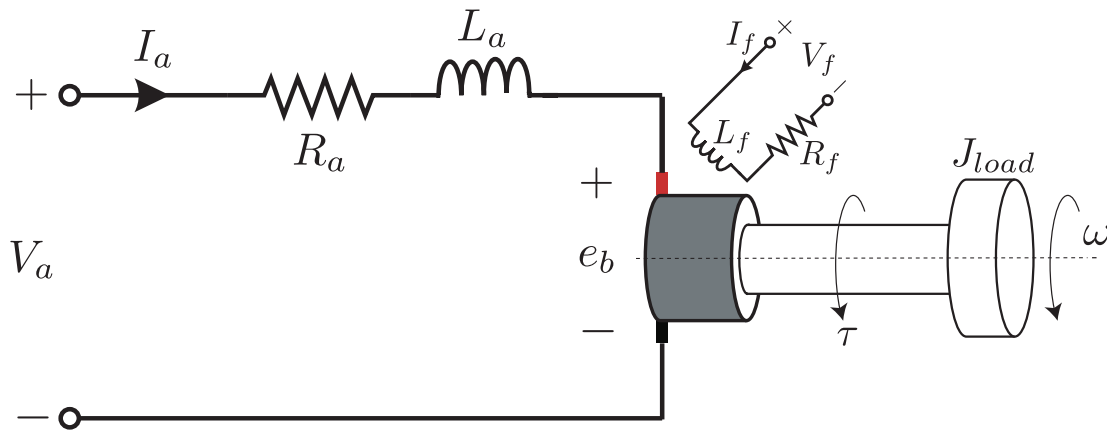


## Lecture 5

Lecturer: Asst. Prof. M. Mert Ankarali

## 5.1 DC Motor Modeling

A general “ideal” DC motor can be modeled as in the Figure below.



The dependent and “independent” variables associated with the idealized DC motor model and important relations/equations regarding the electro-mechanical interactions are given below.

$V_a$	Armature voltage
$i_a$	Armature current
$V_f$	“Field voltage”
$i_f$	“Field current”
$V_b$	Back emf
$\omega$	Rotor angular velocity
$\tau$	Generated torque
$\Phi$	Air-gap magnetic flux

$$\begin{aligned}\Phi(t) &= K_f I_f(t) \\ \tau(t) &= K_m \Phi(t) I_a(t) \\ e_b(t) &= K_b \omega(t)\end{aligned}$$

Note that if both  $i_f(t)$  and  $i_a(t)$  are non-constant the electric-motor model won't be LTI. In order to have an LTI representation, there are two options

- Armature controlled DC motor:  $\Phi$  is kept constant

$$\tau(t) = K_m \Phi I_a(t) = K_\tau^a I_a(t)$$

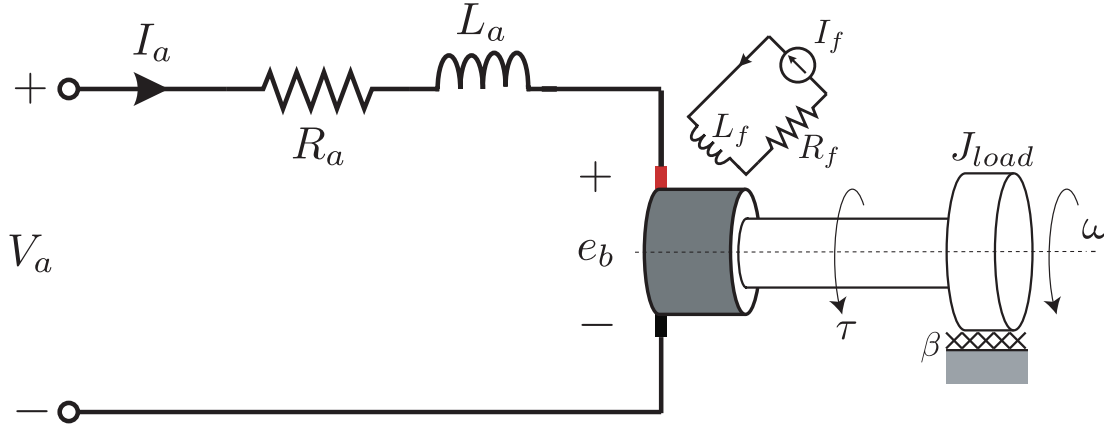
- Field controlled DC motor:  $i_a$  is kept constant

$$\tau(t) = K_m K_f I_a I_f(t) = K_\tau^f I_f(t)$$

### 5.1.1 Armature Controlled DC Motor

Majority of “DC” Motors are controlled (and indeed manufactured) with this approach. Either there is a permanent magnet which satisfies the constant  $\Phi$  or a constant current is supplied through the coils that generates magnetic field.

