

4: Linear Time Invariant Systems

- LTI Systems
- Convolution Properties
- BIBO Stability
- Frequency Response
- Causality
- Convolution Complexity
- Circular Convolution
- Frequency-domain convolution
- Overlap Add
- Overlap Save
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4: **Linear Time Invariant Systems**

two properties

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Convolution Operator is Non-Associative

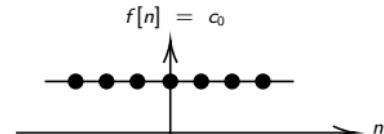
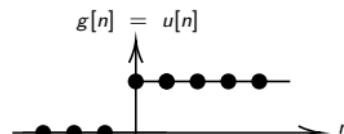
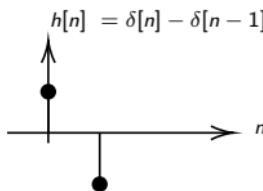
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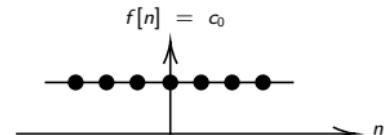
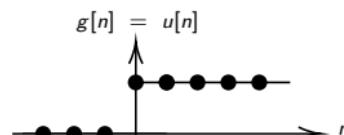
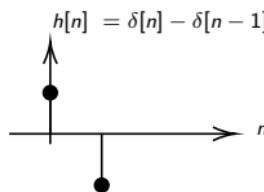


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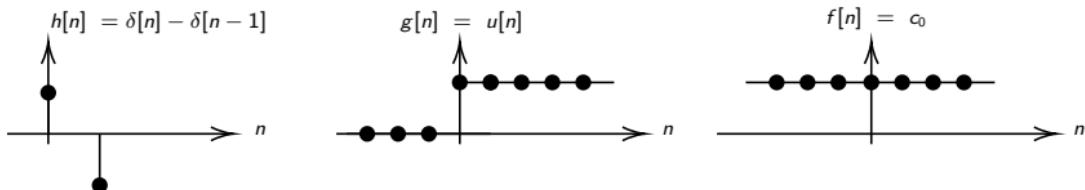
$$(f * g * h)[n] = ((f * g) * h)[n] = (f * (g * h))[n] = ((f * h) * g)[n].$$

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- ③ From the figure above.



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BIBO Stability

BIBO Stability: Bounded Input, $x[n] \Rightarrow$ Bounded Output, $y[n]$

The following are equivalent:

- (1) An LTI system is **BIBO stable**
- (2) $h[n]$ is **absolutely summable**, i.e. $\sum_{n=-\infty}^{\infty} |h[n]| < \infty$
- (3) $H(z)$ **region of absolute convergence includes $|z| = 1$.**

Proof (1) \Rightarrow (2):

Define $x[n] = \begin{cases} 1 & h[-n] \geq 0 \\ -1 & h[-n] < 0 \end{cases}$
then $y[0] = \sum x[0 - n]h[n] = \sum |h[n]|$.
But $|x[n]| \leq 1 \forall n$ so BIBO $\Rightarrow y[0] = \sum |h[n]| < \infty$.

Proof (2) \Rightarrow (1):

Suppose $\sum |h[n]| = S < \infty$ and $|x[n]| \leq B$ is bounded.

