

# MIMO-OFDM Wireless Communications

## Introduction to OFDM

**Edward Kwao**

Hanbat National University

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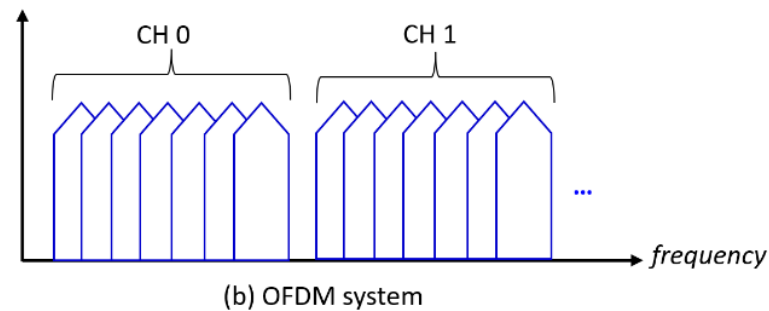
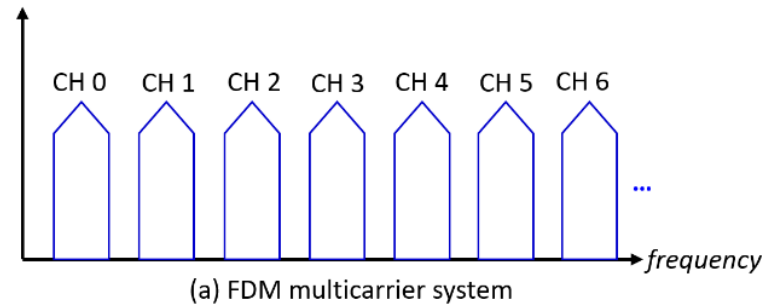
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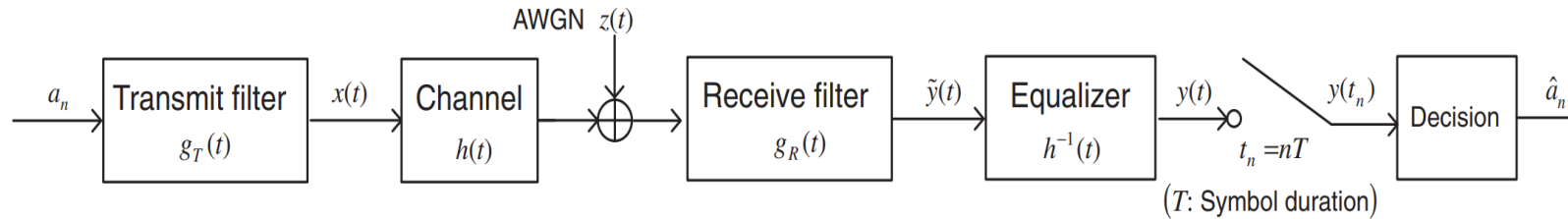
# Introduction to OFDM

- Problem with high bandwidth systems → Frequency Selective Fading
- Orthogonal Frequency Division Multiplexing
- Special case of multicarrier FDM transmission →

