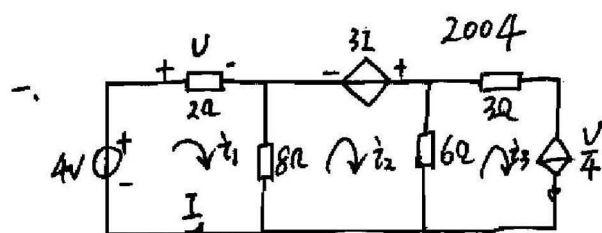
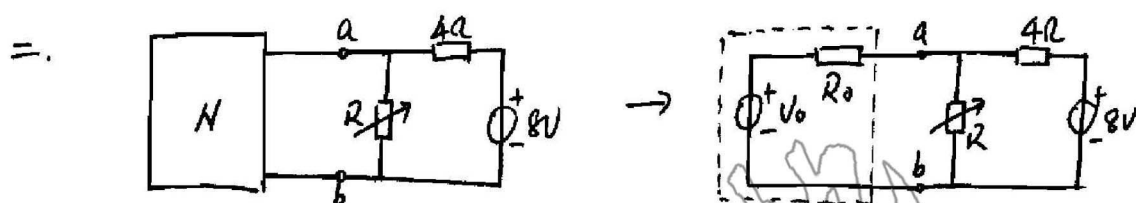


西南交通大学电路分析历年考研真题参考答案



解：

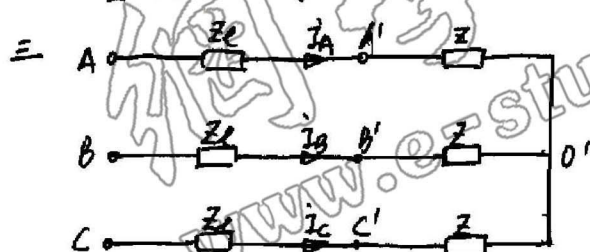
$$\begin{cases} 2i_1 + (i_1 - i_2) \times 8 = 4 \\ -31 + 6(i_2 - i_3) + 8(i_2 - i_1) = 0 \\ i_3 = \frac{V}{4} \\ V = 2i_1 \end{cases} \Rightarrow \begin{cases} i_1 = 2A \\ i_2 = 2A \\ i_3 = 1A \\ I = 2A \end{cases}$$



解：∵ N 为直流线性有源网络，可等效戴维南电路。

当 $R=0$ 时 $i=8A \Rightarrow \frac{U_0}{R_0} = 8\Omega$

当 $R=4\Omega$ 时 $\Rightarrow U_0 - 4R_0 = 8 + 4 \Rightarrow U_0 = 24V, R_0 = 3\Omega$



解： $\dot{U}_{A'B'} = 380 \angle 0^\circ V$

$P = \sqrt{3} U_L I_L \cos \varphi \Rightarrow$

$I_L = \frac{P}{\sqrt{3} U_L \cos \varphi} = 10 A$

$\dot{U}_{A'} = 220 \angle -30^\circ V$

$\dot{U}_B = 220 \angle -150^\circ V$

$\dot{I}_A = 10 \angle \varphi_u - \varphi_{I_2} = 10 \angle 30^\circ - 60^\circ = 10 \angle -90^\circ A$

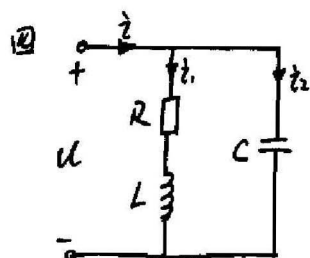
$\dot{I}_B = 10 \angle -210^\circ A, \dot{I}_C = 10 \angle 30^\circ A$

$\dot{U}_A = \dot{I}_A Z_1 + \dot{U}_{A'}, \dot{U}_B = \dot{I}_B Z_1 + \dot{U}_B$

$\dot{U}_{AB} = \dot{U}_A - \dot{U}_B = (\dot{I}_A - \dot{I}_B) Z_1 + (\dot{U}_{A'} - \dot{U}_{B'})$

$= 378.38 \angle 11.54^\circ V$

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解: $\dot{U} = \frac{200}{\sqrt{2}} \angle -90^\circ \text{ V}$ $\dot{I} = \frac{5}{\sqrt{2}} \angle -90^\circ \text{ A}$

$Z = \frac{\dot{U}}{\dot{I}} = \frac{200}{5} = 40 \Omega$

$\therefore R, L, C$ 发生并联谐振.

