

Announcement

Updated lecture notes, handouts and a practice sheet with problems has been uploaded to the course website. <http://alumni.media.MIT.edu/~ayush/course>

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- Organization
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- Sequences
- Time Scaling
- z-Transform
- Region of Convergence
- z-Transform examples
- Rational z-Transforms
- Rational example
- Inverse z-Transform
- MATLAB routines
- Summary

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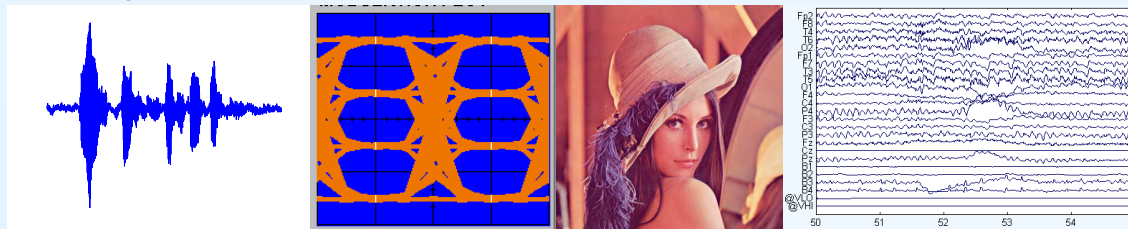
Signals

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- A signal is a numerical quantity that is a function of one or more independent variables such as time or position.

Examples:



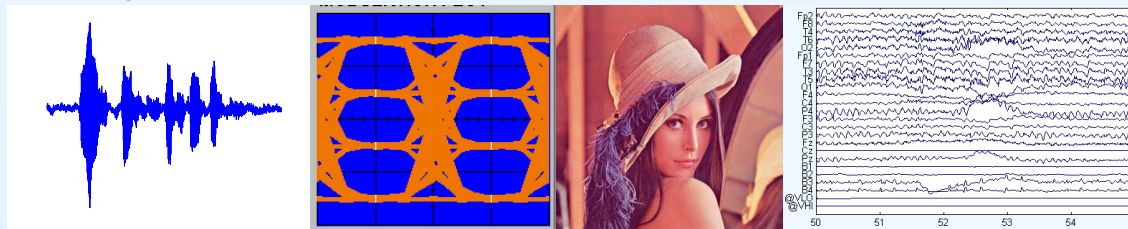
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- A signal is a numerical quantity that is a function of one or more independent variables such as time or position.
- Real-world signals are analog and vary continuously and take continuous values.

