

Transform-domain Approaches.

Rev

$$e^{-t} u(t) \leftrightarrow \frac{1}{s+1}$$

$$X(s) = e^{-t} u(t) \leftrightarrow ?$$

$$\tilde{X}(s) = X(s) e^{-\sigma t} \leftrightarrow \text{收敛域 ROC: region of convergence}$$

$$\tilde{X}(j\omega) = \int_{-\infty}^{+\infty} X(t) e^{-\sigma t} e^{-j\omega t} dt = \int_{-\infty}^{+\infty} X(t) e^{-(\sigma+j\omega)t} dt$$

$s = j\omega$ → FT
 $\sigma = 0$ Fourier Transform

$$\stackrel{s=\sigma+j\omega}{=} \int_{-\infty}^{+\infty} X(t) e^{-st} dt$$

Laplace transform

$$\begin{cases} e^{\alpha t} u(t) \leftrightarrow \frac{1}{s-\alpha}, \forall s \in \text{ROC}_X = \{s: \text{Re}(\alpha) < \text{Re}(s)\} \\ -e^{\alpha t} u(t) \leftrightarrow \frac{1}{s-\alpha}, \forall s \in \text{ROC}_X = \{s: \text{Re}(s) < \text{Re}(\alpha)\} \end{cases}$$

虚轴上的 Laplace → 连续时间的 FT.
单位圆上的 ZF → 离散时间的 FT.

$$X(s) = e^{-\beta t} \Rightarrow \begin{cases} X_L(s) = e^{\beta t} u(t) \leftrightarrow \frac{1}{s-\beta} & \text{ROC}_L = \{s: \text{Re}(s) < \beta\} \\ X_R(s) = e^{-\beta t} u(t) \leftrightarrow \frac{1}{s+\beta} & \text{ROC}_R = \{s: -\beta < \text{Re}(s)\} \end{cases}$$

$$X(s) = X_L(s) + X_R(s) = \frac{-2\beta}{s^2 - \beta^2} \quad \text{ROC}_X = \text{ROC}_L \cap \text{ROC}_R = \{s: -\beta < \text{Re}(s) < \beta\}$$

Properties:

① right-sided signal

② Left-sided signal

③ two-sided signal

$\begin{cases} \text{Poles 极点} \\ \text{zeros 零点} \end{cases} \quad X(s) = \frac{N(s)}{D(s)} \quad \begin{matrix} D(s)=0 \Rightarrow \text{Poles} \\ N(s)=0 \Rightarrow \text{zeros} \end{matrix}$
 ROC { 不能包含 Poles
可包含 zeros

Theorem 10:

• 初值定理 If $\frac{d^k X(t)}{dt^k} \Big|_{t=0+}$ exist for all $k=0,1,\dots$, then (0和+∞在收敛域内能用)

$$X(0+) = \lim_{s \rightarrow \infty} s X(s)$$

• 终值定理 If $\text{Re}(s) = 0 \in \text{ROC}_X$, then

$$X(+\infty) = \lim_{s \rightarrow 0} s X(s)$$

The z-Transform.

$$x[n] \leftrightarrow X(z) \triangleq \sum_{n=-\infty}^{+\infty} x[n] z^{-n}$$

$$\sum_{n=-\infty}^{+\infty} \boxed{x[n] r^{-n}} e^{-j\omega n} = \sum_{n=-\infty}^{+\infty} x[n] (r e^{j\omega})^{-n} \xrightarrow{z = r e^{j\omega}} \sum_{n=-\infty}^{+\infty} x[n] z^{-n} = X(z)$$

$$\begin{cases} \alpha^n u[n] & \leftrightarrow \frac{1}{1 - \alpha z^{-1}} & |\alpha| < |z| & \text{半径比较} \\ -\alpha^n u[-n-1] & \leftrightarrow \frac{1}{1 - \alpha z^{-1}} & |\alpha| > |z| \end{cases} \quad \delta[n] \leftrightarrow 1$$

$$\begin{cases} n \alpha^n u[n] & \leftrightarrow \frac{\alpha z^{-1}}{(1 - \alpha z^{-1})^2} & |\alpha| < |z| \\ -n \alpha^n u[-n-1] & \leftrightarrow \frac{\alpha z^{-1}}{(1 - \alpha z^{-1})^2} & |\alpha| > |z| \end{cases}$$

Transform function of LTI system.

$$x(t) = e^{pt} \rightarrow y(t) = H(p) e^{pt} \quad , \quad \forall p \in \text{ROC}_h$$

$$x[n] = c^n \rightarrow y[n] = H(c) c^n \quad , \quad \forall c \in \text{ROC}_h$$

Revisiting LTI system's stability and causality

① Causality:

