

15: Subband Processing

- Subband processing
- 2-band Filterbank
- Perfect Reconstruction
- Quadrature Mirror Filterbank (QMF)
- Polyphase QMF
- QMF Options
- Linear Phase QMF
- IIR Allpass QMF
- Tree-structured filterbanks
- Summary
- Merry Xmas

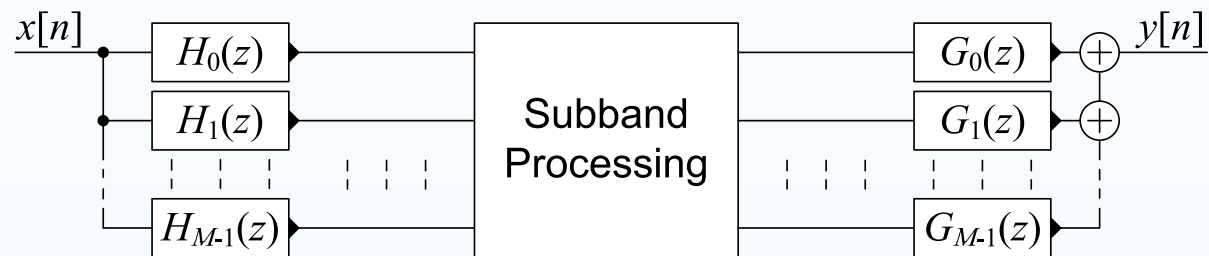
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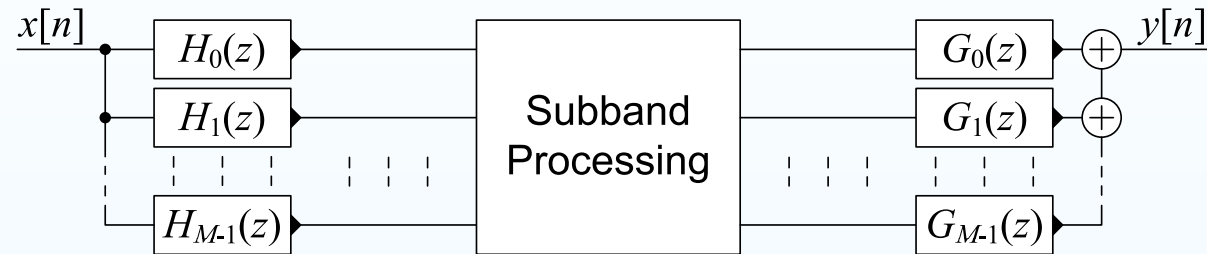


