EE402 - Discrete Time Systems

Spring 2018

Lecture 13

Lecturer: Asst. Prof. M. Mert Ankarali

Matrix Exponential, e^{At}

Let's first review the matrix exponential, e^{At} . Let $t \in \mathbb{R}$ and $A \in \mathbb{R}^{n \times n}$, then e^{At} defined as

$$e^{At} := I + At + \frac{1}{2!}A^2t^2 + \frac{1}{3!}A^3t^3 + \cdots$$
$$e^{At} := \sum_{k=0}^{\infty} \frac{t^k}{k!}A^k$$

which converges for all $t \in \mathbb{R}$ and $A \in \mathbb{R}^{n \times n}$.

Now let's review some properties

• Claim:

$$\frac{d}{dt}e^{At} = Ae^{At} = e^{At}A$$

Proof:

$$\frac{d}{dt}e^{At} = \frac{d}{dt}\left(\sum_{k=0}^{\infty} \frac{t^k}{k!} A^k\right) = \sum_{k=0}^{\infty} k \frac{t^{k-1}}{k!} A^k = \sum_{k=1}^{\infty} \frac{t^{k-1}}{(k-1)!} A^k$$

$$= \sum_{n=0}^{\infty} \frac{t^n}{n!} A^{n+1} = A \sum_{n=0}^{\infty} \frac{t^n}{n!} A^n = \left(\sum_{n=0}^{\infty} \frac{t^n}{n!} A^n\right) A$$

$$\frac{d}{dt} e^{At} = A e^{At} = e^{At} A$$

• Claim: Let $t_1, t_2 \in \mathbb{R}$ then

$$e^{At_1}e^{At_2} = e^{At_2}e^{At_1} = e^{A(t_1+t_2)}$$

Proof:

$$e^{At_1}e^{At_2} = \left(\sum_{k=0}^{\infty} \frac{t_1^k}{k!} A^k\right) \left(\sum_{j=0}^{\infty} \frac{t_2^j}{j!} A^j\right) = \sum_{k=0}^{\infty} \sum_{j=0}^{\infty} \frac{t_1^k}{k!} \frac{t_2^j}{j!} A^{k+j}$$

Let n = k + j and j = n - k, then

$$e^{At_1}e^{At_2} = \sum_{k=0}^{\infty} \sum_{n=k}^{\infty} \frac{t_1^k \ t_2^{n-k}}{k!(n-k)!} A^n = \sum_{k=0}^{\infty} \sum_{n=k}^{\infty} \frac{t_1^k \ t_2^{n-k}}{n!} \frac{n!}{k!(n-k)!} A^n = \sum_{k=0}^{\infty} \sum_{n=k}^{\infty} \frac{t_1^k \ t_2^{n-k}}{n!} \left(\begin{array}{c} n \\ k \end{array}\right) A^n$$

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Since for
$$n < k$$
, $\binom{n}{k} = 0$,

$$e^{At_1}e^{At_2} = \sum_{k=0}^{\infty} \sum_{n=0}^{\infty} \frac{t_1^k t_2^{n-k}}{n!} \binom{n}{k} A^n = \sum_{n=0}^{\infty} \frac{A^n}{n!} \sum_{k=0}^{\infty} t_1^k t_2^{n-k} \binom{n}{k}$$

Using binomial theorem we find

$$e^{At_1}e^{At_2} = \sum_{n=0}^{\infty} \frac{A^n}{n!} (t_1 + t_2)^n$$
$$e^{At_1}e^{At_2} = e^{A(t_1 + t_2)}$$

Now let $t_1 = t$ and $t_2 = -t$, then we have

$$e^{At}e^{-At} = e^{A(t-t)} = I \quad \rightarrow \quad (e^{At})^{-1} = e^{-At}$$

• Claim: Let $A, B \in \mathbb{R}^{n \times n}$ and AB = BA, then

$$e^{At}e^{Bt} = e^{Bt}e^{At} = e^{(A+B)t}$$

Proof: Possible mini project question

Note that if $AB \neq BA$ then

$$e^{At}e^{Bt} \neq e^{(A+B)t}$$

• Claim: Let $P \in \mathbb{R}^{n \times n}$ and $\det(P) \neq 0$, then

$$e^{\left(P^{-1}AP\right)t} = P^{-1}e^{At}P$$

Proof: Possible mini project question

Solution of CT State-Space Equations

CT state-space representation has the form

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$\dot{y}(t) = Cx(t) + Du(t)$$

where Let $x(t) \in \mathbb{R}^n$, $y(t) \in \mathbb{R}^p$, $u(t) \in \mathbb{R}^q$

