

Microelectronics Circuit Analysis and Design

Homework(8th)

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- 5.24 (a) For the circuit in Figure P5.24, determine V_B and I_E such that $V_B = V_C$. Assume $\beta = 90$.
 (b) What value of V_B results in $V_{CE} = 2V$?

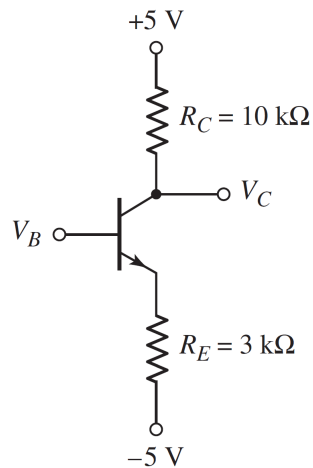


Figure 1: Problem 5.24

Solution:

- (a) Assume the BJT works in the forward-active region:

$$\begin{cases} I_C = \frac{5 - V_C}{R_C} \\ I_E = \frac{V_B - V_{BE(on)} - (-5)}{R_E} \\ I_E = \frac{1 + \beta}{\beta} I_C \\ V_B = V_C \end{cases} \Rightarrow \begin{cases} V_B = -2.14V \\ I_E = 0.72mA \end{cases}$$

(b) Obviously, the BJT work in the active region, so we have equation:

$$\begin{cases} V_C - V_B + V_{BE(on)} = V_{CE} \\ I_C = \frac{5 - V_C}{R_C} \\ I_E = \frac{V_B - V_{BE(on)} - (-5)}{R_E} \\ I_E = \frac{1 + \beta}{\beta} I_C \end{cases} \Rightarrow \begin{cases} V_B = -2.44\text{V} \\ I_E = 0.62\text{mA} \end{cases}$$

5.43 The common-emitter current gain of the transistor in Figure P5.43 is $\beta = 80$. Plot the voltage transfer characteristics over the range $0 \leq V_I \leq 5\text{V}$.

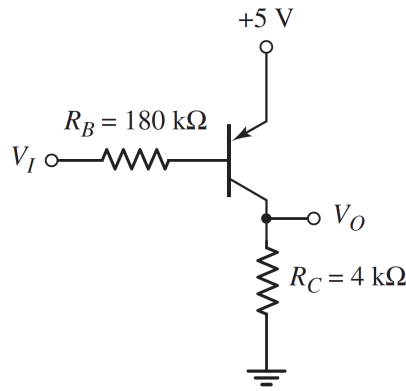


Figure 2: Problem 5.43

Solution:

It's a PNP BJT

case 1: $V_I \in (4.3, 5)$, $V_{EB} < 0.7\text{V}$, the B-E junction isn't conducted, $I_E = 0$, $V_O = 0$

case 2: When BJT works in saturation region(at sat-point), $V_{EC} = 0.2\text{V}$, $V_O = 4.8\text{V}$, $V_I = 5 - V_{EC(ON)} - \frac{V_O}{\beta R_C} R_B = 1.6\text{V}$, so when $V_I < 1.6\text{V}$, $V_O = 4.8\text{V}$

case 3: $V_I \in [1.6, 4.3]$, the BJT works in the active region:

$$\frac{V_O}{R_C} = \beta \frac{5 - V_{BE(ON)} - V_I}{R_B} \Rightarrow V_O = -\frac{16}{9} V_I + \frac{344}{45}$$

5.70 For the circuit in Figure P5.70, let $R_C = 2.2\text{k}\Omega$, $R_E = 2\text{k}\Omega$, $R_1 = 10\text{k}\Omega$, $R_2 = 20\text{k}\Omega$, and $\beta = 60$. (a) Find R_{TH} and V_{TH} for the base circuit. (b) Determine I_{BQ} , I_{CQ} , V_E , and V_C .