- Overlapping subchannels
- Orthogonal subchannels



# Single-carrier Transmission



$$y(t) = \sum_{m=-\infty}^{\infty} a_m g(t - mT) + z(t) \dots \dots \dots (1)$$

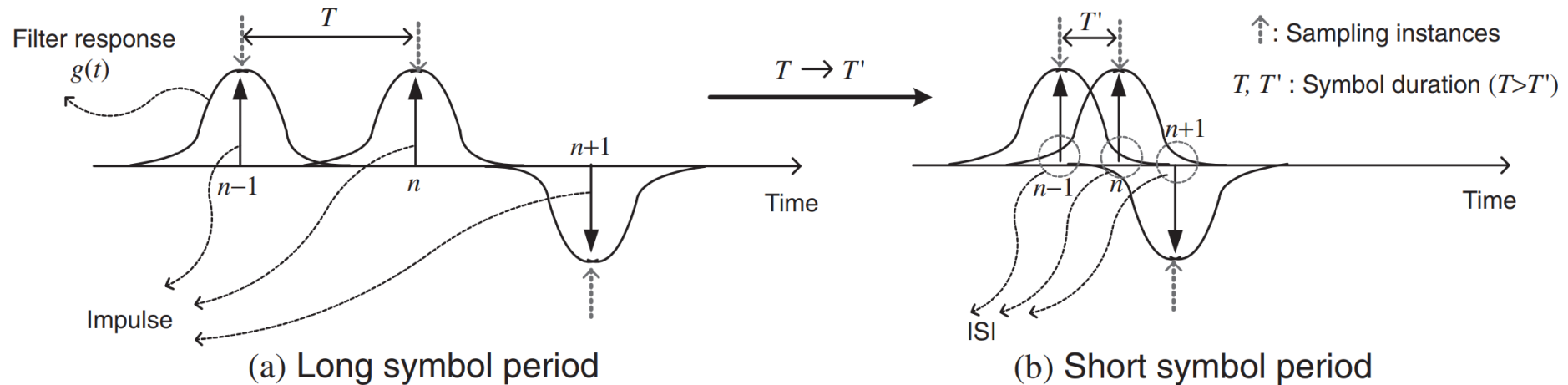
For error-free transmission,  
 $B_{x(t)} < B_{channel}$   
Else, **ISI**

**Sampled output of equalizer:**  $y(t_n) = \sum_{m=-\infty}^{\infty} a_m g((n - m)T) \dots \dots \dots (2)$ , with  $t_n = nT$

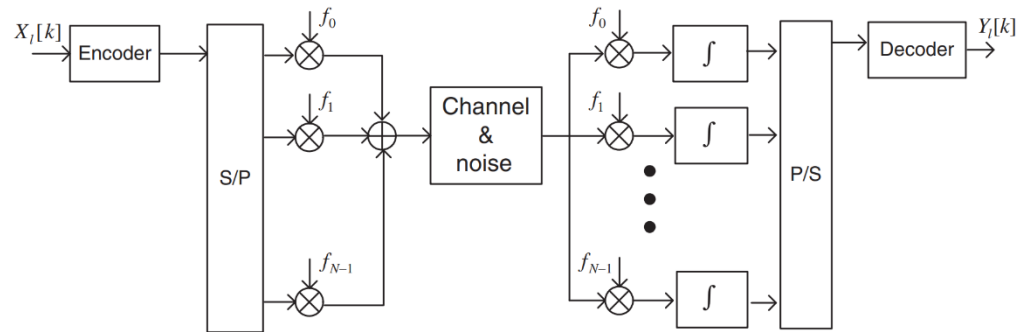
**Isolating  $n^{th}$  sample to Detect  $a_n$  :**

$$y(t_n) = a_n g(0) + \sum_{m=-\infty, m \neq n}^{\infty} a_m g((n - m)T) \dots \dots \dots (3)$$

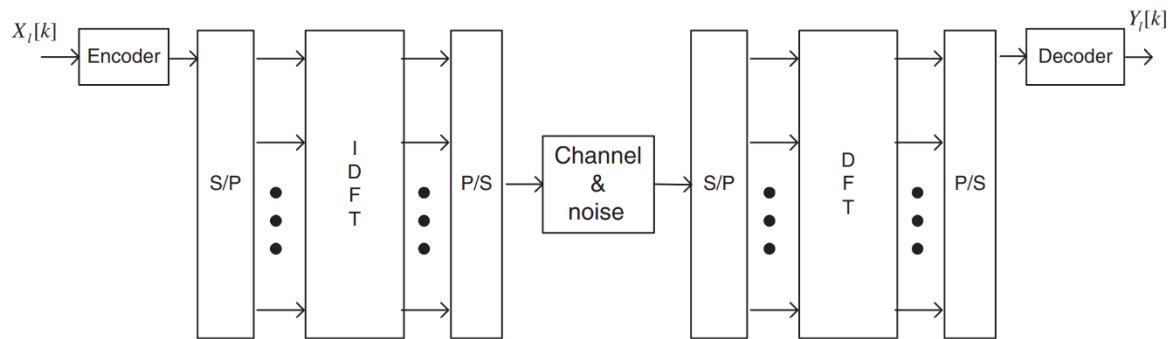
- The extent of ISI depends on the duration of a symbol period,  $T$
- ISI becomes significant as the data rate increases.



