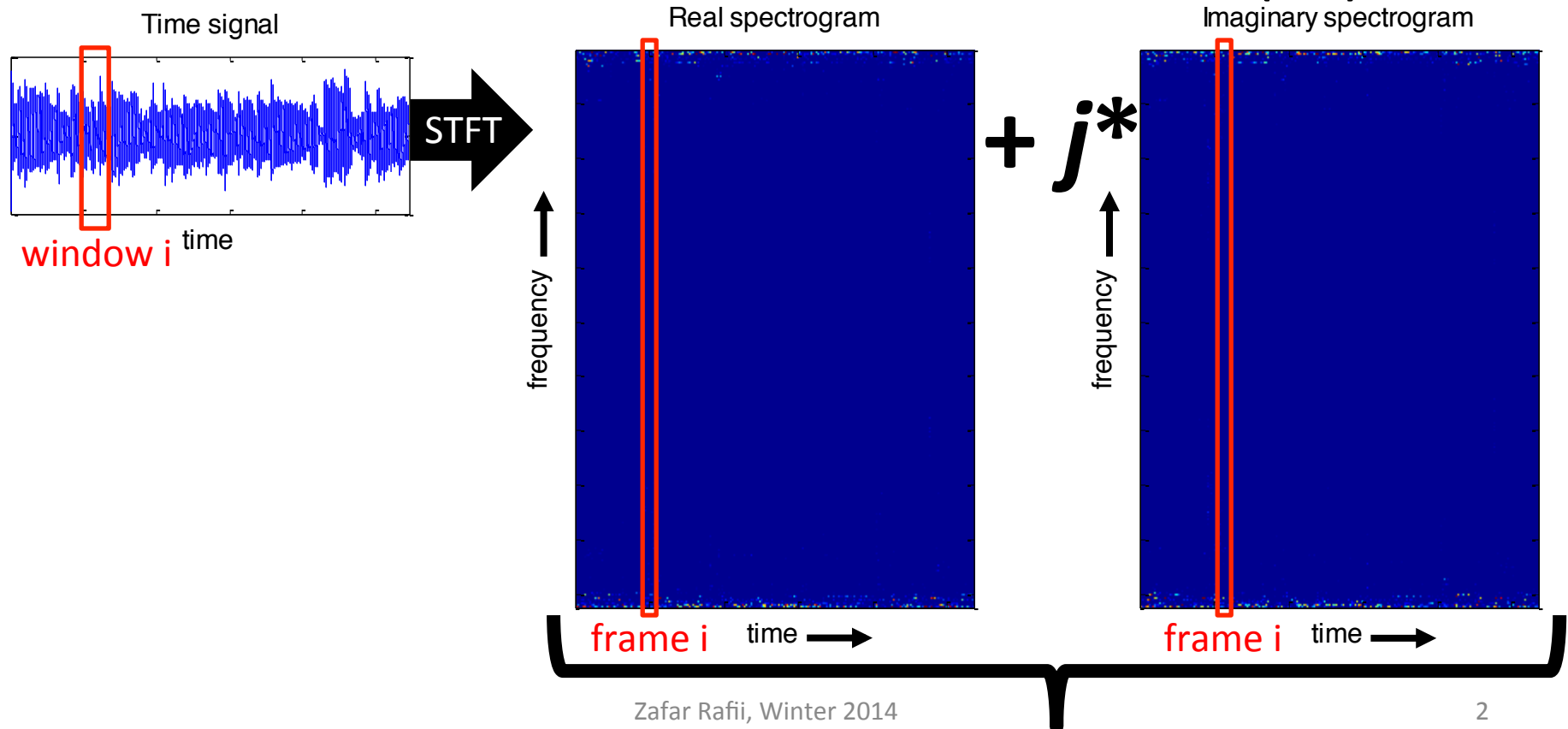


Time-frequency Masking

EECS 352: Machine Perception of
Music & Audio

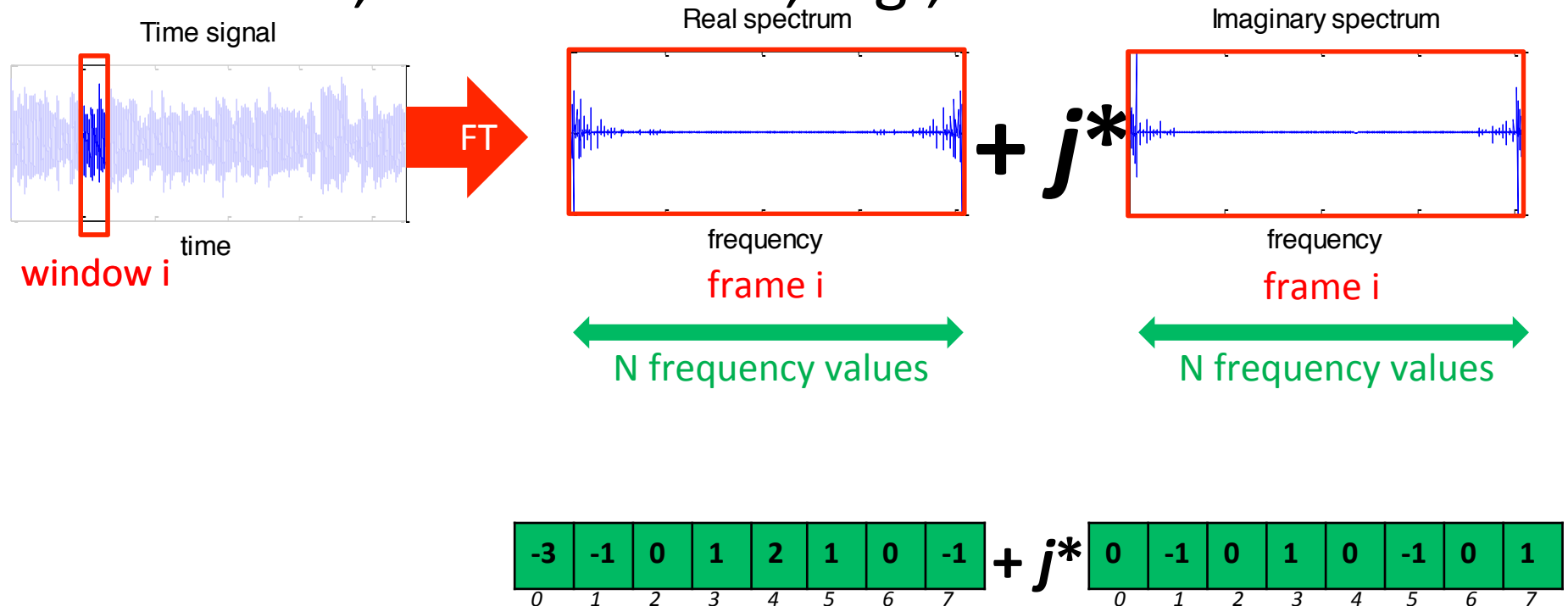
STFT

- The **Short-Time Fourier Transform (STFT)** is a succession of local Fourier Transforms (FT)



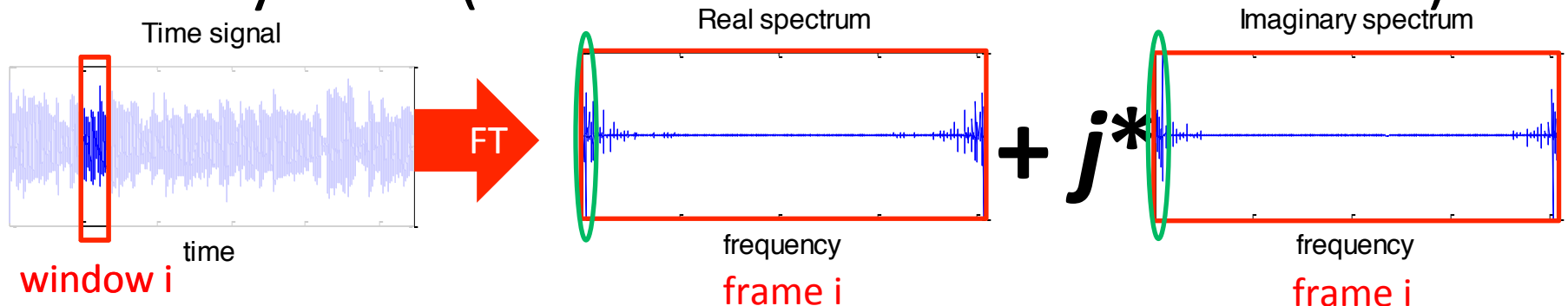
STFT

- If we used a window of N samples, the FT has N values, from 0 to $N-1$; e.g., if $N = 8$...



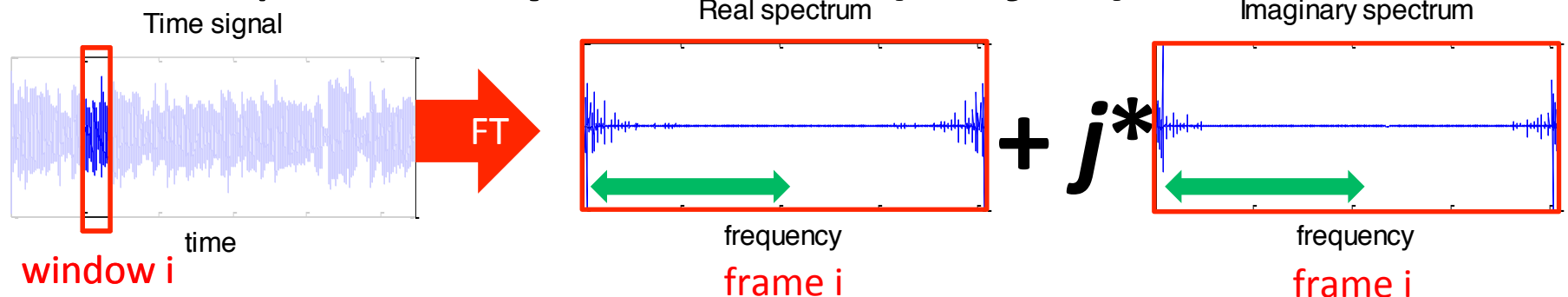
STFT

- Frequency index 0 is the **DC component**; it is always real (it is the sum of the time values!)



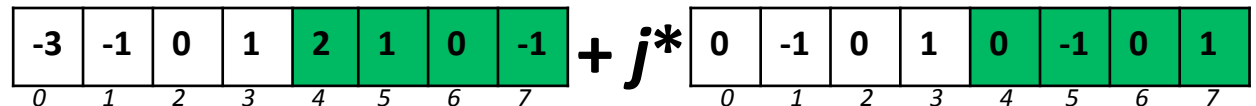
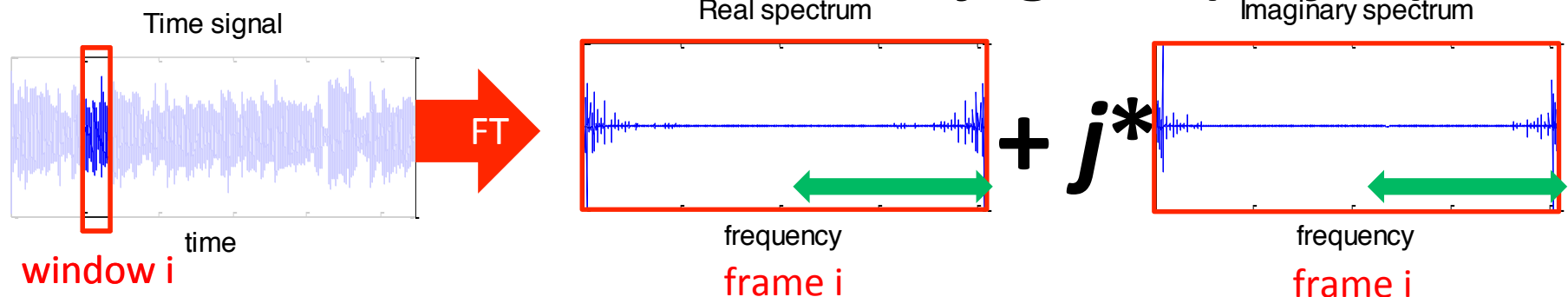
STFT

- Frequency indices from 1 to floor(N/2) are the “unique” complex values $(a + j*b)$



