

11.09.2020

Digital Image Processing (CSE/ECE 478)

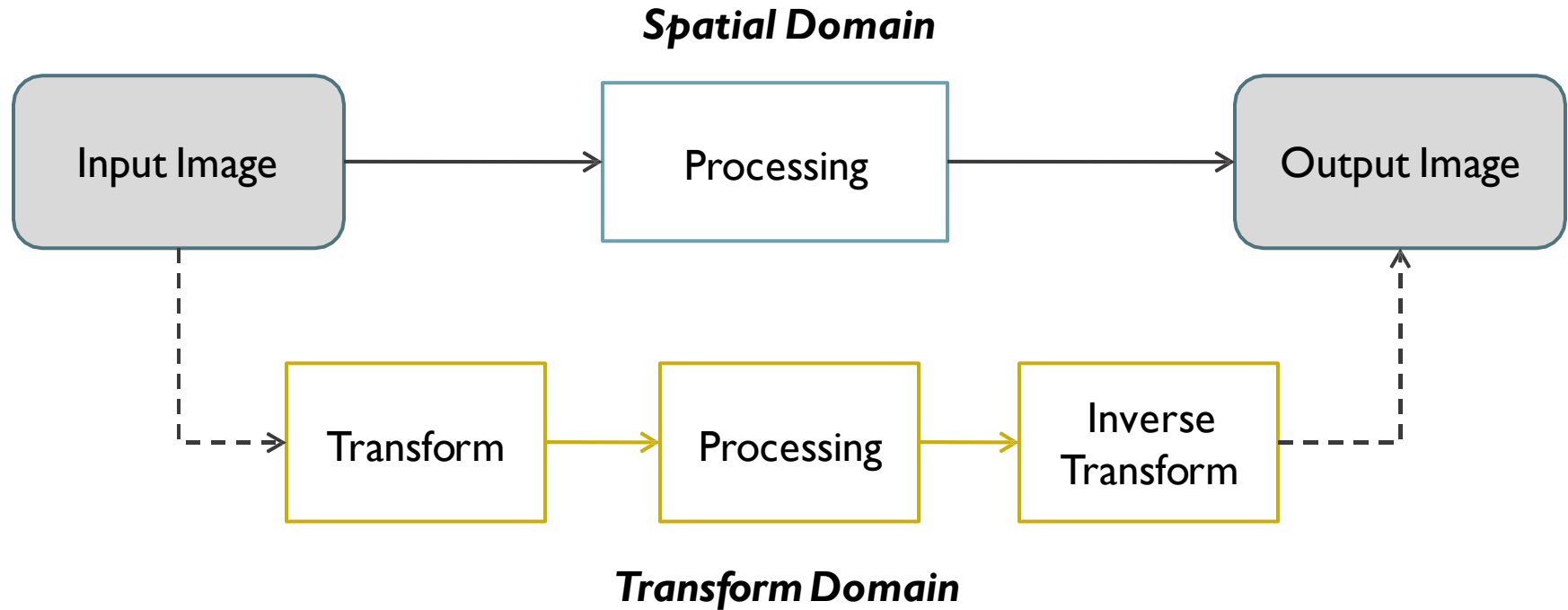
Lecture-10: Frequency Domain Processing

Ravi Kiran

Center for Visual Information Technology (CVIT), IIIT Hyderabad



Spatial vs. Transform Domain Processing

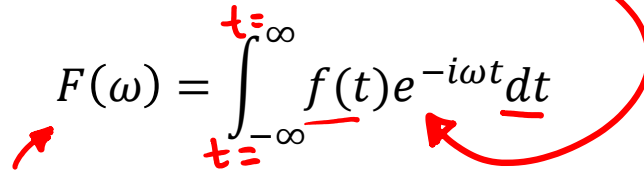


Fourier Transform

Approximate non-periodic signals with complex sinusoids

Intuition for FT

- $f(t)$ = Single number
- How much of frequency ω signal is present for all values of t ?



The diagram shows the Fourier Transform integral $F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-i\omega t} dt$. Red annotations include: a red arrow pointing to $F(\omega)$; a red arrow pointing to the lower limit $-\infty$ with a red $t =$ written below it; a red arrow pointing to the upper limit ∞ with a red $t =$ written above it; and a large red curved arrow pointing from the text 'frequency ω signal' in the list above to the ω in the exponent $e^{-i\omega t}$.

$$F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-i\omega t} dt$$

$$F(\omega) = \mathcal{F}[f(t)]$$

Fourier Transform and Inverse Fourier Transform

- Fourier Transform

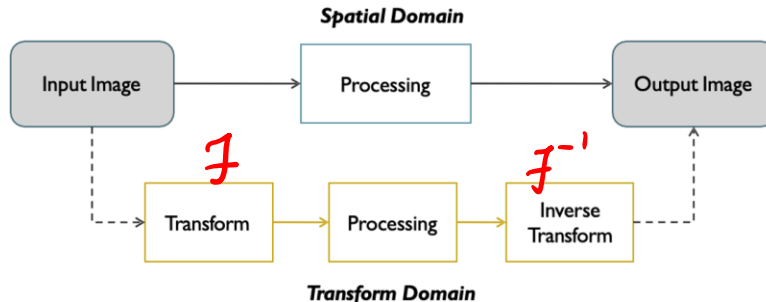
$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-i\omega t} dt$$