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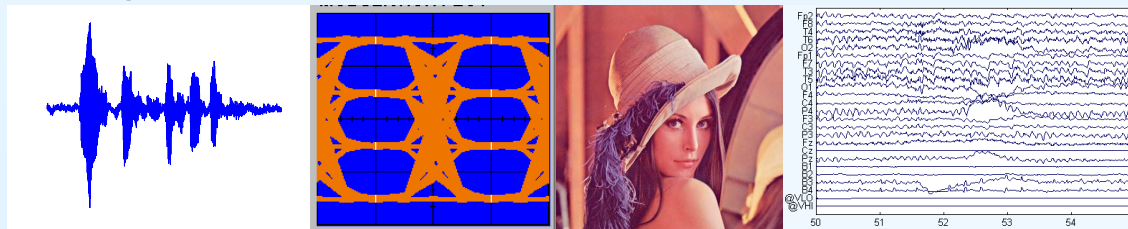
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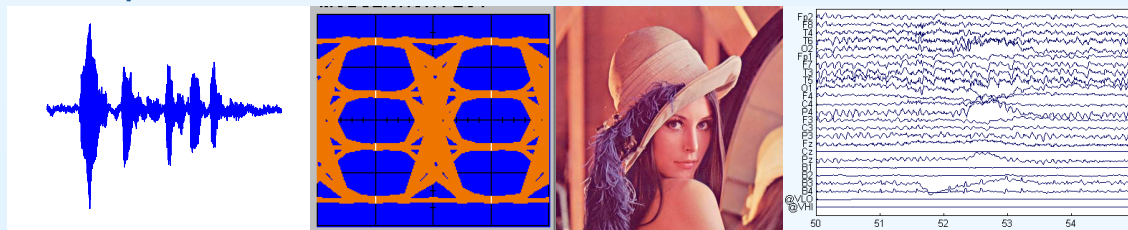
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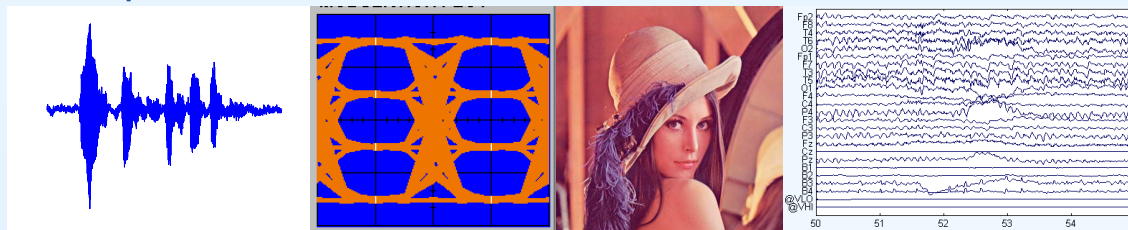
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- Real-world signals are analog and vary continuously and take continuous values.
- Digital signals are sampled at discrete times and are quantized to a finite number of discrete values
- We will mostly consider one-dimensionalal real-valued signals with regular sample instants; except in a few places, we will ignore the quantization.
 - Extension to multiple dimensions and complex-valued signals is straightforward in many cases.

Examples:



Processing

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- Aims to “improve” a signal in some way or extract some information from it
- Examples:
 - Modulation/demodulation
 - Coding and decoding
 - Interference rejection and noise suppression
 - Signal detection, feature extraction
- We are concerned with linear, time-invariant processing

Syllabus

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Main topics:

- Introduction/Revision
- Transforms
- Discrete Time Systems
- Filter Design
 - FIR Filter Design
 - IIR Filter Design
- Multirate systems
 - Multirate Fundamentals
 - Multirate Filters
 - Subband processing

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- **Absolutely Summable:** $\sum_{n=-\infty}^{\infty} |x[n]| < \infty \Rightarrow \text{Finite energy}$

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- Finite Energy: $\sum_{n=-\infty}^{\infty} |x[n]|^2 < \infty$ (e.g. $x[n] = n^{-1}u[n-1]$)
- Absolutely Summable: $\sum_{n=-\infty}^{\infty} |x[n]| < \infty \Rightarrow$ Finite energy

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We usually scale time so that $f_s = 1$: divide all “real” frequencies and angular frequencies by f_s and divide all “real” times by T .

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Warning: Several MATLAB routines scale time so that $f_s = 2$ Hz. Weird, non-standard and irritating.

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- Complex functions are easier to manipulate than sequences

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- Useful operations on sequences correspond to simple operations on the z -transform:
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Why do it?

- Complex functions are easier to manipulate than sequences
- Useful operations on sequences correspond to simple operations on the z -transform:
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- Definition: $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$

Region of Convergence

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The set of z for which $X(z)$ converges is its *Region of Convergence* (ROC).

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$$\lim_{N \rightarrow \infty} \left| x_0 - \sum_{n \leq N} x[n] \right| = 0.$$

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Series converges for $|r| < 1$ because,

$$|r| < 1, \quad \lim_{N \rightarrow \infty} \frac{1 - r^{N+1}}{1 - r} = \frac{1}{1 - r}$$

Convergence: Example

Example: Consider $r = \frac{1}{2}$. The series converges to $x_0 = 2$. Why?