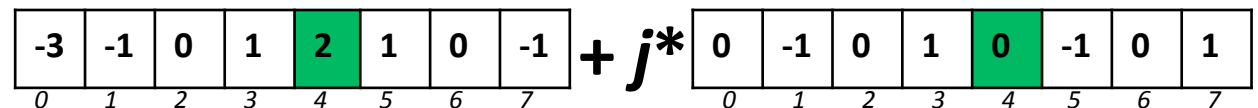
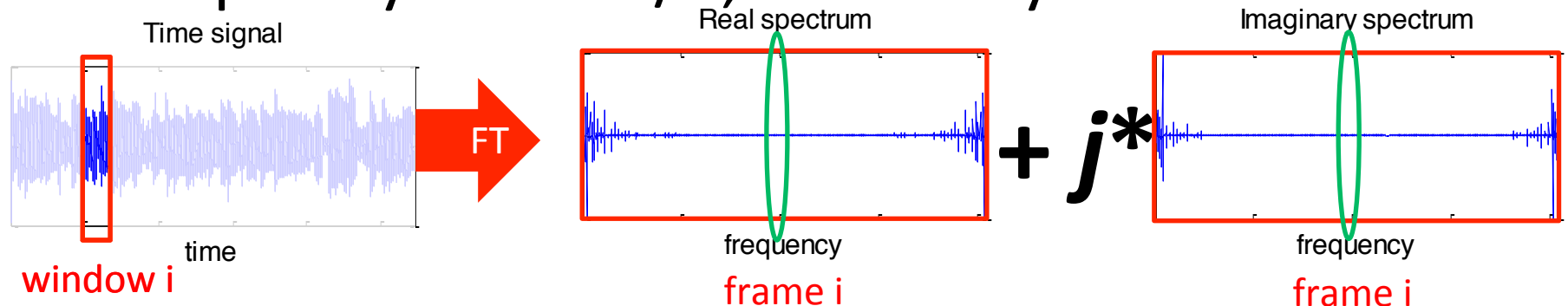
STFT

- Frequency indices from $\text{floor}(N/2)$ to $N-1$ are the “mirrored” **complex conjugates** $(a - j*b)$



STFT

- If N is even, there is a **pivot component** at frequency index $N/2$; it is always real!



STFT

- Summary of the frequency indices and values in the STFT (in colors!)

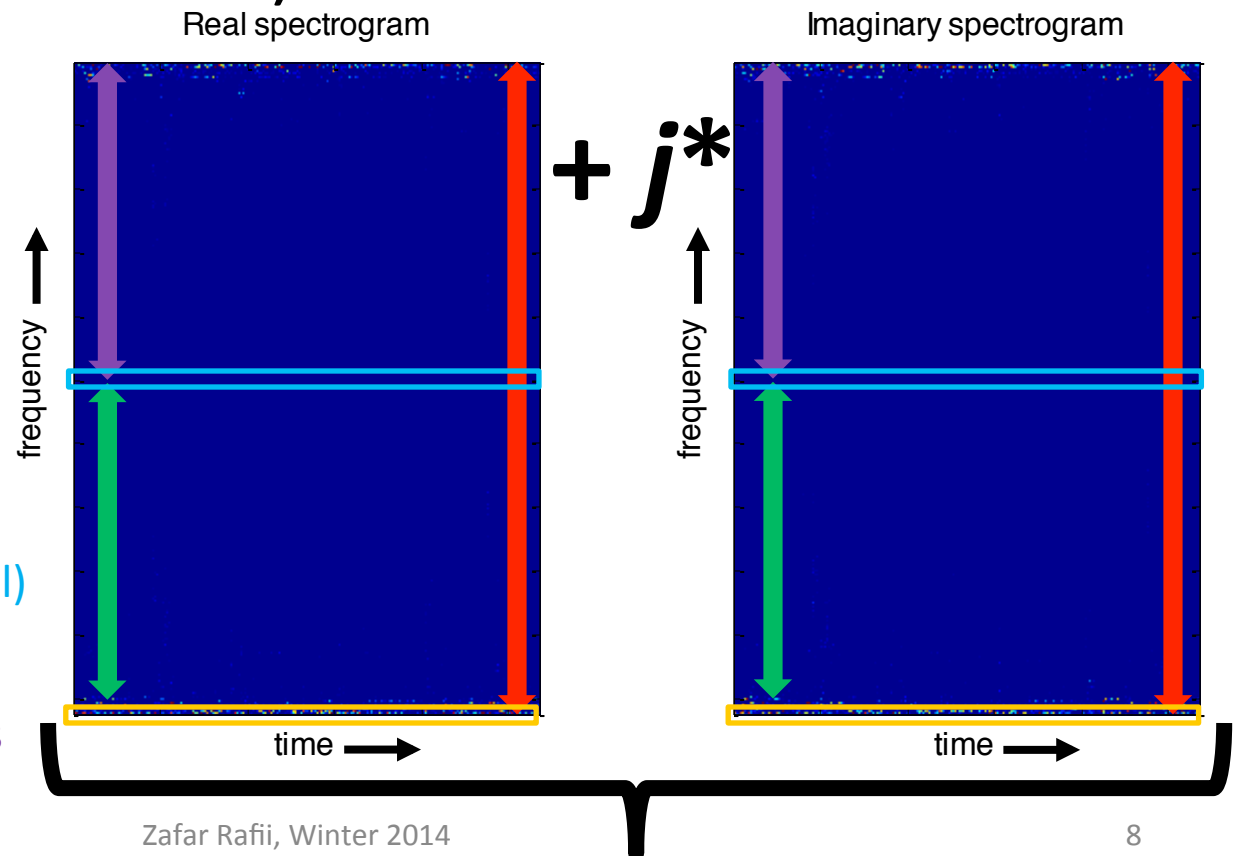
N frequency values =
frequency 0 to N-1

Frequency 0 =
DC component (always real)

Frequency 1 to floor(N/2) =
“unique” complex values

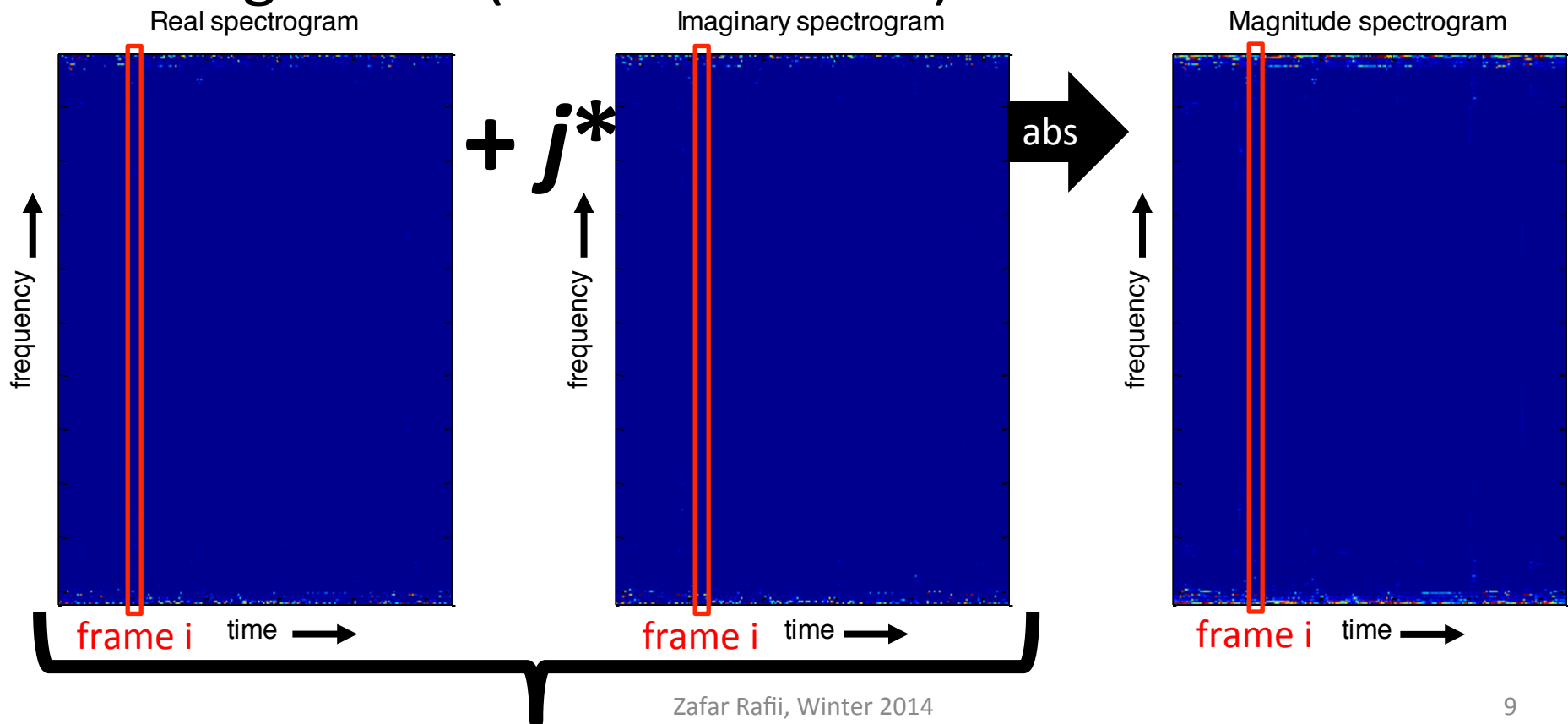
Frequency N/2 =
“pivot” component (always real)

Frequency floor(N/2) to N-1 =
“mirrored” complex conjugates



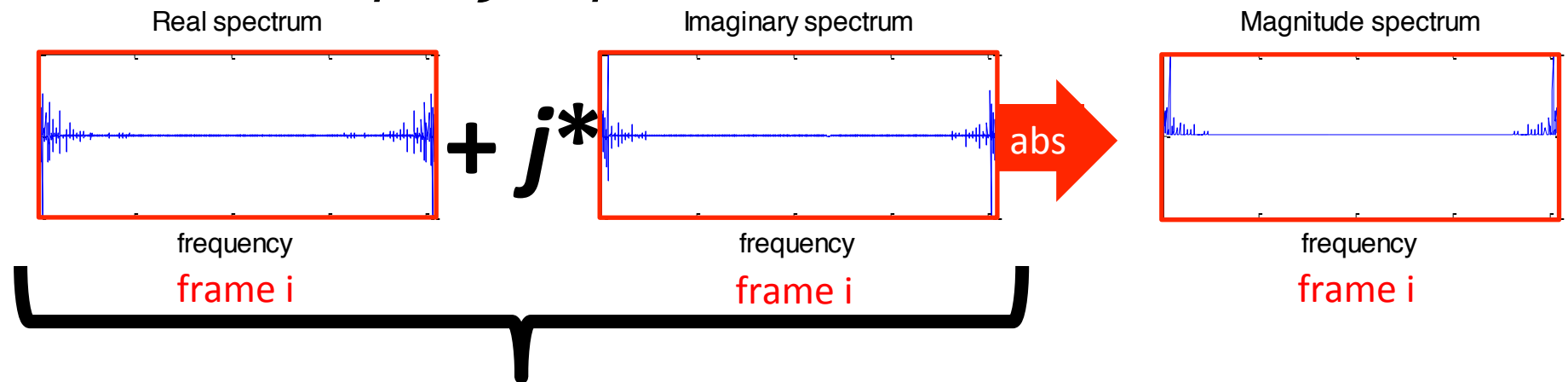
Spectrogram

- The (magnitude) **spectrogram** is the magnitude (absolute value) of the STFT



Spectrogram

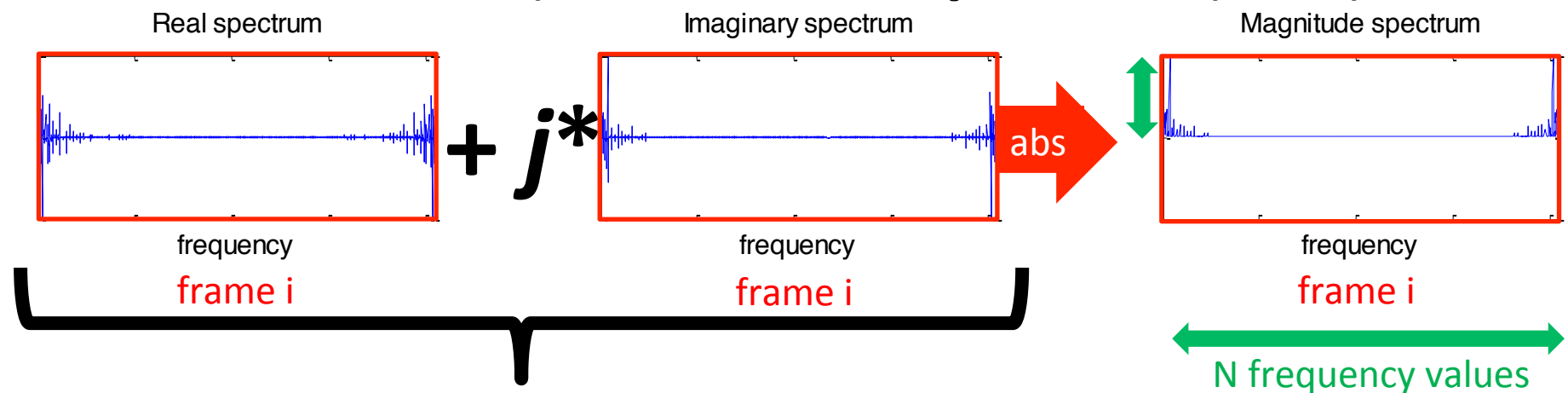
- For a complex number $a+j*b$, the absolute value is $|a+j*b| = \sqrt{a^2 + b^2}$