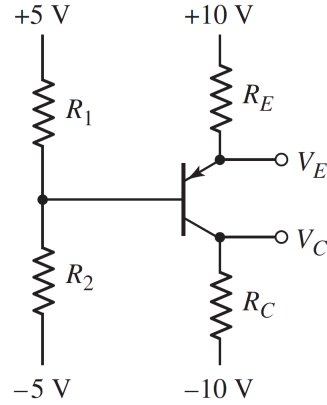


Figure 3: Problem 5.70

Solution:

$$(a) V_{TH} = V^- + \frac{R_2}{R_1 + R_2}(V^+ - V^-) = \frac{5}{3}V, R_{TH} = R_1 || R_2 = \frac{20}{3}k\Omega$$

(b) Equation is as follow:

$$(1 + \beta)I_B R_E + V_{BE(ON)} + I_B R_{TH} + V_{TH} = 10 \Rightarrow I_B = 0.0593mA$$

$$\therefore I_C = \beta I_B = 3.56mA, V_E = V_S - I_E R_E = 2.76V, V_C = -10V + I_C R_C = -2.17V$$

6.8 The parameters of each transistor in the circuits shown in Figure P6.8 are $\beta = 130$, $V_A = 80V$, and $I_{CQ} = 0.2mA$. Determine the output resistance R_o for each circuit.

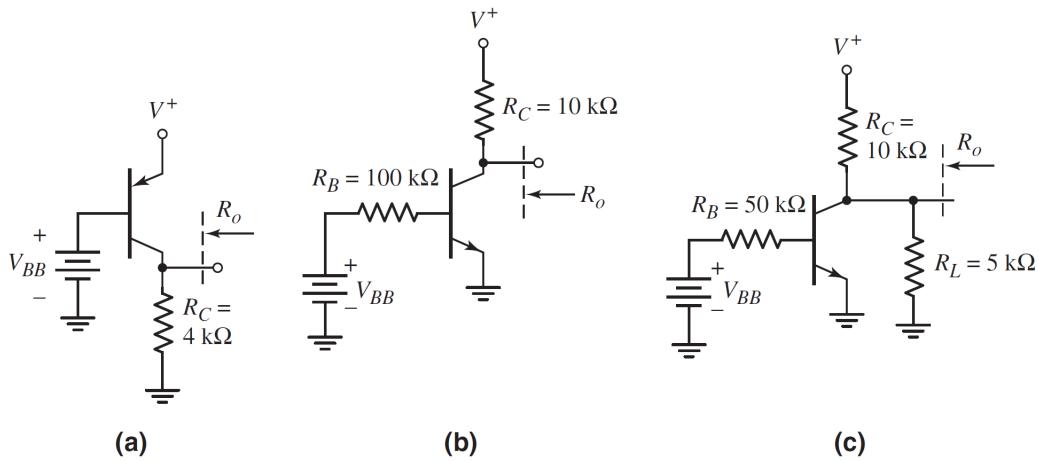


Figure 4: Problem 6.8

Solution:

$$r_o = \frac{V_A}{I_{CQ}} = 400\text{k}\Omega$$

$$(a) R_o = r_o || R_C = 3.96\text{k}\Omega$$

$$(b) R_o = r_o || R_C = 9.76\text{k}\Omega$$

$$(c) R_o = r_o || R_L || R_C = 3.31\text{k}\Omega$$

6.18 The signal source in Figure P6.18 is $v_s = 5 \sin \omega t \text{ mV}$. The transistor parameters are $\beta = 120$ and $V_A = \infty$. (a) (i) Design the circuit such that $I_{CQ} = 0.25 \text{ mA}$ and $V_{CEQ} = 3 \text{ V}$. (ii) Find the small-signal voltage gain $A_v = v_o/v_s$. (iii) Find $v_o(t)$. (b) Repeat part (a) for $R_S = 0$.

