

## Model Paper

Subject:- Basic Signals & Systems

Subject Code:- REE303

Prepared by:- Aditya Chaudhary & Apoorv Vats.

### Section - A.

Q1. Differentiate between Fourier Series and Fourier Transform.

Sol.

Fourier Series	Fourier Transform.
----------------	--------------------

1) It is the representation of non-sinusoidal periodic signals in terms of complex exponential or in terms of sine & cosine terms.

1) It is the representation of non-periodic signals in frequency domain.

2) Trigonometric Fourier Series:-

$$x(t) = a_0 + \sum_{n=1}^{\infty} [a_n \cos(n\omega_0 t) + b_n \sin(n\omega_0 t)]$$

2) Fourier Transform

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt.$$

Q2. Write Dirichlet Conditions for existence of Fourier series.

Sol. Dirichlet Conditions are:-

i) If  $x(t)$  is discontinuous, there are a finite number of discontinuities in the period  $T$ .

ii)  $x(t)$  has a finite average value over the period  $T$ .

iii)  $x(t)$  has a finite number of positive and negative maxima in the period  $T$ .

Q3. Find the Fourier Transform of  $e^{-at} u(t)$ .

Sol.

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt.$$

$$x(t) = e^{-at} u(t)$$

$$\therefore X(\omega) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-j\omega t} dt$$

$$\therefore X(\omega) = \int_0^{\infty} e^{-(a+j\omega)t} dt.$$

$$= \frac{-1}{a+j\omega} \left[ e^{-(a+j\omega)t} \right]_0^{\infty}$$

$$= \frac{1}{a+j\omega}.$$

$$\Rightarrow e^{-at} u(t) \xleftrightarrow{F} \frac{1}{a+j\omega}$$

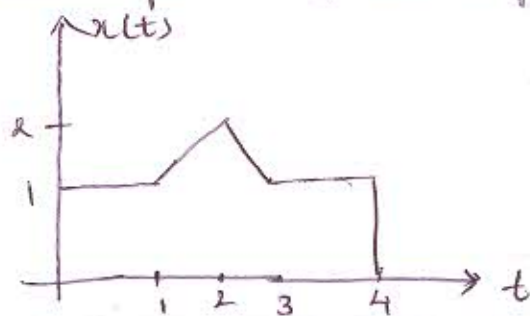
Q4. Find the Laplace transform of  $(1+0.5 \sin(t)) \sin(1000t) u(t)$

Sol.

$$\begin{aligned} x(t) &= (1+0.5 \sin(t)) \sin(1000t) u(t) \\ &= \sin(1000t) u(t) + 0.5 \sin(t) \sin(1000t) u(t) \\ &= \sin(1000t) u(t) + \frac{1}{4} (\cos(999t) - \cos(1001t)) u(t) \\ &= \sin(1000t) u(t) + \frac{1}{4} \cos(999t) u(t) - \frac{1}{4} \cos(1001t) u(t) \end{aligned}$$

$$\therefore L[x(t)] = \frac{1000}{s^2+1000^2} + \frac{1}{4} \frac{s}{s^2+999^2} - \frac{1}{4} \frac{s}{s^2+1001^2}$$

Q5. Find the Laplace transform of the signal shown in figure



Sol.

$$x(t) = u(t) + u(t-1) - 2u(t-2) + u(t-3) - u(t-4)$$

$$\begin{aligned} L[x(t)] &= \frac{1}{s} + \frac{1}{s^2} e^{-s} - \frac{2}{s^2} e^{-2s} + \frac{1}{s^2} e^{-3s} - \frac{1}{s} e^{-4s} \\ &= \left( \frac{1-e^{-4s}}{s} \right) + \left( \frac{e^{-s} - 2e^{-2s} + e^{-3s}}{s^2} \right) \end{aligned}$$



Q6. Prove the frequency shifting property and time scaling property of Fourier Series.

