

▷ **15: Subband
Processing**

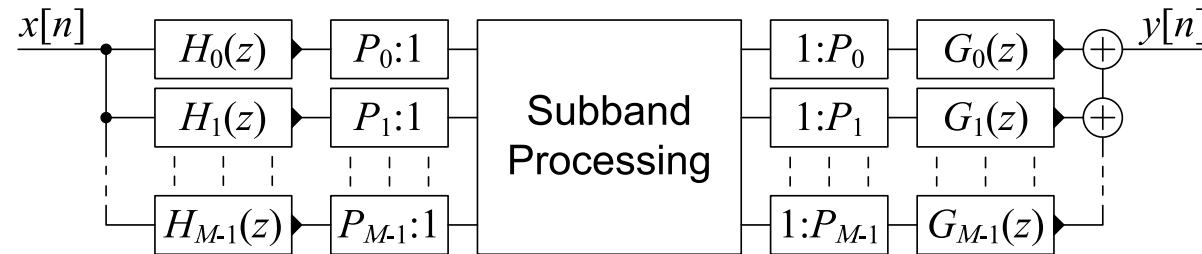
Subband processing
2-band Filterbank
Perfect
Reconstruction
Quadrature Mirror
Filterbank (QMF)
Polyphase QMF
QMF Options
Linear Phase QMF
IIR Allpass QMF
Tree-structured
filterbanks
Summary
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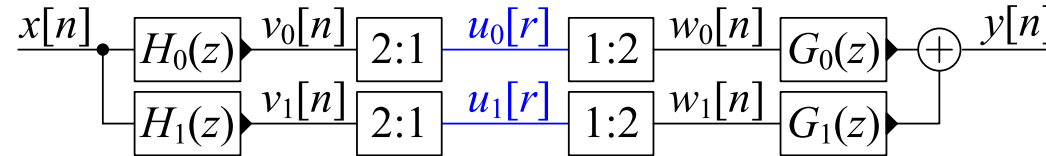
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- The $H_m(z)$ are bandpass *analysis filters* and divide $x[n]$ into frequency bands
- Subband processing often processes frequency bands independently
- The $G_m(z)$ are *synthesis filters* and together reconstruct the output
- The $H_m(z)$ outputs are bandlimited and so can be subsampled without loss of information
 - Sample rate multiplied overall by $\sum \frac{1}{P_i}$
 - $\sum \frac{1}{P_i} = 1 \Rightarrow$ *critically sampled*: good for coding
 - $\sum \frac{1}{P_i} > 1 \Rightarrow$ *oversampled*: more flexible
- **Goals:**
 - (a) good frequency selectivity in $H_m(z)$
 - (b) *perfect reconstruction*: $y[n] = x[n - d]$ if no processing
- **Benefits:** Lower computation, faster convergence if adaptive

2-band Filterbank

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$$V_m(z) = H_m(z)X(z) \quad [m \in \{0, 1\}]$$

$$U_m(z) = \frac{1}{K} \sum_{k=0}^{K-1} V_m(e^{-j\frac{2\pi k}{K}} z^{\frac{1}{K}}) = \frac{1}{2} \left\{ V_m \left(z^{\frac{1}{2}} \right) + V_m \left(-z^{\frac{1}{2}} \right) \right\}$$

$$W_m(z) = U_m(z^2) = \frac{1}{2} \{ V_m(z) + V_m(-z) \} \quad [K = 2]$$

$$= \frac{1}{2} \{ H_m(z)X(z) + H_m(-z)X(-z) \}$$

$$Y(z) = \begin{bmatrix} W_0(z) & W_1(z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix}$$

$$= \begin{bmatrix} X(z) & X(-z) \end{bmatrix} \begin{bmatrix} T(z) \\ A(z) \end{bmatrix} \quad [X(-z)A(z) \text{ is "aliased" term}]$$

We want (a) $T(z) = \frac{1}{2} \{ H_0(z)G_0(z) + H_1(z)G_1(z) \} = z^{-d}$
 and (b) $A(z) = \frac{1}{2} \{ H_0(-z)G_0(z) + H_1(-z)G_1(z) \} = 0$

Perfect Reconstruction

For **perfect reconstruction without aliasing**, we require

$$\frac{1}{2} \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix} \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} = \begin{bmatrix} z^{-d} \\ 0 \end{bmatrix}$$

$$\begin{aligned} \text{Hence: } \begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} &= \begin{bmatrix} H_0(z) & H_1(z) \\ H_0(-z) & H_1(-z) \end{bmatrix}^{-1} \begin{bmatrix} 2z^{-d} \\ 0 \end{bmatrix} \\ &= \frac{2z^{-d}}{H_0(z)H_1(-z) - H_0(-z)H_1(z)} \begin{bmatrix} H_1(-z) & -H_1(z) \\ -H_0(-z) & H_0(z) \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \\ &= \frac{2z^{-d}}{H_0(z)H_1(-z) - H_0(-z)H_1(z)} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix} \end{aligned}$$

For **all filters to be FIR**, we need the denominator to be

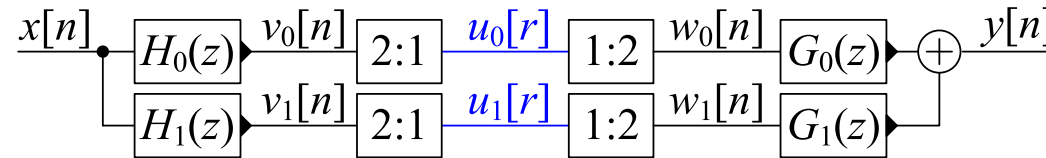
$$H_0(z)H_1(-z) - H_0(-z)H_1(z) = cz^{-k}, \text{ which implies}$$

$$\begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} = \frac{2}{c} z^{k-d} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix} \stackrel{d=k}{=} \frac{2}{c} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix}$$

Note: c just scales $H_i(z)$ by $c^{\frac{1}{2}}$ and $G_i(z)$ by $c^{-\frac{1}{2}}$.