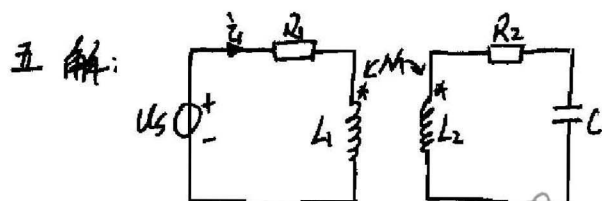
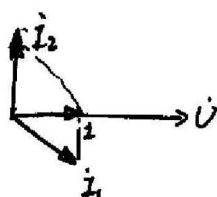
$Y = \frac{1}{R + j\omega L} + j\omega C = \frac{R}{R^2 + (\omega L)^2} + j(\omega C - \frac{\omega L}{R^2 + (\omega L)^2})$

$\frac{R}{R^2 + (\omega L)^2} = \frac{1}{40} = \frac{20}{400 + 10^4 L^2} \Rightarrow L = 0.2 \text{ H}$

$\omega C = \frac{\omega L}{400 + \omega^2 L^2} \Rightarrow C = \frac{L}{400 + \omega^2 L^2} = 250 \text{ nF}$

$j\omega L = j \times 100 \times 0.2 = 20j$ $\frac{1}{j\omega C} = -40j$

$\dot{I}_1 = \frac{\dot{U}}{20 + 20j} = 5 \angle -45^\circ \text{ A}$ $\dot{I}_2 = \frac{\dot{U}}{-40j} = \frac{5}{\sqrt{2}} \angle 90^\circ \text{ A}$

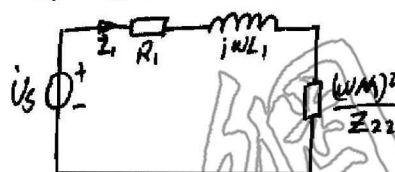


$Z_{11} = R_1 + j\omega L_1 = 4 + j8 \Omega$

$Z_{22} = R_2 + j\omega L_2 + \frac{1}{j\omega C} = 4 \Omega$

$\frac{(\omega M)^2}{Z_{22}} = \frac{16}{4} = 4 \Omega$

原边等效电路如下:



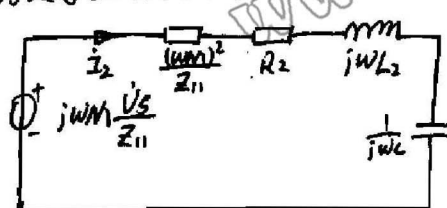
① 原边等效电路直接作用时: $\dot{I}_0 = \frac{12}{4} = 3 \text{ A}$

② 交流作用时: $\dot{U}_s = \frac{4\sqrt{2}}{\sqrt{2}} \angle 45^\circ \text{ V}$

$\dot{I}_1 = \frac{\dot{U}_s}{4 + j8 + 4} = 3 \angle 0^\circ \text{ A}$

$i_1 = 3 + 3\sqrt{2} \cos(2t) \text{ A}$

副边等效电路如下:



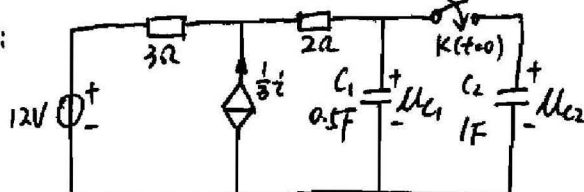
$\frac{(\omega M)^2}{Z_{11}} = \frac{16}{4 + j8} = \frac{4}{1 + 2j}$

$j\omega M \frac{\dot{U}_s}{Z_{11}} = \frac{\dot{U}_s \times 4 \angle 90^\circ}{4 + j8} = \frac{\dot{U}_s \angle 90^\circ}{1 + 2j}$

$\therefore (\frac{4}{1 + 2j} + 4) \dot{I}_2 = \frac{\dot{U}_s \angle 90^\circ}{1 + 2j} \Rightarrow \dot{I}_2 = 3 \angle 90^\circ \text{ A}$