Sol. Frequency Shifting Property:-

$$\text{if } x(t) \longleftrightarrow X_n \text{ then } y(t) = e^{jm\omega_0 t} x(t)$$

has Fourier series co-efficient as  $Y_n = X_{n-m}$

$$\begin{aligned} \text{Proof:- } Y_n &= \frac{1}{T} \int_0^T y(t) e^{-jn\omega_0 t} dt \\ &= \frac{1}{T} \int_0^T e^{jm\omega_0 t} x(t) e^{-jn\omega_0 t} dt \\ &= \frac{1}{T} \int_0^T x(t) e^{-j(n-m)\omega_0 t} dt \\ &= X_{n-m} \end{aligned}$$

Time Scaling Property:-

$$\text{if } x(t) \longleftrightarrow X_n, \text{ then } y(t) = x(at) \longleftrightarrow Y_n = X_n.$$

Proof:- if  $x(t)$  is periodic with period  $T$ , then  $x(at)$  will have a period as  $T/a$  &  $\omega_0 \rightarrow a\omega_0$ .

$$Y_n = \frac{1}{T/a} \int_0^{T/a} x(at) e^{-jn\omega_0 t} dt.$$

$$\text{Let } at = z \quad \& \quad dz = a dt.$$

$$Y_n = \frac{1}{T} \int_0^T x(z) e^{-jn\omega_0 z} dz.$$

$$Y_n = X_n.$$

Q7. State and prove the duality property of Fourier transform. Using duality property, find the Fourier transform of  $g(t) = \frac{1}{1+jt}$ .

Sol. (a) Duality property :- if  $x(t) \longleftrightarrow X(\omega)$   
then  $X(t) \longleftrightarrow 2\pi x(-\omega)$

Proof :- By Fourier Transform definition

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

$$2\pi x(t) = \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

Replacing  $t$  by  $-t \Rightarrow 2\pi x(-t) = \int_{-\infty}^{\infty} X(\omega) e^{-j\omega t} d\omega$

Interchanging the variables  $t$  &  $\omega$

$$\Rightarrow 2\pi x(-\omega) = \int_{-\infty}^{\infty} X(t) e^{-j\omega t} dt$$

$$\Rightarrow F[X(t)] = -2\pi x(-\omega)$$

(b)  $F[e^{-at} u(t)] = \frac{1}{a+j\omega}$

For  $a=1$  ;  $F[e^{-t} u(t)] = \frac{1}{1+j\omega}$

$$e^{-t} u(t) \longleftrightarrow \frac{1}{1+j\omega}$$

$$x(t) = e^{-t} u(t) \longleftrightarrow X(\omega) = \frac{1}{1+j\omega}$$

$$x(-\omega) = e^{\omega} u(-\omega) \longleftrightarrow X(t) = \frac{1}{1+jt}$$

By duality property ;  $F[X(t)] = 2\pi x(-\omega)$

$$F\left[\frac{1}{1+jt}\right] = 2\pi e^{\omega} u(-\omega)$$

Q8. Using Laplace transform solve the following differential

eq.  $\frac{d^2 y(t)}{dt^2} + 5 \frac{dy(t)}{dt} + 4y(t) = 2x(t)$  ;  $x(t) = e^{-2t} u(t)$  &  $y(0^-) = -2$  &  $\frac{dy(0^-)}{dt} = -1$



Sol:  $L[y''(t)] = s^2 Y(s) - s(y(0^-)) - y'(0^-)$

$$L[y'(t)] = s Y(s) - y(0^-)$$

$$\frac{d^2 y(t)}{dt^2} + 5 \frac{dy(t)}{dt} + 4y(t) = 2x(t)$$

Taking Laplace both the sides.

$$s^2 Y(s) - s(-2) - (-1) + 5[s Y(s) - (-2)] + 4Y(s) = 2X(s)$$

$$Y(s) [s^2 + 5s + 4] + 2s + 11 = 2X(s)$$

$$Y(s) = \frac{2X(s)}{s^2 + 5s + 4} - \frac{2s + 11}{s^2 + 5s + 4}$$

$$x(t) = e^{-2t} u(t)$$

$$\therefore X(s) = \frac{1}{s+2}$$

$$\Rightarrow Y(s) = \frac{2}{(s+2)(s^2 + 5s + 4)} - \frac{2s + 11}{(s^2 + 5s + 4)}$$

$$= \frac{2}{(s+2)(s+1)(s+4)} - \frac{(2s+11)}{(s+1)(s+4)}$$

$$= 2 \left[ \frac{1/3}{(s+1)} - \frac{1/2}{(s+2)} + \frac{1/6}{(s+4)} \right] - 2 \left[ \frac{3/2}{(s+1)} - \frac{1/2}{(s+4)} \right]$$

Taking inverse Laplace transform.

$$y(t) = \left[ \frac{2}{3} e^{-t} - e^{-2t} + \frac{1}{3} e^{-4t} \right] + \left[ -3 e^{-t} + e^{-4t} \right]$$

$$y(t) = \left( \frac{4}{3} e^{-4t} - \frac{7}{3} e^{-t} - e^{-2t} \right) u(t)$$

Q9. The output of a linear system is  $y(t) = 10e^{-t} \cos(4t) u(t)$  when the input is  $x(t) = e^{-t} u(t)$ . Find the transfer function of the system and its impulse response.