Let's model the following electro-mechanical system where the DC motor is armature controlled and given that  $y(t) = \omega(t)$  and  $u(t) = V_a(t)$ .



We already know the transfer function of the mechanical sub system:

$$\Omega(s) = \frac{1}{Js + \beta} \mathcal{T}(s)$$

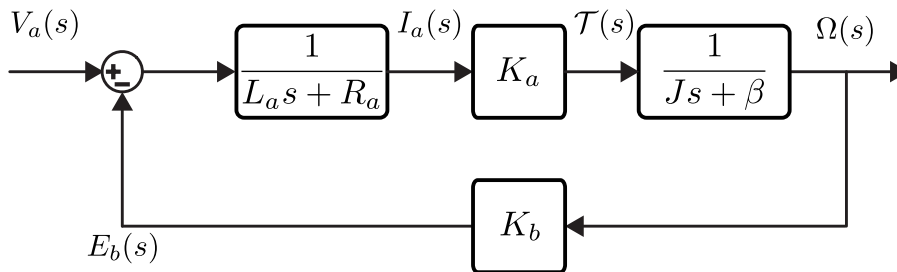
Now let's write the remaining equations in Laplace domain

$$I_a(s) = \frac{1}{L_a s + R_a} (V_a(s) - E_b(s))$$

$$\mathcal{T}(s) = K_a I_a(s)$$

$$E_b(s) = K_b \Omega(s)$$

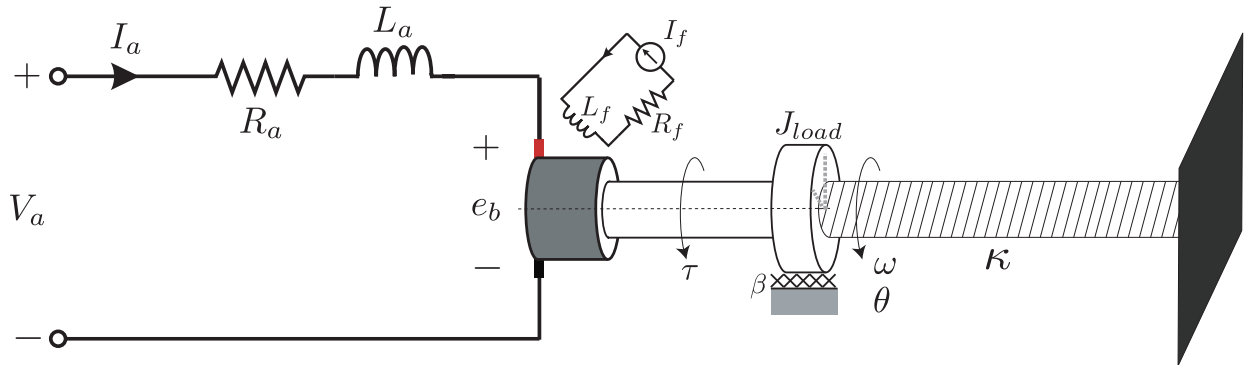
where  $K_a \equiv K_\tau^a$ . Now let's build a block-diagram topology



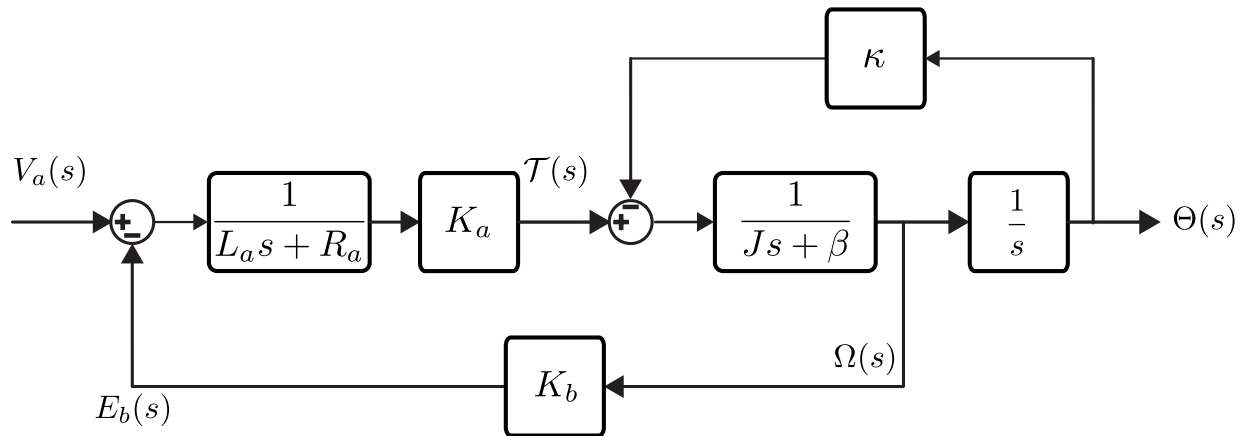
If we simplify the block diagram, we obtain the transfer function form