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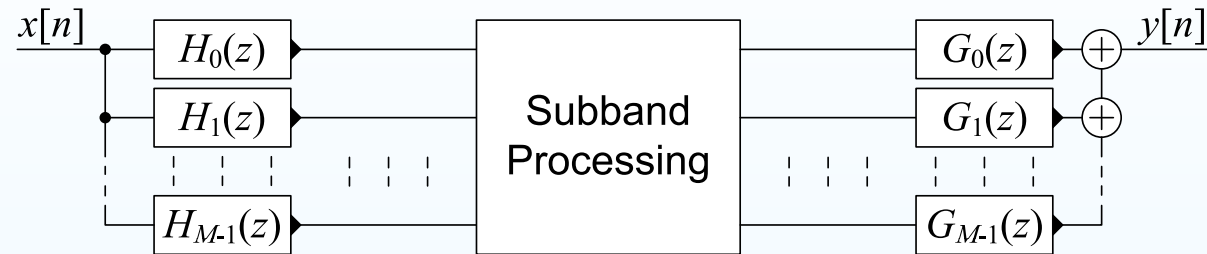
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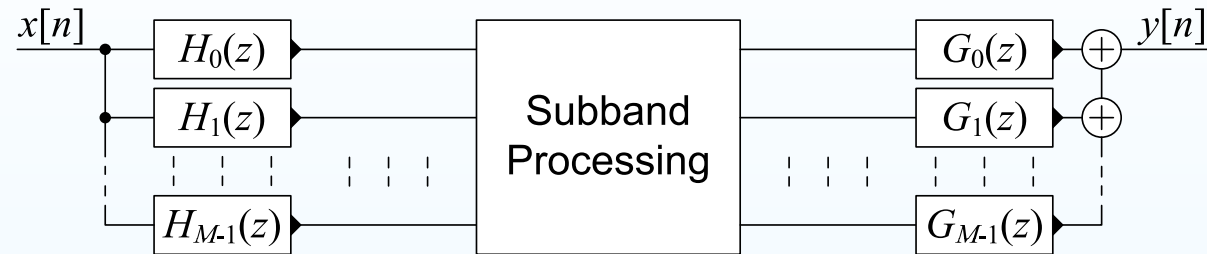
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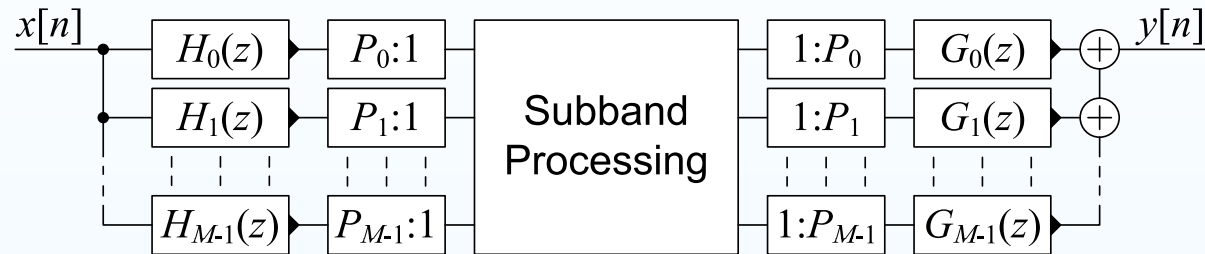
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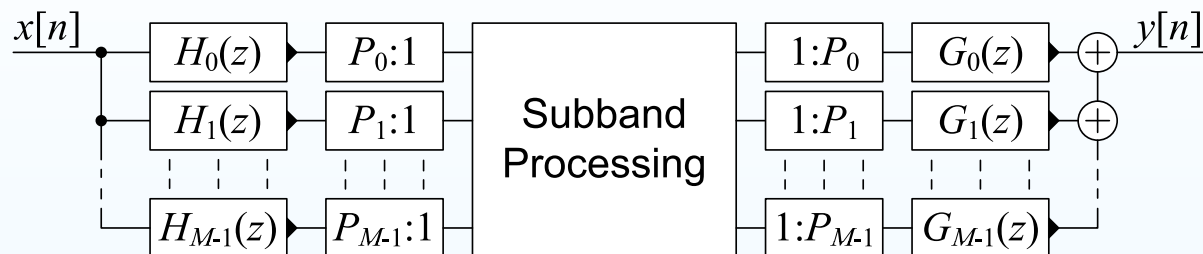
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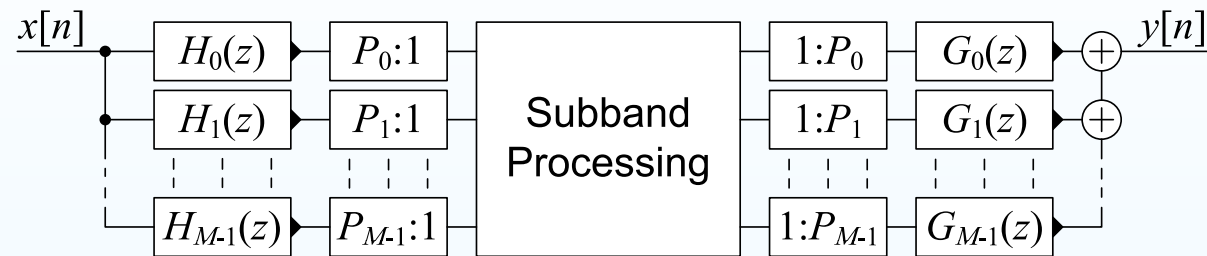
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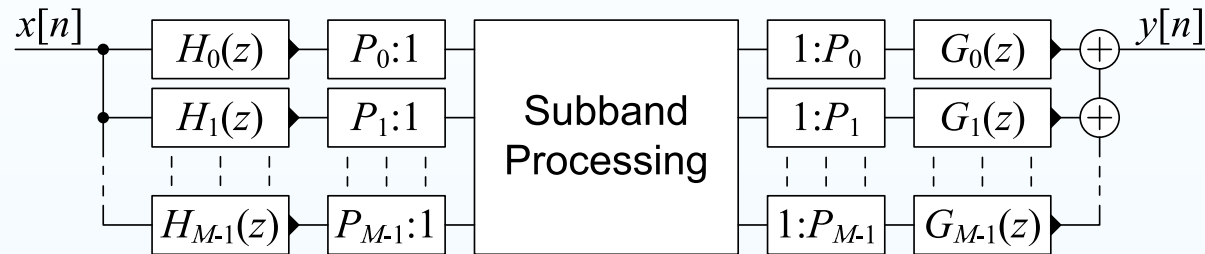
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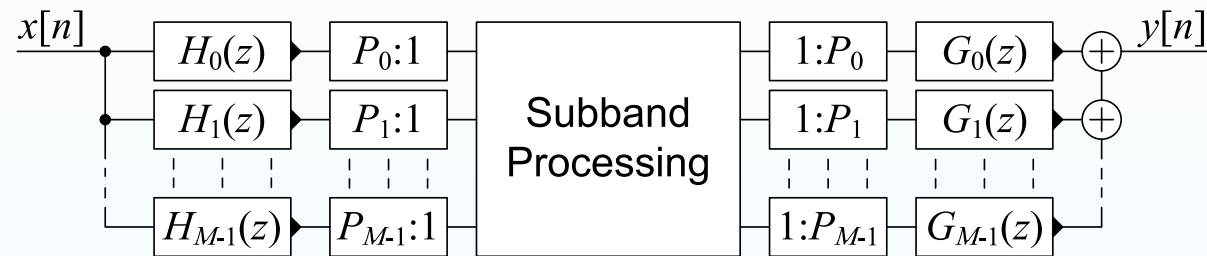
$$\sum \frac{1}{P_i} > 1 \Rightarrow \text{oversampled: more flexible}$$

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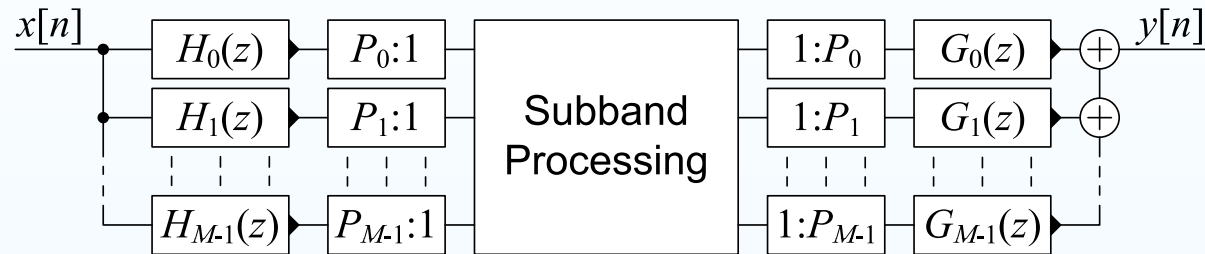
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- Goals:
 - (a) good frequency selectivity in $H_m(z)$