$$F(\omega) = \mathcal{F}[f(t)]$$

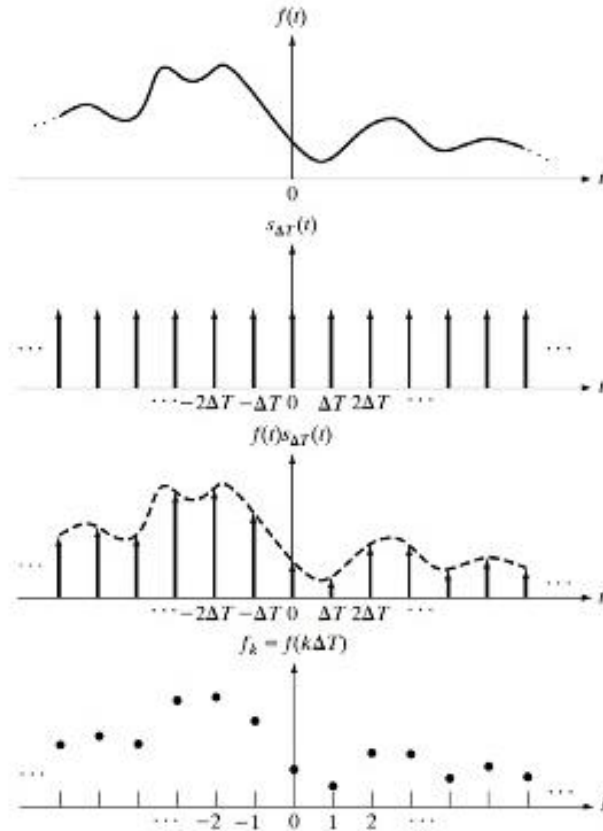
- Inverse Fourier Transform

$$f(t) = \int_{\omega=-\infty}^{\omega=\infty} \frac{1}{2\pi} F(\omega) e^{i\omega t} d\omega$$

$$f(t) = \mathcal{F}^{-1}[F(\omega)]$$



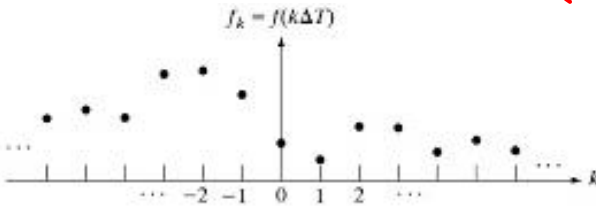
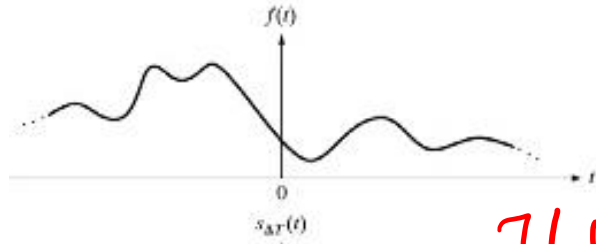
Sampling = $f(t)$ x Impulse Train



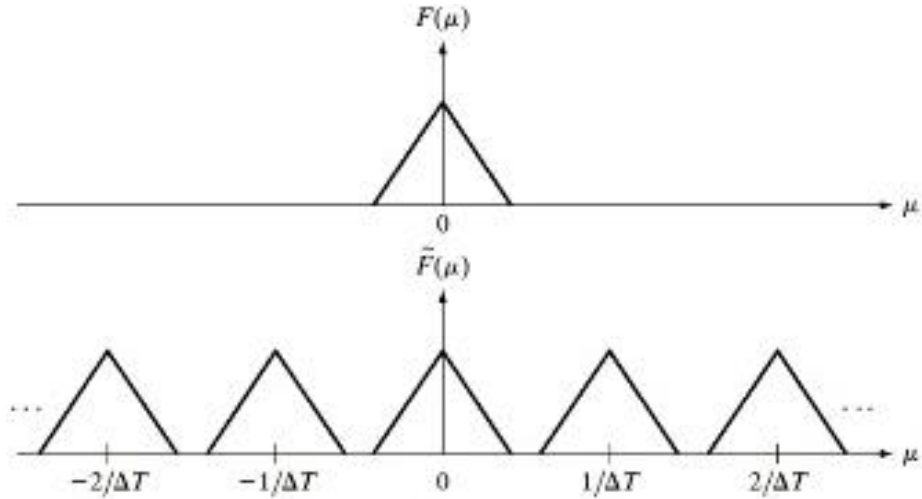
$$s_{\Delta T}(t) = \sum_{n=-\infty}^{n=\infty} \delta(t - n\Delta T)$$

$$\underline{\tilde{f}(t)} = \sum_{n=-\infty}^{n=\infty} \underbrace{f_n}_{\text{red}} \delta(t - n\Delta T)$$

FT of sampled function (G&W 4.2.4)



$$f(f(t) s_{\Delta}(t))$$

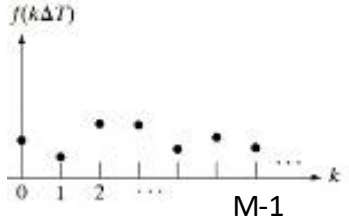


- Continuous
- Periodic (copies of $f(t)$'s FT)

$$\tilde{f}(t) = \sum_{n=-\infty}^{n=\infty} f_n \delta(t - n\Delta T)$$

$$\tilde{F}(\mu) = \frac{1}{\Delta T} \sum_{n=-\infty}^{n=\infty} F\left(\mu - \frac{n}{\Delta T}\right) \quad \frac{1}{\Delta T}$$