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Frequency Response

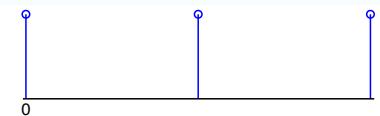
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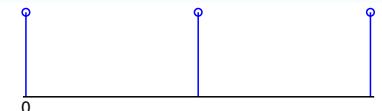
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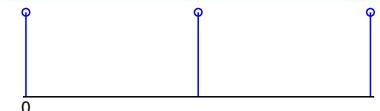
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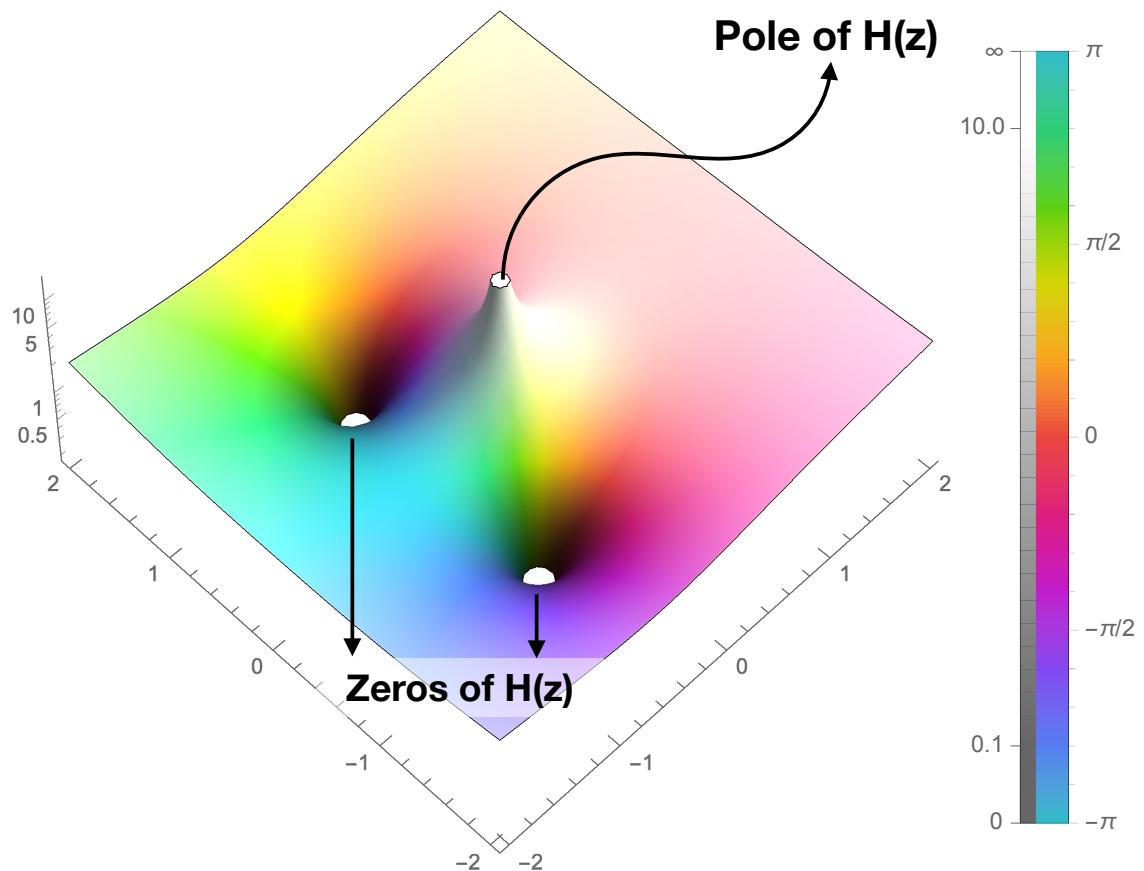
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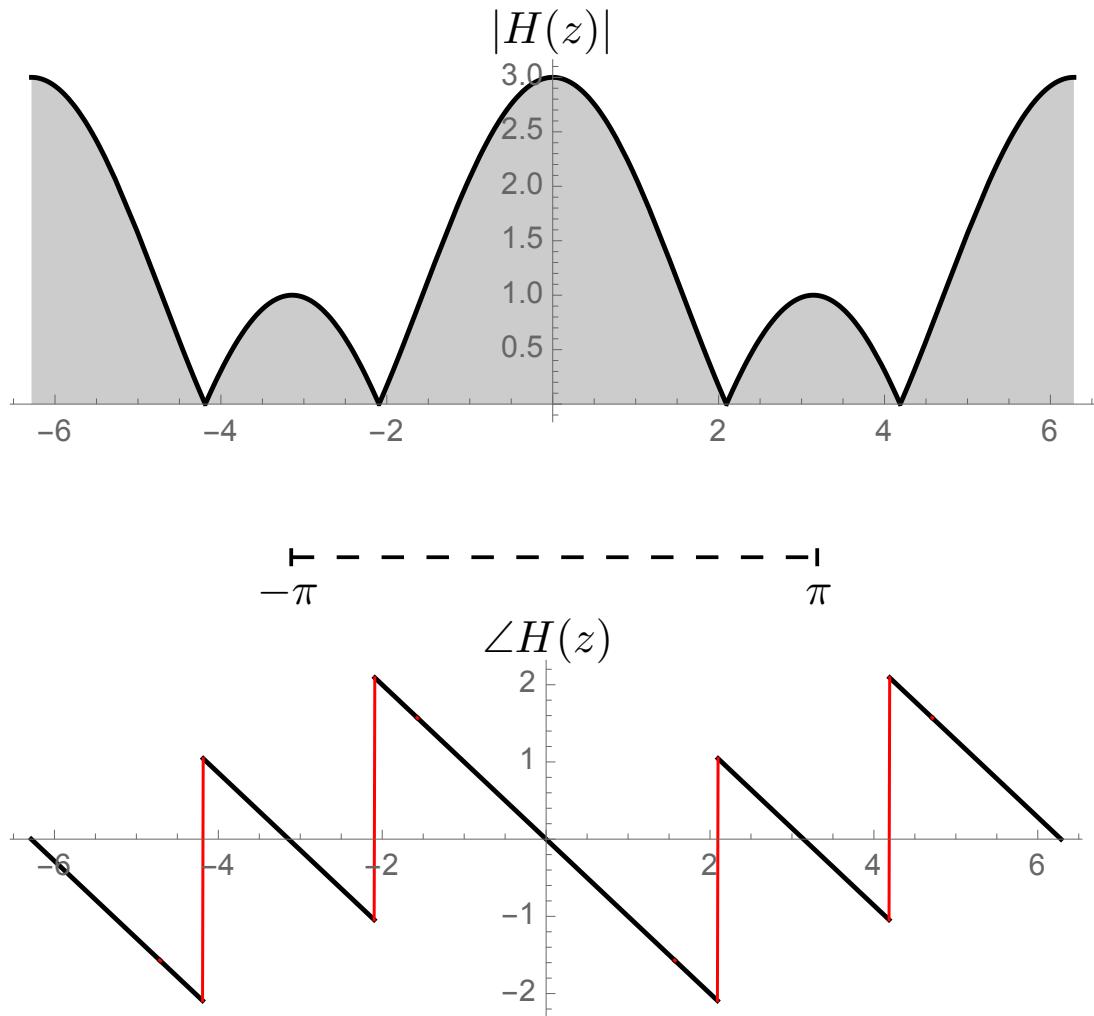
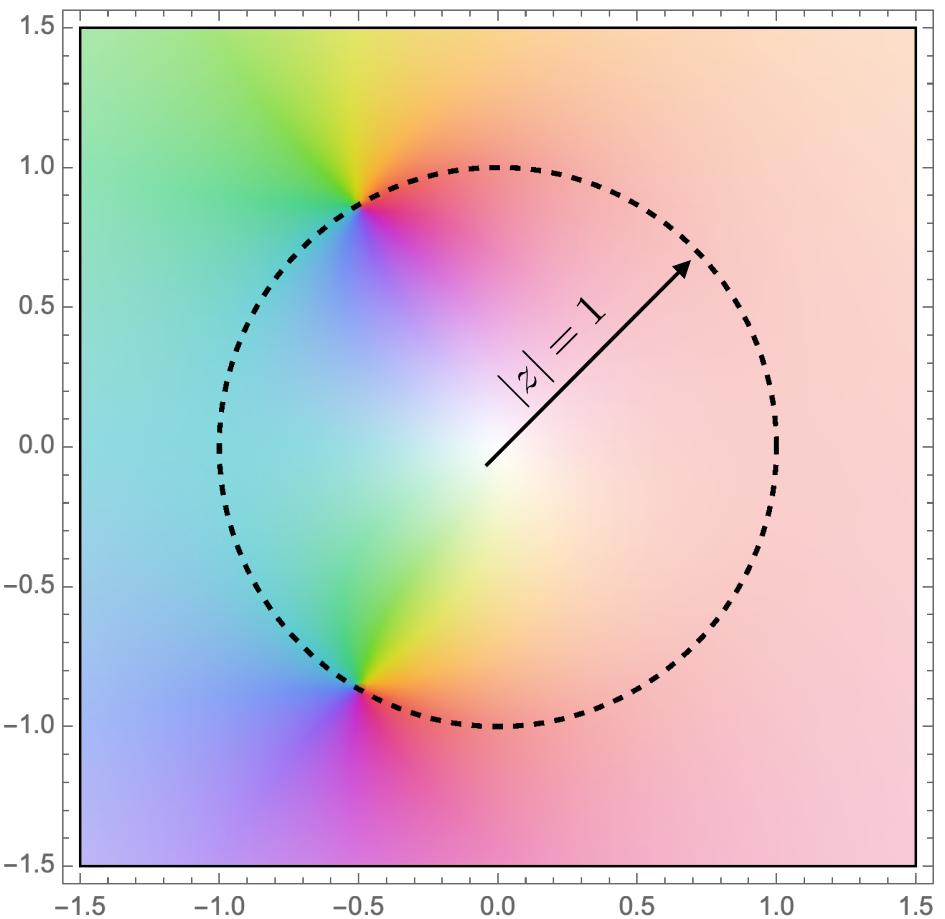
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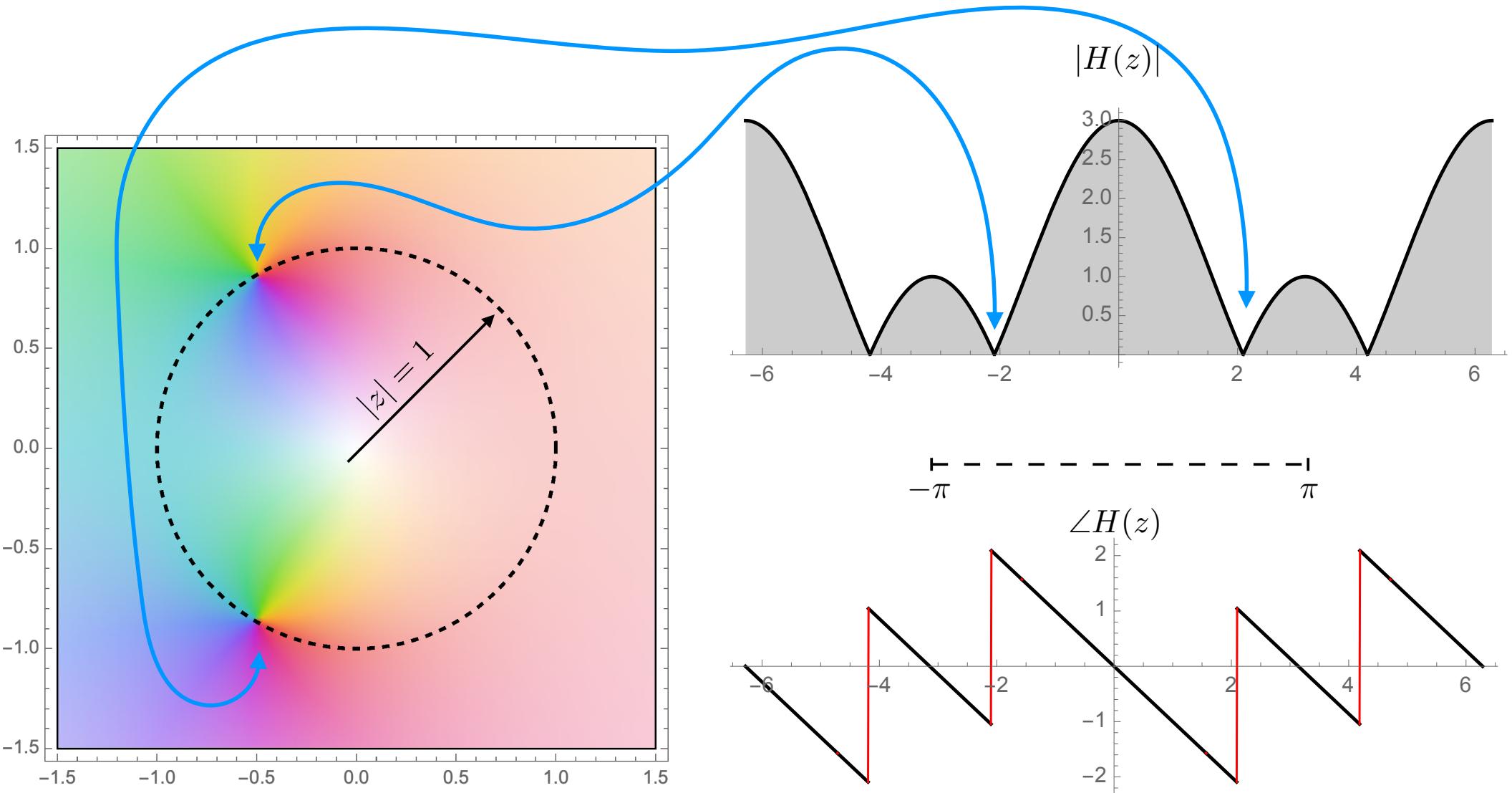
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$$h[n] = \delta[n-1] + \delta[n] + \delta[n+1] \longleftrightarrow H(z) = z + 1 + z^{-1}$$