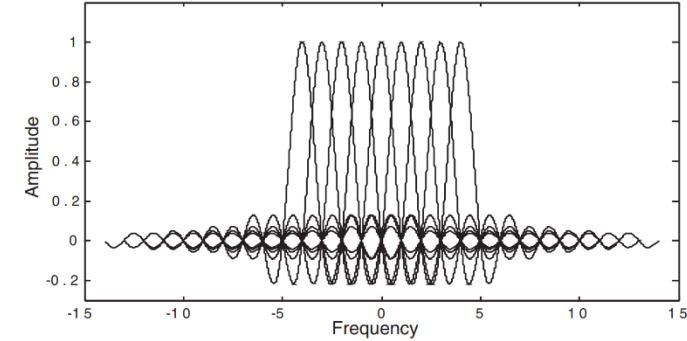
# Multi-carrier Transmission



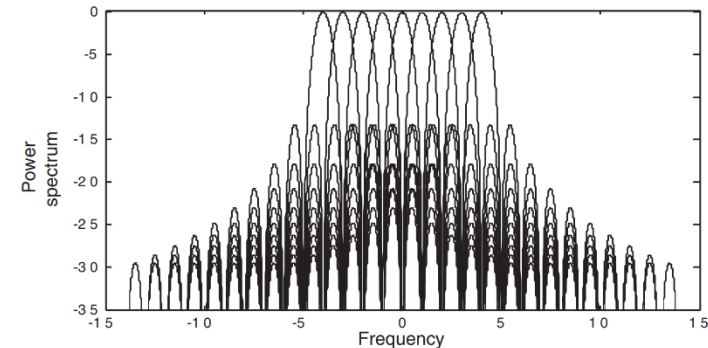
(a) Outline of OFDM transmission scheme



(b) OFDM transmission scheme implemented using IDFT/DFT



(c) The spectrum of OFDM signal (linear scale)



(d) Power spectrum of OFDM signal (dB)

DFT and IDFT are implemented efficiently using **FFT** and **IFFT** respectively

- **Guard bands** reduce out-of-band radiation.
- **Guard interval** called Cyclic Prefix mitigates ISI

# Basic Principles of OFDM

## OFDM Modulation and Demodulation

- How is *orthogonality* achieved in OFDM?

Consider time-limited complex exponential signal:  $\{e^{2\pi f_k t}\}_{k=0}^{N-1}$

At different subcarriers,  $f_k = \frac{k}{T_{sym}}$ , where  $0 \leq t \leq T_{sym}$

$$\frac{1}{T_{sym}} \int_0^{T_{sym}} e^{j2\pi f_k t} e^{-j2\pi f_i t} dt = \frac{1}{T_{sym}} \int_0^{T_{sym}} e^{j2\pi \frac{k}{T_{sym}} t} e^{-j2\pi \frac{i}{T_{sym}} t} dt$$

$$= \frac{1}{T_{sym}} \int_0^{T_{sym}} e^{j2\pi \frac{(k-i)}{T_{sym}} t} dt$$

$$= \begin{cases} 1, & \forall \text{ integer } k = i \\ 0, & \text{otherwise} \end{cases} \longrightarrow$$

Orthogonality  
condition

- Taking the discrete samples with the sampling instances at  $t = nT_s = \frac{nT_{sym}}{N}$ ,  
 $n = 0, 1, 2, \dots, N - 1$