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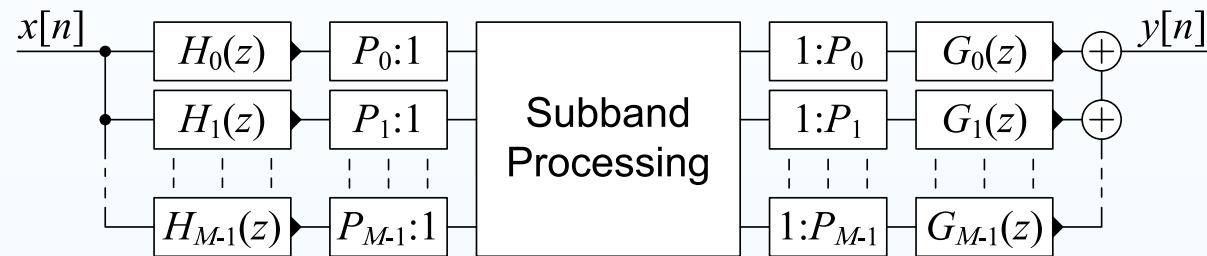
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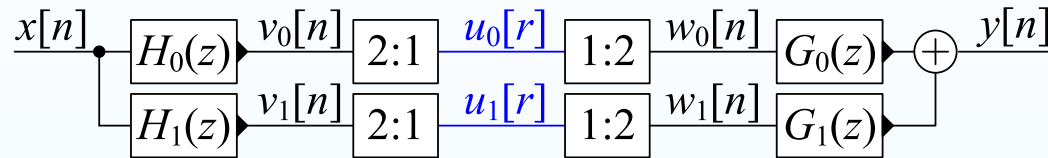
$$\sum \frac{1}{P_i} = 1 \Rightarrow \text{critically sampled: good for coding}$$

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- **Goals:**
 - (a) good frequency selectivity in $H_m(z)$
 - (b) *perfect reconstruction*: $y[n] = x[n - d]$ if no processing
- **Benefits:** Lower computation, faster convergence if adaptive

2-band Filterbank

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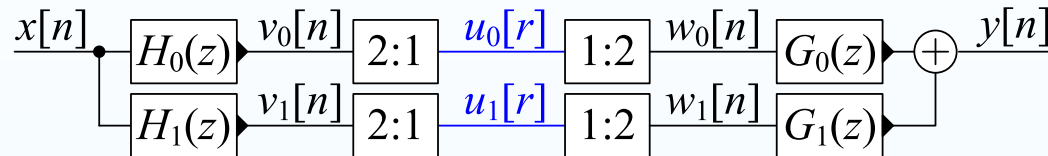
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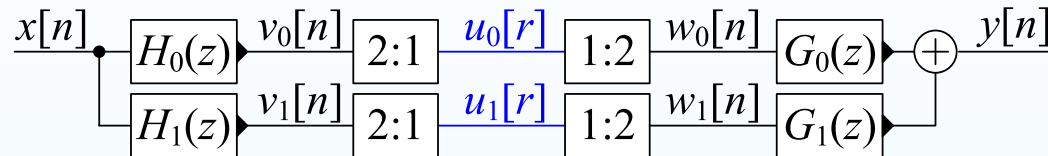
$$V_m(z) = H_m(z)X(z)$$

$$[m \in \{0, 1\}]$$

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$$V_m(z) = H_m(z)X(z)$$

$$U_m(z) = \frac{1}{K} \sum_{k=0}^{K-1} V_m(e^{-j\frac{2\pi k}{K}} z^{\frac{1}{K}})$$

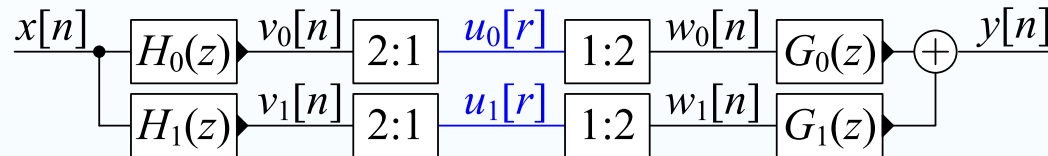
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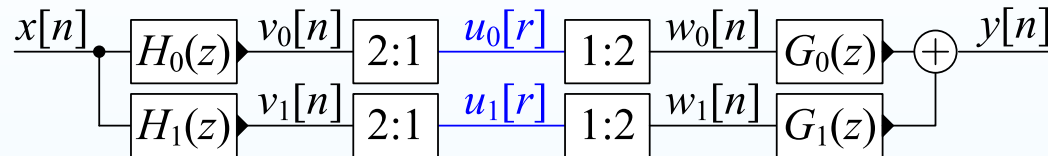
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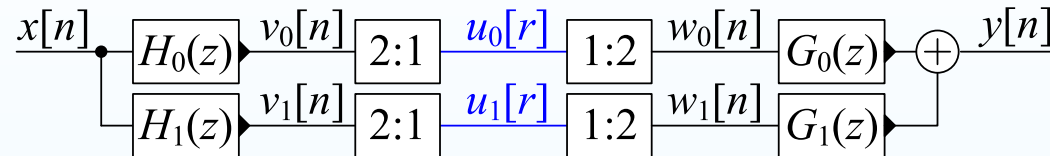
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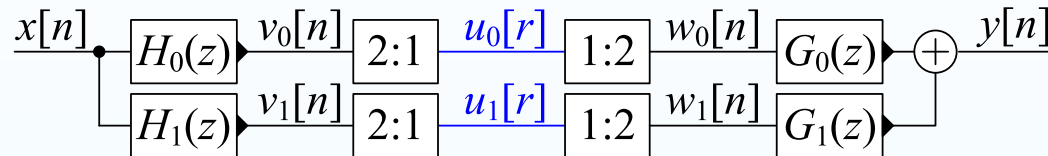