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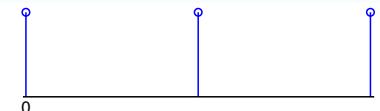
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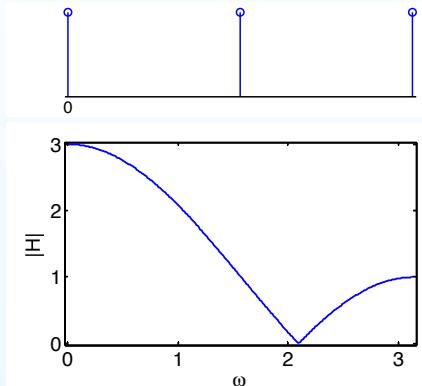
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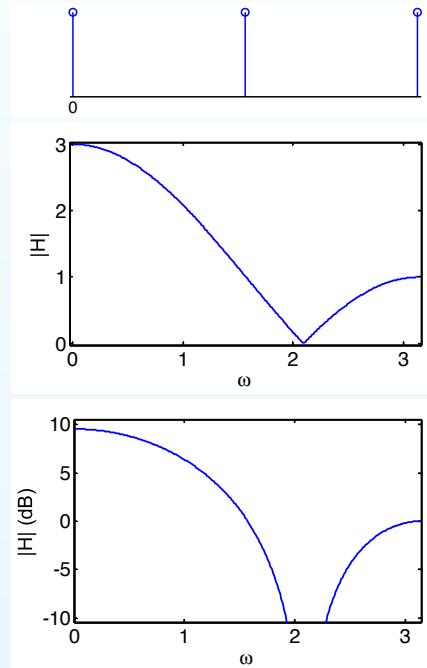
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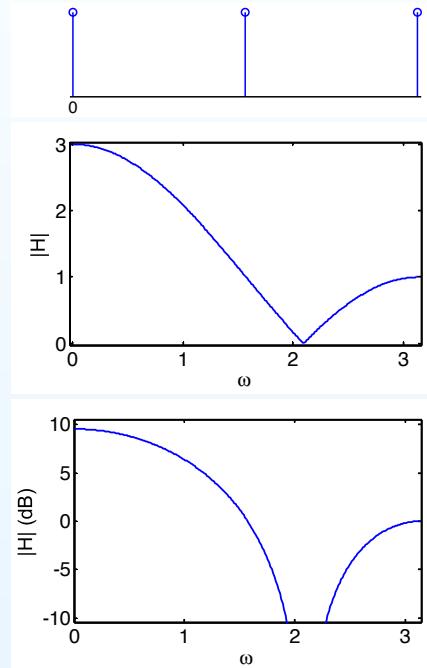
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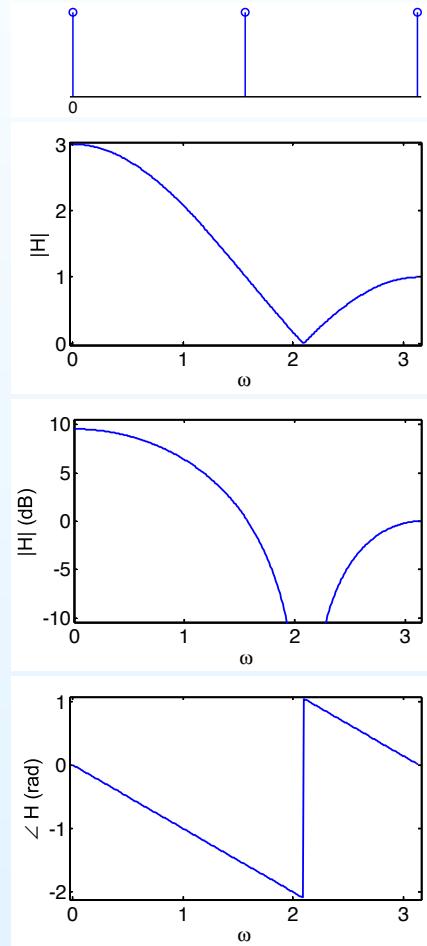
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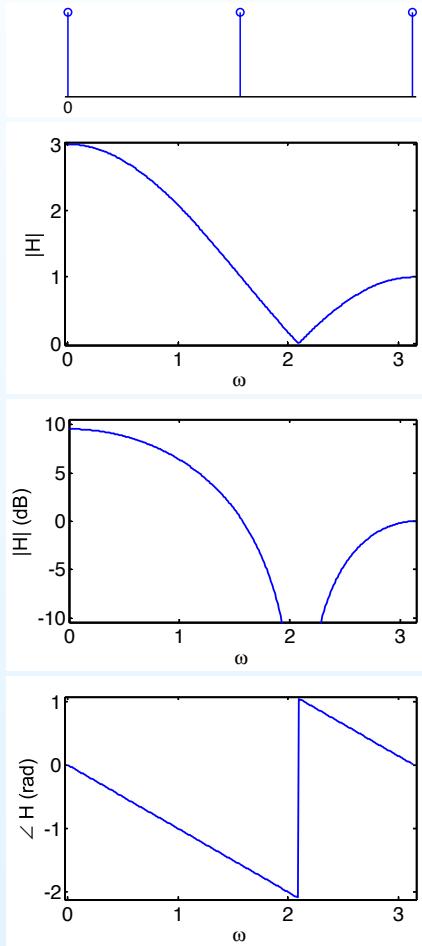
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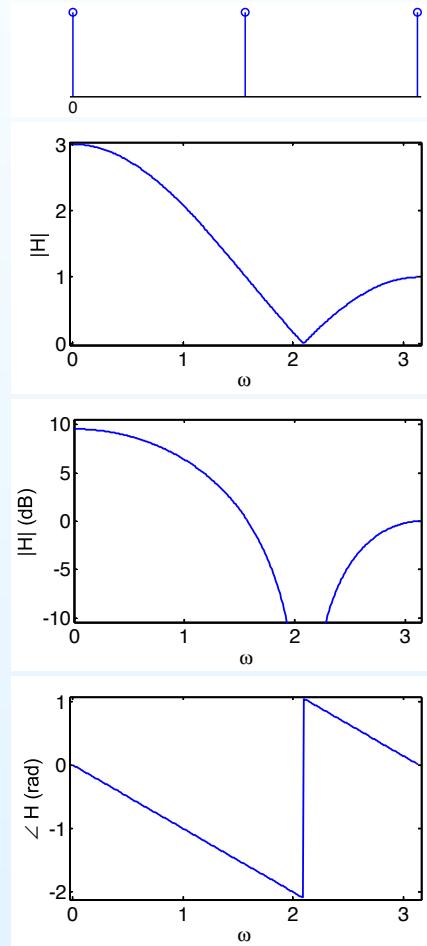
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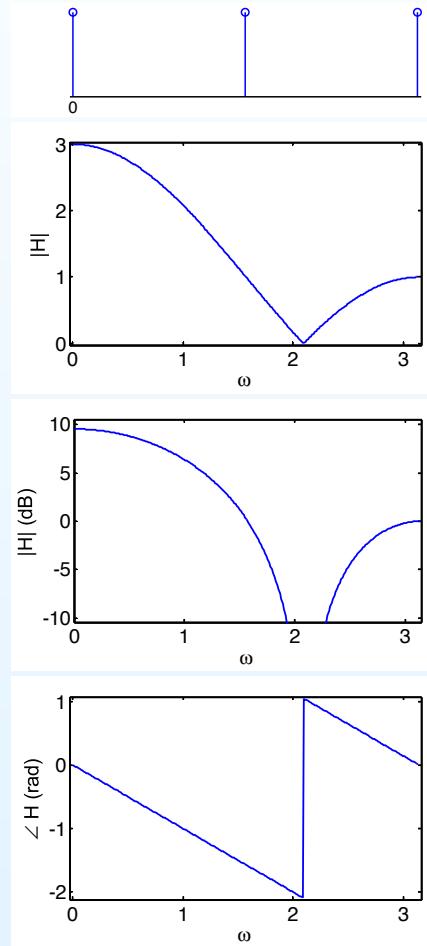
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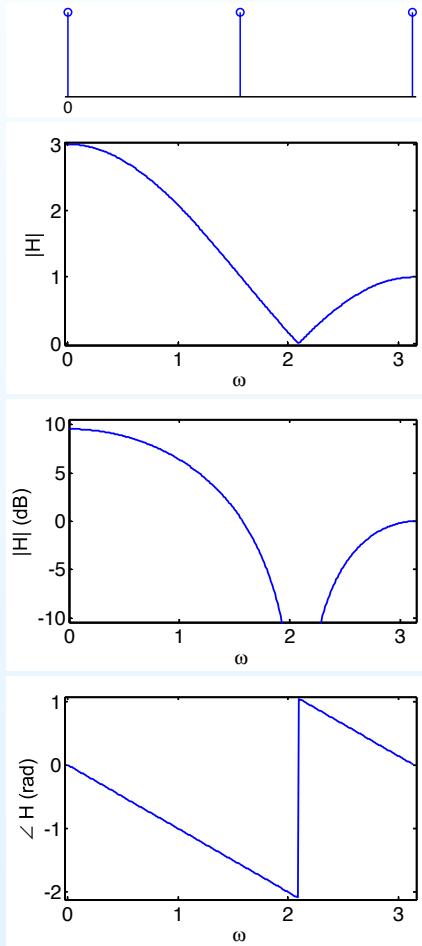
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Any right-sided sequence can be made causal by adding a delay.