$$\begin{aligned} G(s) &= \frac{\Omega(s)}{V_a(s)} = \frac{\frac{K_a}{(L_a s + R_a)(J s + \beta)}}{1 + \frac{K_a K_b}{(L_a s + R_a)(J s + \beta)}} \\ &= \frac{K_a}{L_a J s^2 + (L_a \beta + R_a J) s + R_a \beta} \end{aligned}$$

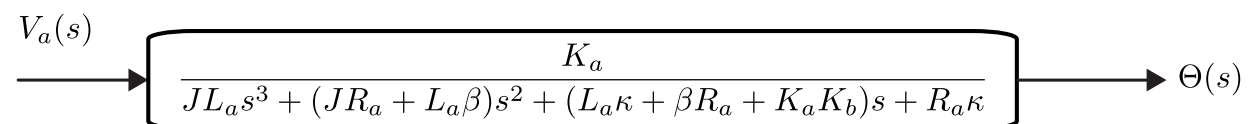
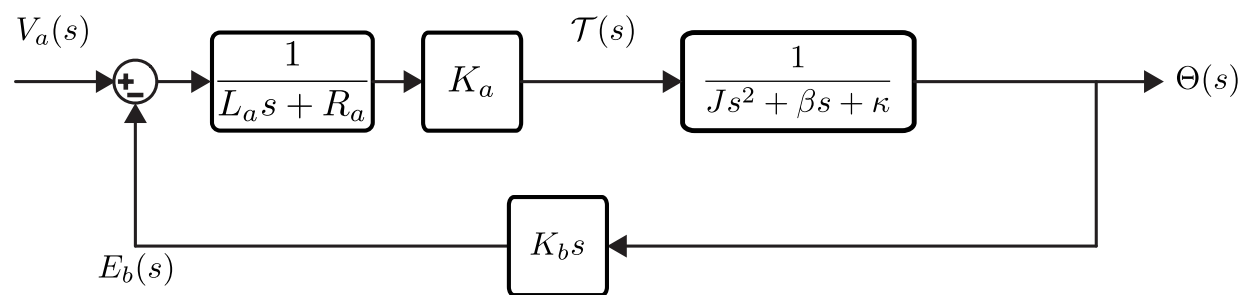
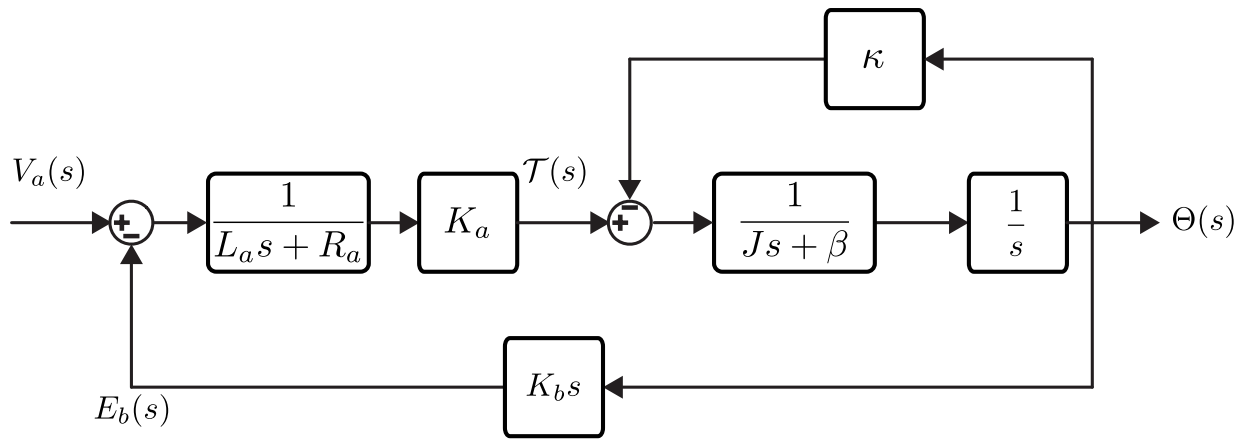
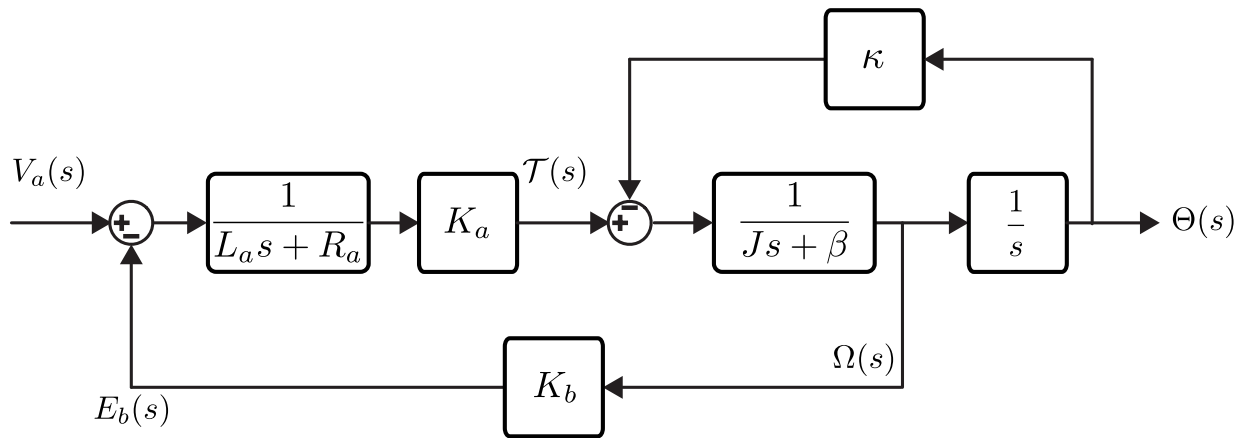
**Example 1:** Given that  $V_a$  is the input and  $\theta$  is the output, construct a block-diagram for the following electro mechanical system and then compute the transfer function.