Sol.  $y(t) = 10e^{-t} \cos(4t) u(t)$

$$Y(s) = 10 \cdot \frac{(s+1)}{(s+1)^2 + 16} = \frac{10(s+1)}{s^2 + 2s + 17}$$

$$x(t) = e^{-t} u(t)$$

$$X(s) = \frac{1}{s+1}$$

$$\therefore \text{Transfer function, } H(s) = \frac{Y(s)}{X(s)} = \frac{10(s+1)/(s^2+2s+17)}{1/(s+1)}$$

$$H(s) = \frac{10(s+1)^2}{s^2+2s+17}$$

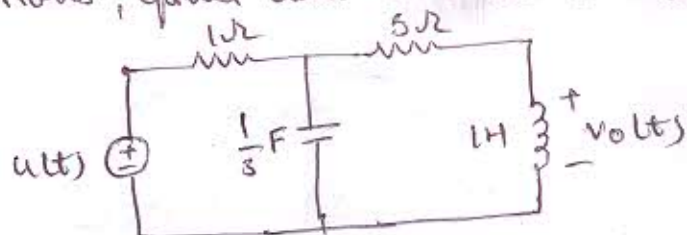
$$\text{Impulse response, } h(t) = \mathcal{L}^{-1}[H(s)]$$

$$= \mathcal{L}^{-1} \left[ \frac{10(s+1)^2}{(s+1)^2 + 4^2} \right] = \mathcal{L}^{-1} \left[ 10 - 40 \cdot \frac{4}{s^2 + 2s + 17} \right]$$

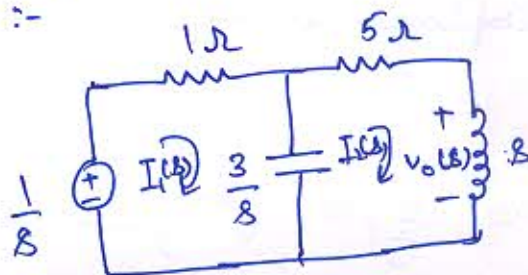
$$= \mathcal{L}^{-1} \left[ 10 - \frac{40 \cdot 2}{(s+1)^2 + 4^2} \right]$$

$$h(t) = 10\delta(t) - 40e^{-t} \sin(4t) u(t)$$

Q10. Assuming zero initial conditions, find the voltage in the circuit of figure 2.



Sol. Taking Laplace Transform of the given circuit :-



From loop 1:-

$$\frac{1}{s} = (1 + 3/s) I_1(s) - \frac{3}{s} I_2(s) \quad \text{--- (1)}$$

From loop 2:-

$$0 = -\frac{3}{s} I_1(s) + (5 + \frac{3}{s} + s) I_2(s) \quad \text{--- (2)}$$

From (1) & (2)

$$I_2 = \frac{\begin{vmatrix} (1 + 3/s) & 1/s \\ (-3/s) & 0 \end{vmatrix}}{\begin{vmatrix} (1 + 3/s) & -3/s \\ (-3/s) & s + 3/s + s \end{vmatrix}}$$



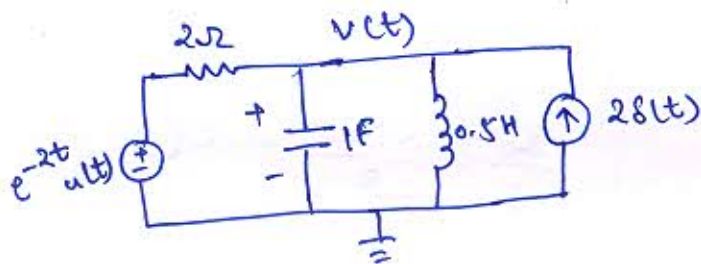
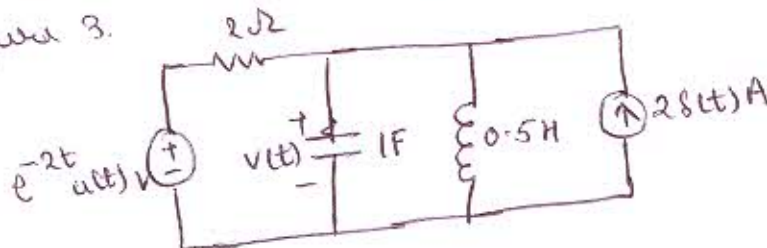
$$\therefore I_2 = \frac{3}{s^3 + 8s^2 + 18s}$$

$$\therefore V_0(s) = s I_2 = \frac{3}{s^2 + 8s + 18} = \frac{3}{\sqrt{2}} \frac{\sqrt{2}}{(s+4)^2 + (\sqrt{2})^2}$$

$$\therefore V_0(t) = \left( \frac{3}{\sqrt{2}} e^{-4t} \sin \sqrt{2} t \right) V.$$