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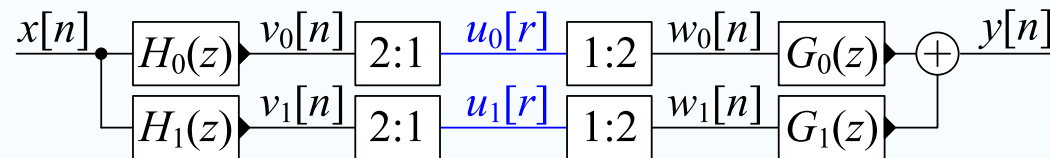
$$= \frac{1}{2} \{H_m(z)X(z) + H_m(-z)X(-z)\}$$

$$Y(z) = \begin{bmatrix} W_0(z) & W_1(z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix}$$

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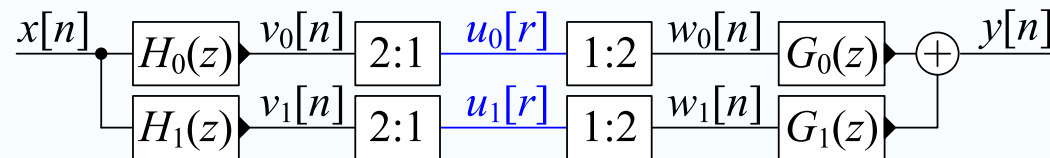
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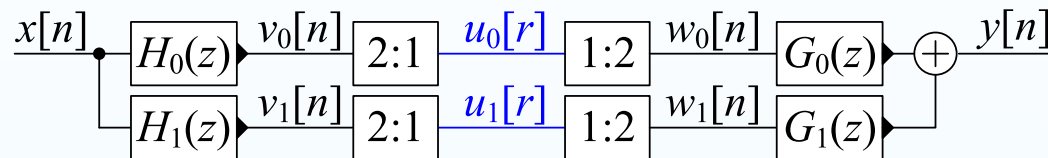
$$= \frac{1}{2} \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix}$$

$$= \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} T(z) \\ A(z) \end{bmatrix}$$

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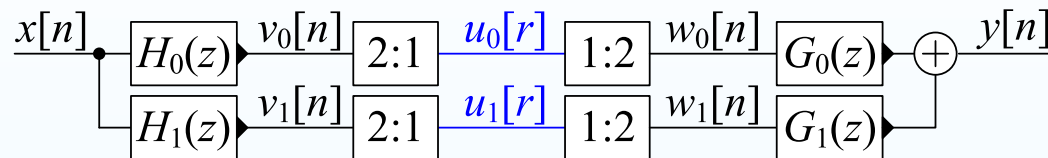
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 Y(z) &= \begin{bmatrix} W_0(z) & W_1(z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} \\
 &= \frac{1}{2} \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} \\
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$$\text{We want (a) } T(z) = \frac{1}{2} \{ H_0(z)G_0(z) + H_1(z)G_1(z) \} = z^{-d}$$

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 &= \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} T(z) \\ A(z) \end{bmatrix} & [X(-z)A(z) \text{ is "aliased" term}]
 \end{aligned}$$

$$\begin{aligned}
 \text{We want (a) } T(z) &= \frac{1}{2} \{ H_0(z)G_0(z) + H_1(z)G_1(z) \} = z^{-d} \\
 \text{and (b) } A(z) &= \frac{1}{2} \{ H_0(-z)G_0(z) + H_1(-z)G_1(z) \} = 0
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Perfect Reconstruction

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For perfect reconstruction without aliasing, we require

$$\frac{1}{2} \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} = \begin{bmatrix} z^{-d} \\ 0 \end{bmatrix}$$

Perfect Reconstruction

15: Subband Processing

- Subband processing
- 2-band Filterbank
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Perfect Reconstruction

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For all filters to be FIR, we need the denominator to be

$$H_0(z)H_1(-z) - H_0(-z)H_1(z) = cz^{-k}$$

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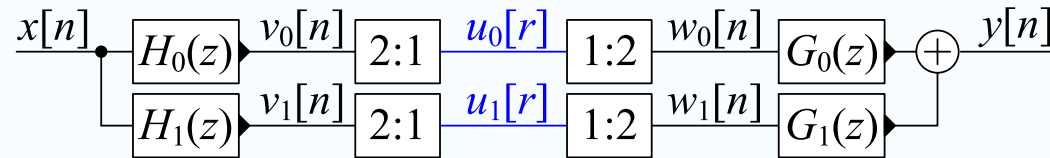
$$\begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} = \frac{2}{c} z^{k-d} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix} \stackrel{d=k}{=} \frac{2}{c} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix}$$

Note: c just scales $H_i(z)$ by $c^{\frac{1}{2}}$ and $G_i(z)$ by $c^{-\frac{1}{2}}$.