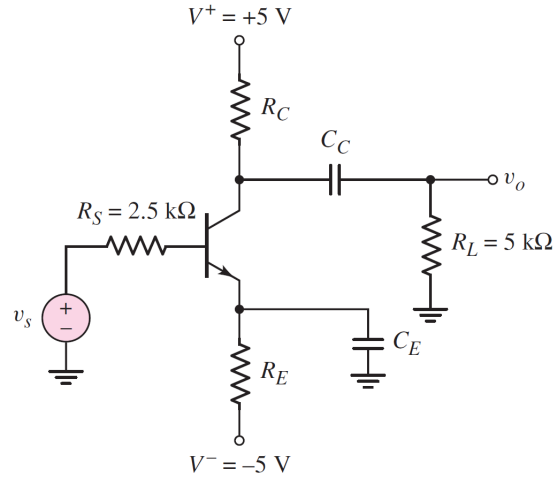


Figure 5: Problem 6.18

Solution:

(a)(i) $I_{BQ} = \frac{I_{CQ}}{\beta} = 0.00661 \text{ mA}$, then because of KVL, we have equation:

$$\begin{cases} I_{BQ}R_S + V_{BEON} + (\beta + 1)I_B R_E = 0 - V^- \\ I_C R_C + V_{CEQ} + (\beta + 1)I_B = V^+ - V^- \end{cases} \Rightarrow \begin{cases} R_C = 10.9\text{k}\Omega \\ R_E = 17.0\text{k}\Omega \end{cases}$$

$$(ii) A_v = \frac{-g_m \left(\frac{r_\pi}{R_S + r_\pi} \cdot v_s \right) (R_L || R_C)}{v_s} = -27.5$$

$$(iii) v_o(t) = A_v v_s = -0.14 \sin \omega t \text{ V}$$

(b)(i) because of KVL, we have equation:

$$\begin{cases} \beta I_{BQ} R_C + V_{CEQ} + (\beta + 1)I_B R_E = V^+ - V^- \\ V_{BEON} + (\beta + 1)I_B R_E = 0 - V^- \end{cases} \Rightarrow \begin{cases} R_C = 10.8\text{k}\Omega \\ R_E = 17.1\text{k}\Omega \end{cases}$$

$$(ii) A_v = \frac{-g_m(\cdot v_s)(R_L || R_C)}{v_s} = -32.9$$

$$(iii) v_o(t) = A_v v_s = -0.16 \sin \omega t \text{ V}$$

6.33 For the circuit in Figure P6.15, let $\beta = 100$, $V_A = \infty$, $R_E = 12.9 \text{ k}\Omega$, and $R_C = 6 \text{ k}\Omega$. Determine the maximum undistorted swing in the output voltage if the total instantaneous C–E voltage is to remain in the range $1 \leq v_{CE} \leq 9 \text{ V}$ and if the total instantaneous collector current is to remain greater or equal to $50 \mu \text{ A}$.

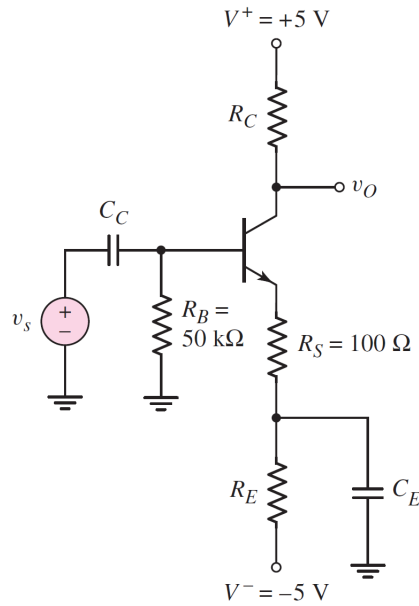


Figure 6: Problem 6.33

6.39 For the circuit in Figure P6.24, the transistor parameters are $\beta = 100$ and $V_A = \infty$. (a) Determine the maximum undistorted swing in the output voltage if the total instantaneous E–C voltage is to remain in the range $1 \leq v_{EC} \leq 9 \text{ V}$. (b) Using the results of part (a), determine the range of collector current.