$i_2 = 3\sqrt{2} \cos(2t + 90^\circ) \text{ A}$ $P = U_0 I_0 + U_1 I_1 \cos \varphi = 12 \times 3 + \frac{4\sqrt{2}}{\sqrt{2}} \times 3 \times \frac{\sqrt{2}}{2} = 108 \text{ W}$

六 解:



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$$u_{C1}(0^-) = 12V \quad u_{C2}(0^-) = 0V$$

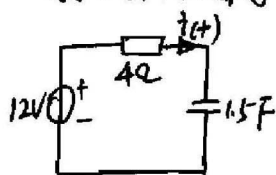
根据电荷守恒定律： $C_1 u_{C1}(0^-) + C_2 u_{C2}(0^-) = C_1 u_{C1}(0^+) + C_2 u_{C2}(0^+)$

$$u_{C1}(0^+) = u_{C2}(0^+)$$

$$\therefore 0.5 \times 12 = (C_1 + C_2) u_{C1}(0^+) = 1.5 u_{C1}(0^+)$$

$$\therefore u_{C1}(0^+) = u_{C2}(0^+) = 4V \quad u_{C1}(\infty) = u_{C2}(\infty) = 12V$$

求 u_{C1} 两端等效电路：① 开路电压 $V_{oc} = 12V$



② 短路电流 i_{sc} ：

$$3(i_{sc} - \frac{1}{3}i_{sc}) + 2i_{sc} = 12 \Rightarrow i_{sc} = 3A$$

$$R_{eq} = \frac{12}{3} = 4\Omega$$

$$\tau = RC = 4 \times 1.5 = 6s$$

$$u_{C2}(t) = 12 + (4 - 12)e^{-\frac{t}{6}} = 12 - 8e^{-\frac{t}{6}} V (t \geq 0)$$

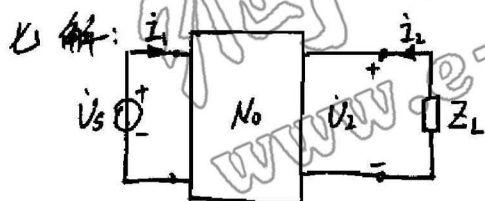
$$i(t) = \frac{12 - u_{C2}(t)}{4} = 2e^{-\frac{t}{6}} A (t \geq 0)$$

求 $i(t)$ 的替代解法：

$$i_{C1}(t) = C_1 \frac{du_{C1}}{dt} = 0.5 \times \frac{4}{6} e^{-\frac{t}{6}} = \frac{2}{3} e^{-\frac{t}{6}} (t \geq 0) \quad \times$$

$$i_{C2}(t) = C_2 \frac{du_{C2}}{dt} = \frac{4}{6} e^{-\frac{t}{6}} (t \geq 0) \quad \times \quad \text{why? 用拉氏变换求解便知.}$$

$$\therefore i(t) = i_{C1}(t) + i_{C2}(t) = 2e^{-\frac{t}{6}} (t \geq 0)$$



① 当 $Z_L = 0$ 时 $i_1 = 18 \angle 0^\circ A \quad i_2 = 9 \angle 180^\circ A$

$$\dot{U}_1 = Z_{11} \dot{I}_1 + Z_{12} \dot{I}_2$$

$$Z_{11} \cdot 18 \angle 0^\circ + Z_{12} \cdot 9 \angle 180^\circ = 90 \angle 0^\circ$$

② $\dot{U}_1 = \dot{U}_s = 90 \angle 0^\circ V \quad \dot{I}_1 = 18 \angle 0^\circ A \quad \dot{U}_2 = 0 \quad \dot{I}_2 = 9 \angle 180^\circ A$

$$\hat{U}_1 = 90 \angle 0^\circ V \quad \hat{I}_2 = ? \quad \hat{U}_2 = 30 \angle 0^\circ V \quad \hat{I}_2 = 5 \angle 180^\circ A$$