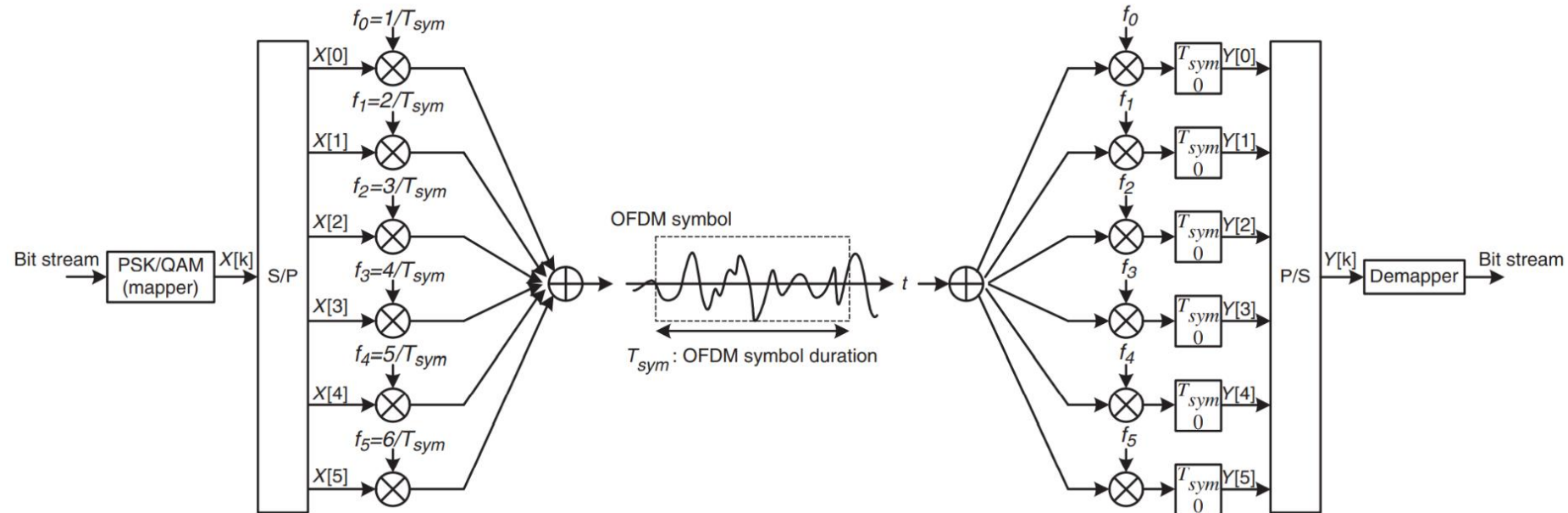
Discrete-time domain:

$$\begin{aligned}
 \frac{1}{N} \sum_{n=0}^{N-1} e^{j2\pi \frac{k}{T_{sym}} \cdot nT_s} e^{-j2\pi \frac{i}{T} \cdot nT_s} &= \frac{1}{N} \sum_{n=0}^{N-1} e^{j2\pi \frac{k}{T_{sym}} \cdot \frac{nT}{N}} e^{-j2\pi \frac{i}{T_{sym}} \cdot \frac{nT_{sym}}{N}} \\
 &= \frac{1}{N} \sum_{n=0}^{N-1} e^{j2\pi \frac{(k-i)}{N} n} \\
 &= \begin{cases} 1, & \forall \text{ integer } k = i \\ 0, & \text{otherwise} \end{cases} \longrightarrow \boxed{\text{Orthogonality condition}}
 \end{aligned}$$

- The above orthogonality is an essential condition for the OFDM signal to be ICI-free



## Block diagram of OFDM modulation and demodulation: $N = 6$ symbols



(a) OFDM modulation/demodulation

# OFDM Guard Interval

Transmitted  $l^{th}$  OFDM signal:  $x_l(t) = \sum_{k=0}^{N-1} X_l[k] e^{j2\pi f_k(t-lT_{sym})} \dots \dots \dots (1)$

