

Fundamentals of Wireless Communication

Information Theory From First Principles

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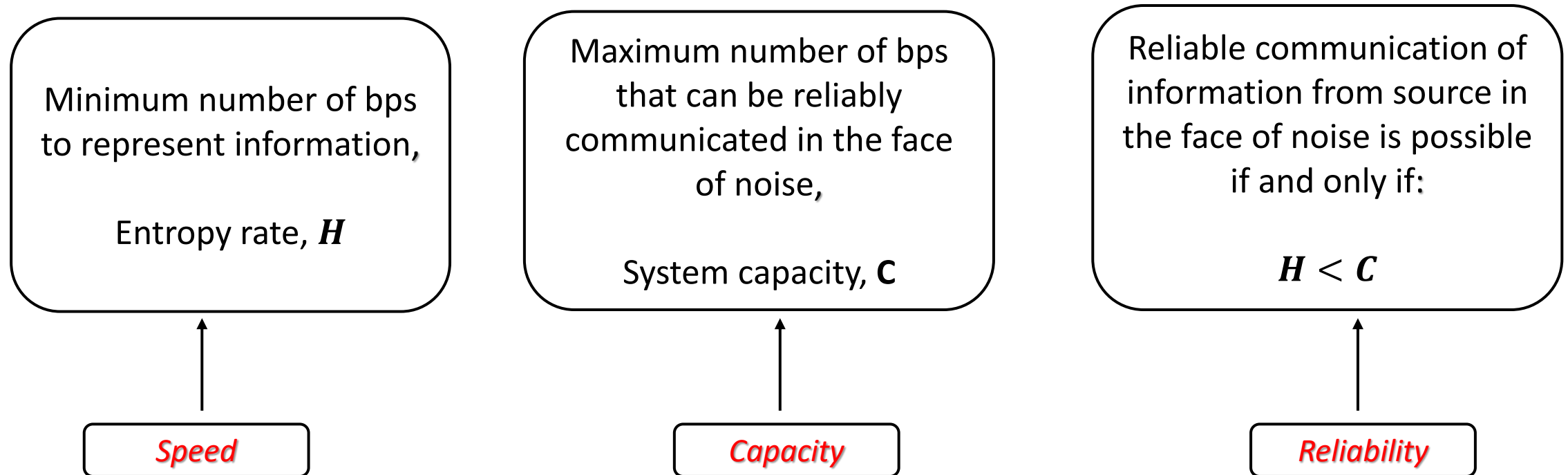
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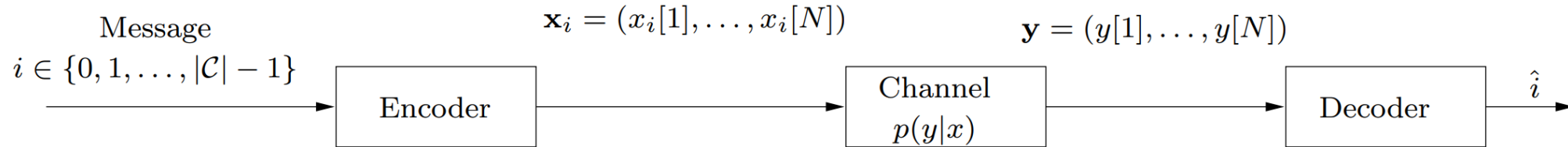
Introduction to Information Theory

- Information theory is the probabilistic treatment of the concepts, parameters and rules governing the transmission of messages through communication systems.

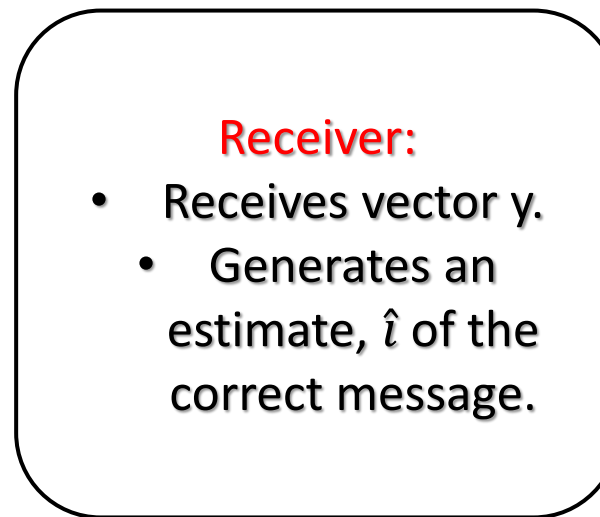
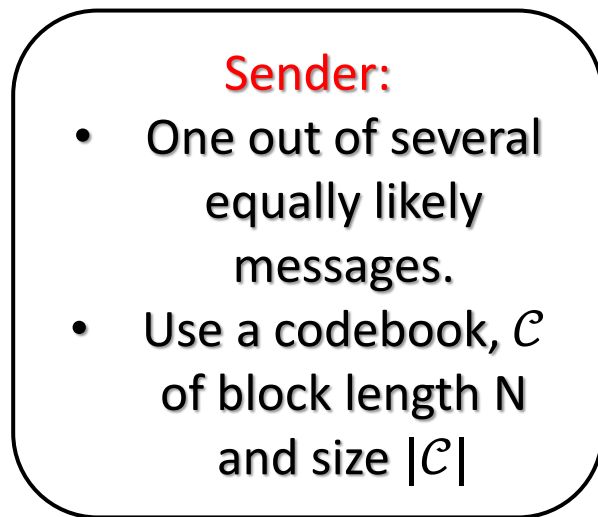
Key idea is to: increase *speed*, *capacity* and *reliability*



Discrete Memoryless Channels



- $P(y|x)$, the channel transition probability, indicates each possible input-output paths.