Quadrature Mirror Filterbank (QMF)

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QMF satisfies:

- (a) $H_0(z)$ is causal and real
- (b) $H_1(z) = H_0(-z)$: i.e. $|H_0(e^{j\omega})|$ is reflected around $\omega = \frac{\pi}{2}$
- (c) $G_0(z) = 2H_1(-z) = 2H_0(z)$
- (d) $G_1(z) = -2H_0(-z) = -2H_1(z)$

QMF is alias-free:

$$\begin{aligned} A(z) &= \frac{1}{2} \{H_0(-z)G_0(z) + H_1(-z)G_1(z)\} \\ &= \frac{1}{2} \{2H_1(z)H_0(z) - 2H_0(z)H_1(z)\} = 0 \end{aligned}$$

QMF Transfer Function:

$$\begin{aligned} T(z) &= \frac{1}{2} \{H_0(z)G_0(z) + H_1(z)G_1(z)\} \\ &= H_0^2(z) - H_1^2(z) = H_0^2(z) - H_0^2(-z) \end{aligned}$$

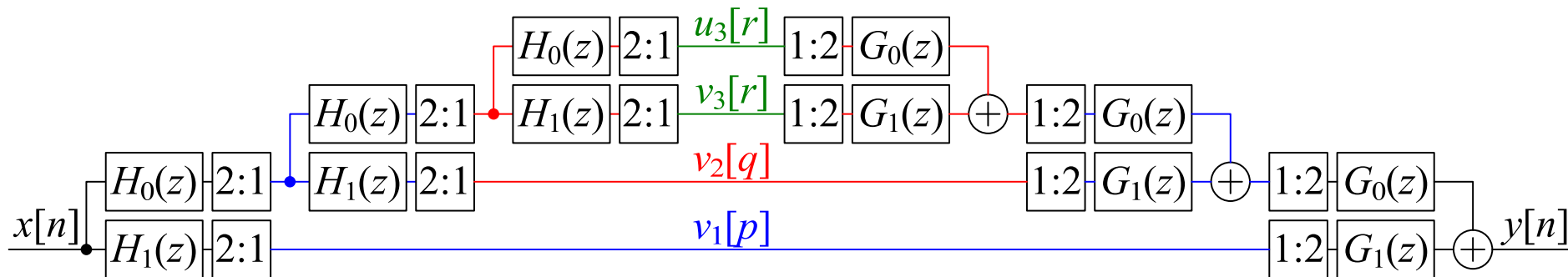
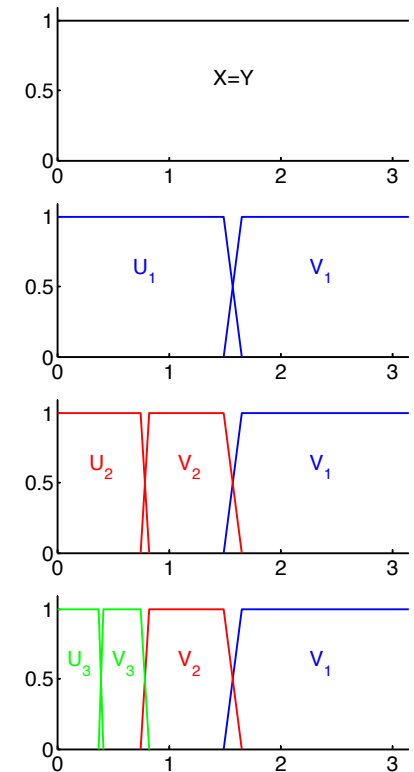
Tree-structured filterbanks

A *half-band filterbank* divides the full band into two equal halves.

You can repeat the process on either or both of the signals $u_1[p]$ and $v_1[p]$.

Dividing the lower band in half repeatedly results in an *octave band filterbank*. Each subband occupies one octave (= a factor of 2 in frequency) except the first subband.

The properties “perfect reconstruction” and “allpass” are preserved by the iteration.



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- **Half-band filterbank:**
 - Reconstructed output is $T(z)X(z) + A(z)X(-z)$
 - Unwanted alias term is $A(z)X(-z)$
- **Perfect reconstruction:** imposes strong constraints on analysis filters $H_i(z)$ and synthesis filters $G_i(z)$.
- **Quadrature Mirror Filterbank (QMF)** adds an additional symmetry constraint $H_1(z) = H_0(-z)$.
 - Perfect reconstruction now impossible except for trivial case.
 - Neat polyphase implementation with $A(z) = 0$
 - Johnston filters: Linear phase with $T(z) \approx 1$
 - Allpass filters: Elliptic or Butterworth with $|T(z)| = 1$
- Can iterate to form a tree structure with equal or unequal bandwidths.

See Mitra chapter 14 (which also includes some perfect reconstruction designs).