Quadrature Mirror Filterbank (QMF)

15: Subband Processing

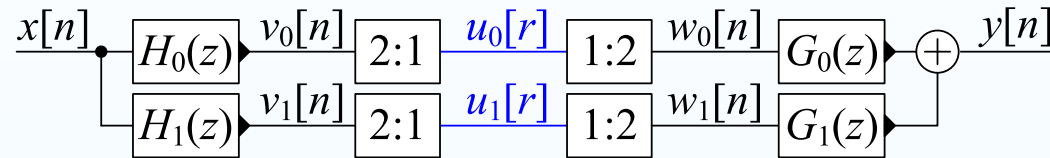
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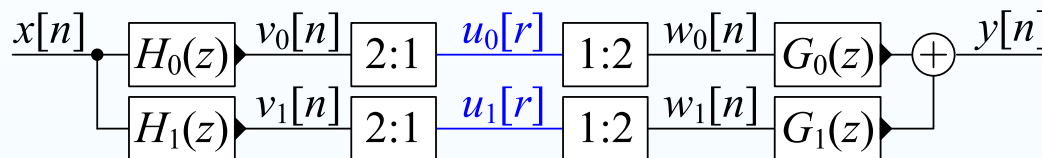
QMF satisfies:

(a) $H_0(z)$ is causal and real

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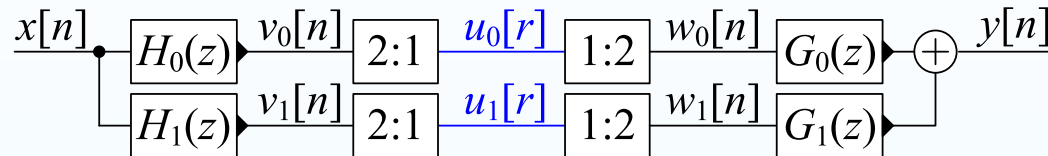
(a) $H_0(z)$ is causal and real

(b) $H_1(z) = H_0(-z)$: i.e. $|H_0(e^{j\omega})|$ is reflected around $\omega = \frac{\pi}{2}$

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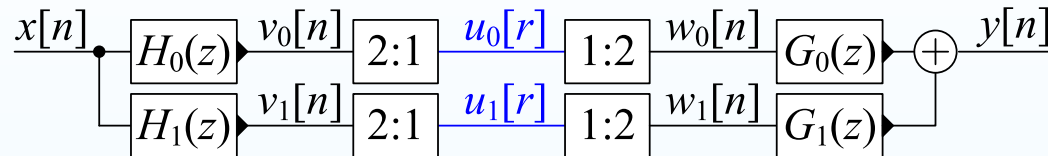
(b) $H_1(z) = H_0(-z)$: i.e. $|H_0(e^{j\omega})|$ is reflected around $\omega = \frac{\pi}{2}$

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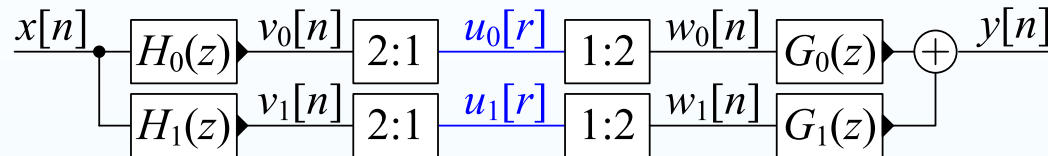
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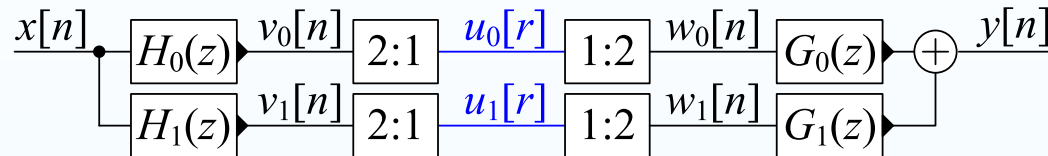
QMF is alias-free:

$$A(z) = \frac{1}{2} \{H_0(-z)G_0(z) + H_1(-z)G_1(z)\}$$

Quadrature Mirror Filterbank (QMF)

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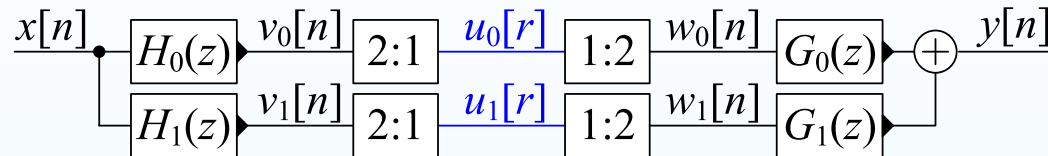
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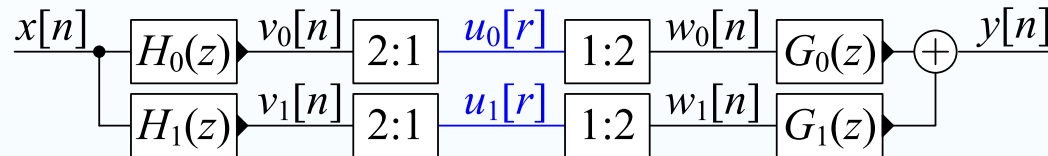
QMF Transfer Function:

$$T(z) = \frac{1}{2} \{H_0(z)G_0(z) + H_1(z)G_1(z)\}$$

Quadrature Mirror Filterbank (QMF)

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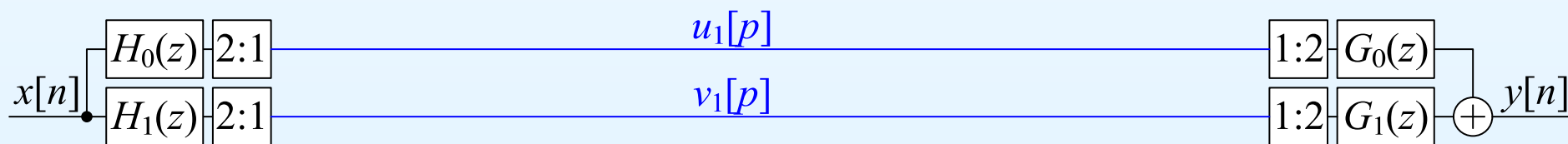
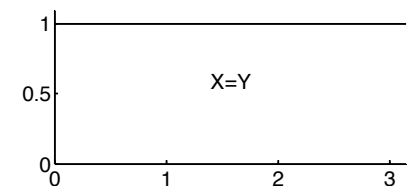
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QMF Transfer Function:

