

Remember - Signal Energy:

- The Energy of a signal is defined by:

$$E_g = \int_{-\infty}^{\infty} |g(t)|^2 dt \quad * \text{ If } g(t) \text{ is complex, then } |g(t)|^2 \text{ is equivalent to } \rightarrow |g(t)|^2 = g(t) \cdot g^*(t)$$

← complex conjugate

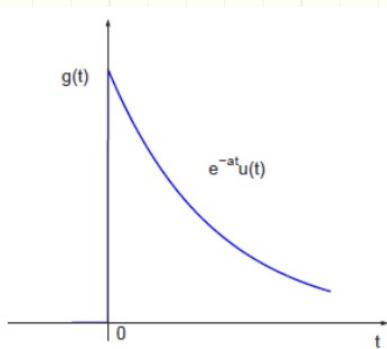
- If we know our Fourier Transform, we can also relate Energy to the Signal Spectrum

$$E_g = \int_{-\infty}^{\infty} |g(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |G(\omega)|^2 d\omega \quad \text{Parseval's Theorem}$$

Parseval's Theorem - The total energy of a signal calculated in the time domain is equal to the total energy calculated in the frequency domain.

Example ① : Parseval's Theorem

Consider the signal $g(t) = e^{-\alpha t} u(t)$ for ($\alpha > 0$), calculate its signal energy in both time and frequency domains.



$$|U(t)|^2 \rightarrow 1 \text{ for } t > 0 \\ 0 \text{ for } t \leq 0$$

Step ①: Find E_g in the time-domain

$$E_g = \int_{-\infty}^{\infty} |g(t)|^2 dt = \int_{-\infty}^{\infty} |e^{-\alpha t} u(t)|^2 dt = \int_0^{\infty} e^{-2\alpha t} dt$$

$$E_g = -\frac{1}{2\alpha} e^{-2\alpha t} \Big|_0^{\infty} = -\frac{1}{2\alpha} [e^{-\infty} - e^0] = -\frac{1}{2\alpha} [0 - 1]$$

$E_g = \frac{1}{2\alpha}$ Time - Domain Energy
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Example ① : Parseval's Theorem - ContinuedStep ②: Find E_g in the frequency-domain

$$E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} |G(\omega)|^2 d\omega$$

$$g(t) = e^{-\alpha t} u(t) \rightarrow G(\omega) = \frac{1}{\alpha + j\omega}. (\text{when } \alpha > 0)$$

$$E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left| \frac{1}{\alpha + j\omega} \right|^2 d\omega \neq |\alpha + j\omega|^2 = (\alpha + j\omega)(\alpha - j\omega) \\ \neq |\alpha + j\omega|^2 = \alpha^2 + \omega^2$$

$$E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{1}{\alpha^2 + \omega^2} d\omega \quad \neq \int \frac{1}{\alpha^2 + \omega^2} d\omega = \frac{1}{\alpha} \arctan\left(\frac{\omega}{\alpha}\right)$$

$$E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{1}{\alpha^2 + \omega^2} d\omega = \frac{1}{2\pi} \cdot \frac{1}{\alpha} \arctan\left(\frac{\omega}{\alpha}\right) \Big|_{-\infty}^{+\infty}$$

$$E_g = \frac{1}{2\pi\alpha} \left[\frac{\pi}{2} - \left(-\frac{\pi}{2} \right) \right] = \frac{1}{2\pi\alpha} [\cancel{\pi}] = \frac{1}{2\alpha}$$

$$E_g = \frac{1}{2\alpha} \quad \text{Frequency Domain}$$

$$E_g = \frac{1}{2\alpha} \quad \text{Time-Domain Energy}$$

Energy Spectral Density - How Is Energy Distributed Across Frequency?

