Fundamentals of Wireless Communication

Capacity of Wireless Channels

Edward Kwao

Hanbat National University

3rd February, 2022



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Introduction

Problem of communication over wireless fading channels

Optimal performance achievable on a given channel

Techniques to achieve such optimal performance

• The basic measure of performance is the *capacity* of a channel.

AWGN Channel Capacity and Channel Resources

Consider a continuous-time AWGN Channel with bandwidth W Hz, power constraint \bar{P} watts, AWGN with psd $^{N_0}/_2$. For sampling at $^1/_W$, the discrete-time complex baseband channel:

$$y[m] = x[m] + w[m]$$
, where $w[m]$ is $C\mathcal{N}(0, N_0)$ and is i. i. d over time

The capacity of the channel is:

Given
$$SNR := \frac{P}{N_0W}$$

$$C_{awan} = log(1 + SNR) \dots (2)$$
 bps/Hz



- (1) suggests that capacity of the channel depends on the basic resources:
- Received power $\overline{m{P}}$
- Bandwidth $m{W}$

Bandwidth limited regime SNR \gg 1: Capacity logarithmic in received power but approximately linear in bandwidth.

$$W log \left(1 + \frac{\overline{P}}{N_0 W}\right) \approx W \left(\frac{\overline{P}}{N_0 W}\right) log_2 e = \frac{\overline{P}}{N_0} log_2 e \dots \dots \dots \dots (3)$$

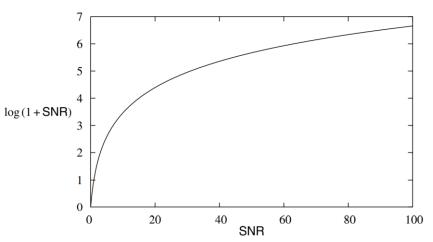
Power limited regime SNR $\ll 1$: Capacity linear in received power but insensitive to bandwidth.

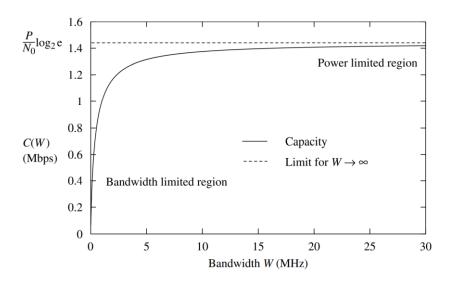
■ From (4), the capacity is finite even if there is no bandwidth constraint.

The main objective is to minimize the required energy per bit ε_b rather than maximizing spectral efficiency in some communication applications.

The minimum ε_b , $\frac{P}{C_{awgn}(\bar{P},W)}$ is achieved when $\bar{P} \to 0$

$$\Rightarrow \left(\frac{\varepsilon_b}{N_0}\right)_{min} = \lim_{\bar{P}\to 0} \frac{\bar{P}}{C_{awgn}(\bar{P},W)} = \frac{\bar{P}}{Wlog\left(1 + \frac{\bar{P}}{N_0W}\right)} = \frac{1}{log_2e} = -1.59dB$$





Frequency – Selective AWGN Channel

Consider $a \ time - invariant \ L - tap$ frequency-selective AWGN channel:

$$y[m] = \sum_{l=0}^{L-1} h_l x[m-l] + w[m]$$

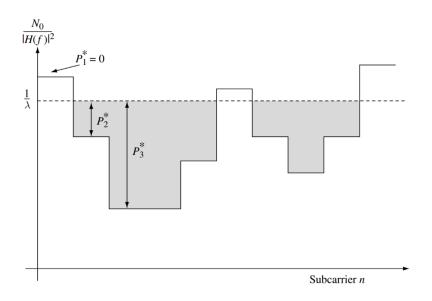
OFDM converts it into a parallel channel:

$$\dot{\tilde{y}}_n = \tilde{h}_n \tilde{d}_n + \tilde{w}_n, \qquad n = 0, 1, \dots, N_c - 1$$

The maximum rate of reliable communication with this scheme is:

Where, p_n^* is the waterfilling power allocation, defined as: $\left(\frac{1}{\lambda} - \frac{N_0}{|\tilde{h}_n|^2}\right)^{\mathsf{T}}$

λ, the Lagrange multiplier is chosen such that the power constraint is met.



- The values $\frac{N_0}{|\tilde{h}_n|^2}$ plotted as a function of the sub-carrier index $n=0,1,\dots,N_c-1$
- P units of water per sub-carrier are filled into the vessel, the depth of the water at sub-carrier n is the power allocated to that sub-carrier.
- $\frac{1}{\lambda}$ is the height of the water surface.