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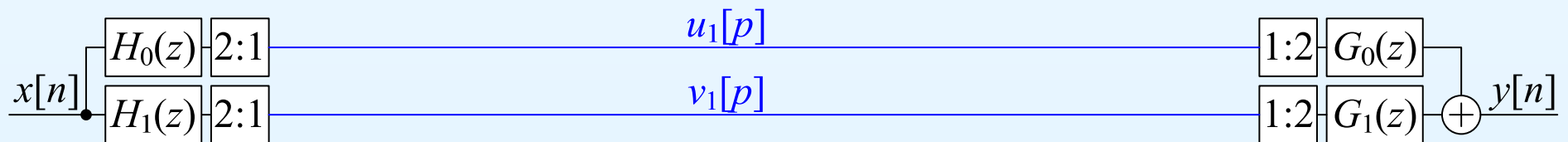
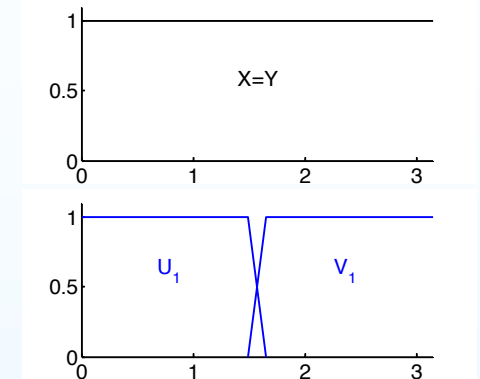
Tree-structured filterbanks

A *half-band filterbank* divides the full band into two equal halves.



Tree-structured filterbanks

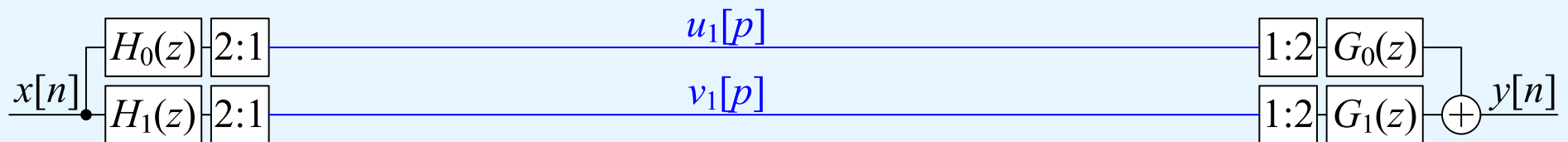
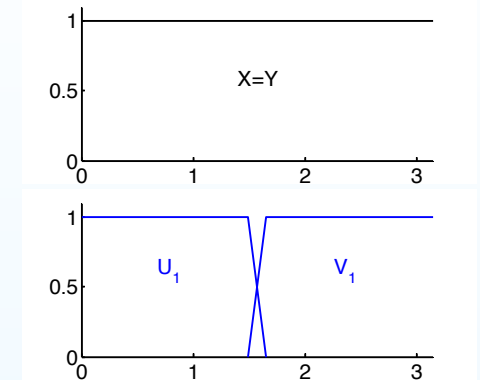
A *half-band filterbank* divides the full band into two equal halves.



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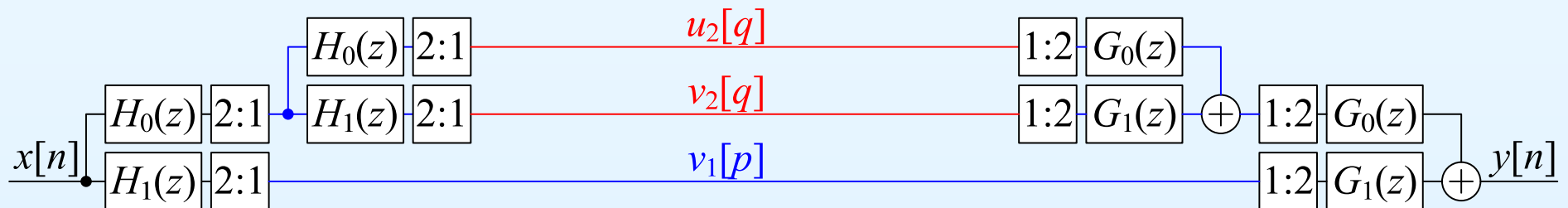
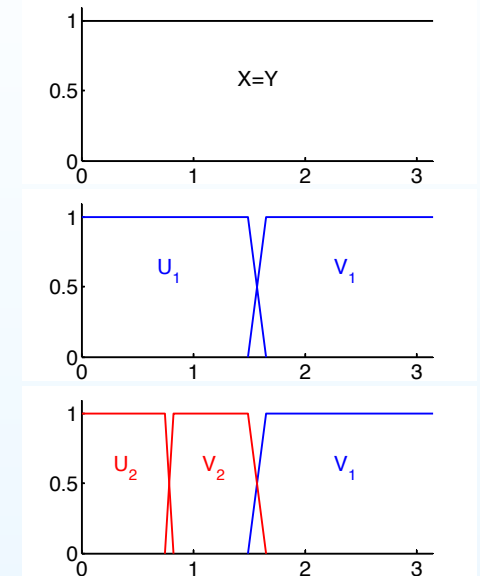
You can repeat the process on either or both of the signals $u_1[p]$ and $v_1[p]$.