$$\left| \begin{array}{c|c|c|c|c|c|c|c} -3 & -1 & 0 & 1 & 2 & 1 & 0 & -1 \\ \hline 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{array} \right| + j * \left| \begin{array}{c|c|c|c|c|c|c|c} 0 & -1 & 0 & 1 & 0 & -1 & 0 & 1 \\ \hline 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{array} \right| = \begin{array}{c|c|c|c|c|c|c|c} 3 & 1.4 & 0 & 1.4 & 2 & 1.4 & 0 & 1.4 \\ \hline \end{array}$$

Spectrogram

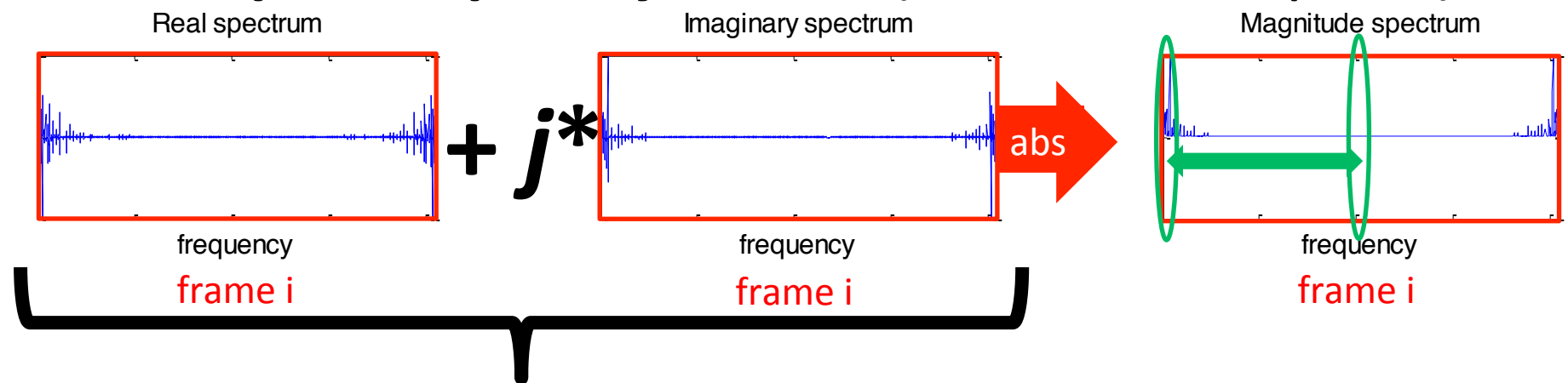
- All the N frequency values (frequency indices from 0 to $N-1$) are **real and positive** (abs!)



$$\begin{bmatrix} -3 & -1 & 0 & 1 & 2 & 1 & 0 & -1 \\ 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{bmatrix} + j * \begin{bmatrix} 0 & -1 & 0 & 1 & 0 & -1 & 0 & 1 \\ 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{bmatrix} = \begin{bmatrix} 3 & 1.4 & 0 & 1.4 & 2 & 1.4 & 0 & 1.4 \\ 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{bmatrix}$$

Spectrogram

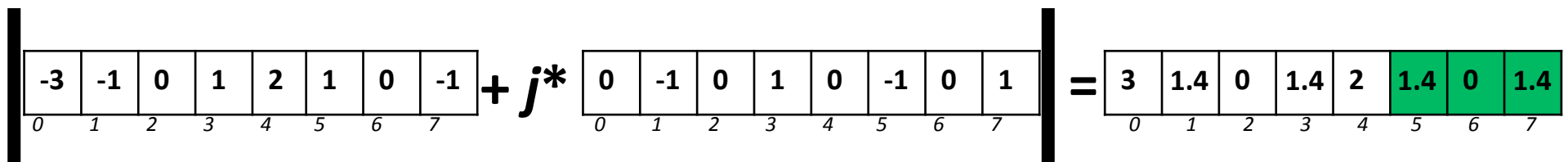
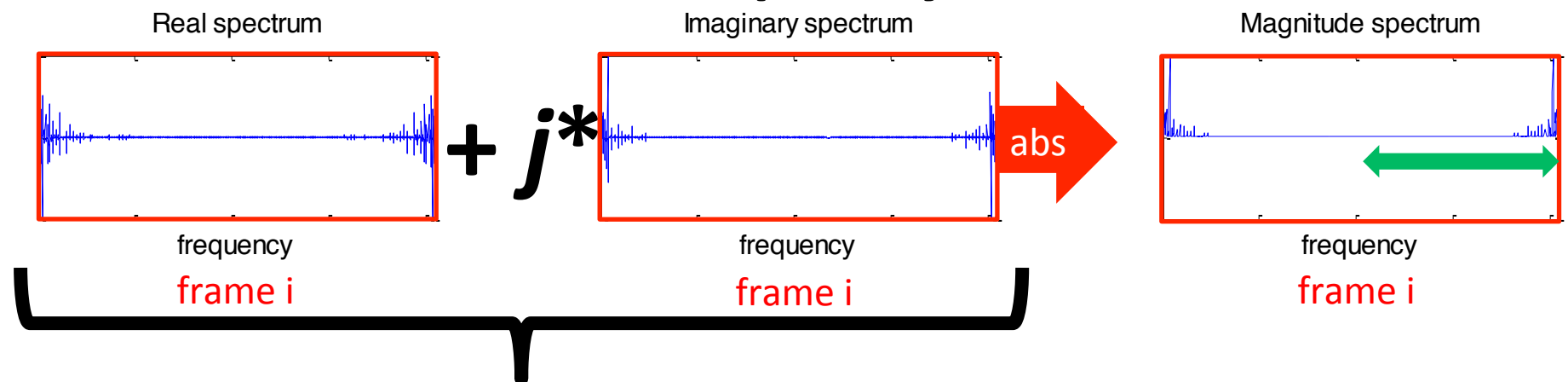
- Frequency indices from 0 to floor(N/2) are the **unique frequency values** (with DC and pivot)



$$\left| \begin{array}{c|c|c|c|c|c|c|c} -3 & -1 & 0 & 1 & 2 & 1 & 0 & -1 \\ \hline 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{array} \right| + j * \left| \begin{array}{c|c|c|c|c|c|c|c} 0 & -1 & 0 & 1 & 0 & -1 & 0 & 1 \\ \hline 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{array} \right| = \begin{array}{c|c|c|c|c|c|c|c} 3 & 1.4 & 0 & 1.4 & 2 & 1.4 & 0 & 1.4 \\ \hline 0 & 1 & 2 & 3 & 4 & 5 & 6 & 7 \end{array}$$

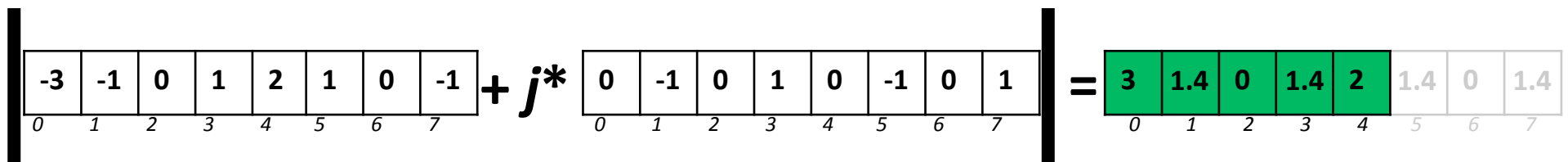
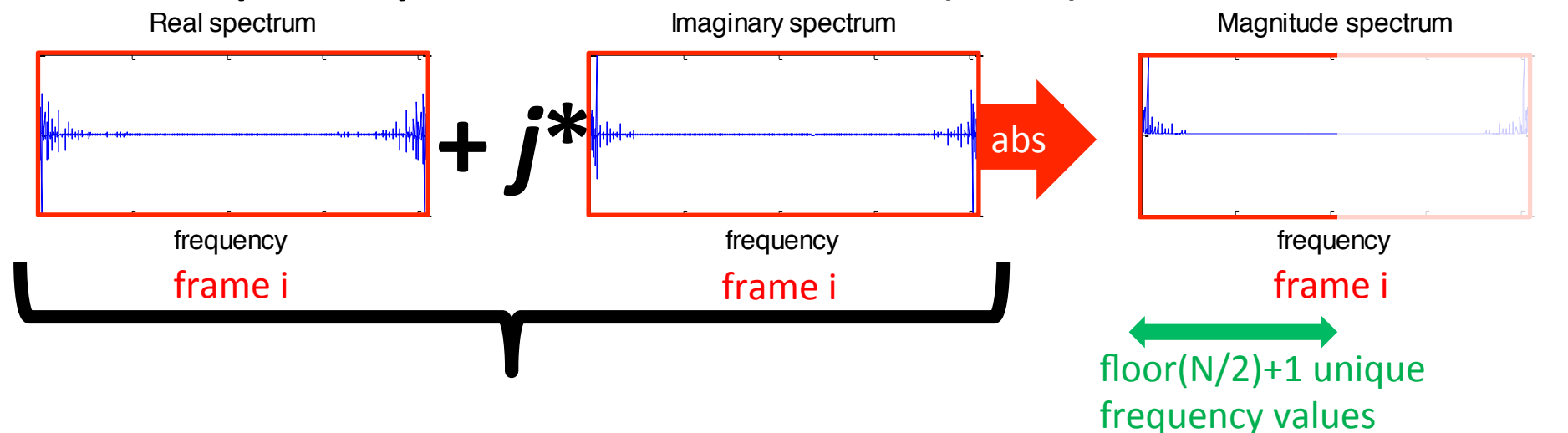
Spectrogram

- Frequency indices from $\text{floor}(N/2)+1$ to $N-1$ are the **mirrored frequency values**