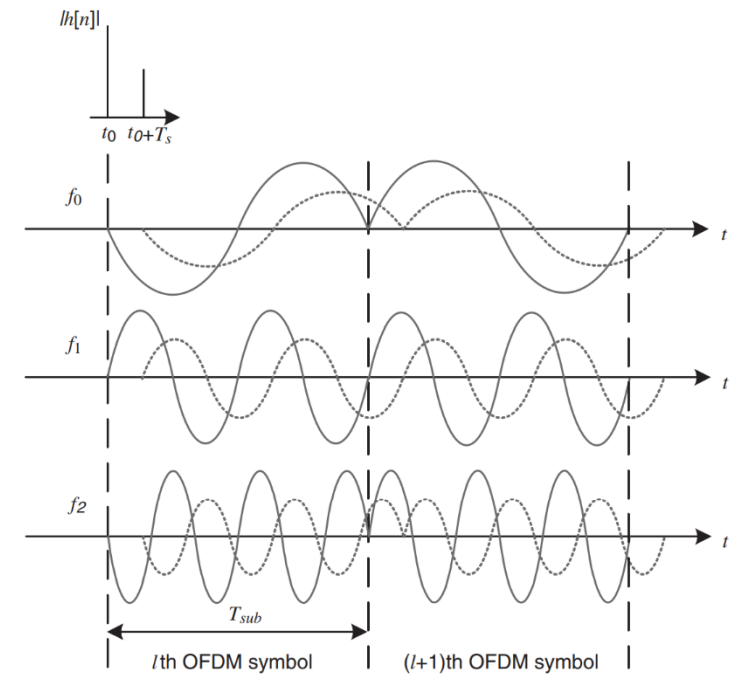
$$lT_{sym} < t \leq lT_{sym} + nT_s$$

DT form of sampled received signal:

$$y_l[n] = h_l[n] * x_l[n] + z_l[n] = \sum_{m=0}^{\infty} h_l[m] x[n-m] + z_l[n] \dots \dots \dots (2)$$



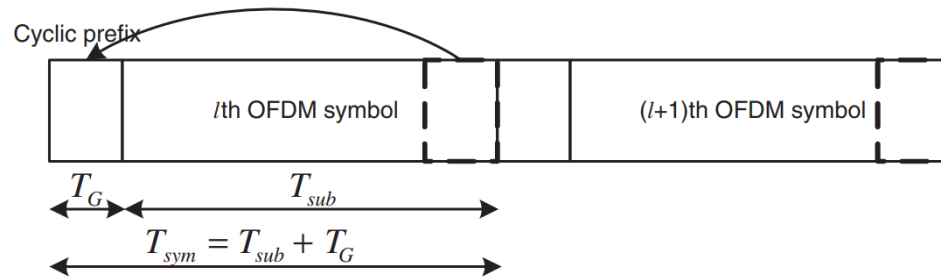
(a) OFDM symbols without guard interval



(b) ISI effect of a multipath channel on the received signal

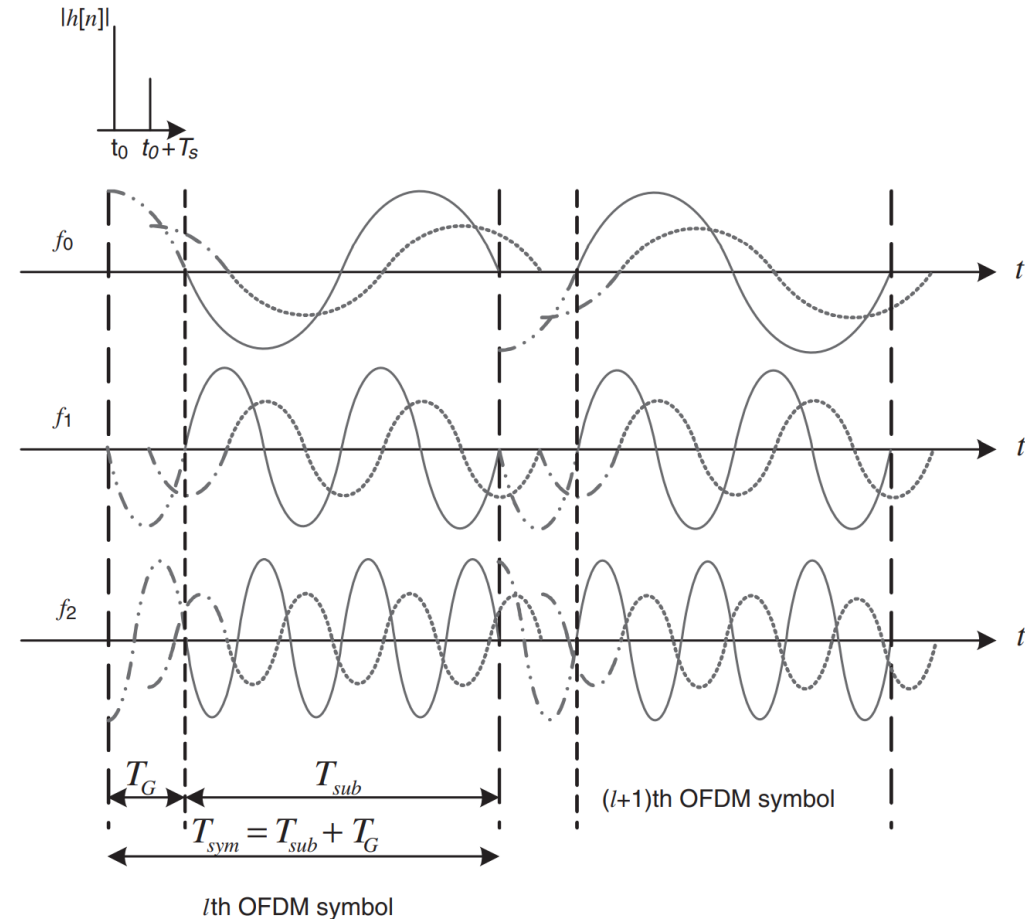
- Let  $T_G = \text{length of CP in terms of samples}$