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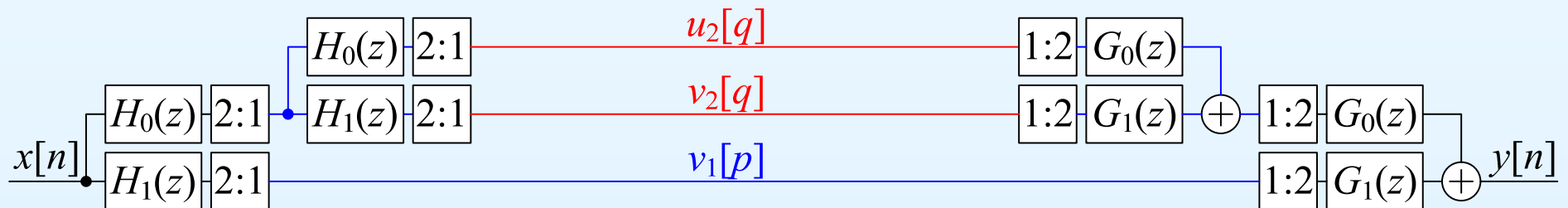
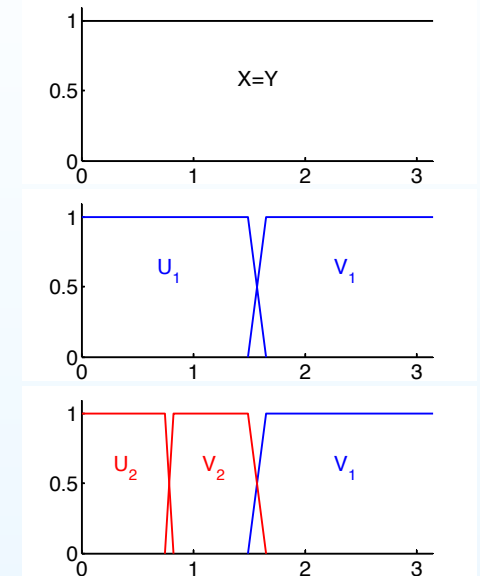


Tree-structured filterbanks

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You can repeat the process on either or both of the signals $u_1[p]$ and $v_1[p]$.

Dividing the lower band in half repeatedly results in an *octave band filterbank*.

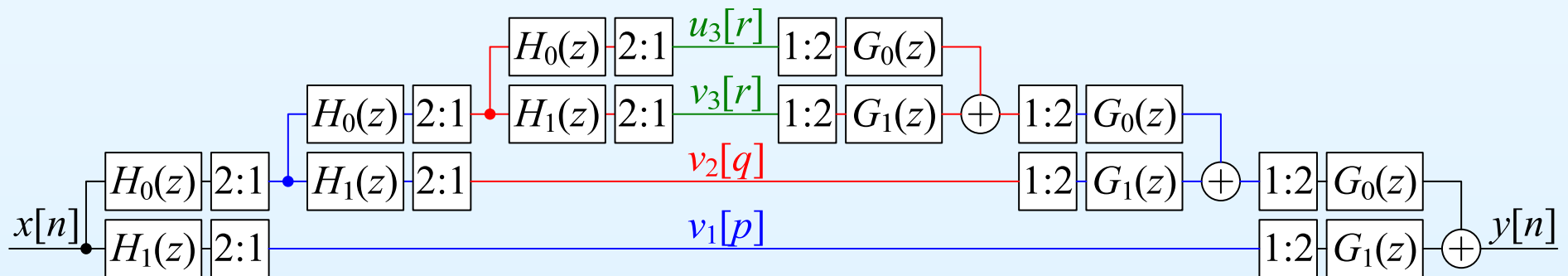
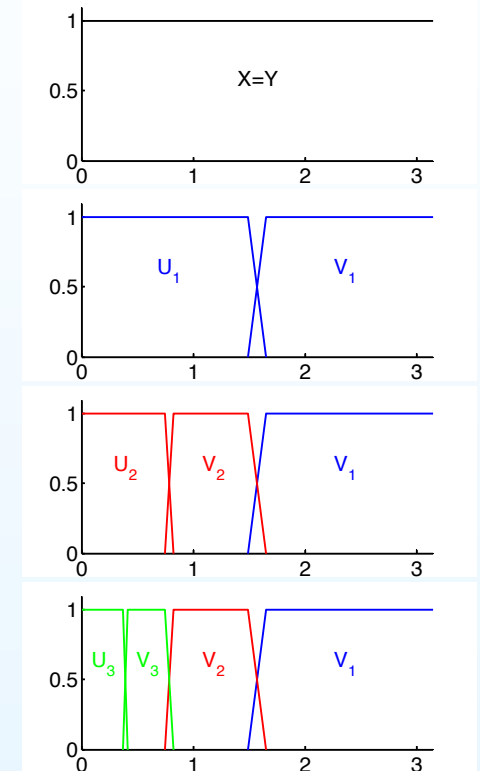