**Solution:** A block diagram topology can be constructed by modifying the previous block diagram (armature controlled DC motor without torsional spring).



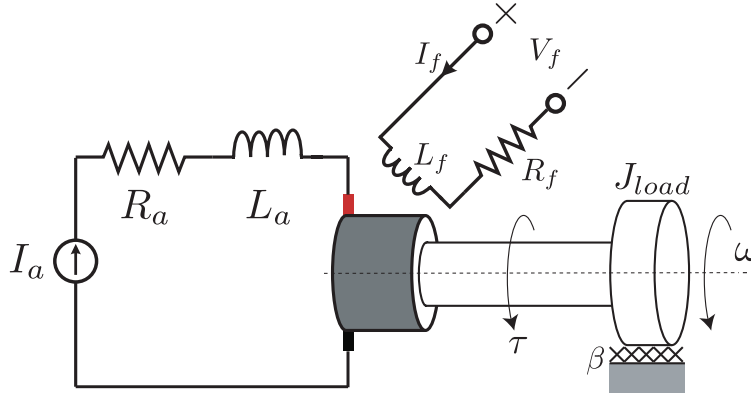
Then the transfer function can be derived using block-diagram simplification methods as given in the next page



### 5.1.2 Field Controlled DC Motor

In the field controlled DC motors, magnetic flux is actively controlled by adjusting electrical current/voltage. We assume that  $I_a$  is constant (LTI constraints). Since, there is no “feedback” in this field controlled DC motor model, the electrical circuit is isolated from the mechanical one.

Let's model the following electro-mechanical system where the DC motor is field controlled and given that  $y(t) = \omega(t)$  and  $u(t) = V_f(t)$ .