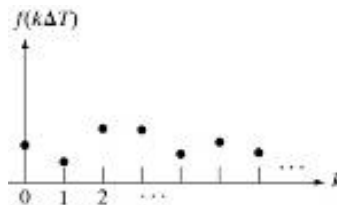
FT of sampled function



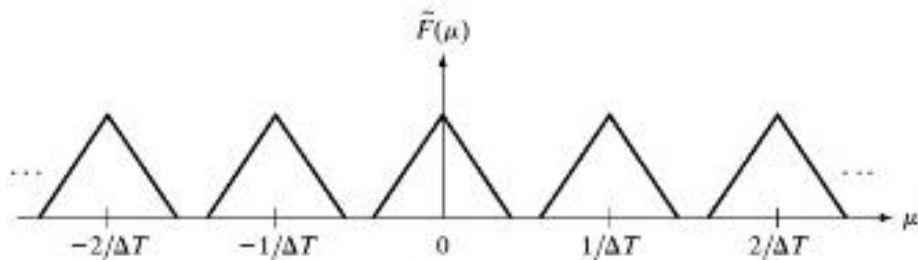
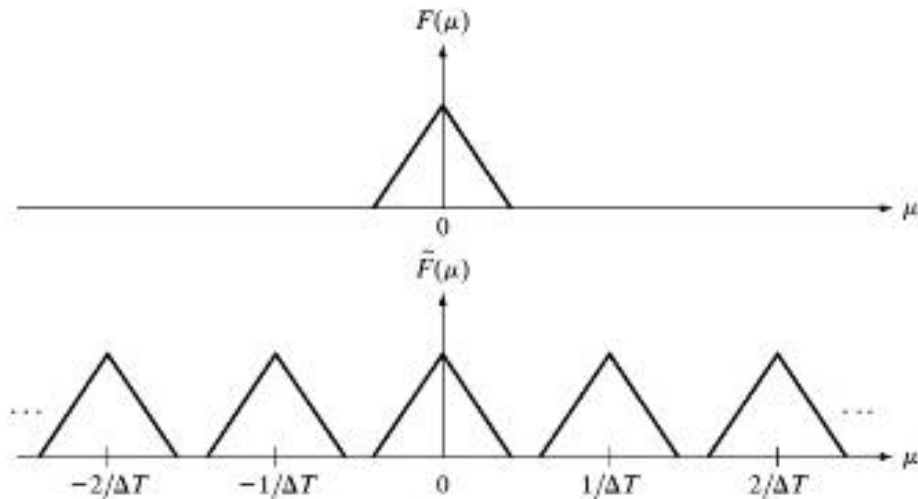
$$\tilde{f}(t) = \sum_{n=0}^{n=(M-1)\Delta T} f_n \delta(t - n\Delta T)$$

$$\underline{\tilde{F}(\mu)} = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T}$$

Digital processing of frequencies



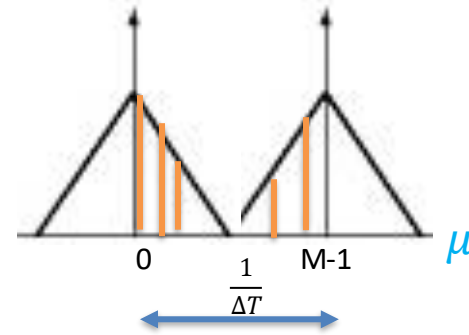
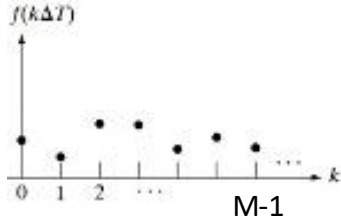
$$\tilde{f}(t) = \sum_{n=-\infty}^{n=\infty} f_n \delta(t - n\Delta T)$$



$$\tilde{F}(\mu) = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T} \quad \mu \in \mathbb{R}$$

- Need discrete frequency samples, but FT of sampled function is continuous
- OBSERVATION: Characterizing one period ($\frac{1}{\Delta T}$) is enough
- How do we get frequency 'samples' ?

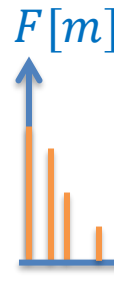
FT of sampled function (G&W 4.4.1)



M sample

$$\tilde{f}(t) = \sum_{n=0}^{n=(M-1)\Delta T} f_n \delta(t - n\Delta T)$$

$$\tilde{F}(\mu) = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T}, \mu \in R$$



• Substituting $\mu = \frac{m}{M\Delta T}$ $m = 0, 1, 2, \dots, M-1$

$$F[m] = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\frac{mn}{M}}, m = 0, 1, \dots, (M-1)$$

NOTE: No direct dependence on ΔT



DFT and IDFT

$$\underline{F[m]} = \sum_{n=0}^{n=(M-1)} f_n e^{\frac{-j2\pi nm}{M}}, m = 0, 1, \dots (M-1)$$

$$f_n = \frac{1}{M} \sum_{m=0}^{m=(M-1)} F_m e^{\frac{j2\pi nm}{M}}, n = 0, 1, \dots (M-1)$$

$F[m]$

$$\text{Re}\{F[m]\} \quad \text{Im}\{F[m]\}$$

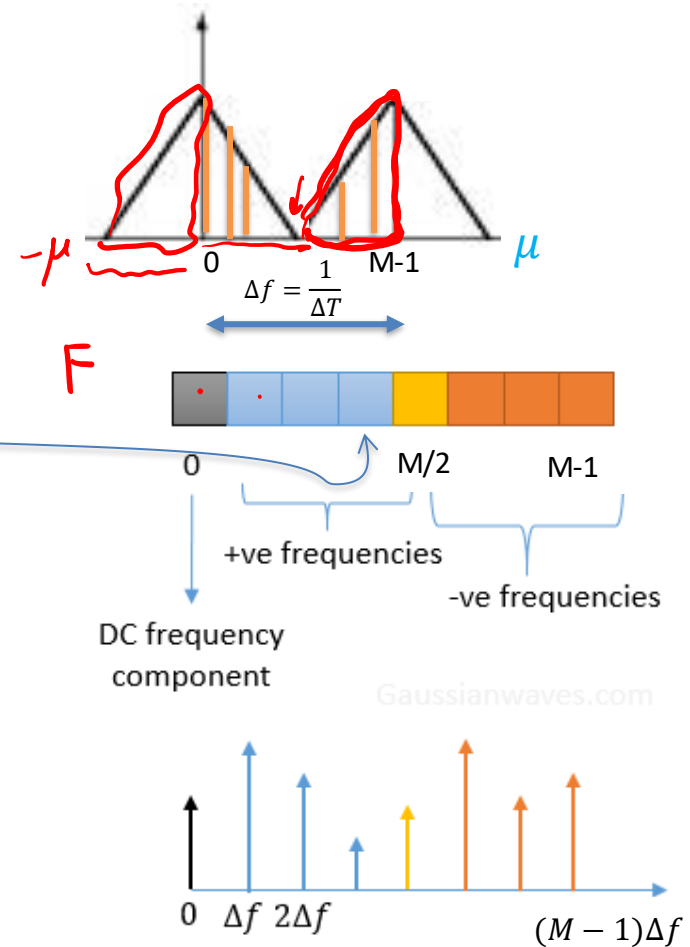
- A complex value
- Represents amplitude, phase of function $f[.]$'s content at angular frequency $2\pi m/M$