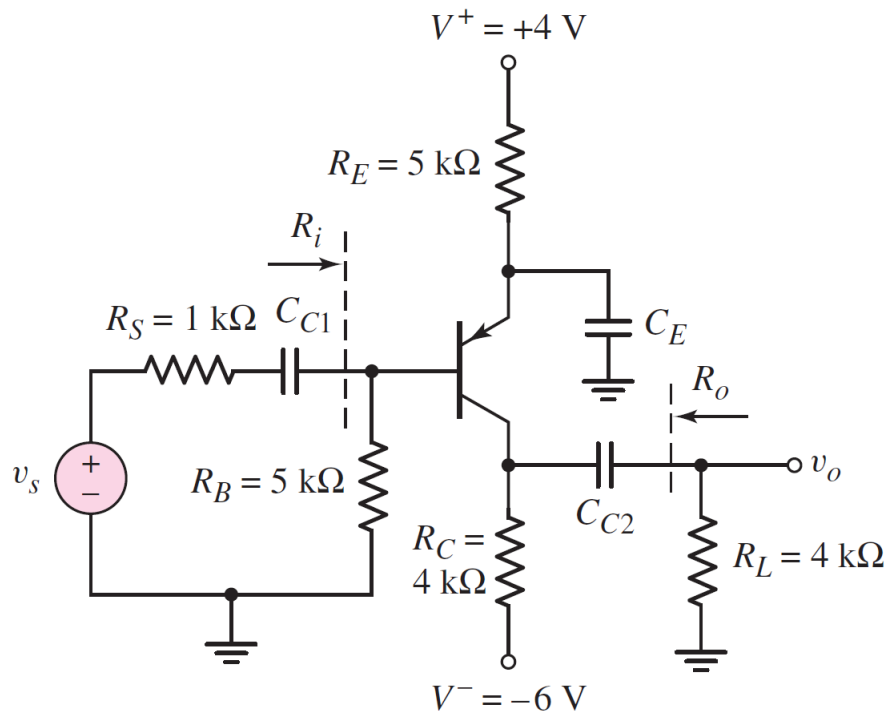


Figure 7: Problem 6.39

6.65 Consider the circuit shown in Figure P6.69. The transistor has parameters $\beta = 60$ and $V_A = \infty$. (a) Determine the quiescent values of I_{CQ} and V_{CEQ} . (b) Determine the small-signal voltage gain $A_v = v_o/v_s$.

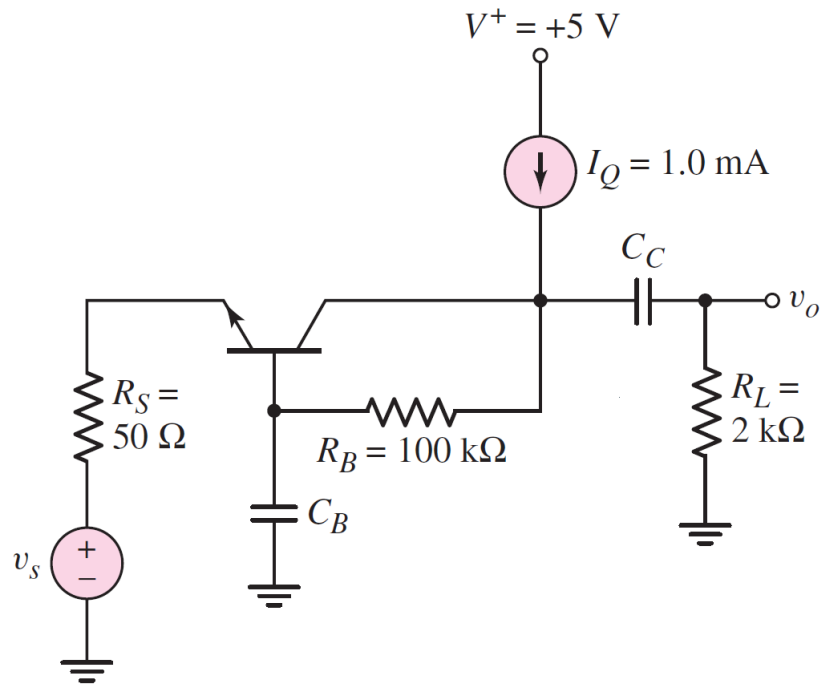


Figure 8: Problem 6.69