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- (3) $H(z)$ converges for $z = \infty$

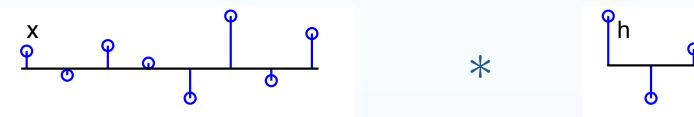
Any right-sided sequence can be made causal by adding a delay.

All the systems we will deal with are causal.

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- MATLAB routines

Convolution Complexity

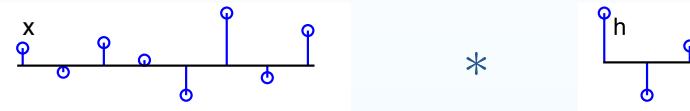
$y[n] = x[n] * h[n]$: convolve $x[0 : N - 1]$ with $h[0 : M - 1]$



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Convolution sum:

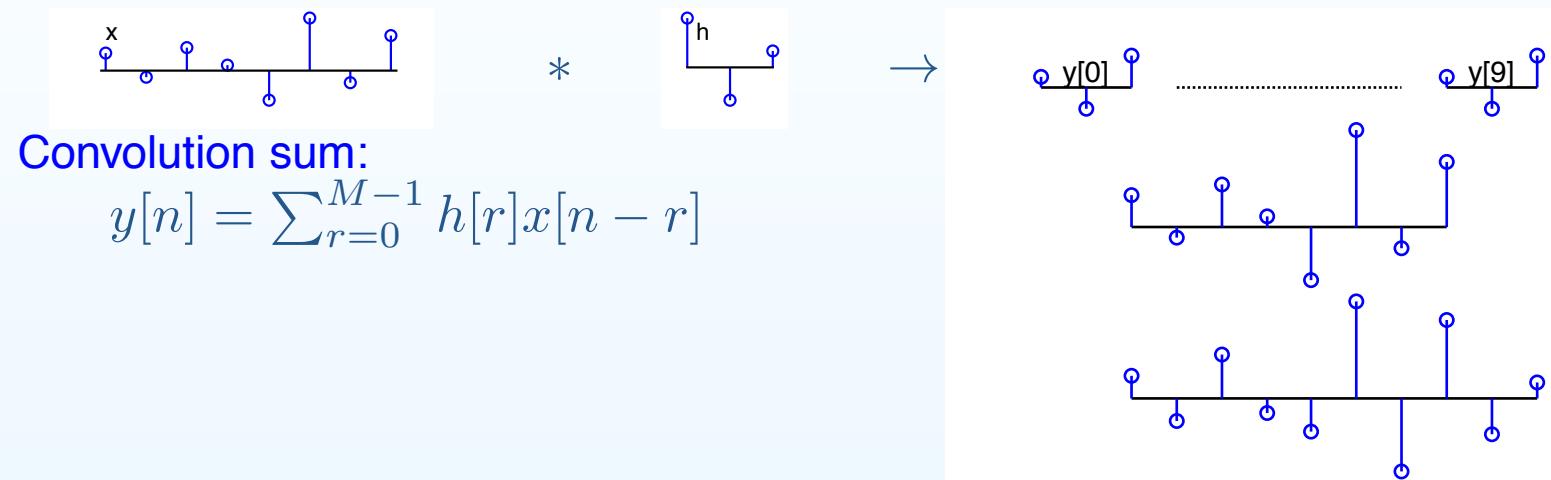
$$y[n] = \sum_{r=0}^{M-1} h[r]x[n-r]$$

4: Linear Time Invariant Systems

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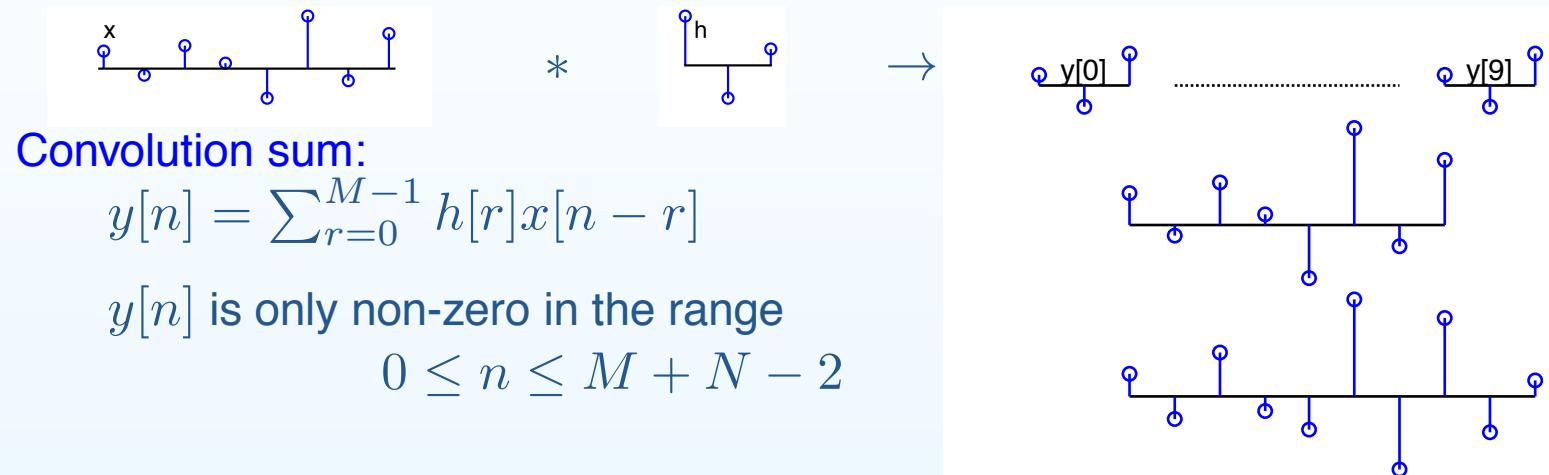
$$y[n] = \sum_{r=0}^{M-1} h[r]x[n-r]$$

$$\begin{aligned}N &= 8, M = 3 \\M + N - 1 &= 10\end{aligned}$$

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$y[n] = x[n] * h[n]$: convolve $x[0 : N - 1]$ with $h[0 : M - 1]$



Convolution sum:

$$y[n] = \sum_{r=0}^{M-1} h[r]x[n-r]$$

$y[n]$ is only non-zero in the range

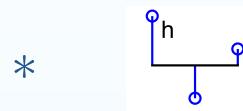
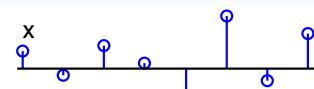
$$0 \leq n \leq M + N - 2$$

$$\begin{aligned}N &= 8, M = 3 \\M + N - 1 &= 10\end{aligned}$$

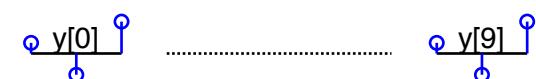
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→