- We can think of the **signal Energy (E_g)** as the result of energies contributed by all **spectral components**

$$E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} |G(\omega)|^2 d\omega$$

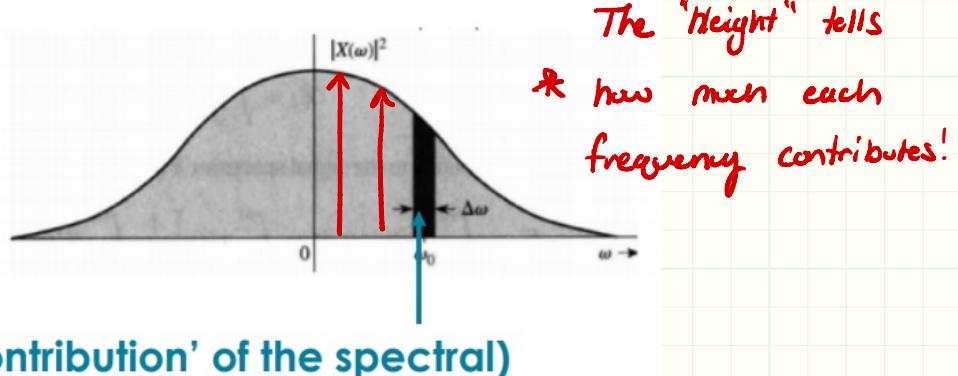
Resulting [↑]
 Total
 Energy

Energy of each individual
 Frequency Component
 "Spectrum"

- We say that **ESD** is the energy per unit bandwidth of the spectral components of $g(t)$ centered at frequency (ω)

$$\Psi(\omega) = |G(\omega)|^2 \quad \text{and} \quad E_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} \Psi(\omega) d\omega$$

- We can also call **ESD** the Spectral Contribution



- If $x(t)$ and $y(t)$ are the input and output of LTI system

$$Y(\omega) = h(\omega) X(\omega)$$

$$|Y(\omega)|^2 = |h(\omega)|^2 |X(\omega)|^2$$

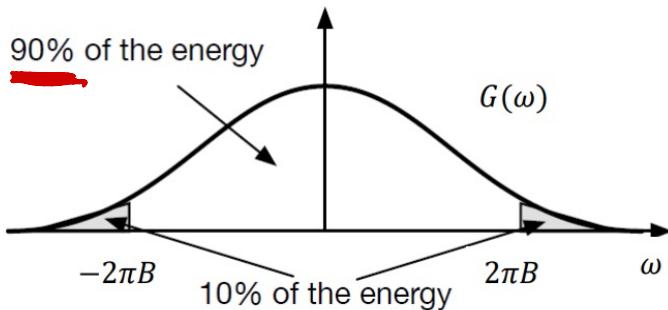
$$\Psi_y(\omega) = |h(\omega)|^2 \Psi_x(\omega)$$

$\Psi(\omega)$ - ESD

The ESD of an output signal is the multiplication of $|h(\omega)|^2$ and the ESD of the system input signal

Essential Bandwidth

- The essential bandwidth is the range of frequencies where most of the signal's energy is held.
 - ↳ "most" depends on the application (could be 90%, 95%, etc)
 - ↳ The percentage is with respect to your overall energy



- Using the picture above, if $G(\omega)$ is a low pass signal and E_B is the Energy from $-2\pi B$ to $2\pi B$, then