The extended OFDM symbols now have the duration of:  $T_{sym} = T_{sub} + T_G$



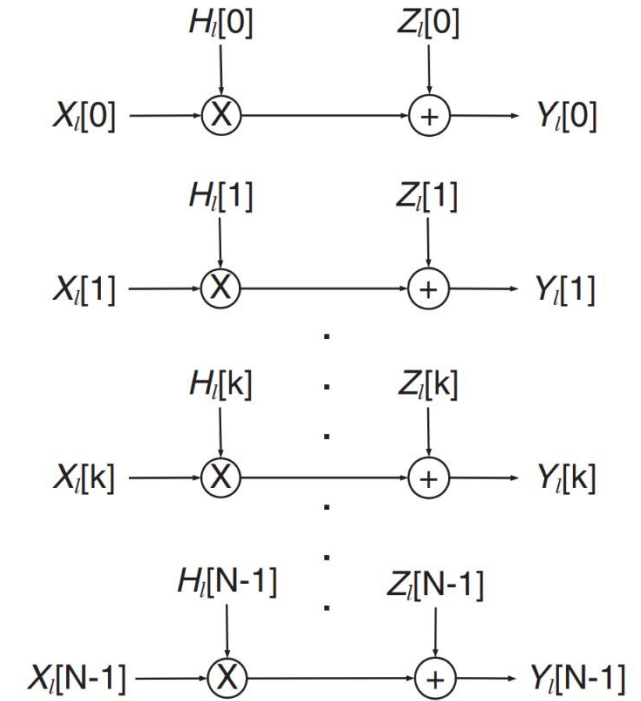
(a) OFDM symbols with CP

If  $T_G \geq \sigma_{\tau MAX}$ ,  
effect of ISI is  
confined within the  
guard interval



(c) ISI effect of a multipath channel for each subcarrier

$$\begin{aligned}
Y_l[k] &= \sum_{n=0}^{N-1} y_l[n] e^{-j2\pi kn/N} \\
&= \sum_{n=0}^{N-1} \left\{ \sum_{m=0}^{\infty} h_l[m] x_l[n-m] + z_l[n] \right\} e^{-j2\pi kn/N} \\
&= \sum_{n=0}^{N-1} \left\{ \sum_{m=0}^{\infty} h_l[m] \left\{ \frac{1}{N} \sum_{i=0}^{N-1} X_l[i] e^{j2\pi i(n-m)/N} \right\} \right\} e^{-j2\pi kn/N} + Z_l[k] \\
&= \frac{1}{N} \sum_{i=0}^{N-1} \left\{ \left\{ \sum_{m=0}^{\infty} h_l[m] e^{-j2\pi im/N} \right\} X_l[i] \sum_{n=0}^{\infty} e^{-j2\pi(k-i)n/N} \right\} e^{-j2\pi kn/N} + Z_l[k] \\
&= H_l[k] X_l[k] + Z_l[k]
\end{aligned}$$