First consider the homogeneous solution, i.e. u(t) = 0 and $x(0) = x_0$.

$$\dot{x}(t) = Ax(t) \quad , \quad x(0) = x_0$$

$$\dot{y}(t) = Cx(t)$$

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Let's test if $x(t) = e^{At}x_0$ is a solution of the homogeneous equation

$$x(0) = e^{A0}x_0 = x_0$$
$$\dot{x}(t) - Ax(t) = (Ae^{At})x_0 - Ae^{At}x_0 = 0$$

Now let's compute the forced response. First let's analyze the following derivative

$$\frac{d}{dt} \left[e^{-At} x(t) \right] = e^{-At} \dot{x}(t) - e^{-At} A x(t) = e^{-At} \left[\dot{x}(t) - A x(t) \right]$$

Now using this relation let's solve the state-space equations

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$\dot{x}(t) - Ax(t) = Bu(t)$$

$$e^{-At} \left[\dot{x}(t) - Ax(t) \right] = e^{-At} Bu(t)$$

$$\frac{d}{dt} \left[e^{-At} x(t) \right] = e^{-At} Bu(t)$$

$$e^{-At} x(t) = x(0) + \int_0^t e^{-A\tau} Bu(\tau) d\tau$$

$$x(t) = e^{-At} x_0 + \int_0^t e^{A(t-\tau)} Bu(\tau) d\tau$$

Thus the solution of a system in state-space form can be written as

$$x(t) = e^{-At}x_0 + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau$$

$$y(t) = Ce^{-At}x_0 + \int_0^t Ce^{A(t-\tau)}Bu(\tau)d\tau + Du(t)$$

The function $\Psi(t) = e^{At}$ is called the state-transition matrix of the system.

Example: Let's assume that system is a SISO system and $u(t) = \delta(t)$ (unit-impulse function) and $x_0 = 0$, compute the impulse response of the system, i.e. y(t) = h(t),

$$h(t) = \int_{0}^{t} Ce^{A(t-\tau)}B\delta(\tau)d\tau + D\delta(t)$$
$$= Ce^{At}B + D\delta(t)$$

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S-Domain Solution of CT State-Space Equations

First take the Laplace transform of state evaluation equation

$$\mathcal{L}[\dot{x}(t)] = \mathcal{L}[Ax(t) + Bu(t)]$$

$$sX(s) - x(0) = AX(s) + BU(s)$$

$$[sI - A]X(s) = x(0) + BU(s)$$

$$X(s) = [sI - A]^{-1}x_0 + [sI - A]^{-1}BU(s)$$

$$Y(s) = C[sI - A]^{-1}x_0 + [C[sI - A]^{-1}B + D]U(s)$$

If we relate time and s-domain solutions we obtain

$$e^{At} = \mathcal{L}^{-1} [[sI - A]^{-1}]$$

 $h(t) = \mathcal{L}^{-1} [C[sI - A]^{-1}B + D]$

Discretization of CT State-Space Equations

Consider the CT system with the given state-space representation

$$\dot{x}(t) = Ax(t) + Bu(t)$$
$$y(t) = Cx(t) + Du(t)$$

and suppose that the input is piece-wise constant over intervals of length T. That is

$$u(t) = u[k]$$
 , $t \in (kT (k+1)T)$

i.e. input of the system is the output of a ZOH operator. Let's derive the DT state-space equations with respect to the sampled-state x[k] = x(kT).

Let's start with the state evolution equation

$$x(t) = e^{At}x(0) + \int_{0}^{t} e^{A(t-\tau)}Bu(\tau)d\tau$$

It is obvious that due to time-invariant the initial time of the equation above can be generalized as

$$x(t) = e^{A(t-t_0)}x(t_0) + \int_{t_0}^t e^{A(t-\tau)}Bu(\tau)d\tau$$
 , $t > t_0$

Now let $t_0 = kT$ and t = (k+1)T,

$$x((k+1)T) = e^{AT}x(kT) + \int_{kT}^{(k+1)T} e^{A((k+1)T - \tau)} Bu(\tau)d\tau$$