DFT: Record of 'energy' portion at various frequency bands present in input function $f[.]$

$$C = A + iB \quad |C| = \sqrt{A^2 + B^2}$$
$$\phi = \tan^{-1}\left(\frac{B}{A}\right)$$

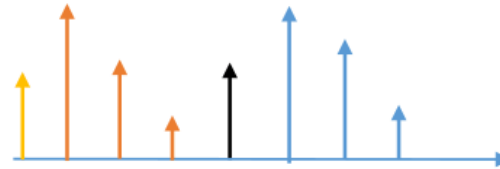
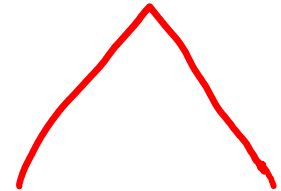
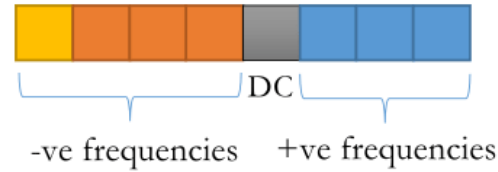
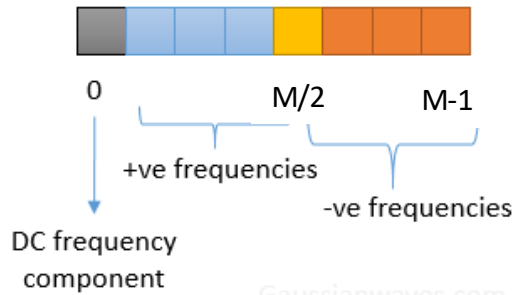
DFT (in practice)

$$F[m] = \sum_{n=0}^{n=(M-1)} \underline{f[n]} e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$



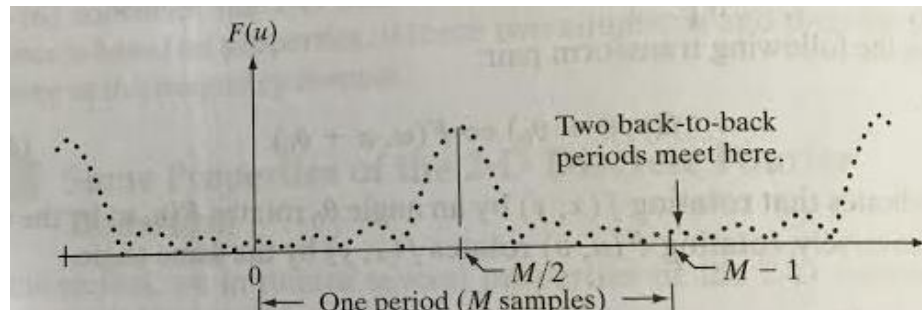
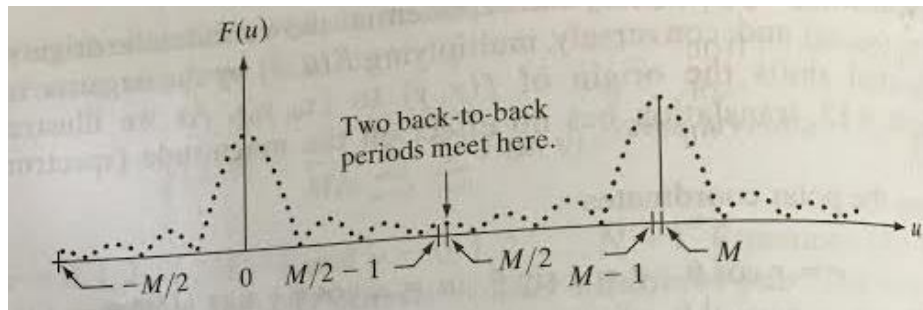
DFT – center shifted (for plotting)

$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$



Shifting origin

1-D



$$F(u) \quad F(u - u_o)$$

$$\underbrace{f[x] e^{\frac{j2\pi u_o x}{M}}}_{\text{red underline}} \leftrightarrow \underbrace{F(u - u_o)}_{\text{red underline}}$$

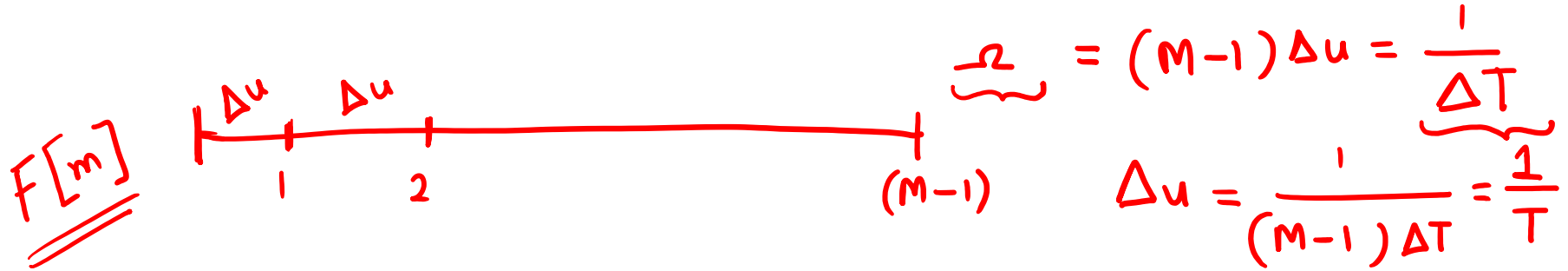
$$\underbrace{f[n](-1)^n}_{\text{red underline}}$$