Capacity of Fading Channels Slow Fading Channels

Considering a channel where the channel gain is random but remains constant for all time:

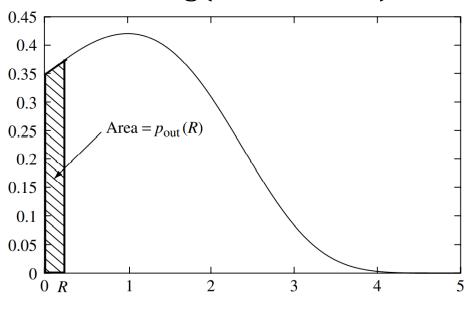
$$y[m] = hx[m] + w[m]$$

- For slow fading situation: $coherence\ time \gg latency\ requirement$
- It can be observed that there's no definite capacity. The performance is a function of h. Good $h \Rightarrow$ Good performance and vice versa.
- In this situation, we can define the notion of outage probability:

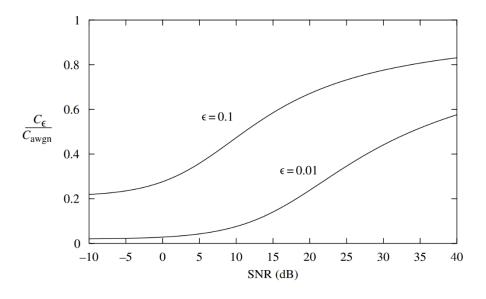
$$P_{out}(R) = \mathbb{P}\{\log(1 + |h|^2 \text{SNR} < R)\}....(6)$$

- A good performance measure is: ϵ outage capacity
- This is the largest rate of transmission R such that $P_{out}(R)$ is less than ϵ .

PDF: $log(1 + |h|^2 SNR)$

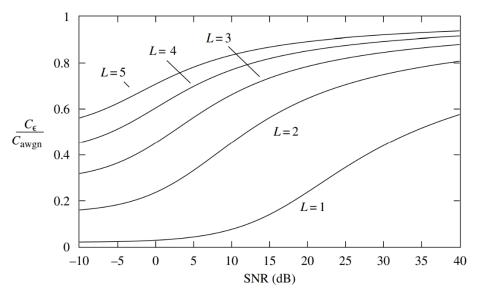


$$P_{out}(R) \approx \frac{2^R - 1}{SNR}$$



To assess the impact of fading, the ϵ -outage capacity is plotted as a fraction of the AWGN capacity at the same SNR

- If we increase the diversity of the channel by using L receive antennas
- Outage occurs whenever: $P_{out}^{rx}(R) = \mathbb{P}\left\{\log\left(1+\left||h|\right|^2 \text{SNR} < R\right)\right\}....(8)$



- This causes dramatic improvement in the gain
- At ϵ = 0.01 and L = 2, the outage capacity is increased to 14% of the AWGN capacity (as opposed to 1% for L = 1).

Fast Fading Channel

Model:
$$y[m] = h[m]x[m] + w[m]$$

■ When the codeword length spans many coherence periods, we are in the so — called fast fading regime.

Block fading model: $h[m] = h_l$ remains constant over l^{th} coherence period of T_c Symbols and is i, i, d across different coherence period.

Suppose coding is done over L coherence periods. If $T_c\gg 1$, L parallel subchannels fade independently.

$$P_{out}(R) = \mathbb{P}\left\{\frac{1}{L}\sum_{l=1}^{L} log(1 + |h_l|^2 SNR < R)\right\}....(9)$$

■ For finite L, $P_{out}(R) = \frac{1}{L} \sum_{l=1}^{L} log(1 + |h_l|^2 SNR)$

As $L \to \infty$, and by the law of large numbers:

$$\frac{1}{L} \sum_{l=1}^{L} log(1 + |h_l|^2 SNR) \to \mathbb{E}[log(1 + |h|^2 SNR)] \dots \dots \dots \dots (10)$$

$$\therefore C = \mathbb{E}[log(1 + |h|^2 SNR)] \text{ bps/Hz...} (11)$$



• In the case of ideal interleaving, we assume h[m]s are i.i.d.

For a large block length N and a given realization of fading gains h[1], ..., h[N],

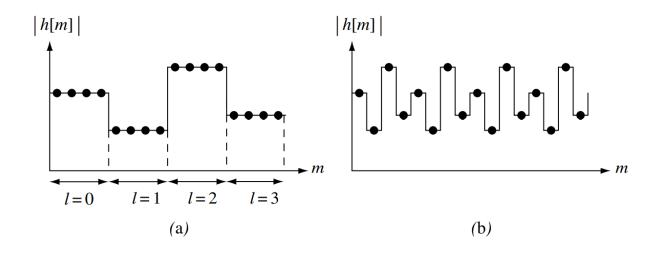
The maximum achievable rate through this interleaved channel is:

$$\frac{1}{N} \sum_{m=1}^{N} \log(1 + |h[m]|^2 SNR) bps/Hz$$

As $N \to \infty$, and by the law of large numbers:

$$\frac{1}{N} \sum_{m=1}^{N} log(1 + |h[m]|^2 SNR) \to \mathbb{E}[log(1 + |h|^2 SNR)]$$

 Trajectory of the channel strength as a function of symbol time under block fading model and after interleaving



- Even with interleaving the capacity of the fast-fading channel can be achieved.
- The benefit however is: Capacity is achieved with a much shorter block length