$$\sum_{k=0}^n \dot{U}_k \dot{I}_k = 0 \quad (\text{特勒根定理})$$

$$-90 \angle 0^\circ \cdot \hat{I}_1 = -90 \angle 0^\circ \cdot 18 \angle 0^\circ - 30 \angle 0^\circ \cdot 9 \angle 180^\circ$$

$$\therefore \hat{I}_1 = 15 \angle 0^\circ A$$

$$\begin{cases} Z_{11} \cdot 18 - 9Z_{12} = 90 \\ Z_{11} \cdot 15 - 5Z_{12} = 90 \end{cases}$$

$$\Rightarrow \begin{cases} Z_{11} = 8\Omega \\ Z_{12} = 6\Omega \end{cases}$$

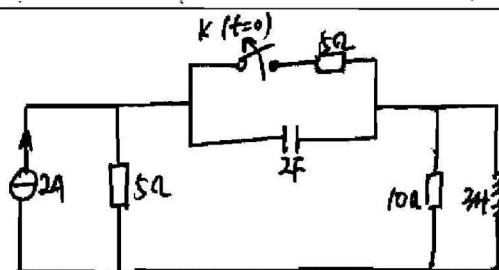
$$\begin{cases} Z_{21} \cdot 18 - 9Z_{22} = 0 \\ Z_{21} \cdot 15 - 5Z_{22} = 30 \end{cases}$$

$$\Rightarrow Z_{21} = 6\Omega \quad Z_{22} = 12\Omega$$

$$\therefore Z = \begin{bmatrix} 8 & 6 \\ 6 & 12 \end{bmatrix} \Omega \quad Y = Z^{-1} = \begin{bmatrix} \frac{1}{5} & -\frac{1}{10} \\ -\frac{1}{10} & \frac{2}{15} \end{bmatrix} S$$

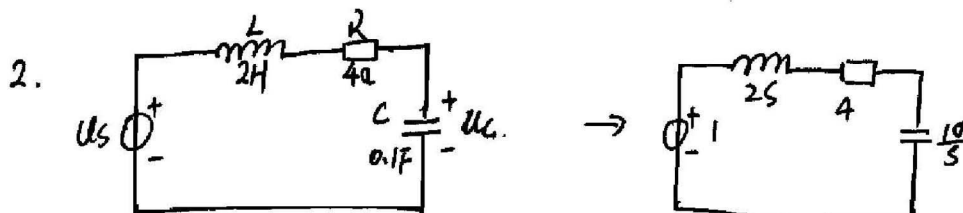
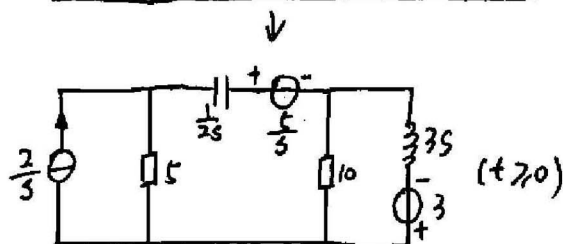
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18. 解：



$$u_C(0_-) = 5V$$

$$i_L(0_-) = 2A$$



$$\text{方法一：} \quad I(s) = \frac{1}{2s+4+\frac{10}{s}} \quad u_C(s) = \frac{\frac{10}{s}}{2s+4+\frac{10}{s}} = \frac{5}{s^2+2s+5}$$

$$K_1 = \frac{D(s)}{Q'(s)} \Big|_{s=-1+2j} = \frac{5}{2s+2} \Big|_{s=-1+2j} = \frac{5}{4} \angle 90^\circ$$

$$\therefore u_C(t) = 2 \times \frac{5}{4} e^{-t} \cos(2t - 90^\circ) \quad i(t) = \frac{5}{2} e^{-t} \sin 2t \quad \epsilon(t)$$

$$\text{方法二：} \quad LC \frac{d^2 u_C(t)}{dt^2} + RC \frac{du_C(t)}{dt} + u_C = \epsilon(t) \quad u_C(0_-) = 0 \quad i_L(0_-) = 0$$