Frequency-domain equivalent model of OFDM system

$$y_l[n] = h_l[n] * x_l[n] + z_l[n] \longrightarrow y_l[n] = h_l[n] \odot x_l[n] + z_l[n] \quad Y_l[k] = H_l[k] X_l[k] + z_l[k]$$

# BER of OFDM Scheme

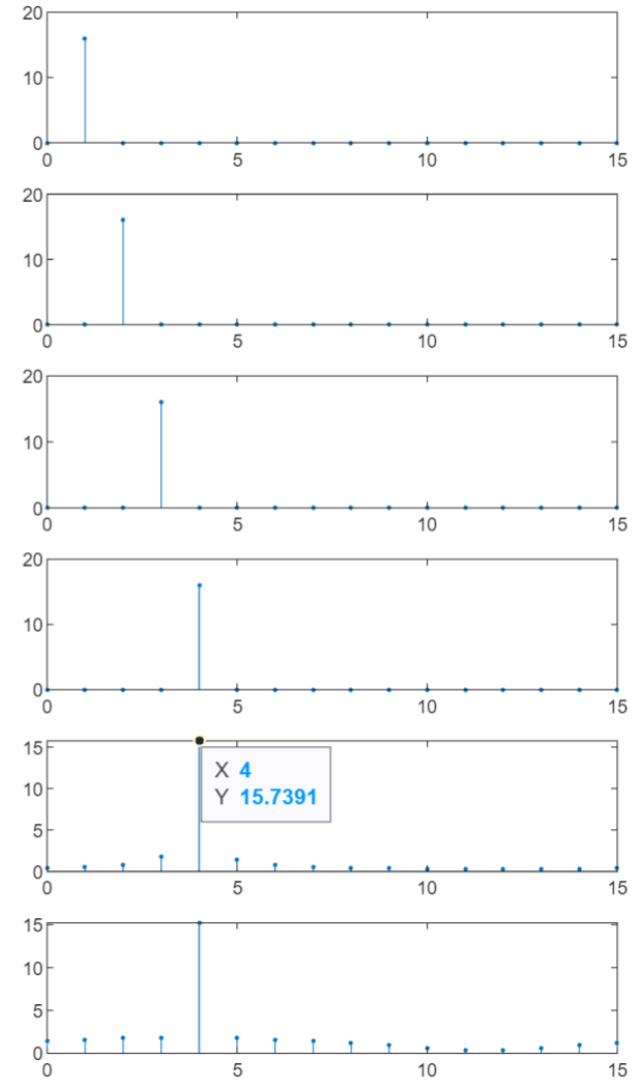
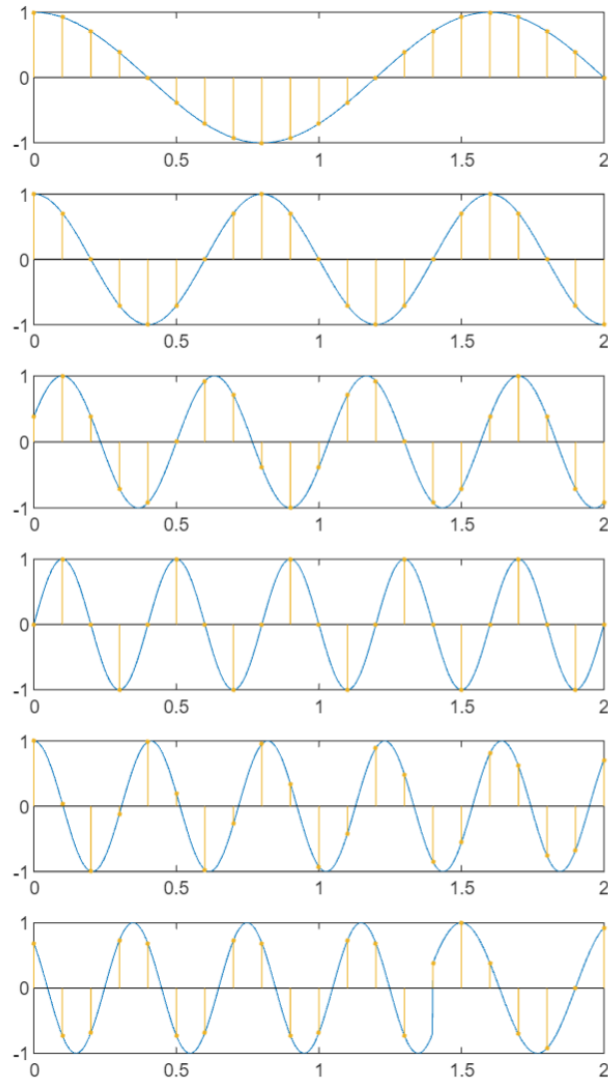
BER for M-ary QAM signalling in AWGN and Rayleigh Fading channels