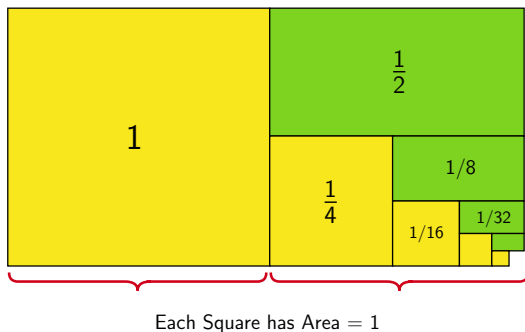
$$\left| 2 - \sum_{n \leq N} \left(\frac{1}{2}\right)^n \right| = \left| 2 - \frac{1 - \left(\frac{1}{2}\right)^{N+1}}{1 - \frac{1}{2}} \right| = \left(\frac{1}{2}\right)^N, \quad \lim_{N \rightarrow \infty} \left(\frac{1}{2}\right)^N = 0.$$

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Visual Proof: Slicing “Squares”.



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There are examples of series that converge at ∞ .

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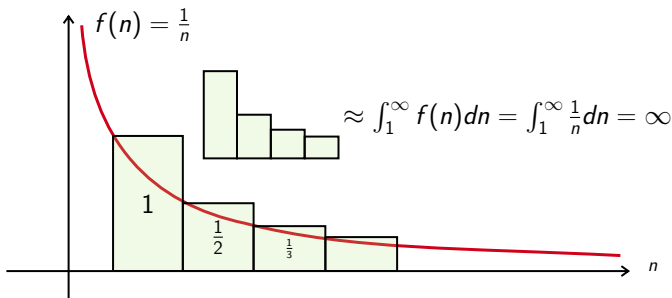
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However, this series converges at ∞ .



ROC and its Shape

On the z -plane, the ROC is always an annulus, that is,

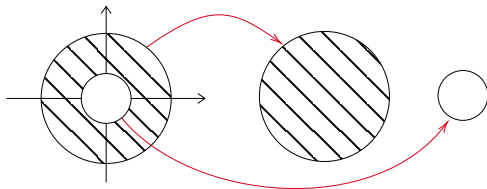
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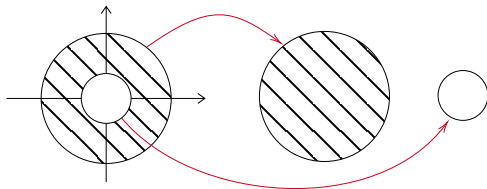


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Why?

ROC and its Shape

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Why? Because... Let $z = re^{j\theta}$ (polar co-ordinates).

$$\begin{aligned} |X(z)| &= \left| \sum_n x[n] z^{-n} \right| \leq \left| \sum_n |x[n]| r^{-n} \right| \\ &= \underbrace{\sum_{n \geq 1} |x[-n]| r^n}_{\text{Anti-causal Part}} + \underbrace{\sum_{n \geq 0} \frac{|x[n]|}{r^n}}_{\text{Causal Part}} \end{aligned}$$

- On the z -plane, ROC is the region where both parts of the sum are finite.
- When the causal part is finite, the series converges on the exterior of some circle.
- When the anti-causal part is finite, the series converges on the interior of some circle.