$$E_B = \frac{1}{2\pi} \int_{-2\pi B}^{2\pi B} |G(\omega)|^2 d\omega$$

$$\frac{E_B}{E_g} = 0.9$$

- These two equations help you find B if you know all other information

Correlation Function

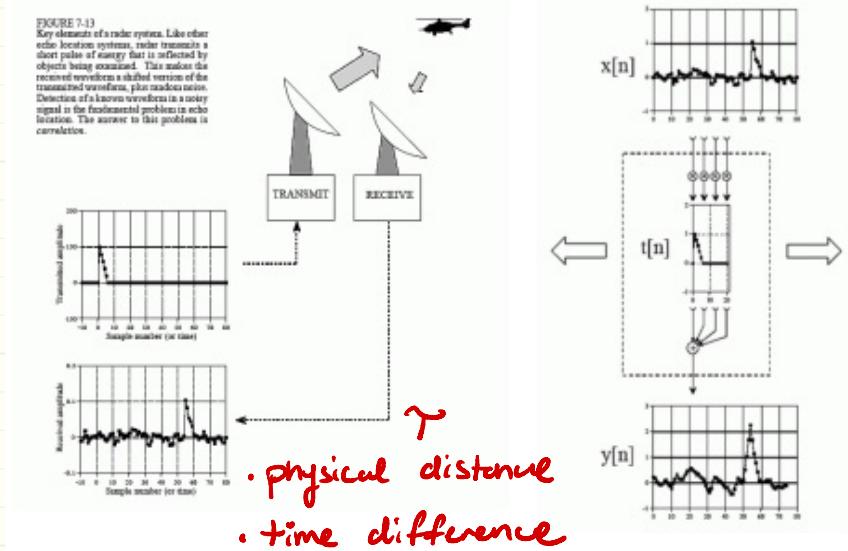
- Correlation is the measurement of **similarity** between two signals

$$R_{12} = \int_{-\infty}^{\infty} x_1(t) x_2(t-\tau) dt$$

- This function has two use cases, to find the **Time Delay** between two signals or to examine the **Similarity** of two functions

↳ If τ is zero \rightarrow Peak Correlation

- In a **RADAR** example, τ can be thought of as a **range variable**:



- This **Correlation Function** is extremely similar to Energy and ESD

$$\Psi_g(\tau) = \int_{-\infty}^{\infty} g^*(t) g(t+\tau) dt$$

Auto Correlation

$$x_1(t) = x_2(t)$$

Cross Correlation
 $x_1(t) \neq x_2(t)$

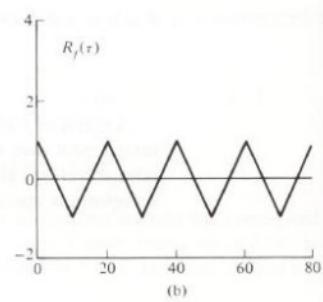
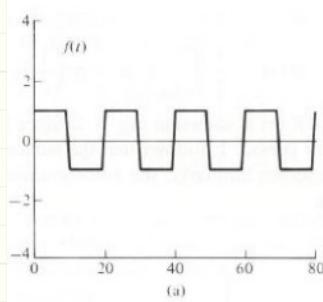
Time Delay

* more complex *

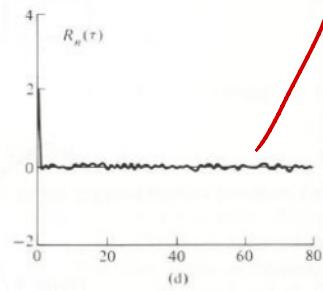
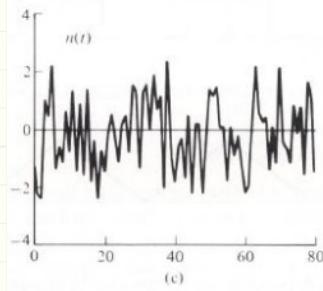
- In this class, we will mostly deal with Auto Correlation

Correlation Example

- Original Signal

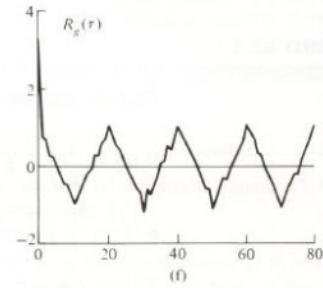
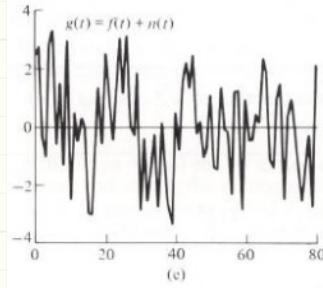


- Channel Noise



- Resulting Signal

↳ Very hard to extract any good info!