Key performance measures

$$R = \frac{1}{N} \log |\mathcal{C}|$$

$$p_e = \mathbb{P} \left\{ \hat{i} \neq i \right\}$$

Where $\mathcal{C} = \{\mathbf{x}_1, \dots, \mathbf{x}_{|\mathcal{C}|}\}$, \mathbf{x}_i = codewords, p_e = ML error probability

Entropy, Conditional Entropy and Mutual Information

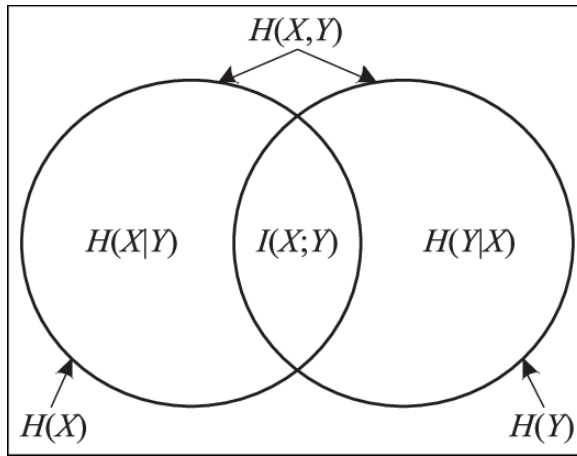
- Let x be a discrete random variable taking on values in X and with a probability mass function P_x

$$\text{Entropy of } x, H(x) := \sum_{i \in X} P_x(i) \log \left(\frac{1}{P_x(i)} \right) \dots \dots \dots (1)$$

The amount of uncertainty associated with x is $H(x)$

For two discrete-random variables x and y , the entropy of x conditioned on $y = j$ is:

$$H(x|y = j) := \sum_{i \in X} P_{x|y}(i|j) \log \left(\frac{1}{P_{x|y}(i|j)} \right) \dots \dots \dots (2)$$



Total uncertainty in x and y :

$$H(x, y) = H(x) + H(y|x) = H(y) + H(x|y) \dots \dots (3)$$

If x and y are independent, $H(x) = H(x|y)$, hence $H(x, y) = H(x) + H(y) \dots \dots \dots (4)$

Conditioning reduces uncertainty: $H(x) \geq H(x|y) \Rightarrow (3) \leq (4)$

With equality and independence of x and y : $H(x) - H(x|y)$

Hence mutual information, $I(x; y) := H(y) - H(y|x) = H(x) - H(x|y) \dots \dots \dots (5)$

Noisy Channel Coding Theorem

- From (5), to decode the transmitted message correctly with high probability, the conditional entropy, $H(x|y)$ must be close to zero.
- $H(x)$ is then = the number of bits conveyed ie. $\log|\mathcal{C}| = NR$, where $R = \text{data rate}$

For reliable communication, $H(x|y) \approx 0$, which implies

$$R \approx \frac{1}{N} I(x; y) \dots \dots \dots (6)$$

- $I(x; y)$ depends on the distribution of the random input x and this distribution is in turn a function of the code \mathcal{C} . By optimizing over all input distribution instead of all codes,
- **Upper bound on reliable rate of communication:** $\bar{C} := \max_{P_x} \frac{1}{N} I(x; y)$

- If the inputs are made independent over time,

$$\bar{C} = \frac{1}{N} \sum_{m=1}^N \max_{P_{x[m]}} I(x[m]; y[m]) = \max_{P_{x[1]}} I(x[1]; y[1]) \dots \dots \dots (7)$$

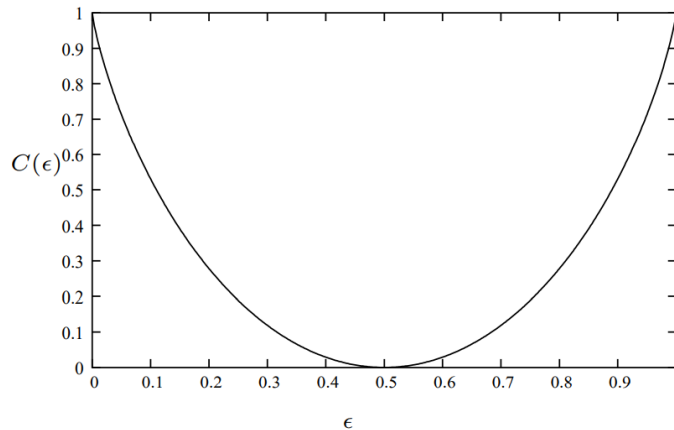
- To achieve \bar{C} , one must find a code whose $\frac{I(x;y)}{N}$ is close to \bar{C} to satisfy (6)
- These codes exist ***if the block length N is chosen sufficiently large.***