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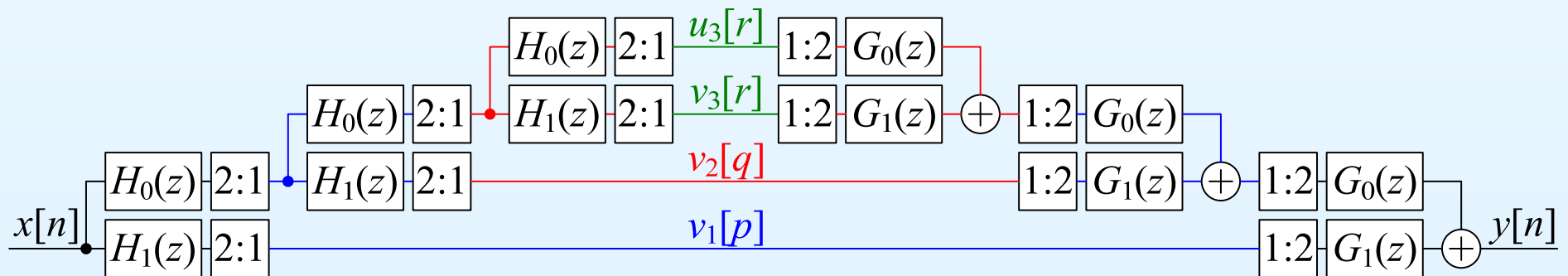
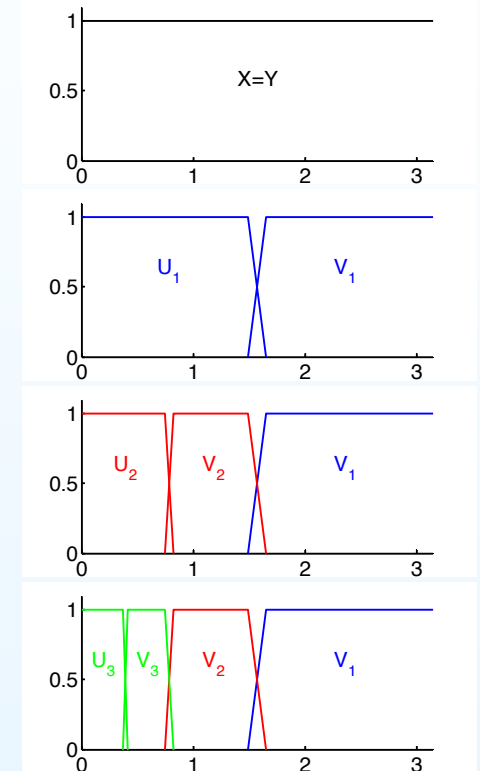


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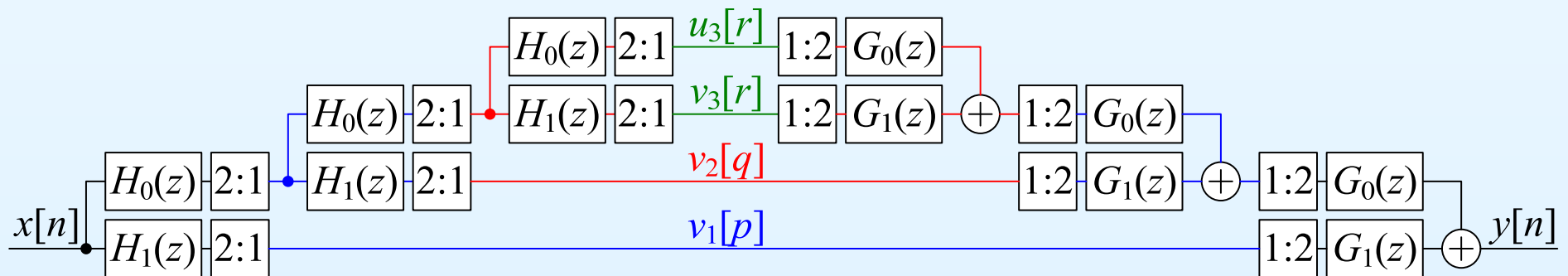
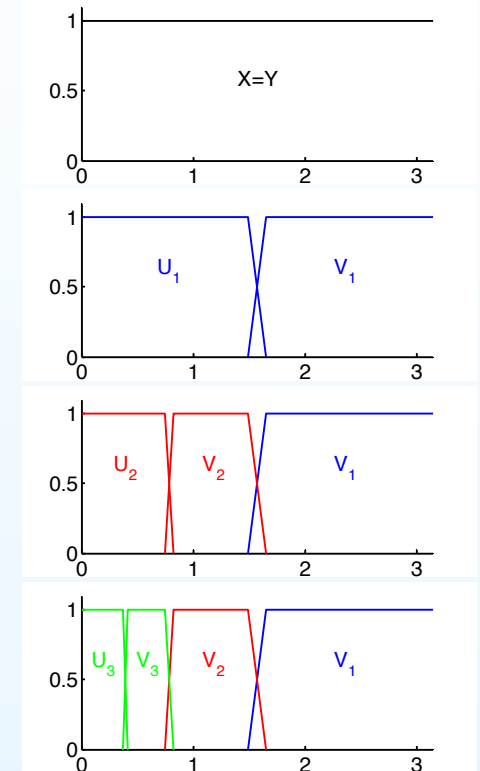
Tree-structured filterbanks

A *half-band filterbank* divides the full band into two equal halves.

You can repeat the process on either or both of the signals $u_1[p]$ and $v_1[p]$.

Dividing the lower band in half repeatedly results in an *octave band filterbank*. Each subband occupies one octave (= a factor of 2 in frequency) except the first subband.

The properties “perfect reconstruction” and “allpass” are preserved by the iteration.



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- Perfect Reconstruction
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- QMF Options
- Linear Phase QMF
- IIR Allpass QMF
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- **Summary**
- Merry Xmas

- **Half-band filterbank:**
 - Reconstructed output is $T(z)X(z) + A(z)X(-z)$
 - Unwanted alias term is $A(z)X(-z)$

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- **Quadrature Mirror Filterbank (QMF)** adds an additional symmetry constraint $H_1(z) = H_0(-z)$.
 - Perfect reconstruction now impossible except for trivial case.
 - Neat polyphase implementation with $A(z) = 0$
 - Johnston filters: Linear phase with $T(z) \approx 1$
 - Allpass filters: Elliptic or Butterworth with $|T(z)| = 1$

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See Mitra chapter 14 (which also includes some perfect reconstruction designs).