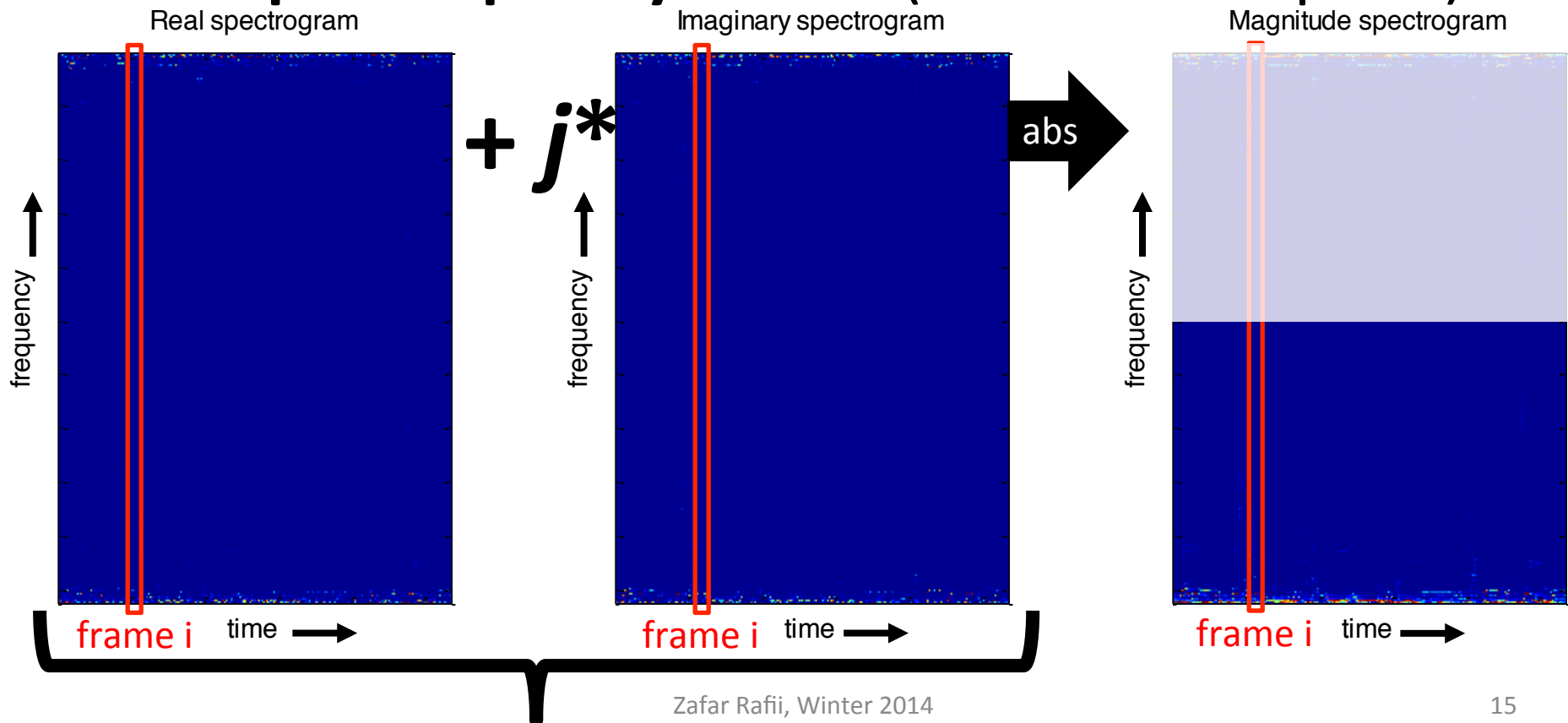
Spectrogram

- Since they are redundant, we can discard the frequency values from $\text{floor}(N/2)+1$ to $N-1$



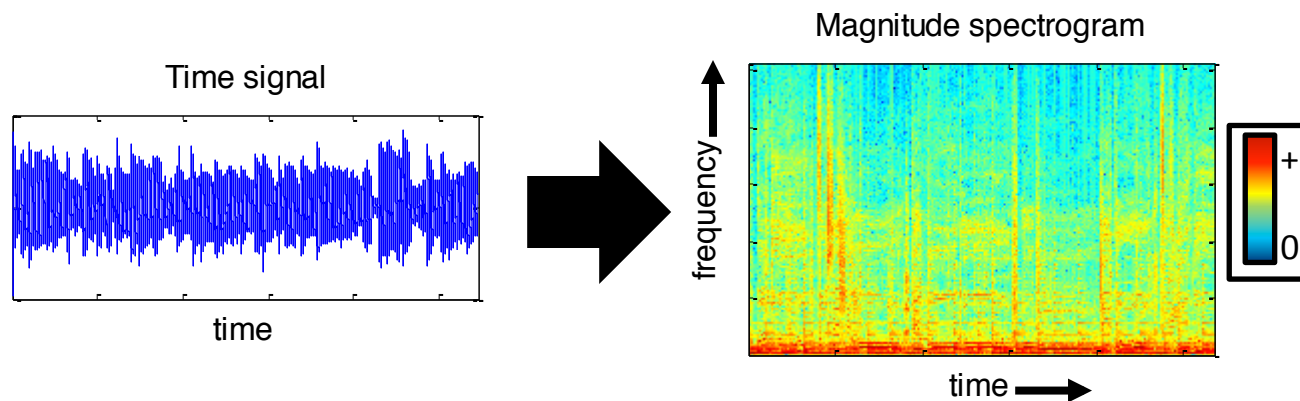
Spectrogram

- The spectrogram has therefore **$\text{floor}(N/2)+1$ unique frequency values** (with DC and pivot)



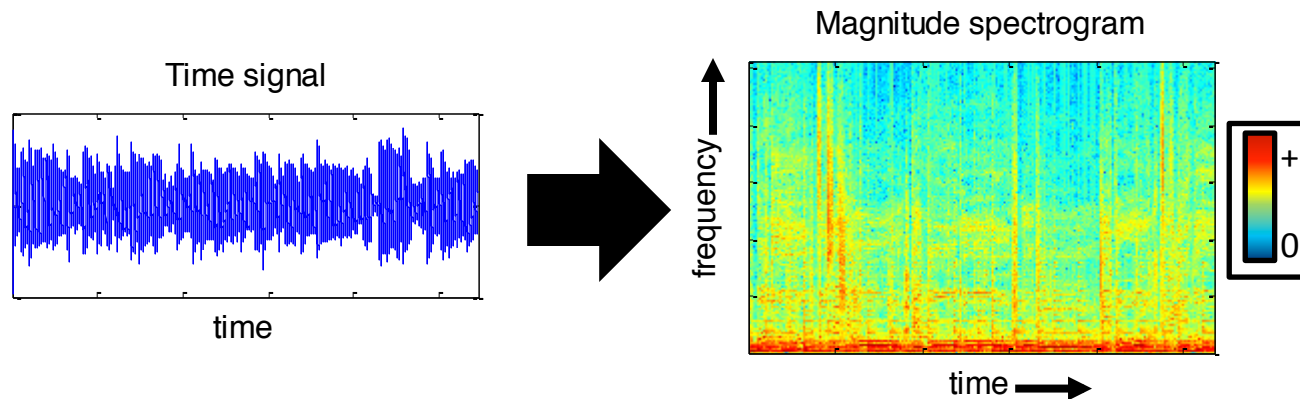
Spectrogram

- Why the magnitude spectrogram?
 - Easy to visualize (compare with the STFT)
 - Magnitude information more important
 - Human ear less sensitive to phase



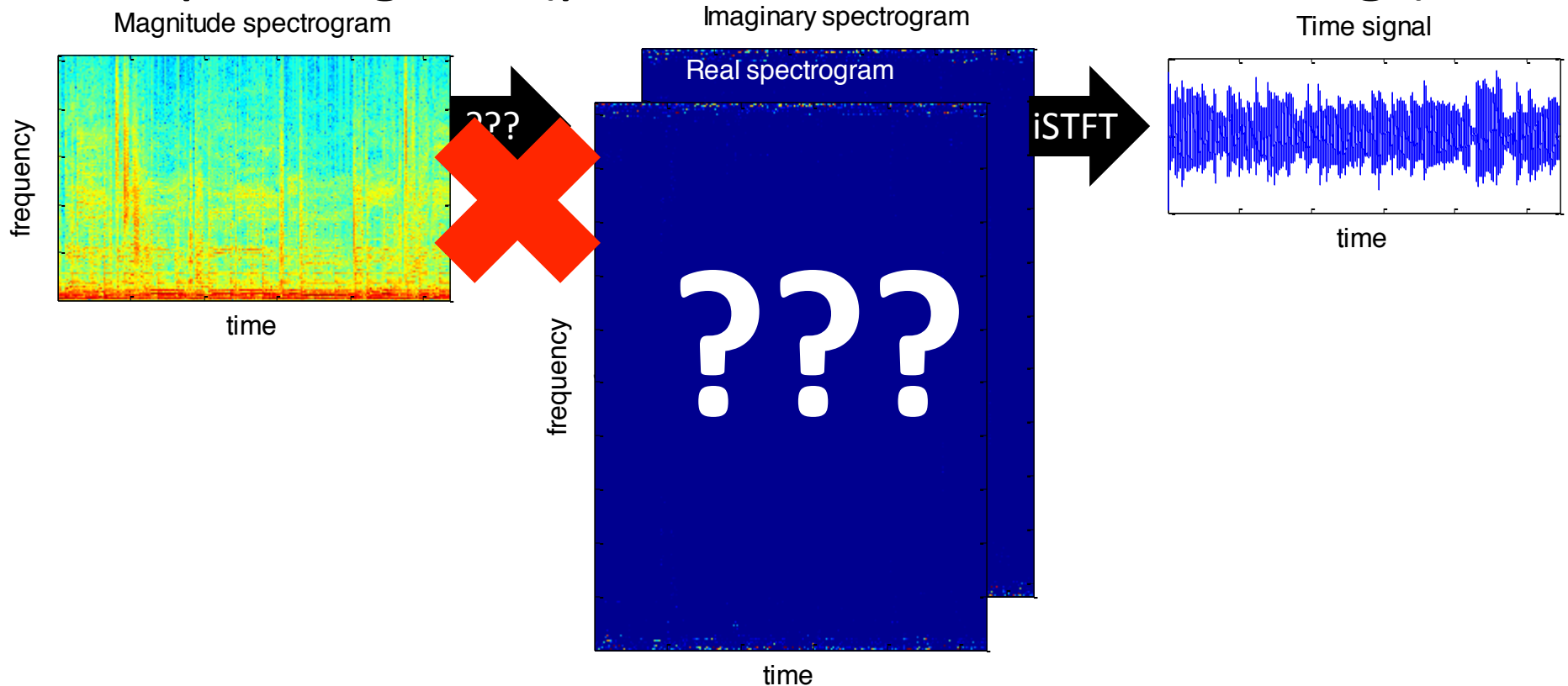
Spectrogram

- When you display a spectrogram in Matlab...
 - *imagesc*: data is scaled to use the full colormap
 - $10*\log_{10}(V)$: magnitude spectrogram in dB
 - *set(gca,'YDir','normal')*: y-axis from bottom to top



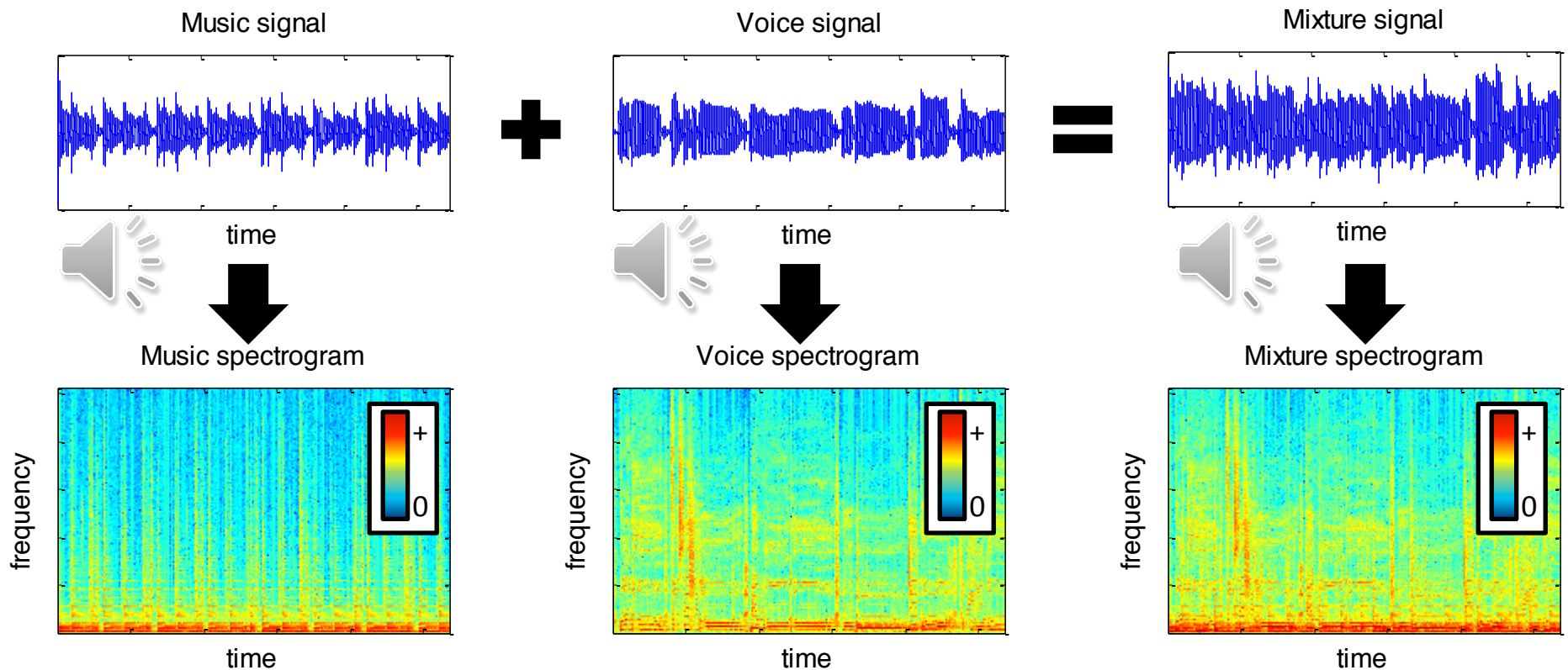
Spectrogram

- The signal **cannot be reconstructed** from the spectrogram (phase information is missing!)