$$u_o = \frac{M}{2}$$

Relationship between Sampling and Frequency Intervals

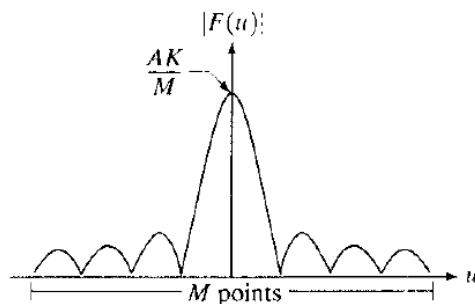
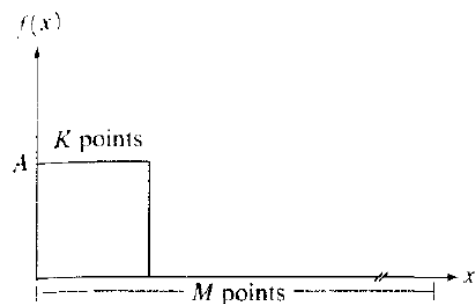


$$T = (M-1) \Delta T$$



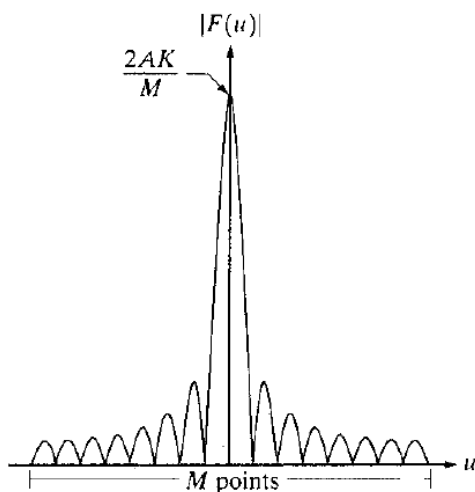
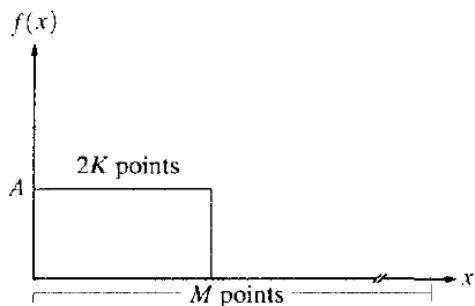
- Ω (Range of frequencies) depends inversely on sampling interval ΔT
- Δu (Frequency Resolution of DFT) depends inversely on duration T over which $f(t)$ is sampled

Relationship between u and x



a b
c d

FIGURE 4.2 (a) A discrete function of M points, and (b) its Fourier spectrum. (c) A discrete function with twice the number of nonzero points, and (d) its Fourier spectrum.



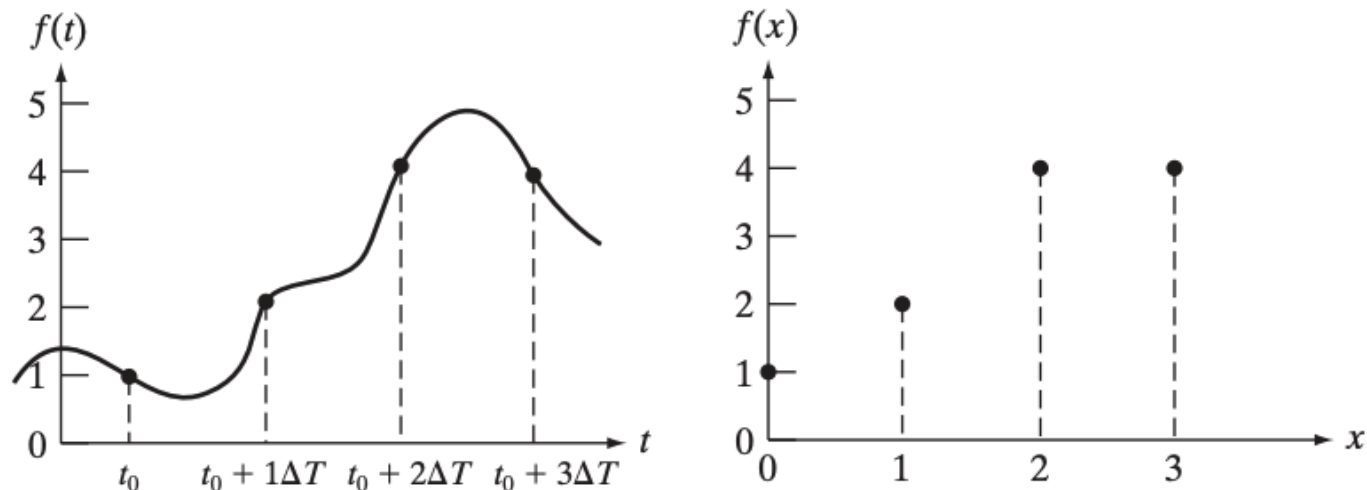
$$\Delta u = \frac{1}{M \Delta x}$$

1-D DFT example

a b

FIGURE 4.11

(a) A function, and (b) samples in the x -domain. In (a), t is a continuous variable; in (b), x represents integer values.



$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{-j \frac{2\pi mn}{M}}, m = 0, 1, \dots, (M-1)$$

$$F[1] = \sum_{n=0}^3 f[n] e^{-j 2\pi \frac{1n}{4}} = -3 + 2j$$

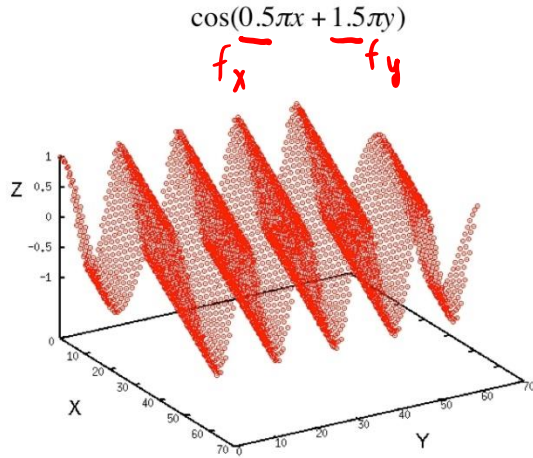
$$F[0] = 1 + 2 + 4 + 4$$

$$\begin{aligned} 0 &\rightarrow e^0 \rightarrow 1 \times 1 \\ 1 &\rightarrow e^{-j 2\pi/4} \rightarrow -j \times 2 \\ 2 &\rightarrow e^{-j 4\pi/4} \rightarrow -1 \times 4 \\ 3 &\rightarrow e^{-j 6\pi/4} \rightarrow +j \times 4 \end{aligned}$$