- AWGN:  $P_e = \frac{2(M-1)}{M \log_2 M} Q \left( \sqrt{\frac{6E_b}{N_o} \cdot \frac{\log_2 M}{M^2-1}} \right) \dots\dots\dots(1)$

- Rayleigh:  $P_e = \frac{(M-1)}{M \log_2 M} Q \left( 1 - \sqrt{\frac{3\gamma \log_2 M / (M^2-1)}{\frac{3\gamma \log_2 M}{M^2-1} + 1}} \right) \dots\dots\dots(2)$

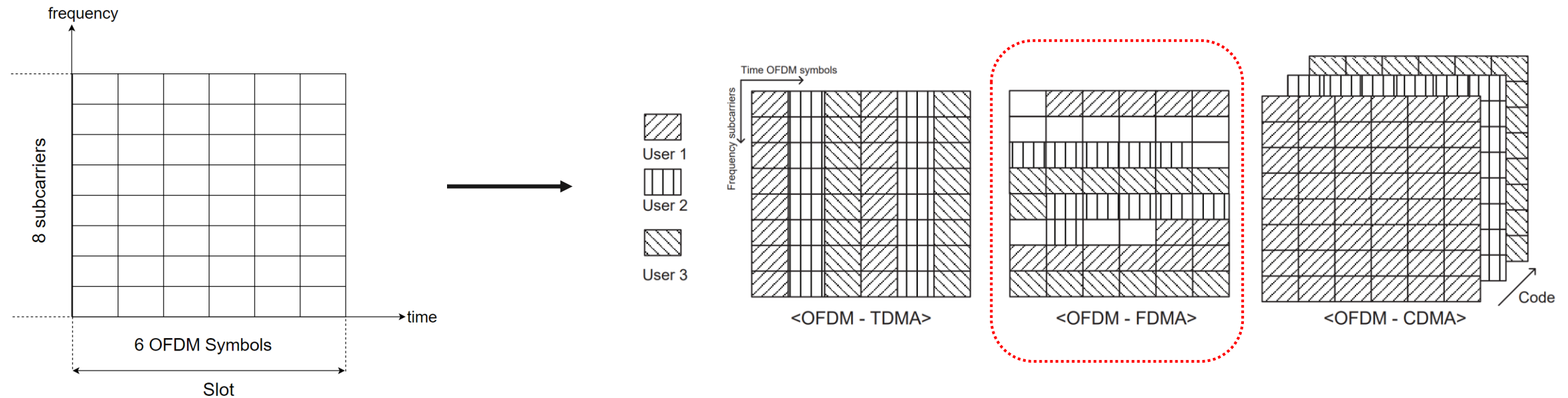
- $SNR_t = SNR_f + 10 \log \left( \frac{N_{used}}{N} \right) \text{ [dB]} \dots\dots\dots(3)$

# MATLAB Simulations



# OFDMA: Multiple Access Extension of OFDM

- In OFDM, ***all*** subcarriers are used for transmitting the symbols of ***a single user***.
- OFDM can be combined with TDMA, FDMA, and CDMA for a multi-user system.



# Any Questions?