### Section C.

Q11. Using the Fourier transform, find  $v(t)$  in the circuit shown in figure 3.



Applying Nodal Analysis

$$\frac{v(t) - e^{-2t} u(t)}{2} + \frac{v(t)}{1/j\omega} + \frac{v(t)}{j\omega 0.5} - 28(t) = 0.$$

$$v(t) - e^{-2t} u(t) + 2j\omega v(t) + \frac{v(t)}{j\omega} = 48(t)$$

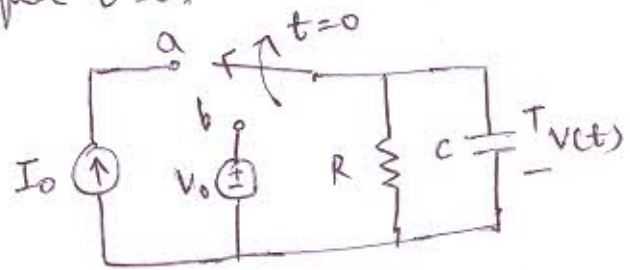
Taking Fourier Transform,

$$V(j\omega) - \frac{1}{2+j\omega} + \left( 2j\omega + \frac{1}{j\omega} \right) V(j\omega) = 4$$

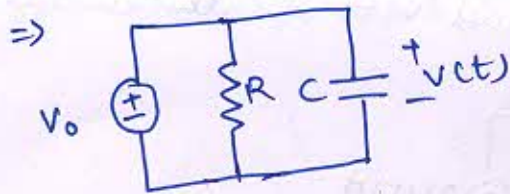
$$V(j\omega) = \frac{j\omega (4j\omega + 9)}{(2+j\omega)(4-2\omega^2+j\omega)}$$

$$V(j\omega) = \frac{2j\omega (4.5 + 2j\omega)}{(2+j\omega)(4-2\omega^2+j\omega)}$$

Q12. The switch in figure 4 has been in position b for a long time. It is moved to position a at  $t=0$ . Using Laplace transform, determine the value of  $v(t)$  for  $t > 0$ .

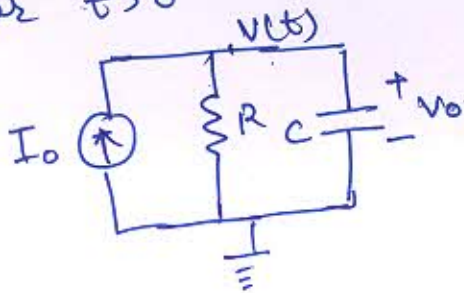


Sol. For  $t < 0$



Initial voltage at Capacitor  
 $V(0^-) = V_0$

For  $t > 0$



$$\Rightarrow I_0 = \frac{V(t)}{R} + C \frac{dV(t)}{dt}$$

Taking Laplace transform

$$\frac{I_0}{s} = \frac{V(s)}{R} + C[sV(s) - V_0]$$

$$\frac{I_0}{s} + CV_0 = \left[ \frac{1}{R} + sC \right] V(s)$$

$$\Rightarrow V(s) = \frac{I_0}{s \left[ \frac{1}{R} + sC \right]} + \frac{CV_0}{\frac{1}{R} + sC}$$

$$= \frac{I_0}{sC \left[ \frac{1}{RC} + s \right]} + \frac{V_0}{\left[ \frac{1}{RC} + s \right]}$$

$$V(s) = I_0 R \left[ \frac{1}{s} - \frac{1}{s + 1/RC} \right] + \frac{V_0}{s + 1/RC}$$

Taking inverse Laplace transform.

$$v(t) = \mathcal{L}^{-1}[V(s)]$$

$$= I_0 R (1 - e^{-t/RC}) + V_0 e^{-t/RC}$$

$$v(t) = (V_0 - I_0 R) e^{-t/RC} + I_0 R$$

— x — x —