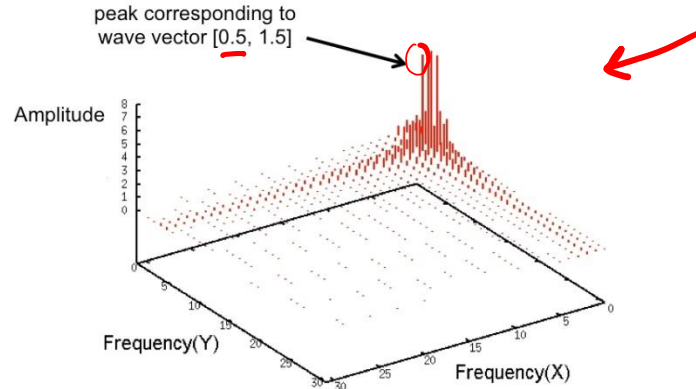
2D DFT and IDFT

$$\underline{F[m,n]} = \sum_{y=0}^{y=(N-1)} \sum_{x=0}^{x=(M-1)} \underline{f[x,y]} \underbrace{e^{-2\pi j\left(\frac{mx}{M} + \frac{ny}{N}\right)}}_{\substack{\downarrow \quad \downarrow \\ f_x \quad f_y}}$$

$$F[3,3] =$$



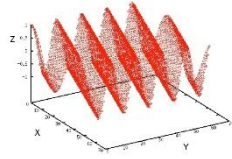
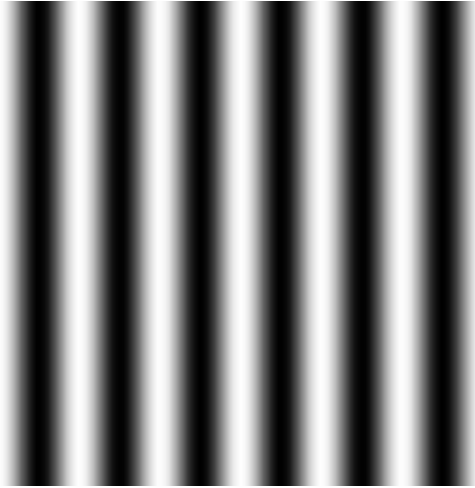
$$\text{FFT}[\cos(0.5\pi x + 1.5\pi y)]$$



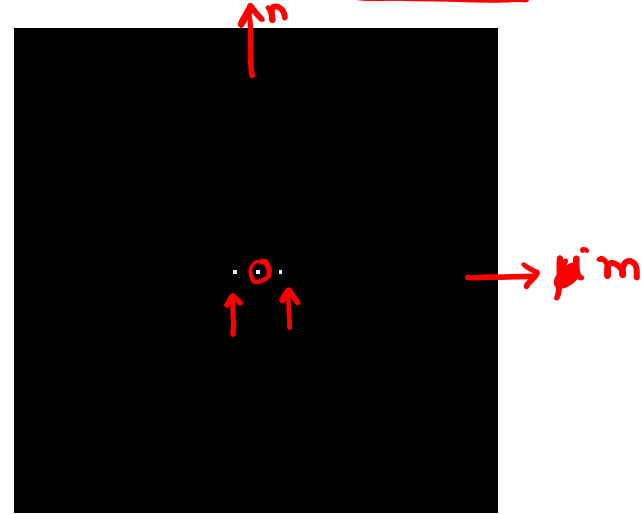
$$\underline{f[x,y]} = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} \underline{F[m,n]} e^{2\pi j\left(\frac{mx}{M} + \frac{ny}{N}\right)}$$