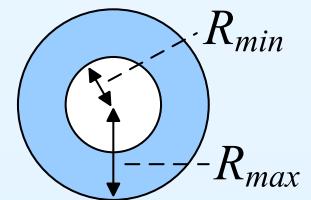
Region of Convergence

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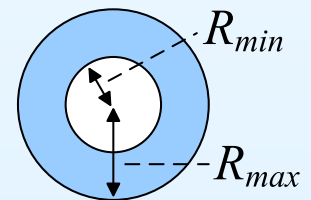
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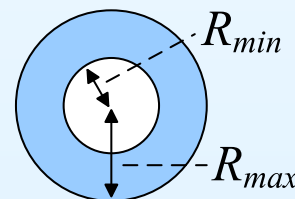
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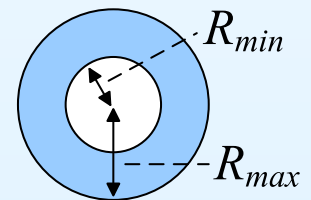
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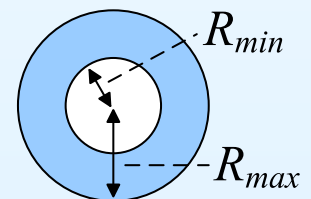
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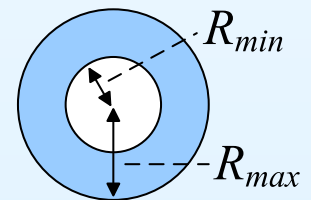
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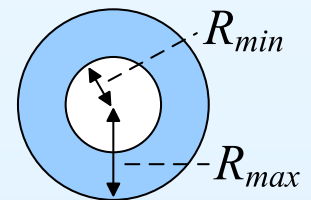
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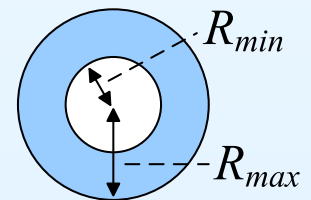
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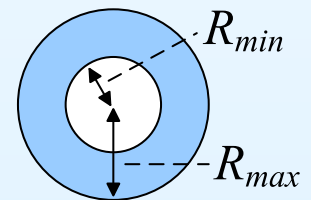
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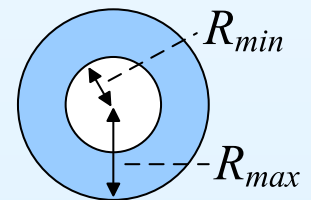
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z-Transform examples

The sample at $n = 0$ is indicated by an open circle.



z-Transform examples

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$$u[n] \quad \cdots \cdot \cdot \cdot \circ \uparrow \uparrow \uparrow \uparrow \cdots$$

$$\frac{1}{1-z^{-1}}$$

$$1 < |z| \leq \infty$$

$$\text{Geometric Progression: } \sum_{n=q}^r \alpha^n z^{-n} = \frac{\alpha^q z^{-q} - \alpha^{r+1} z^{-r-1}}{1 - \alpha z^{-1}}$$

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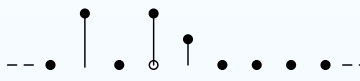
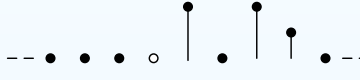
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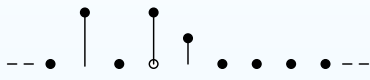
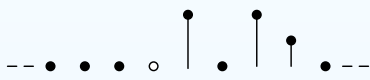
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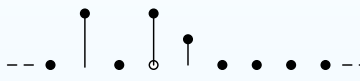
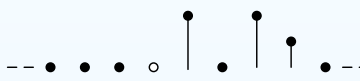

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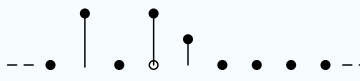
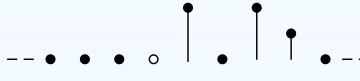

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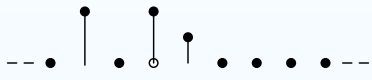
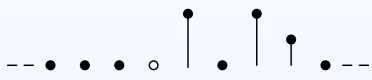


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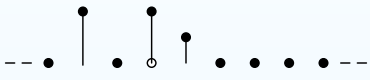
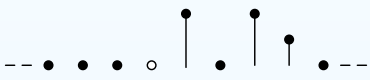


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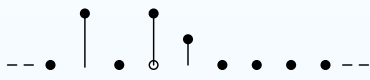
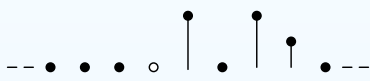


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$$\text{Geometric Progression: } \sum_{n=q}^r \alpha^n z^{-n} = \frac{\alpha^q z^{-q} - \alpha^{r+1} z^{-r-1}}{1 - \alpha z^{-1}}$$

z-Transform examples

The sample at $n = 0$ is indicated by an open circle.