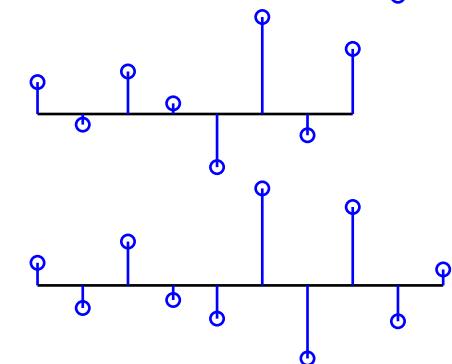


Convolution sum:

$$y[n] = \sum_{r=0}^{M-1} h[r]x[n-r]$$

$y[n]$ is only non-zero in the range
 $0 \leq n \leq M + N - 2$

Thus $y[n]$ has only
 $M + N - 1$ non-zero values



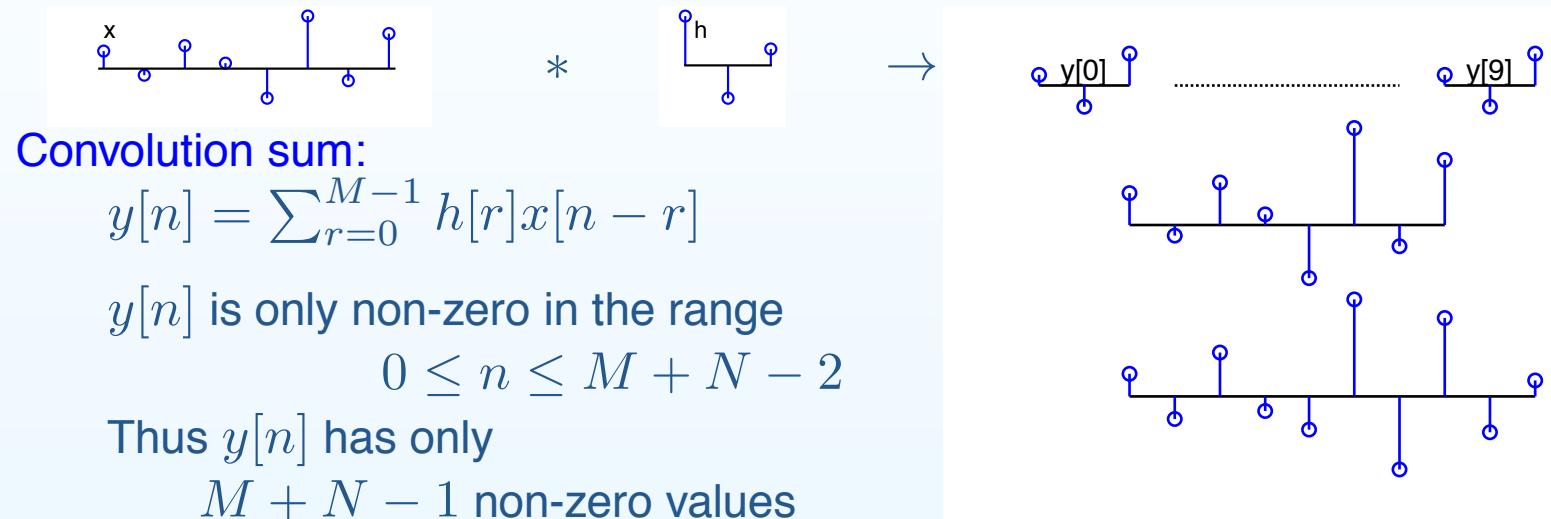
$$N = 8, M = 3$$

$$M + N - 1 = 10$$

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$y[n] = x[n] * h[n]$: convolve $x[0 : N - 1]$ with $h[0 : M - 1]$



Algebraically:

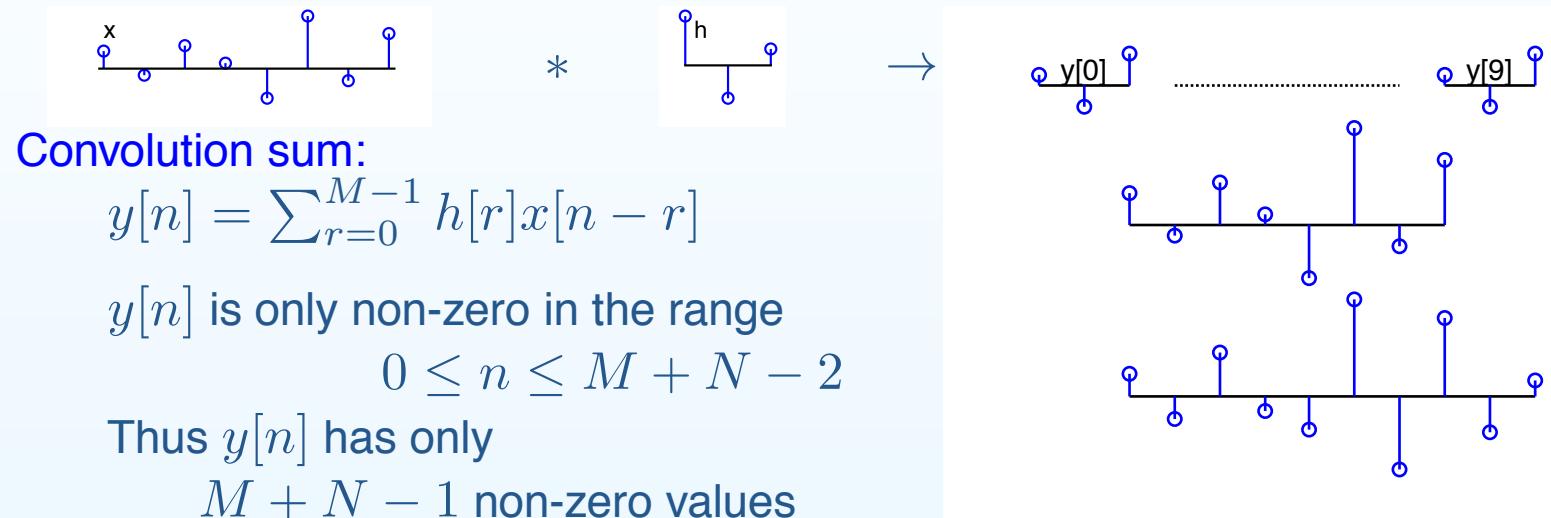
$$x[n - r] \neq 0 \Rightarrow 0 \leq n - r \leq N - 1$$

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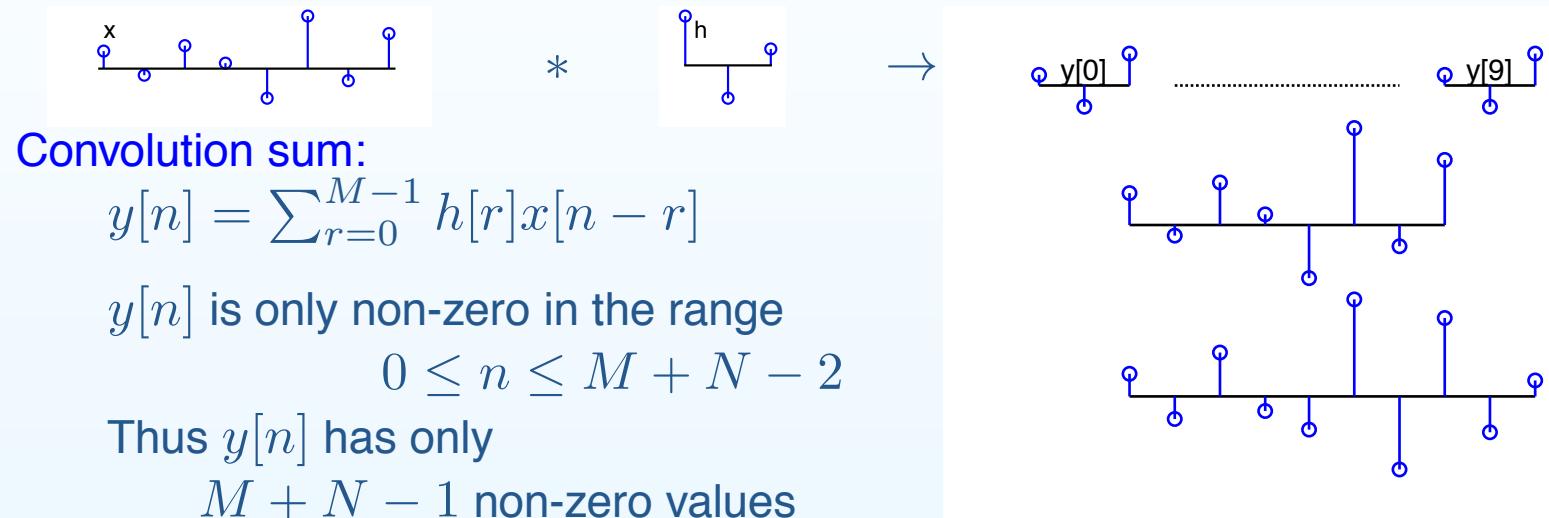
$$\begin{aligned} x[n - r] \neq 0 &\Rightarrow 0 \leq n - r \leq N - 1 \\ &\Rightarrow n + 1 - N \leq r \leq n \end{aligned}$$

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Algebraically:

$$\begin{aligned} x[n-r] \neq 0 \Rightarrow 0 &\leq n-r \leq N-1 \\ &\Rightarrow n+1-N \leq r \leq n \end{aligned}$$