Random Noise &
low average
distribution

Auto
Correlation

Allows us
to examine
meaningful
Similarities

- Remember, Auto - Correlation and ESD \rightarrow Very Similar

$$\hat{\psi}_g(\tau)$$

Auto - Correlation

$$\hat{\psi}(w)$$

ESD

]

Transform Pair

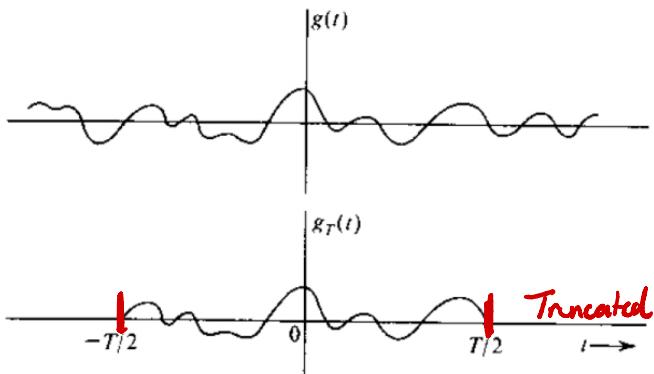
- Both of these are wanting to examine a $|G(w)|^2$ signal!

Power Spectral Density - How Is My Power Distributed Across Frequency

Recall that :

$$E_g = \int_{-\infty}^{\infty} |g(t)|^2 dt$$

- Let's say I have some truncated signal $g_T(t)$, shown below:



$$E_{gT} = \int_{-\infty}^{\infty} |g_T(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |G_T(\omega)|^2 d\omega$$

* Remember → Spectral Density means the contributions @ certain frequency

- We say that Power Spectral Density (PSD) is given as:

$$S_g(\omega) = \lim_{T \rightarrow \infty} \frac{|G_T(\omega)|^2}{T} \quad (\text{PSD})$$

* As T increases, $|G_T(\omega)|^2$ also increases

* Since P_g is finite (power signal), the increase rates of T and $|G_T(\omega)|^2$ must be the same

$$P_g = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_g(\omega) d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \lim_{T \rightarrow \infty} \frac{|G_T(\omega)|^2}{T} d\omega$$

Auto - Correlation of Power Signal

$$R_g(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\infty}^{\infty} g^*(t) g(t + \tau) dt$$

* Having a plus τ shift
in the Auto - Correlation
function is okay due
to Hermitian Symmetry
 $R_x(\tau) = R_x(-\tau)$

$$R_g(\tau) = \frac{1}{T_0} \int_0^{T_0} g^*(t) g(t + \tau) dt$$

* what we will use
the most in our course

- As with ESD previously, the auto-correlation function and PSD are a **Transform Pair**

$$R_g(\tau) \leftrightarrow S_g(\omega)$$

:-)

- Remember, we use transform pairs to help simplify !

Example ②: PSD vs. Power

Find the power spectral density (PSD) and the power of the function $x(t) = A \cos(\omega_0 t + \Theta)$ Theta

* Try and work this problem for next class *

Next class we will work examples and I
will assign Homework ① *

Example ②: PSD vs. Power

Find the power spectral density (PSD) and the power of the function $x(t) = A \cos(\omega_0 t + \theta)$ \rightarrow Theta

Step ①: Remember the PSD \leftrightarrow Autocorrelation Fourier Transform Pair

$$\text{PSD : } S_x(\omega) = \lim_{T \rightarrow \infty} \frac{|G_x(\omega)|^2}{T}$$

$$\text{Auto Correlation : } R_x(\tau) = \frac{1}{T} \int_0^{T_0} x^*(t) x(t+\tau) dt$$