The capacity of Binary symmetric and Binary erasure Channels

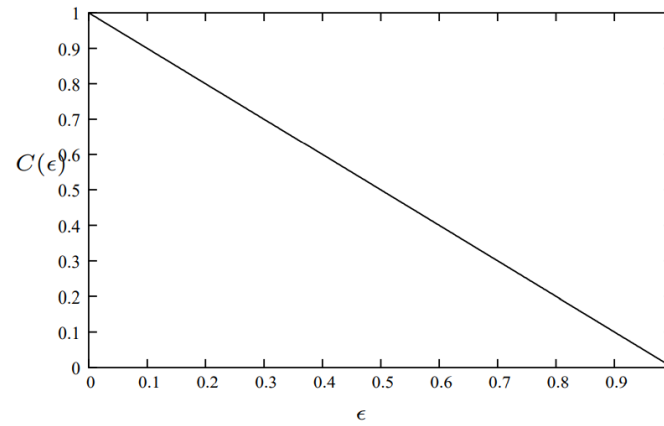
- The optimal input distribution for the BSC is uniform because of the symmetry.
- Similar symmetry exists in the BEC.

$$\begin{aligned} C &= \max_{P_x} H(y) - H(y|x) \\ &= \max_{P_x} H(y) - H(\epsilon) \\ &= 1 - H(\epsilon) \text{ bits per channel use} \end{aligned}$$

$$C = 1 - \epsilon \text{ bits per channel use}$$



(a)



(a)

Analog Memoryless Channels and AWGN Capacity

- For continuous random variable x with pdf f_x , differential entropy of x :

$$h(x) := \int_{-\infty}^{\infty} f_x(u) \log \left(\frac{1}{f_x(u)} \right) du \dots \dots \dots (8)$$

- The conditional differential entropy of x given y is:

$$h(x|y) := \int_{-\infty}^{\infty} f_{x,y}(u,v) \log \left(\frac{1}{f_{x|y}(u|v)} \right) dudv \dots \dots \dots (9)$$

The mutual information is again defined as:

$$I(x; y) := h(x) - h(x|y)$$

$$\text{Capacity: } \bar{C} := \max_{f_x} I(x; y)$$

- Every codeword, x_n in the codebook must satisfy:

$$\frac{1}{N} \sum_{m=1}^N c(x_n[m]) \leq A,$$

where $c: X \rightarrow \mathfrak{R}$ is the cost constraint function on the transmitted codewords for the channel

- The minimum rate of reliable communication subject to this constraint on the codewords is:

$$C = \max_{f_x: E[c(x)] \leq A} I(x; y) \dots \dots \dots (10)$$

- For real power-constrained AWGN channel: $y = x + w$, with cost function $c(x) = x^2$,

The differential entropy of w , $h(w) = \frac{1}{2} \log(2\pi e \sigma^2)$

- $I(x; y) = h(y) - h(y|x) = h(y) - \frac{1}{2} \log(2\pi e \sigma^2)$

$$\text{Capacity: } C = \max_{f_x: E[x^2] \leq P} I(x; y)$$

$h(y)$ is maximized when y is $\mathcal{N}(0, P + \sigma^2)$ which is achieved by choosing x to be $\mathcal{N}(0, P)$

$$\therefore C = \frac{1}{2} \log(2\pi e(P + \sigma^2)) - \frac{1}{2} \log(2\pi e \sigma^2) = \frac{1}{2} \log\left(1 + \frac{P}{\sigma^2}\right) \dots \dots \dots (11)$$

Capacity of the Fast-Fading Channel

- For fast fading channel: $y[m] = h[m]x[m] + w[m]$

The capacity of the power-constrained fast fading channel with receiver CSI:

$$C = \max_{P_x: \mathbb{E}[x^2] \leq P} I(x; y, h)$$

- Since h is independent of the input, $I(x; y|h) = 0$,

$$I(x; y, h) = I(x, h) + I(x; y|h) = I(x; y|h) \dots \dots \dots (12)$$

$$\text{Conditioned on fading coefficient, } h \text{ SNR} = \frac{P|h|^2}{N_0}$$

$$C = \mathbb{E}_h \left[\log \left(1 + \frac{P|h|^2}{N_0} \right) \right] \dots \dots \dots (13)$$

- (13) holds if $h[m]$ is **stationary** and **ergodic**.

Rate of reliable communication is the average rate of flow of $I(x; y)$

$$\frac{1}{N} I(x; y) = \frac{1}{N} \sum_{m=1}^N \log(1 + |h[m]|^2 SNR)$$

- For large N , due to the ergodicity of the fading process,

$$\frac{1}{N} \sum_{m=1}^N \log(1 + |h[m]|^2 SNR) \rightarrow \mathbb{E}[\log(1 + |h|^2 SNR)]$$

Any Questions?