DFT for simple spatial patterns

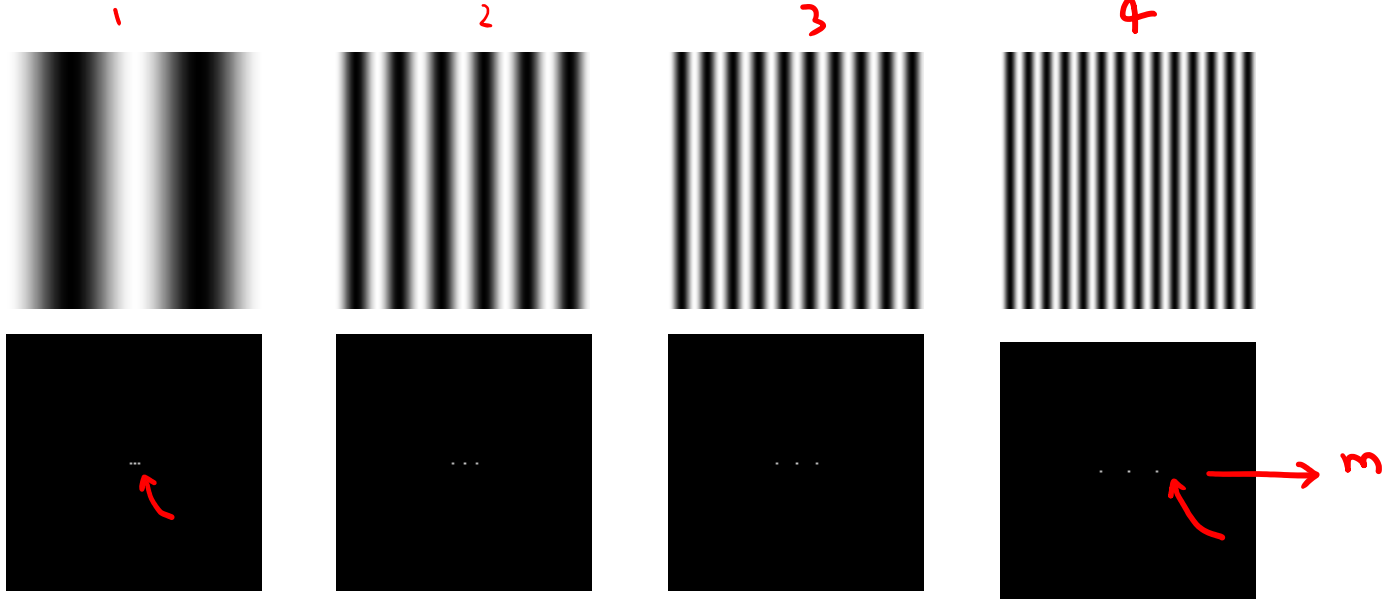
Brightness Image



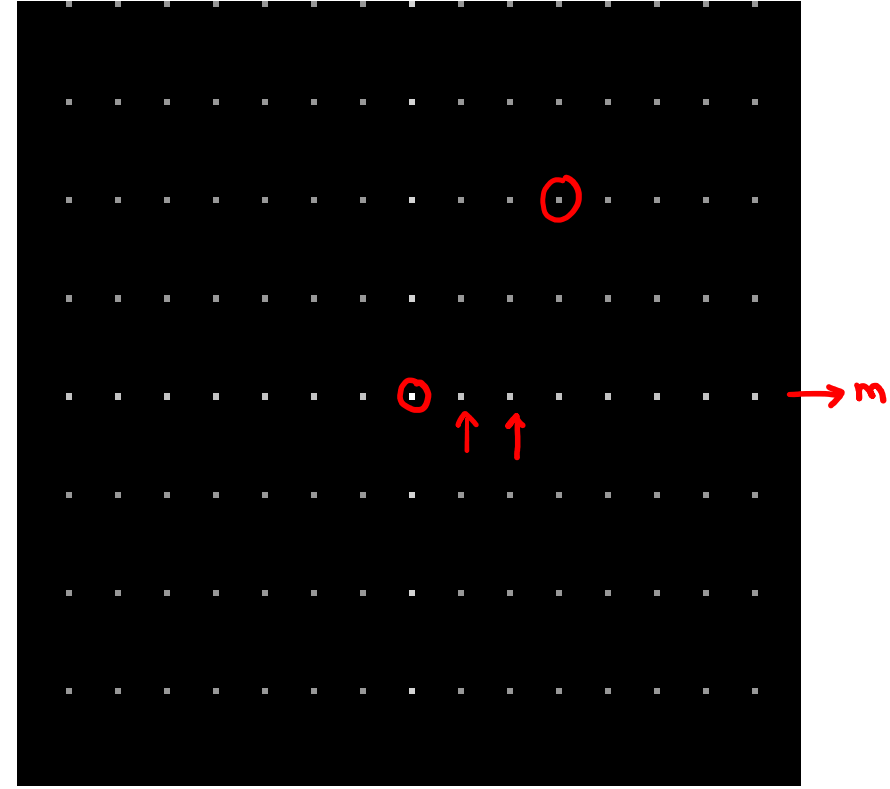
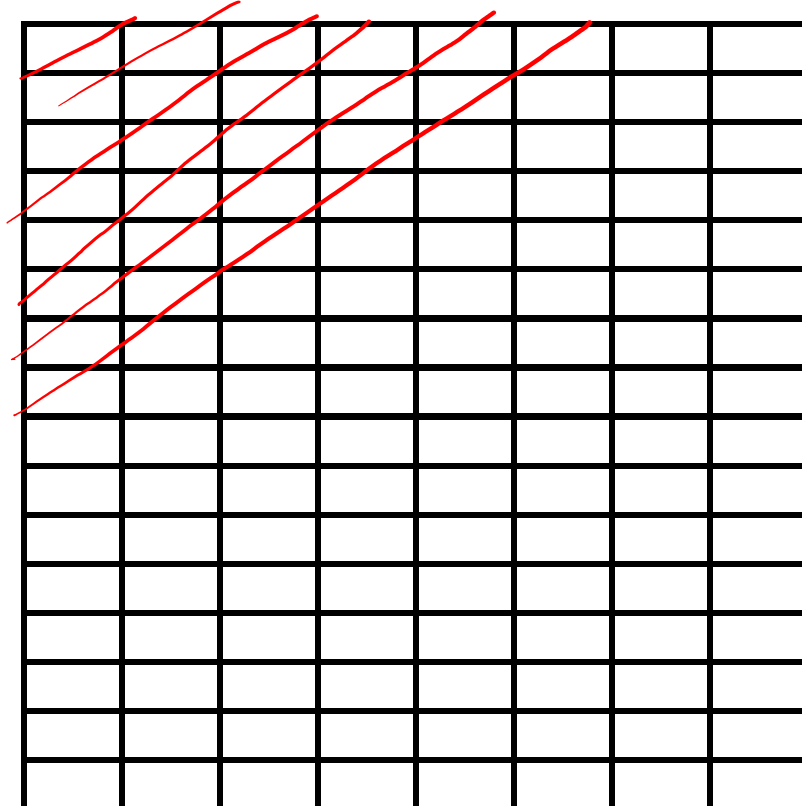
$|F(m,n)|$
Fourier transform spectrum



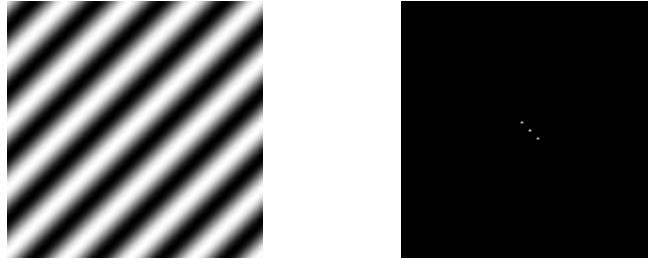
DFT Example



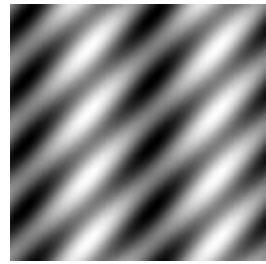
Example



DFT Example (Rotation)