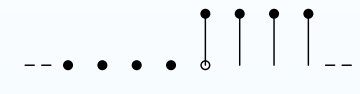
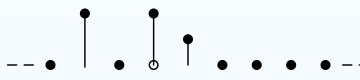


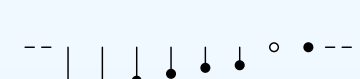

$u[n]$		$\frac{1}{1-z^{-1}}$	$1 < z \leq \infty$
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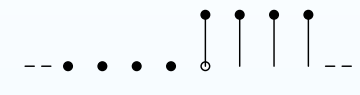
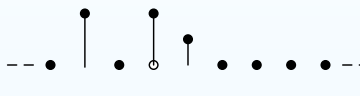
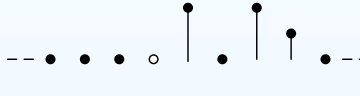



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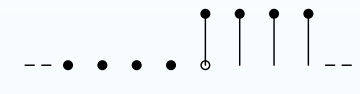
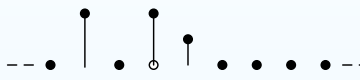


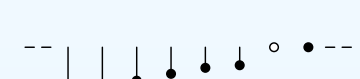


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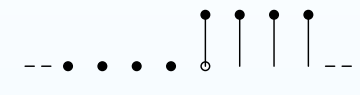
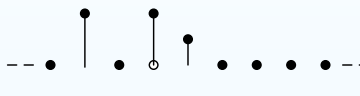
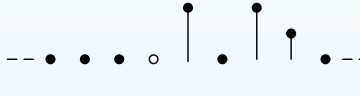



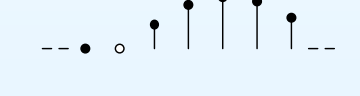
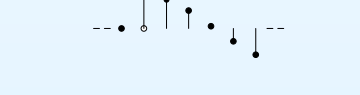
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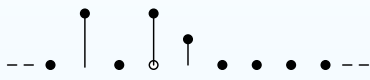
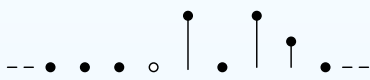





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Examples of Z-transform

- 1 Causal Sinusoid $x[n] = \cos(\omega_0 n + \xi) u[n]$.

Start with,

$$x[n] = \frac{1}{2} (\exp(j(\omega_0 n + \xi)) + \exp(-j(\omega_0 n + \xi))) = \frac{\alpha}{2} \beta^n u[n] + \frac{\alpha^*}{2} (\beta^*)^n u[n].$$

Clearly, $(\frac{\alpha}{2}) \beta^n u[n] \xrightarrow{\text{Z-transform}} (\frac{\alpha}{2}) \frac{1}{1 - \beta z^{-1}}$ and likewise for the conjugate term.
Finally,

$$X(z) = \left(\frac{\alpha}{2}\right) \frac{1}{1 - \beta z^{-1}} + \left(\frac{\alpha^*}{2}\right) \frac{1}{1 - \beta^* z^{-1}} = \frac{\cos \xi - z^{-1} \cos(\omega_0 - \xi)}{1 - 2z^{-1} \cos \omega_0 + z^{-2}}.$$

✗

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- ② Suppose $x[n] \rightarrow X(z)$. Find the z-transform of $nx[n]$. We have,

$$X(z) = \sum_n x[n] z^{-n} \Rightarrow \frac{d}{dz} X(z) = - \sum_n (nx[n]) z^{-n-1} = - \underbrace{\frac{1}{z} \sum_n (nx[n]) z^{-n}}_{Z\{nx[n]\}}.$$

Hence,

$$Z\{nx[n]\} = -z \left(\frac{d}{dz} X(z) \right).$$

Rational z-Transforms

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Most z -transforms that we will meet are **rational polynomials** with real coefficients, usually one polynomial in z^{-1} divided by another.