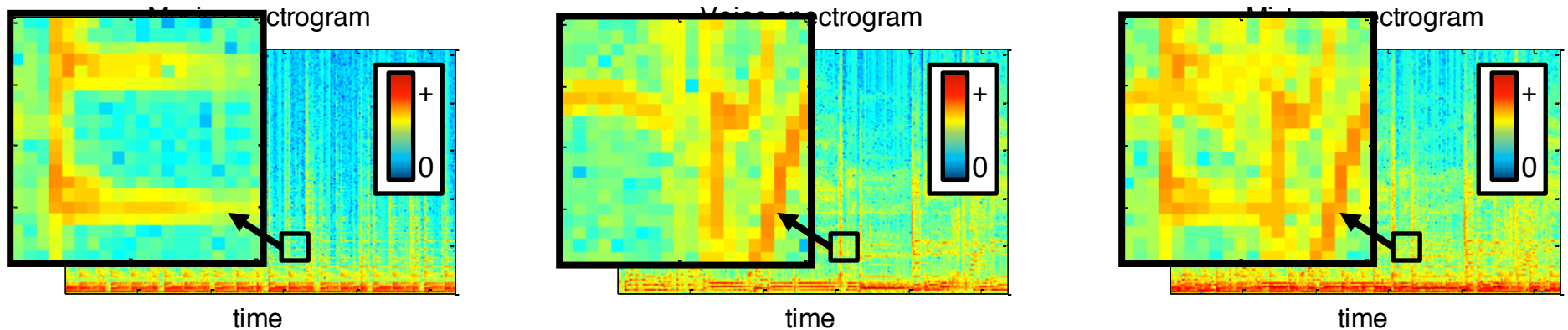
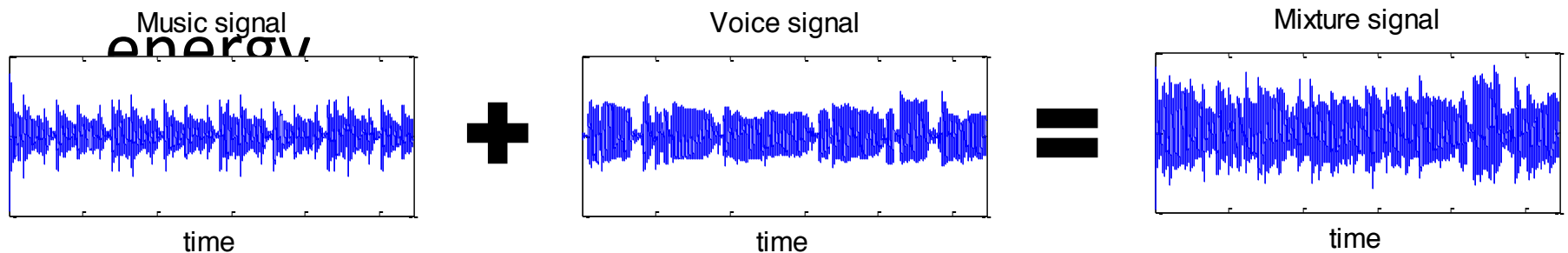
Time-frequency Masking

- Suppose we have a mixture of two sources: a music signal and a voice signal



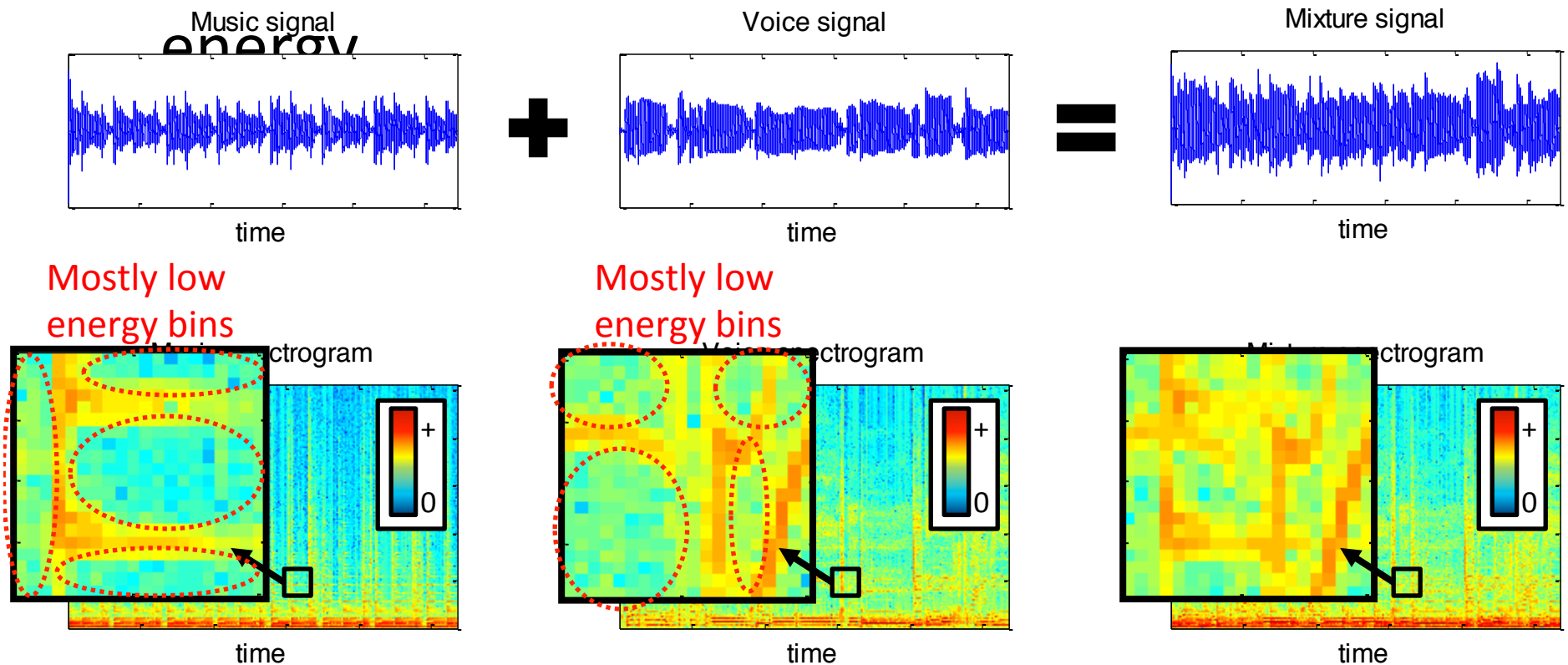
Time-frequency Masking

- We assume that the sources are **sparse** = most of the time-frequency bins have null



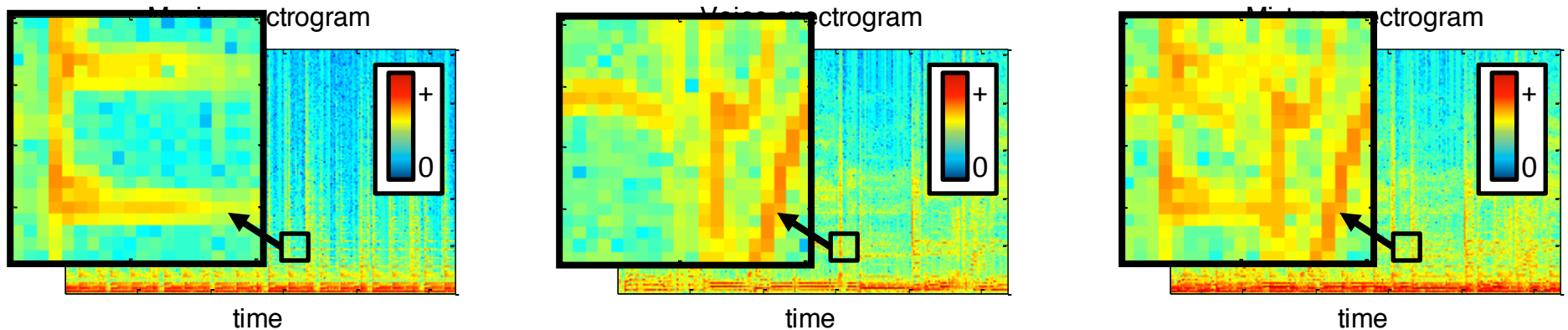
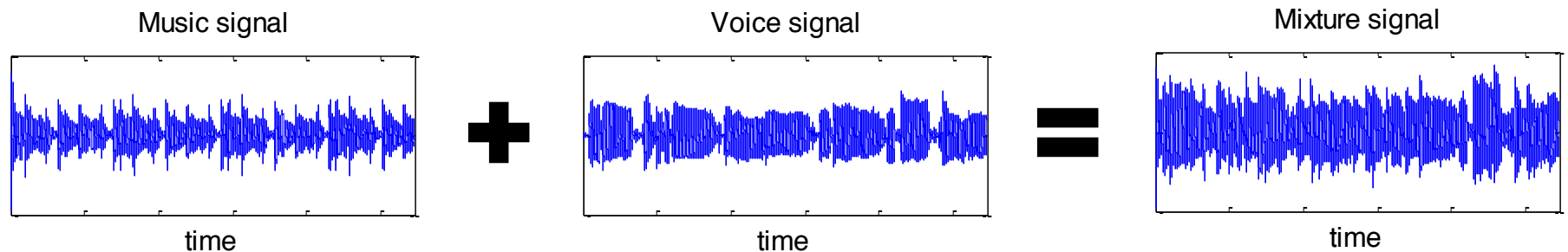
Time-frequency Masking

- We assume that the sources are **sparse** = most of the time-frequency bins have null