Since the u(t) = u[k] in this time interval

$$x[k+1] = e^{AT}x[k] + \int_{kT}^{(k+1)T} e^{A((k+1)T-\tau)}Bd\tau \ u[k]$$

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Let $\lambda = (k+1)T - \tau$, then

$$x[k+1] = e^{AT}x[k] - \int_{T}^{0} e^{A\lambda}Bd\lambda \ u[k]$$
$$= \left[e^{AT}\right]x[k] + \left[\left(\int_{0}^{T} e^{A\lambda}d\lambda\right)B\right]u[k]$$

Given that

$$x[k+1] = Gx[k] + Hu[k]$$

G and H matrices can be extracted as

$$G = e^{AT}$$

$$H = \left(\int_{0}^{T} e^{A\lambda} d\lambda\right) B$$

Claim: If A is invertable then we also have

$$H = A^{-1} (e^{AT} - I) B = (e^{AT} - I) A^{-1} B$$

Claim: Mini Project 6

Now let's consider the output equation

$$y(t) = Cx(t) + Dx(t)$$
$$y(kT) = Cx(kT) + Dx(kT)$$
$$y[k] = Cx[k] + Dx[k]$$

It can be seen that output equation matrices are not affected from the discretization.

Example: Consider the following CT-plant transfer function.

$$\frac{Y(s)}{U(s)} = \frac{1}{s} + \frac{1}{s + \ln(2)}$$

Find a CT state-space representation for this system

Solution:

$$\dot{x}(t) = \begin{bmatrix} 0 & 0 \\ 0 & -\ln(2) \end{bmatrix} x(t) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(t)$$
$$y(t) = \begin{bmatrix} 1 & 1 \end{bmatrix} x(t)$$

Compute the state-transition matrix

Solution:

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$$e^{At} = \begin{bmatrix} e^{0t} & 0\\ 0 & e^{-\ln(2)t} \end{bmatrix}$$
$$= \begin{bmatrix} 1 & 0\\ 0 & 0.5^t \end{bmatrix}$$

Discretize the CT State-Space equation under zero hold operation and ideal sampling of the defined state variables, with T=1s.

Solution:

$$G = e^{AT} = \begin{bmatrix} 1 & 0 \\ 0 & 0.5 \end{bmatrix}$$

$$H = \begin{pmatrix} \int_{0}^{T} e^{A\lambda} d\lambda \\ 0 \end{pmatrix} B = \begin{pmatrix} \int_{0}^{1} \begin{bmatrix} 1 & 0 \\ 0 & 0.5^{\lambda} \end{bmatrix} d\lambda \\ H = \begin{bmatrix} 1 \\ 0.721 \end{bmatrix}$$

Full DT state-space formulation takes the form

$$x[k+1] = \begin{bmatrix} 1 & 0 \\ 0 & 0.5 \end{bmatrix} x[k] + \begin{bmatrix} 1 \\ 0.721 \end{bmatrix} u[k]$$
$$y[k] = \begin{bmatrix} 1 & 1 \end{bmatrix} x[k]$$

Compute the DT pulse transfer function Y(z)/U(z) Solution:

$$\begin{split} \frac{Y(z)}{R(z)} &= C \left[zI - G \right]^{-1} H \\ &= \left[\begin{array}{cc} 1 & 1 \end{array} \right] \left[zI - G \right]^{-1} \left[\begin{array}{cc} 1 \\ 0.721 \end{array} \right] \\ &= \left[\begin{array}{cc} 1 & 1 \end{array} \right] \left[\begin{array}{cc} z - 1 & 0 \\ 0 & z - 0.5 \end{array} \right]^{-1} \left[\begin{array}{cc} 1 \\ 0.721 \end{array} \right] \\ &= \left[\begin{array}{cc} 1 & 1 \end{array} \right] \left[\begin{array}{cc} \frac{1}{z - 1} & 0 \\ 0 & \frac{1}{z - 0.5} \end{array} \right] \left[\begin{array}{cc} 1 \\ 0.721 \end{array} \right] \\ &= \frac{1}{z - 1} + \frac{0.721}{z - 0.5} \end{split}$$

Now discretize Y(s)/U(s) directly under ZOH operation

Solution:

$$\frac{Y(z)}{U(z)} = \mathcal{Z}\left[\frac{1 - e^{-s}}{s} \frac{Y(s)}{U(s)}\right] = \frac{1}{z - 1} + \frac{0.721}{z - 0.5}$$