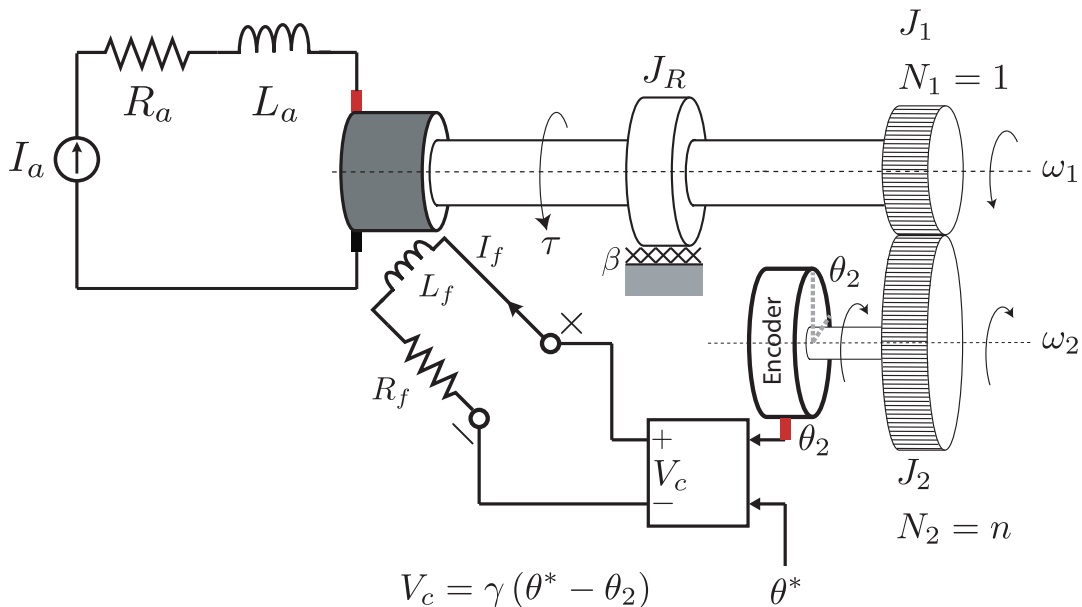


$$\begin{aligned}\frac{\Omega(s)}{\mathcal{T}(s)} &= \frac{1}{Js + \beta} \\ \frac{I_f(s)}{V_f(s)} &= \frac{1}{L_f s + R_f} \\ \mathcal{T}(s) &= K_f I_f(s)\end{aligned}$$

where  $K_f \equiv K_{tau}^f$ . Finally transfer function can be computed as

$$G(s) = \frac{K_f}{JL_f s^2 + (JR_f + \beta L_f)s + (\beta R_f + K_f)}$$

**Example 2:** Consider the following closed-loop field controlled electro-mechanical circuit. It is given that  $\theta^*(t)$ , i.e. reference angle signal, is the input and  $\theta_2$ , angular displacement of the second gear, is the output. In the system, there is an encoder which reads the angular displacement and sends it to a controller box. The other input of this box is the reference signal. The box produces an output voltage,  $V_c = \gamma(\theta^* - \theta_2)$ , and feeds it to the input terminal of the  $V_f$ . Compute the transfer function.



**Solution:** A block diagram topology can be constructed by modifying the previous block diagram (armature controlled DC motor without torsional spring).

Let's first find a transfer function from  $\tau$  to  $\omega_2$  and  $\theta_2$ . The easiest way of computing this is using the concept of reflected inertia, damping, and torque.

$$\Omega_2(s) = \frac{\bar{\mathcal{T}}(s)}{J_T s + \beta_T} = \frac{n}{(n^2 J_R + n^2 J_1 + J_2) s + n^2 \beta} \mathcal{T}(s)$$

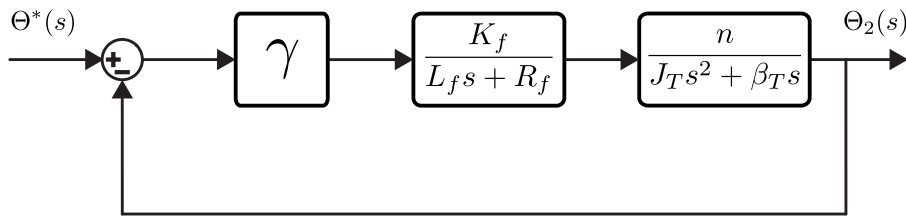
$$\Theta_2(s) = \frac{n \mathcal{T}(s)}{J_T s^2 + \beta_T s}$$

We know that Laplace domain equations for remaining parts take the form

$$\frac{\mathcal{T}(s)}{V_f(s)} = \frac{K_f}{L_f s + R_f}$$

$$V_f(s) = \gamma (\Theta^*(s) - \Theta_2(s))$$

Now let's construct a block diagram representation



Finally, transfer function can be computed as

$$G(s) = \frac{\gamma K_f n}{J_T L_f s^3 + (J_T R_f + \beta_T L_f) s^2 + \beta_T R_f s + \gamma K_f n}$$