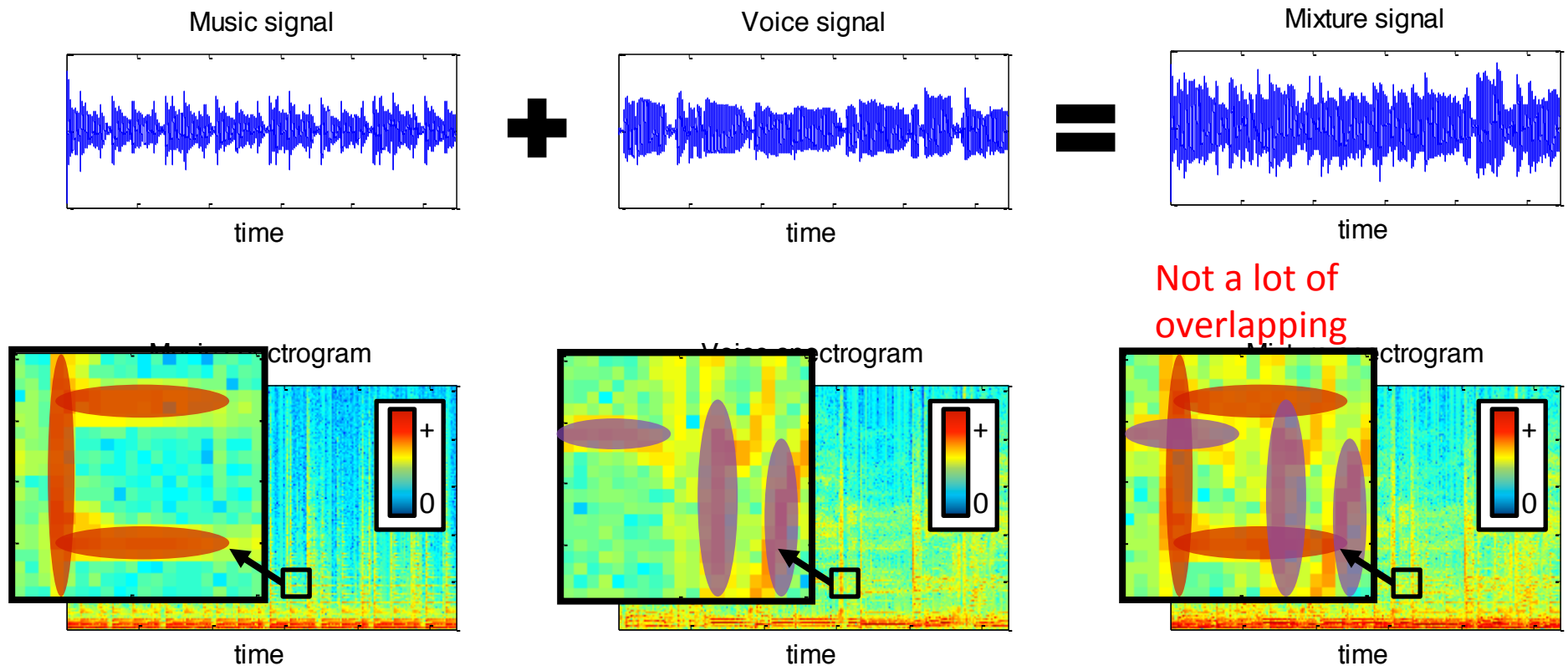
Time-frequency Masking

- We assume that the sources are **disjoint** = their time-frequency bins do not overlap



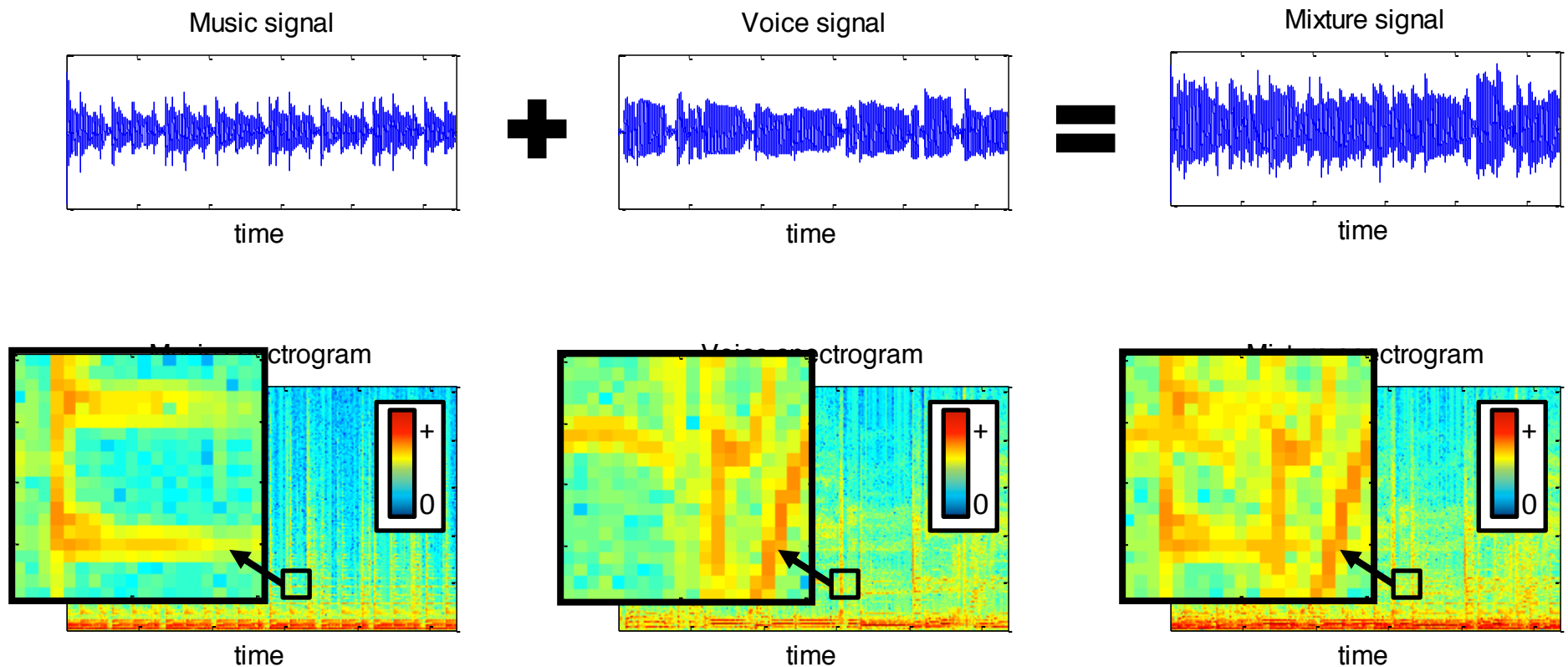
Time-frequency Masking

- We assume that the sources are **disjoint** = their time-frequency bins do not overlap



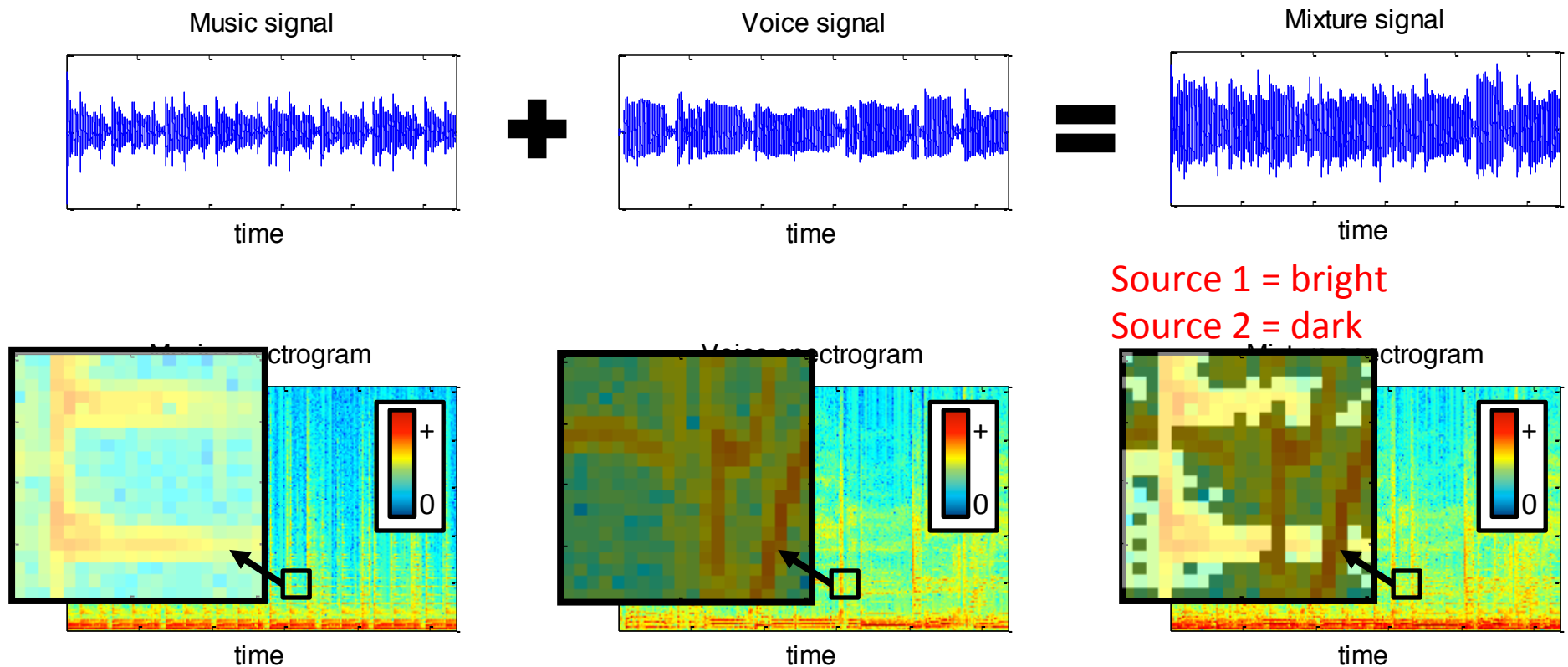
Time-frequency Masking

- Assuming sparseness and disjointness, we can **discriminate** the bins between mixed sources



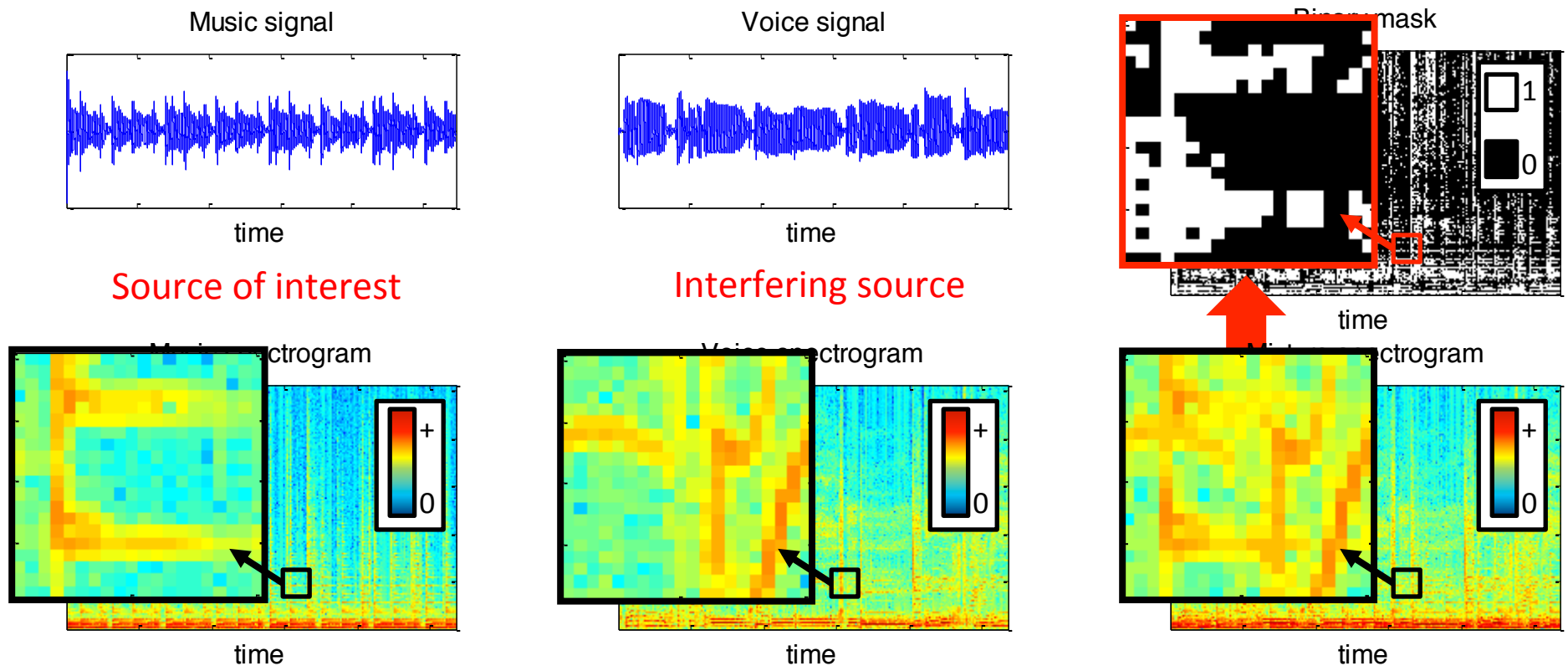
Time-frequency Masking

- Assuming sparseness and disjointness, we can **discriminate** the bins between mixed sources



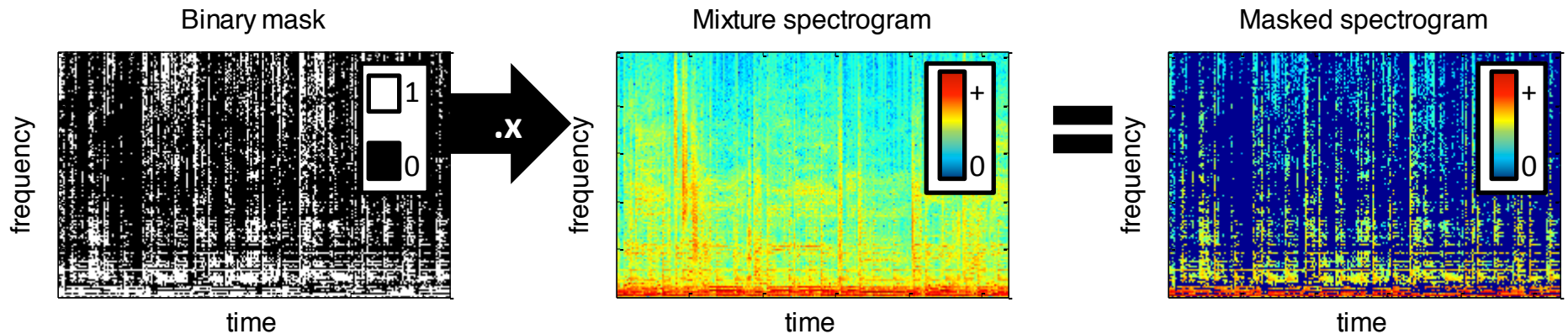
Time-frequency Masking

- Bins that are likely to belong to one source are assigned to 1, the rest to 0 = **binary masking**!