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Most z -transforms that we will meet are **rational polynomials** with real coefficients, usually one polynomial in z^{-1} divided by another.

$$G(z) = g \frac{\prod_{m=1}^M (1 - z_m z^{-1})}{\prod_{k=1}^K (1 - p_k z^{-1})}$$

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$\exists R + 1$ different ROCs

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$$G(z) = g \frac{\prod_{m=1}^M (1 - z_m z^{-1})}{\prod_{k=1}^K (1 - p_k z^{-1})} = g z^{K-M} \frac{\prod_{m=1}^M (z - z_m)}{\prod_{k=1}^K (z - p_k)}$$

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Note: There are $K - M$ zeros or $M - K$ poles at $z = 0$ (**easy to overlook**)

Rational example

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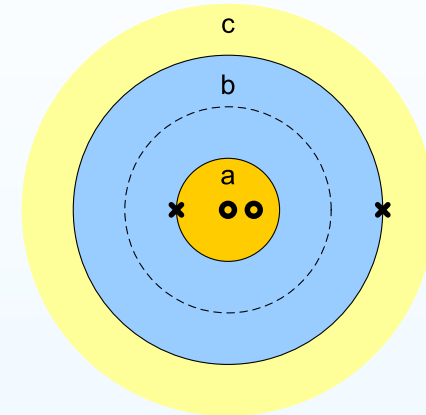
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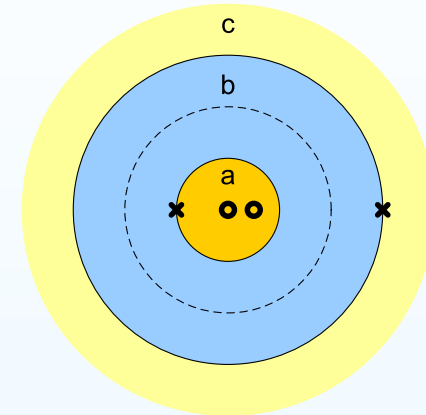
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Partial Fractions: $G(z) = \frac{0.75}{1+0.5z^{-1}} + \frac{1.25}{1-1.5z^{-1}}$



Rational example

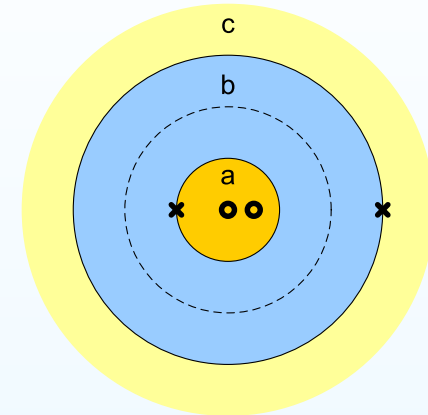
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Partial Fractions: $G(z) = \frac{0.75}{1+0.5z^{-1}} + \frac{1.25}{1-1.5z^{-1}}$

ROC	ROC	$\frac{0.75}{1+0.5z^{-1}}$	$\frac{1.25}{1-1.5z^{-1}}$	$G(z)$
a	$0 \leq z < 0.5$			
b	$0.5 < z < 1.5$			
c	$1.5 < z \leq \infty$			

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Inverse z-Transform

$g[n] = \frac{1}{2\pi j} \oint G(z) z^{n-1} dz$ where the integral is anti-clockwise around a circle within the ROC, $z = Re^{j\theta}$.

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Proof:

$$\frac{1}{2\pi j} \oint G(z) z^{n-1} dz = \frac{1}{2\pi j} \oint \left(\sum_{m=-\infty}^{\infty} g[m] z^{-m} \right) z^{n-1} dz$$

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Inverse z-Transform

$g[n] = \frac{1}{2\pi j} \oint G(z) z^{n-1} dz$ where the integral is anti-clockwise around a circle within the ROC, $z = Re^{j\theta}$.

