz} ll gm_z)}$$

$$K = \frac{i_x}{V_x} = -gm_2$$

$$A_{C,L} = -\frac{gm_1 r_{01} \left(\frac{1}{gm_2} \parallel r_{02}\right)}{1 + gm_1 gm_2 r_{01} \left(\frac{1}{gm_2} \parallel r_{02}\right)}$$

(b) Breaking the feedback network results in the following:



$$RoL = \frac{Oout}{\widehat{c}_{IN}}$$

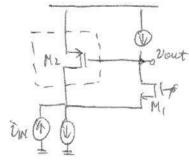
$$= -r_{02} \times g_{M_1} \left[r_{01} || r_{02} || g_{M_2} \right]$$

$$Rin, open = r_{02}$$

$$\frac{1}{12} \sum_{i=1}^{\infty} \frac{1}{v_{i}} = -g_{m_{z}}$$

$$k = \frac{i_x}{v_x} = -g_{m_z}$$

(c) Breaking the feedback network results in the following:



Now the Roll = Vout
$$Rol = \frac{Vout}{\overline{v}_{IN}}$$

$$\approx \left(\frac{1}{g_{M_1}} \| roz \right) \left(1 + g_{M_1} ro_1\right)$$

Rout, open = $roz + ro_1 \left(1 + g_{M_1} roz \right)$