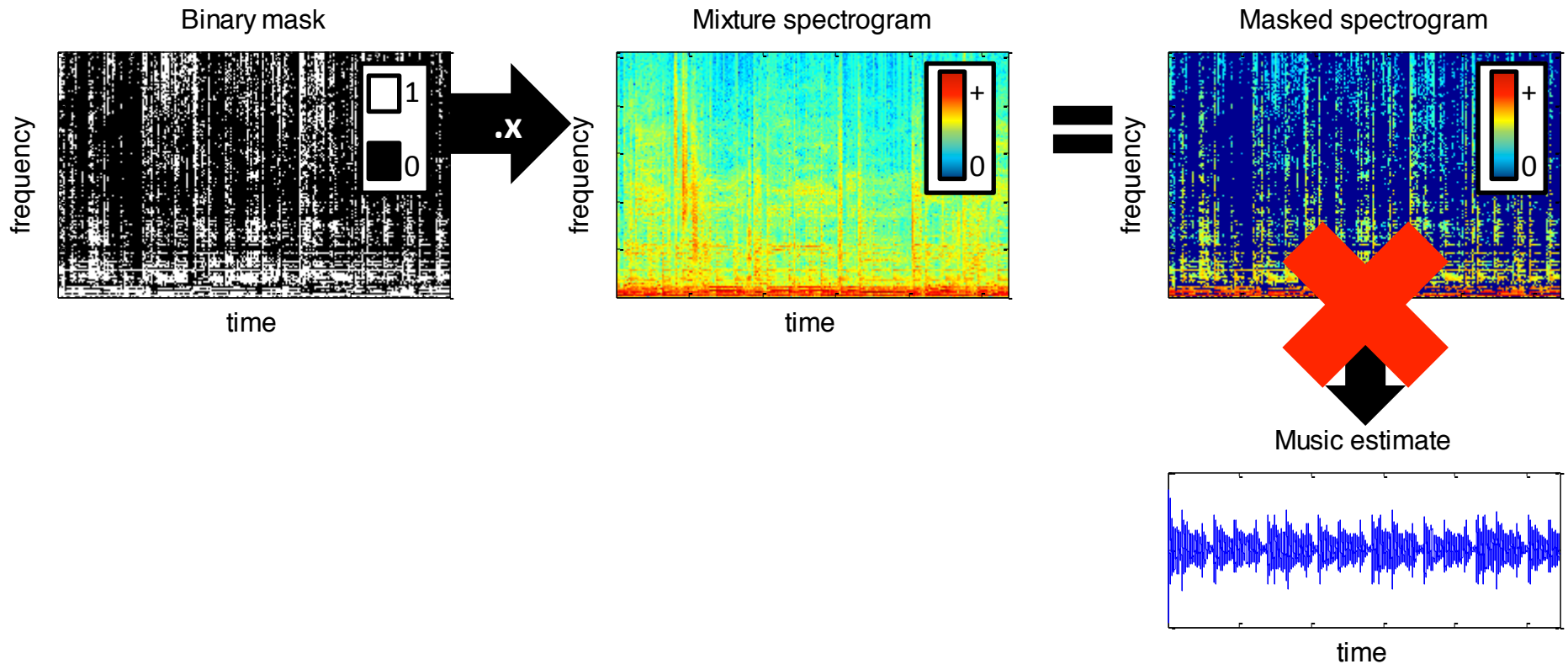
Time-frequency Masking

- By multiplying the binary mask to the mixture spectrogram, we can “preview” the estimate



Time-frequency Masking

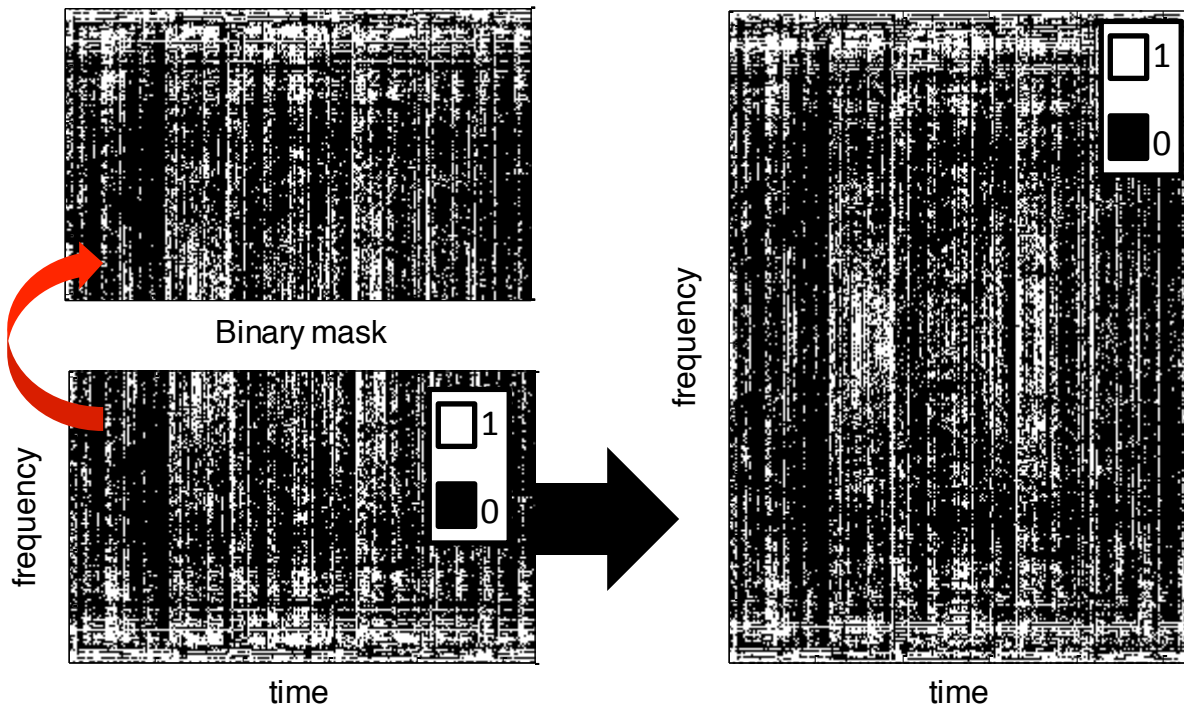
- However, we cannot derive the estimate itself because we cannot invert a spectrogram!



Time-frequency Masking

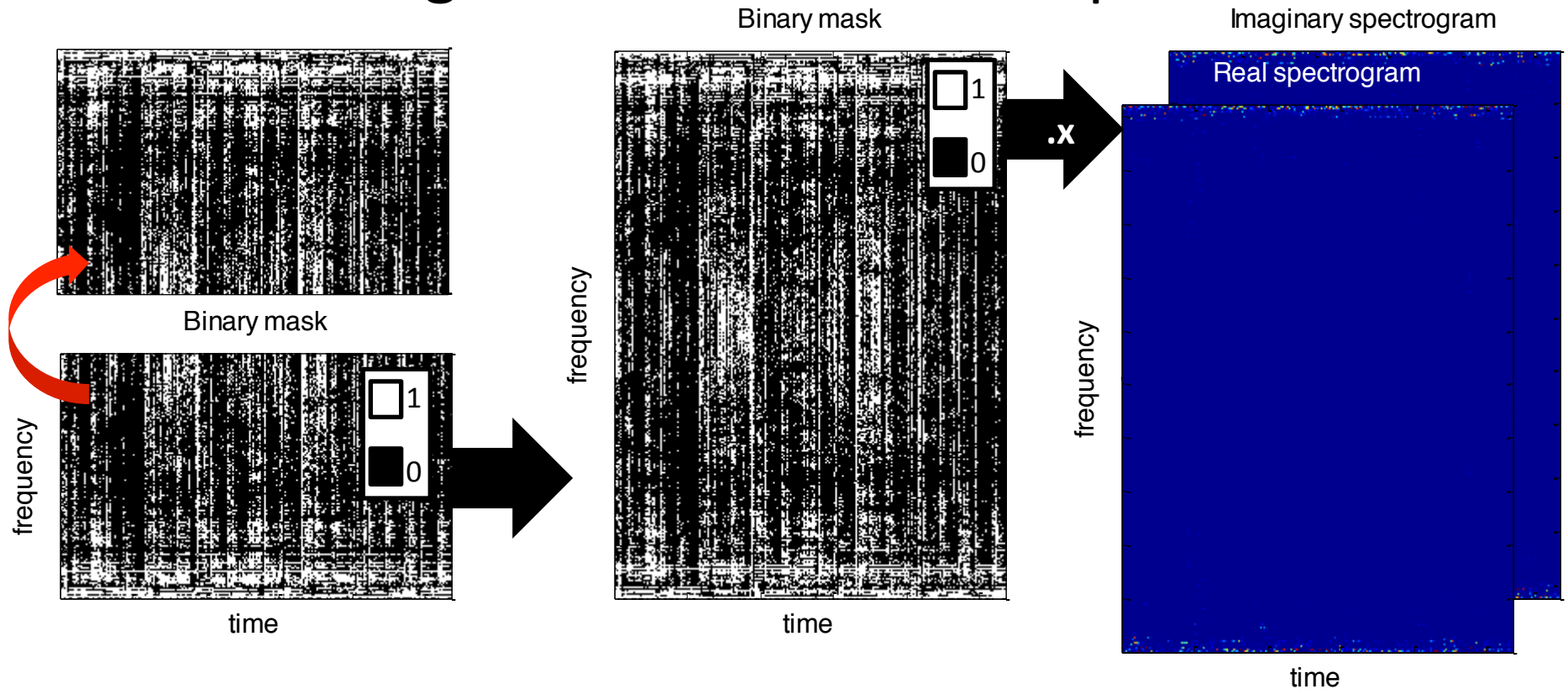
- We mirror the redundant frequencies from the unique frequencies (without DC and pivot)