Step ②: Find the Power Spectral Density (PSD)

$$S_x(\omega) = \lim_{T \rightarrow \infty} \frac{|G_x(\omega)|^2}{T}$$

* We can use $R_x(\tau)$ to find the Power Spectral Density! *
Fourier Transform Pair

$$R_x(\tau) = \frac{1}{T} \int_0^{T_0} [A \cos(\omega_0 t + \theta)] [A \cos(\omega_0 (t+\tau) + \theta)] dt$$

* Remember, $x^*(t)$ of a real function is itself.
 $(\text{Real})^* = \text{Real}$

* We can use the identity $\rightarrow \cos(a) \cos(b) = \frac{1}{2} [\cos(a-b) + \cos(a+b)]$

$$R_x(\tau) = \frac{1}{T} \int_0^{T_0} A^2 \cdot \frac{1}{2} [\cos(a-b) + \cos(a+b)] dt \quad \left| \begin{array}{l} a = [\omega_0 t + \theta] \\ b = [\omega_0 (t+\tau) + \theta] \end{array} \right.$$

$$\begin{aligned} a-b &= [\omega_0 t + \theta] - [\omega_0 (t+\tau) + \theta] \\ &= \cancel{\omega_0 t + \theta} - \cancel{\omega_0 t} - \omega_0 \tau - \cancel{\theta} = -\omega_0 \tau \end{aligned}$$

$$\begin{aligned} a+b &= [\omega_0 t + \theta] + [\omega_0 (t+\tau) + \theta] \\ &= \omega_0 t + \theta + \omega_0 t + \omega_0 \tau + \theta = 2\omega_0 t + \omega_0 \tau + 2\theta \end{aligned}$$

$$R_x(\tau) = \frac{1}{T} \int_0^{T_0} \frac{A^2}{2} [\cos(-\omega_0 \tau) + \cos(2\omega_0 t + \omega_0 \tau + 2\theta)] dt$$

Example ②: PSD vs. Power - Continued

$$R_x(\tau) = \frac{1}{T} \int_0^{T_0} \frac{A^2}{2} [\cos(-\omega_0 \tau) + \cos(2\omega_0 t + \omega_0 \tau + 2\phi)] dt$$

Even Function
 $\cos(\omega_0 \tau)$

$$R_x(\tau) = \frac{1}{T} \int_0^{T_0} \frac{A^2}{2} \cos(\omega_0 \tau) dt + \frac{1}{T} \int_0^{T_0} \frac{A^2}{2} \cos(2\omega_0 t + \omega_0 \tau + 2\phi) dt$$

"Constants" so
integral is 0

$$R_x(\tau) = \frac{1}{T} \cdot \frac{A^2}{2} \cos(\omega_0 \tau) t \Big|_0^{T_0} + \frac{1}{T} \cdot 0$$

Integral is area under the curve will be zero - when looking over one period

$$R_x(\tau) = \frac{A^2}{2T} \cos(\omega_0 \tau) T_0 - \frac{A^2}{2T} \cos(\omega_0 \tau) 0 + 0$$

$$R_x(\tau) = \frac{A^2}{2T} \cos(\omega_0 \tau) \cancel{T_0} = \frac{A^2}{2} \cos(\omega_0 \tau)$$

PSD $S_x(\omega) \leftrightarrow \tilde{R}_x(\tau)$

$$S_x(\omega) = \tilde{\tau} \left[\frac{A^2}{2} \cos(\omega_0 \tau) \right] = \frac{A^2}{2} \cdot \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$S_x(\omega) = \frac{A^2}{2} \cdot \pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

Step ③: Find the Power (P_x)

$$P_x = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_x(\omega) d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{A^2 \pi}{2} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] d\omega$$

$$P_x = \frac{A^2 \pi}{4\pi} \int_{-\infty}^{\infty} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] d\omega = \frac{A^2}{4} (1+1)$$

$$P_x = \frac{A^2}{2}$$

$$\boxed{\text{Power } (P_x) = \frac{A^2}{2}}$$

& Area of $\delta(\omega)$ is 1