<1> $y[n] \rightarrow x[m], m \leq n$

<2> $h(t) = 0, t < 0$
 $h[n] = 0, n < 0$ 小于0时h等于0

<3> $H(s)$ $ROC_h \supseteq \{s: \text{Re}(s) < +\infty\}$

$H(z)$ $ROC_h = \{z: p_r < |z| \}$



> Poles
zeros

② Stability

<1> if $|x(t)| \leq M, |y(t)| \leq N$ 都有

<2> $\int_{-\infty}^{+\infty} |h(t)| dt < M$ 绝对可积 $\sum_{n=-\infty}^{+\infty} |h[n]| < M$ 绝对可和

<3> $H(s): ROC_h \supseteq \text{Re}(s) = 0$ 包含虚轴

$H(z): ROC_h \supseteq \{ |z| = 1 \}$ 包含单位圆

Unilateral transforms 单边LT

$-\infty < t < +\infty \Rightarrow t \geq 0$

complete response $\bar{y}(t) = y(t)u_{-}(t)$

Definition: $x(t) \leftrightarrow X(s) \triangleq \int_0^{+\infty} x(t) e^{-st} dt$

Compute (i) $x_1(t) = e^{-\alpha t} u(t)$

(ii) $x_2(t) = e^{-\alpha(t+1)} u(t+1)$

(iii) $x_3(t) = e^{-\alpha(t-1)} u(t-1)$

Solution: as $x_1(t) = 0, t < 0$

$X_1(s) = X_1(s) = \frac{1}{s+\alpha}, ROC_{x_1} = \{s: -\text{Re}(\alpha) < \text{Re}(s)\}$

$x_2(t) = e^{-\alpha(t+1)} u(t+1)$

$X_2(s) = \int_{-\infty}^{+\infty} e^{-\alpha(t+1)} u(t+1) e^{-st} dt = e^{-\alpha} \int_{-\infty}^{+\infty} e^{-\alpha t} e^{-st} e^{-st} dt$

$= \frac{e^{-\alpha}}{s+\alpha}, ROC_{x_2} = \{s: -\text{Re}(\alpha) < \text{Re}(s)\}$

$\neq X_1(s) = \frac{e^{-s}}{s+\alpha}$

$x_3(t) = e^{-\alpha(t-1)} u(t-1) = 0 = x_1(t-1) \forall t < 0$

we have $X_3(s) = X_1(s) = X_1(s) e^{-s} = X_1(s) e^{-s} ROC_{x_3} = \{s: -\text{Re}(\alpha) < \text{Re}(s)\}$

分部积分 $u'v = (uv)' - uv'$

$\int_0^{+\infty} \frac{dX(t)}{dt} e^{-st} dt = X(t) e^{-st} \Big|_0^{+\infty} + s X(s) = s X(s) - X(0-)$

$(u = X(t), v = e^{-st}) \quad -(-s) \left[e^{-st} \int_0^{+\infty} X(t) dt \right] X(s)$

$\Delta 1. \frac{dX(t)}{dt} \Leftrightarrow s X(s) - X(0-)$

2. $\frac{d^2 X(t)}{dt^2} \Leftrightarrow s[s X(s) - X(0-)] - \frac{dX(t)}{dt} \Big|_{t=0-}$

Unilateral z-transform

$$x[n] \longleftrightarrow X(z) \triangleq \sum_{n=0}^{+\infty} x[n] z^{-n}$$

for a causal signal $x[n]$, i.e. $x[n] = x[n]u[n]$ then $X(z) = X(z)$

$$\star \begin{cases} x[n-1] \longleftrightarrow z^{-1}X(z) + x[-1] \\ x[n+1] \longleftrightarrow z[X(z) - x[0]] \end{cases}$$

当 $x[n]$ 只在 $n \geq 0$ 有非零值与双边相同。
而当 $x[n]$ 在 $n < 0$ 时也有非零值时，双边与单边不同。

Ex. LTI $y[n] + 2y[n-1] = x[n]$ $y[-1] = 1$, $x[n] = ku[n]$ for $n \geq 0$

Solution: $Y(z) - [z^{-1}Y(z) + y[-1]] = \frac{k}{1-z^{-1}}$

$$Y(z) = -\frac{2y[-1]}{1+2z^{-1}} + \frac{k}{(1+2z^{-1})(1-z^{-1})} \triangleq Y_{zi}(z) + Y_{zs}(z)$$

with $ROC_y = \{z: 2 < |z| < \infty\}$ $\hookrightarrow \frac{k}{3} \left(\frac{2}{1+2z^{-1}} + \frac{1}{1-z^{-1}} \right)$

$$\begin{aligned} \bar{y}[n] &= -2 \cdot 1 \cdot (-2)^n u[n] + \frac{k}{3} [2(-2)^n + 1] u[n] \\ &\triangleq y_{zi}[n] + y_{zs}[n] \end{aligned}$$