$$2 \frac{d^2 u_C(t)}{dt^2} + 4 \frac{du_C(t)}{dt} + 10 u_C = 0$$

$$p^2 + 2p + 5 = 0 \quad p = \frac{-2 \pm \sqrt{4-20}}{2} = -1 \pm 2j$$

$$\text{通解：} \quad u_C' = e^{-t} (C_1 \cos 2t + C_2 \sin 2t) V$$

$$\text{特解：} \quad u_C'' = 1V$$

$$\therefore u_C(t) = e^{-t} (C_1 \cos 2t + C_2 \sin 2t) + 1$$

$$\begin{cases} C_1 + 1 = 0 \Rightarrow C_1 = -1 \end{cases}$$

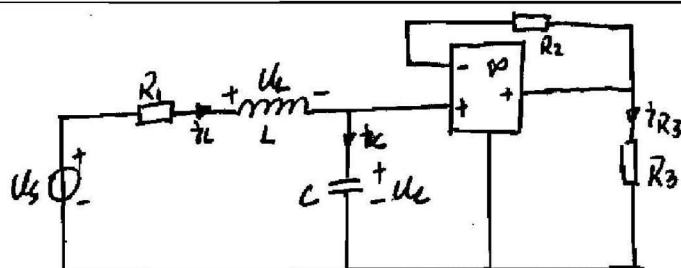
$$\begin{cases} \frac{du_C(t)}{dt} = -e^{-t} (C_1 \cos 2t + C_2 \sin 2t) + e^{-t} (-2C_1 \sin 2t + 2C_2 \cos 2t) \\ \Rightarrow -C_1 + 2C_2 = 0 \Rightarrow C_2 = -\frac{1}{2} \end{cases}$$

$$\therefore u_C(t) = [e^{-t} (-\cos 2t - \frac{1}{2} \sin 2t) + 1] \epsilon(t)$$

$$i(t) = \frac{d u_C(t)}{dt} = \frac{5}{2} e^{-t} \sin(2t) \epsilon(t) V$$

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12. 解:



取 U_C, i_L 为自变量

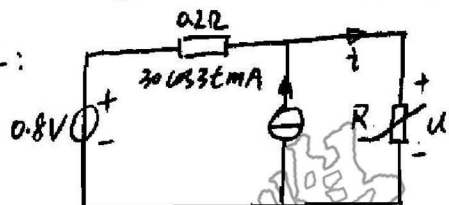
$$i_C = i_L \Rightarrow \frac{dU_C}{dt} = \frac{1}{C} i_L \quad (1)$$

$$R_1 i_L + U_L + U_C = U_s \Rightarrow U_L = U_s - U_C - R_1 i_L$$

$$\frac{di_L}{dt} = -\frac{1}{C} U_C - \frac{R_1}{L} i_L + \frac{1}{L} U_s \quad (2)$$

$$\therefore \begin{bmatrix} \frac{dU_C}{dt} \\ \frac{di_L}{dt} \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{C} \\ -\frac{1}{C} & -\frac{R_1}{L} \end{bmatrix} \begin{bmatrix} U_C \\ i_L \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} U_s$$

+ 解:

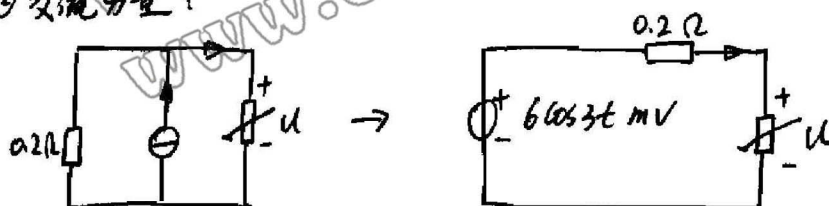


应用叠加原理:

① 直流分量: $0.2i + 0.1i^2 = 0.8 \Rightarrow i = -4A$ (舍) $i = 2A$

$$\therefore U_Q = 0.4V \quad I_Q = 2A$$

② 交流分量:



$$R = \frac{du}{di} \Big|_{i=2A} = 0.2i \Big|_{i=2A} = 0.4\Omega$$

$$i = \frac{u}{0.2+0.4} = 10\cos 3t \text{ mA}$$

$$\therefore U_0 = i \times 0.4 = 4\cos 3t \text{ mV}$$

$$\therefore U = U_Q + u = 0.4 + 0.004\cos 3t \text{ V}$$

$$i = I_Q + i = 2 + 0.01\cos 3t \text{ A}$$