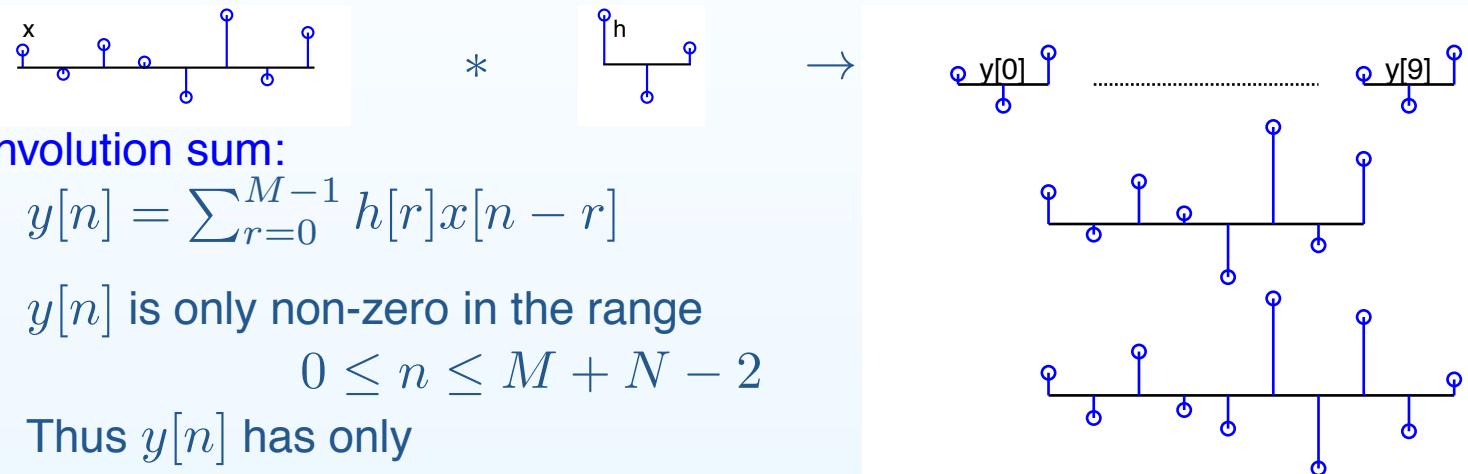
Hence: $y[n] = \sum_{r=\max(0, n+1-N)}^{\min(M-1, n)} h[r]x[n-r]$

$$\begin{aligned} N &= 8, M = 3 \\ M + N - 1 &= 10 \end{aligned}$$

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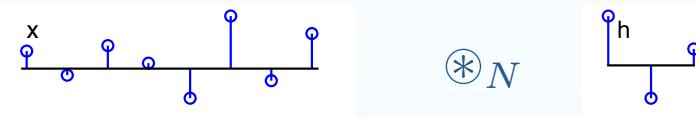
$$\begin{aligned} N &= 8, M = 3 \\ M + N - 1 &= 10 \end{aligned}$$

We must multiply each $h[n]$ by each $x[n]$ and add them to a total
 \Rightarrow total arithmetic complexity (\times or $+$ operations) $\approx 2MN$

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Circular Convolution

$$y_{\circledast}[n] = x[n] \circledast_N h[n]: \text{circ convolve } x[0 : N - 1] \text{ with } h[0 : M - 1]$$



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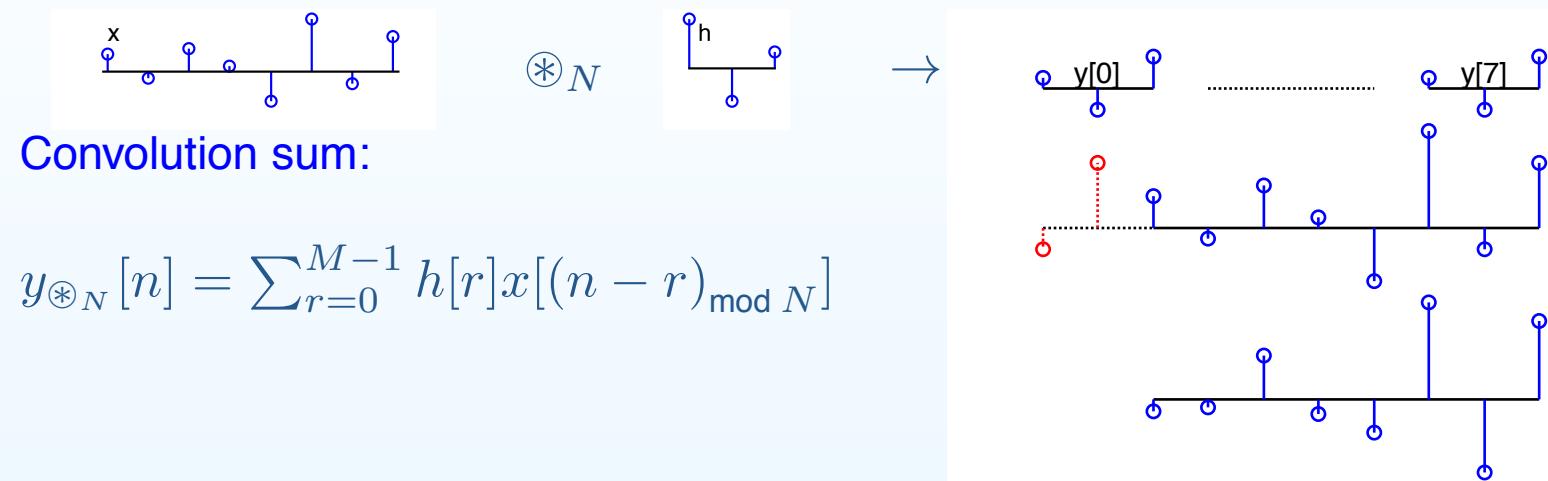
Convolution sum:

$$y_{\circledast_N}[n] = \sum_{r=0}^{M-1} h[r]x[(n - r)_{\text{mod } N}]$$

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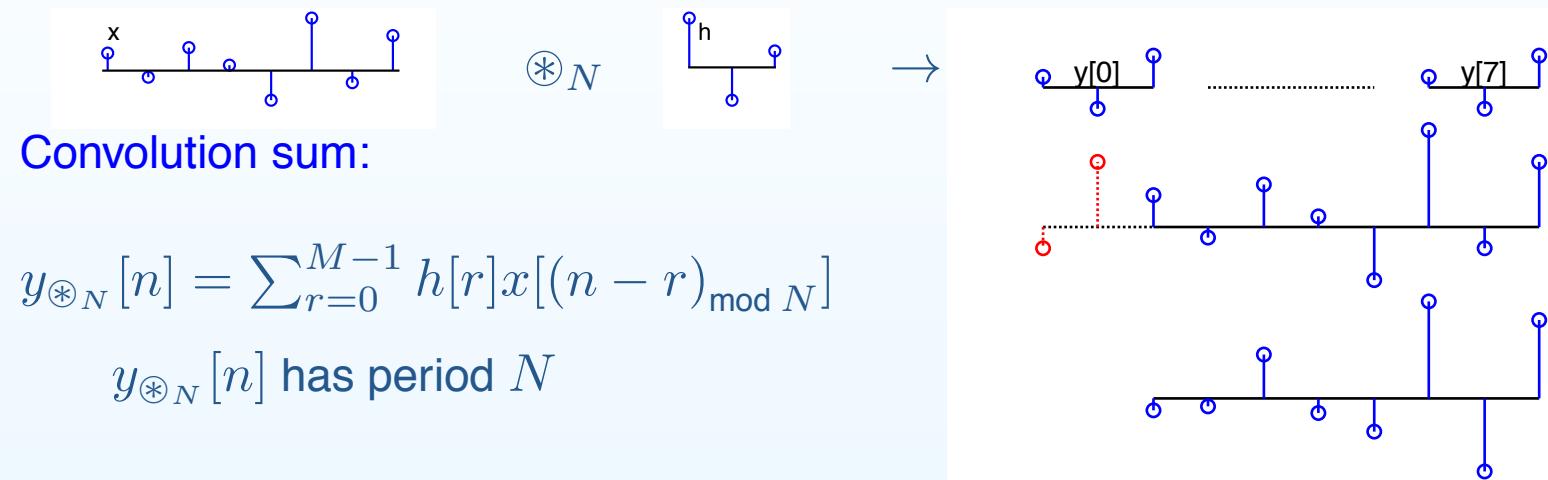