Binary mask



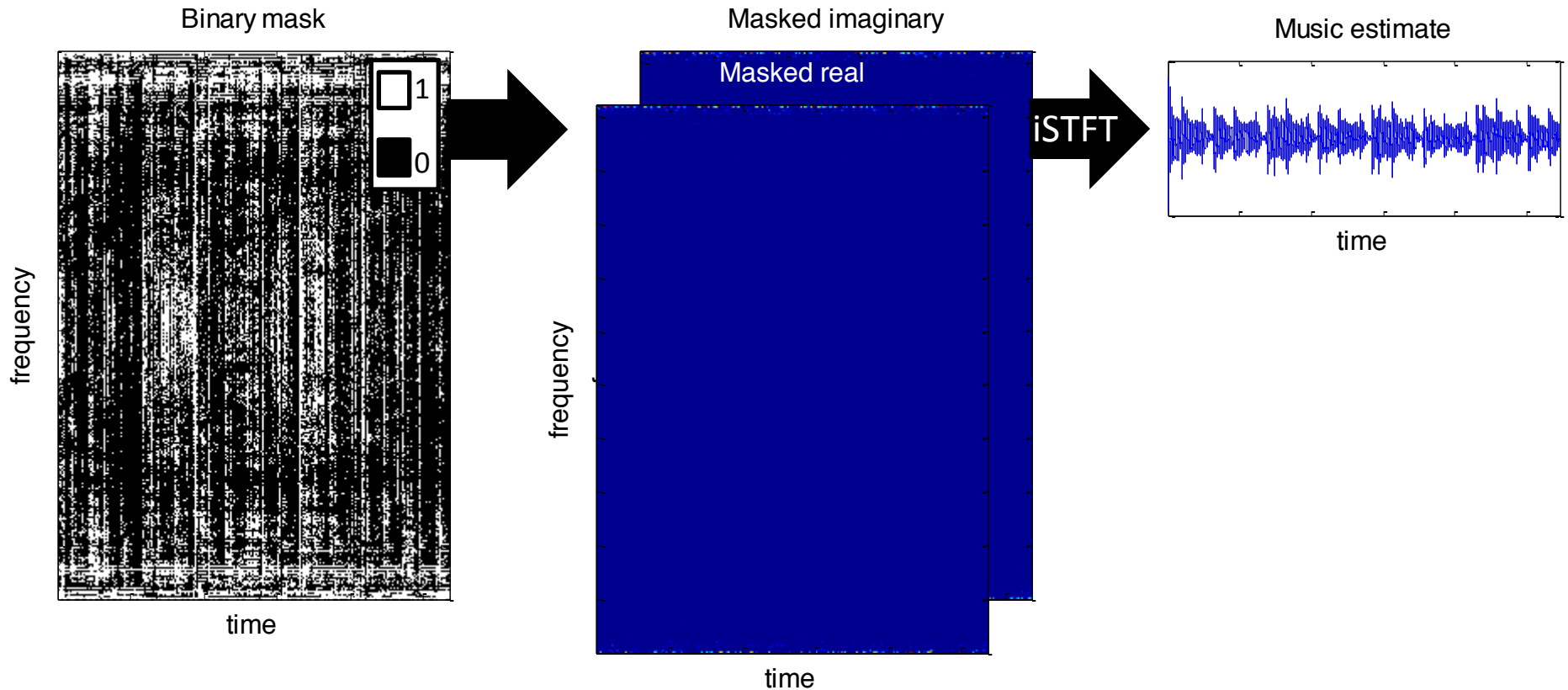
Time-frequency Masking

- We then apply this full binary mask to the STFT using a element-wise multiplication



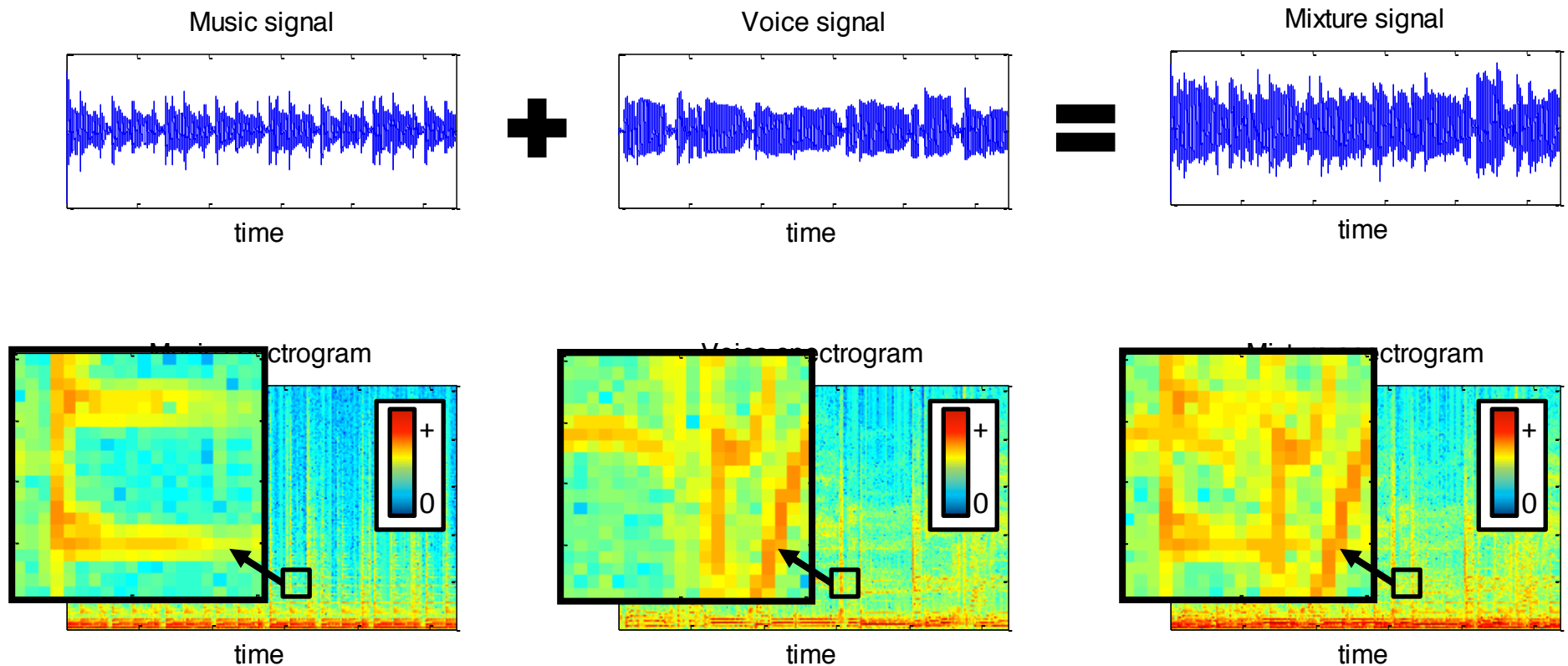
Time-frequency Masking

- The estimate signal can now be reconstructed via inverse STFT



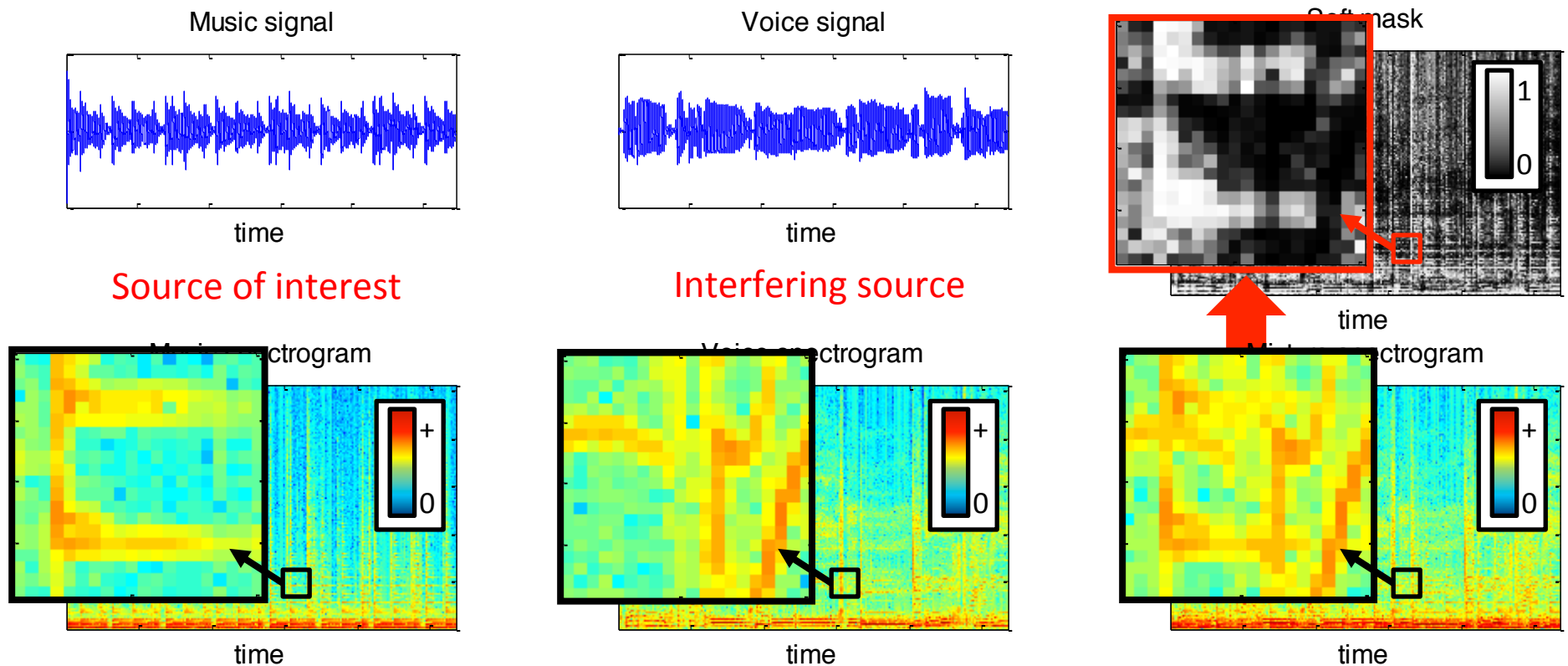
Time-frequency Masking

- Sources are not really sparse or disjoint in time-frequency in the mixture



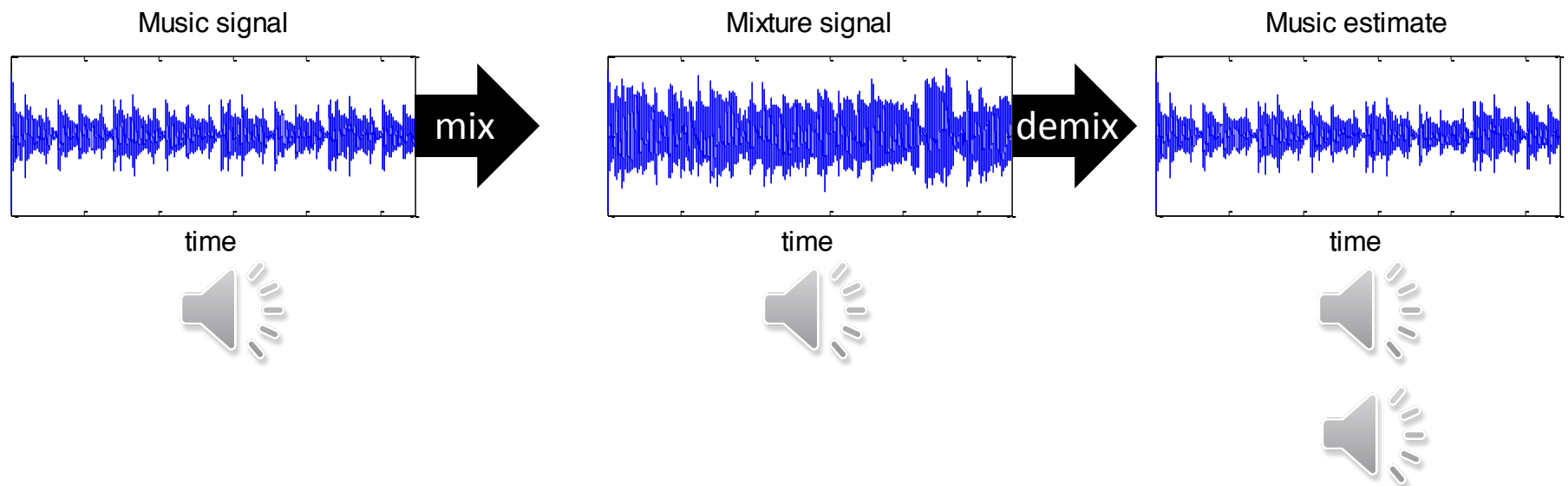
Time-frequency Masking

- Bins that are likely to belong to one source are close to 1, the rest close to 0 = **soft masking**!



Time-frequency Masking

- Let's listen to the results!



Question

- How can we efficiently model a binary/soft time-frequency mask for source separation?...
- To be continued...