Proof:

$$\begin{aligned} \frac{1}{2\pi j} \oint G(z) z^{n-1} dz &= \frac{1}{2\pi j} \oint \left(\sum_{m=-\infty}^{\infty} g[m] z^{-m} \right) z^{n-1} dz \\ &\stackrel{(i)}{=} \sum_{m=-\infty}^{\infty} g[m] \frac{1}{2\pi j} \oint z^{n-m-1} dz \end{aligned}$$

(i) depends on the circle with radius R lying within the ROC

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(ii) Cauchy's theorem: $\frac{1}{2\pi j} \oint z^{k-1} dz = \delta[k]$ for $z = Re^{j\theta}$ anti-clockwise.

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$$= R^k \delta(k) = \delta(k) \quad [R^0 = 1]$$

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In practice use a combination of partial fractions and table of z -transforms.

MATLAB routines

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tf2zp,zp2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \{z_m, p_k, g\}$
residuez	$\frac{b(z^{-1})}{a(z^{-1})} \rightarrow \sum_k \frac{r_k}{1-p_k z^{-1}}$
tf2sos,sos2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \prod_l \frac{b_{0,l}+b_{1,l}z^{-1}+b_{2,l}z^{-2}}{1+a_{1,l}z^{-1}+a_{2,l}z^{-2}}$
zp2sos,sos2zp	$\{z_m, p_k, g\} \leftrightarrow \prod_l \frac{b_{0,l}+b_{1,l}z^{-1}+b_{2,l}z^{-2}}{1+a_{1,l}z^{-1}+a_{2,l}z^{-2}}$
zp2ss,ss2zp	$\{z_m, p_k, g\} \leftrightarrow \begin{cases} x' = Ax + Bu \\ y = Cx + Du \end{cases}$
tf2ss,ss2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \begin{cases} x' = Ax + Bu \\ y = Cx + Du \end{cases}$

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- Time scaling: assume $f_s = 1$ so $-\pi < \omega \leq \pi$

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- Time scaling: assume $f_s = 1$ so $-\pi < \omega \leq \pi$
- z-transform: $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$

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Summary

- **Time scaling:** assume $f_s = 1$ so $-\pi < \omega \leq \pi$
- **z-transform:** $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$
- **ROC:** $0 \leq R_{min} < |z| < R_{max} \leq \infty$

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- z-transform: $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$
- ROC: $0 \leq R_{min} < |z| < R_{max} \leq \infty$
 - Causal: $\infty \in \text{ROC}$
 - Absolutely summable: $|z| = 1 \in \text{ROC}$
- Inverse z-transform: $g[n] = \frac{1}{2\pi j} \oint G(z)z^{n-1}dz$
 - Not unique unless ROC is specified

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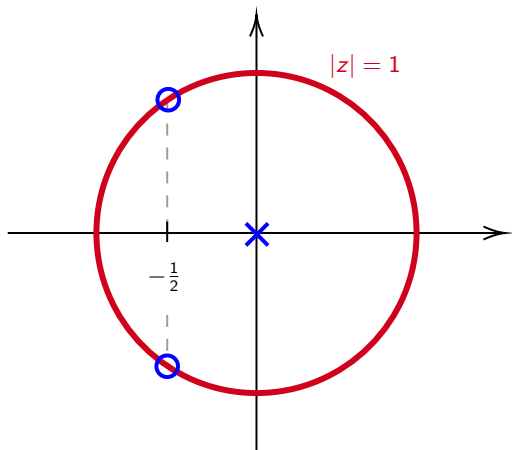
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- **Inverse z-transform:** $g[n] = \frac{1}{2\pi j} \oint G(z)z^{n-1}dz$
 - **Not unique** unless ROC is specified
 - Use **partial fractions** and/or a **table**

Exam 2020

The pole-zero plot of a discrete filter is given below.

When the input $x[n] = 1$ for all n , the output is exactly the same.

What is the impulse response of such a filter?



Further questions on practice sheet announced on course website.

<http://alumni.media.mit.edu/~ayush/course>