$$N = 8, M = 3$$

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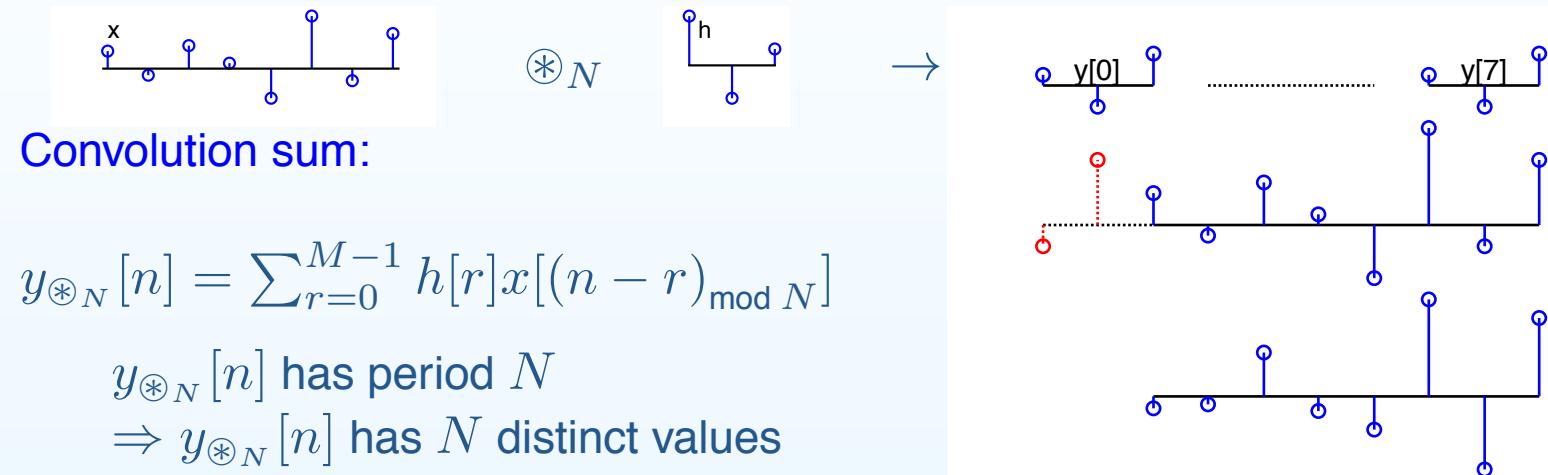
$y_{\circledast_N}[n]$ has period N

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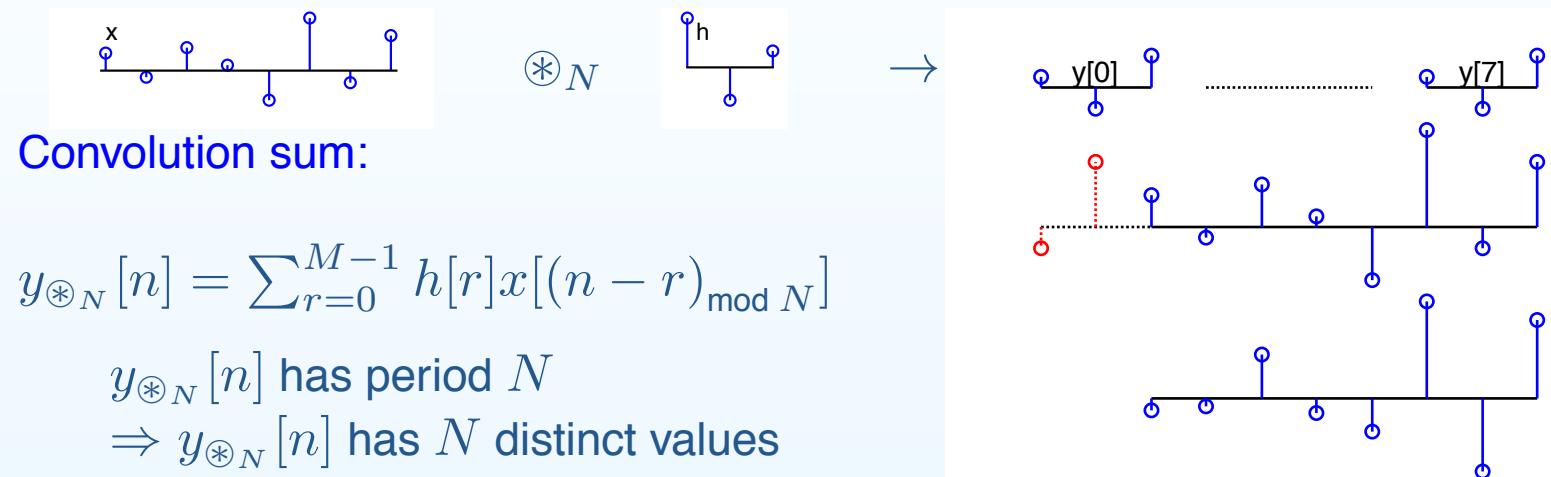


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$\Rightarrow y_{\circledast_N}[n]$ has N distinct values

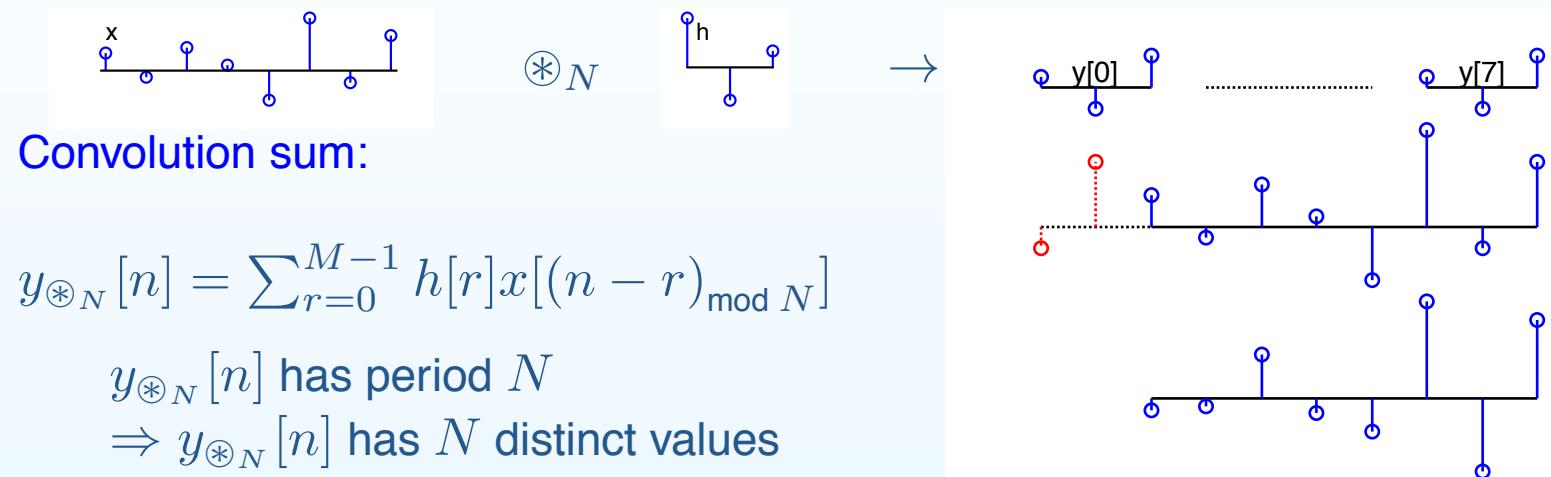
$$N = 8, M = 3$$

- Only the first $M - 1$ values are affected by the circular repetition:

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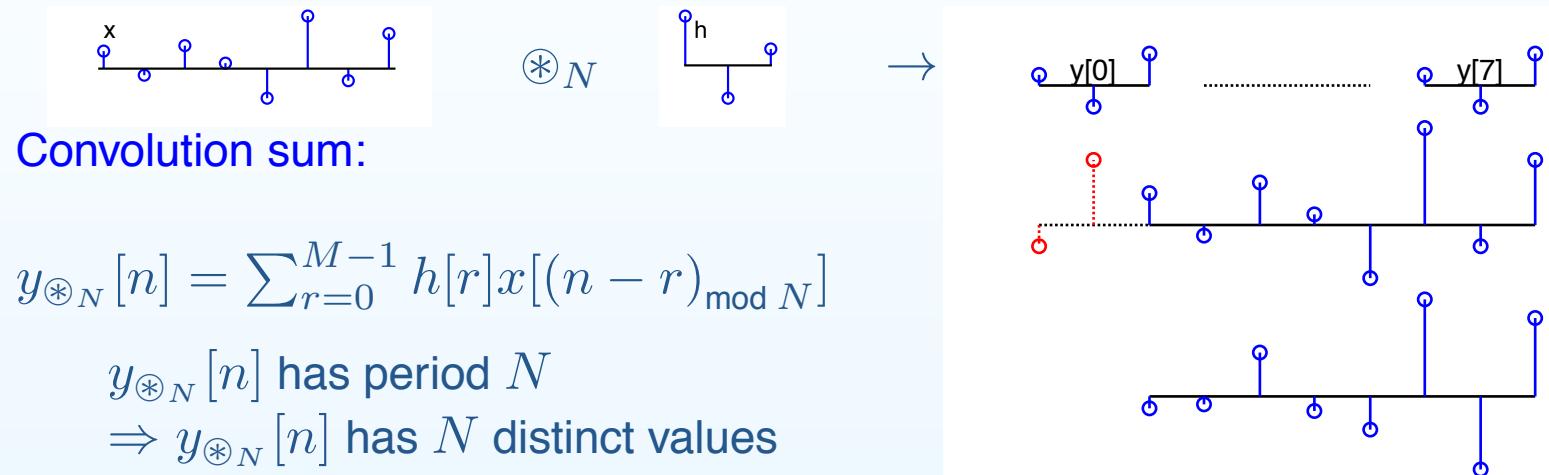
- Only the first $M - 1$ values are affected by the circular repetition:

$$y_{\circledast_N}[n] = y[n] \text{ for } M - 1 \leq n \leq N - 1$$

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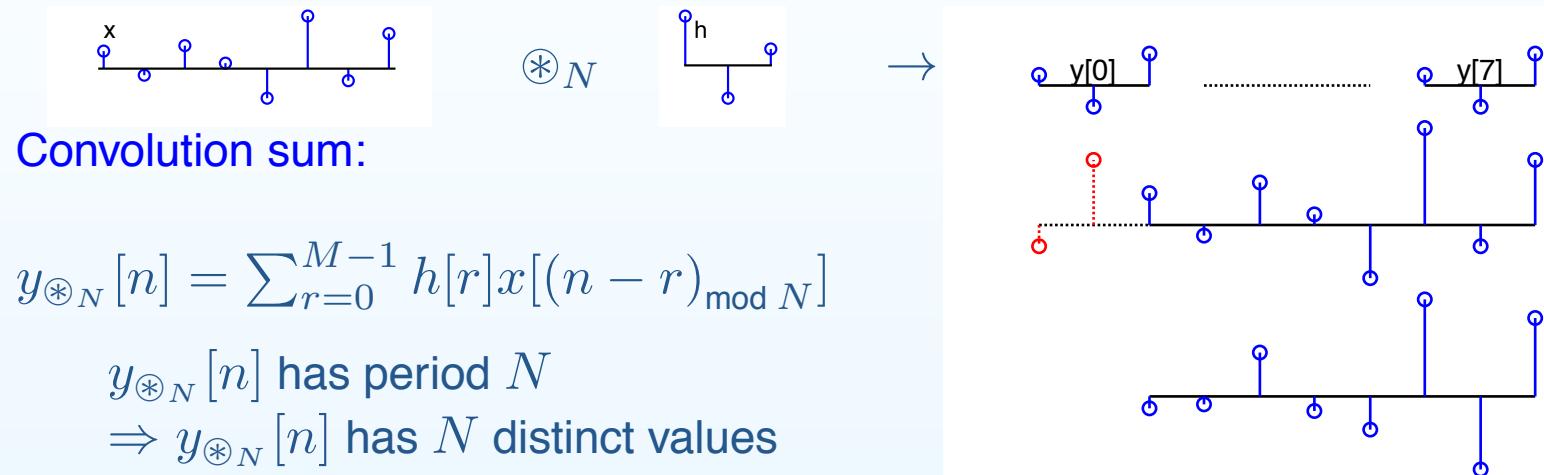
- If we append $M - 1$ zeros (or more) onto $x[n]$, then the circular repetition has no effect at all and:

$$y_{\circledast_{N+M-1}}[n] = y[n] \text{ for } 0 \leq n \leq N + M - 2$$

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Circular convolution is a necessary evil in exchange for using the DFT

Frequency-domain convolution

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Idea: Use DFT to perform circular convolution - less computation

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(1) Choose $L \geq M + N - 1$ (normally round up to a power of 2)

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- (1) Choose $L \geq M + N - 1$ (normally round up to a power of 2)
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Arithmetic Complexity:

DFT or IDFT take $4L \log_2 L$ operations if L is a power of 2
(or $16L \log_2 L$ if not).

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DFT or IDFT take $4L \log_2 L$ operations if L is a power of 2
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Total operations: $\approx 12L \log_2 L \approx 12(M + N) \log_2 (M + N)$

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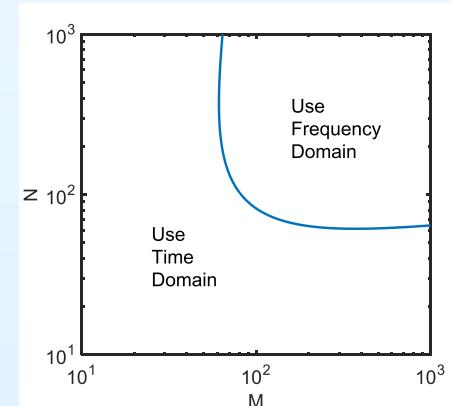
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Beneficial if both M and N are $> \sim 70$.



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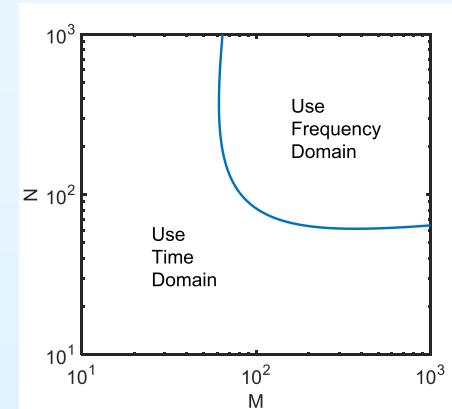
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Example: $M = 10^3, N = 10^4$:



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- (4) $y[n] = \tilde{y}[n]$ for $0 \leq n \leq M + N - 2$.

Arithmetic Complexity:

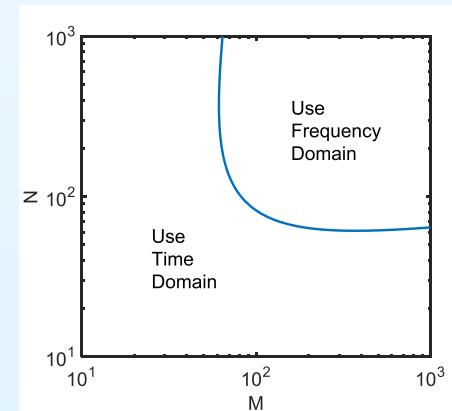
DFT or IDFT take $4L \log_2 L$ operations if L is a power of 2
(or $16L \log_2 L$ if not).

Total operations: $\approx 12L \log_2 L \approx 12(M + N) \log_2 (M + N)$

Beneficial if both M and N are $> \sim 70$.

Example: $M = 10^3, N = 10^4$:

Direct: $2MN = 2 \times 10^7$



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Frequency-domain convolution

Idea: Use DFT to perform circular convolution - less computation

- (1) Choose $L \geq M + N - 1$ (normally round up to a power of 2)
- (2) Zero pad $x[n]$ and $h[n]$ to give sequences of length L : $\tilde{x}[n]$ and $\tilde{h}[n]$
- (3) Use DFT: $\tilde{y}[n] = \mathcal{F}^{-1}(\tilde{X}[k]\tilde{H}[k]) = \tilde{x}[n] \circledast_L \tilde{h}[n]$
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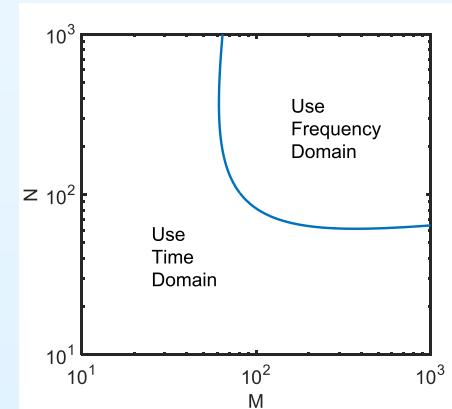
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Example: $M = 10^3, N = 10^4$:

Direct: $2MN = 2 \times 10^7$

with DFT: $= 1.8 \times 10^6$ ☺



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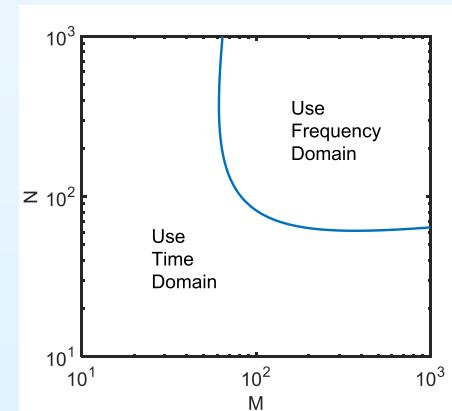
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Example: $M = 10^3, N = 10^4$:

Direct: $2MN = 2 \times 10^7$

with DFT: $= 1.8 \times 10^6$ ☺

But: (a) DFT may be very long if N is large
(b) No outputs until all $x[n]$ has been input.



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Summary

- LTI systems: impulse response, frequency response, group delay

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For further details see Mitra: 4 & 5.

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fftfilt	Convolution using overlap add
$x[n] \circledast y[n]$	<code>real(ifft(fft(x).*fft(y)))</code>