DFT Example (Sum of Signals)



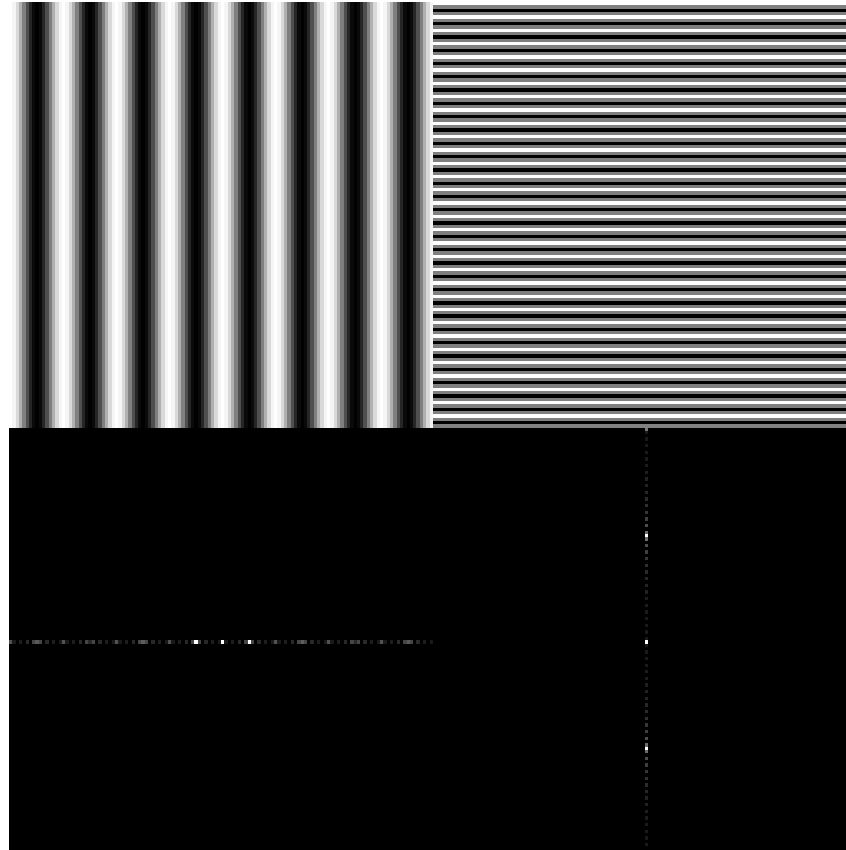
$D + X$



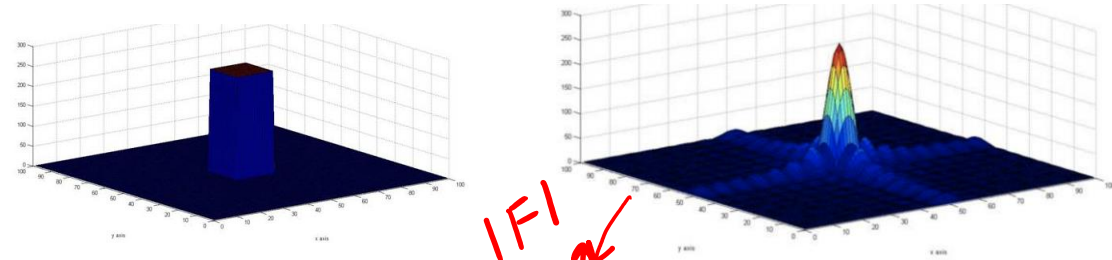
$F(f_1) + F(f_2)$



DFT for simple 'spatial' patterns



DFT Example (Rect, Log Transformation)



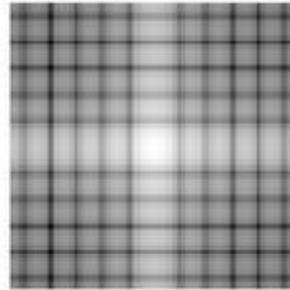
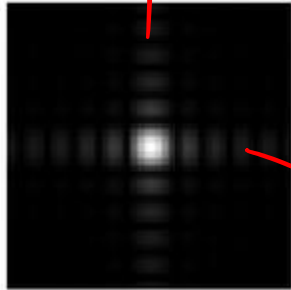
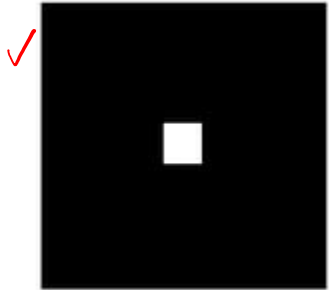
$|F|$

$$\frac{10^6 \text{ to } 10^{-2} : V}{\log(1+V)}$$

$$\log(1+V)$$

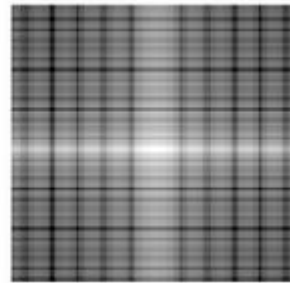
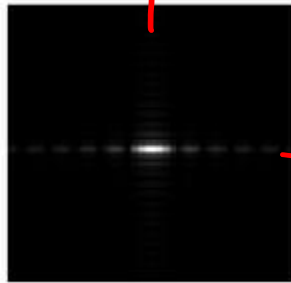
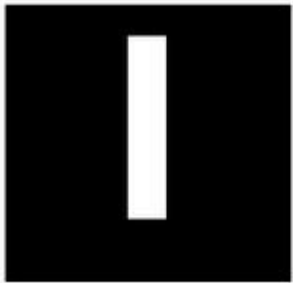
m, n

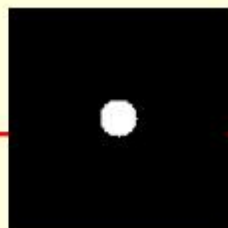
$$0 \dots \frac{m-1}{2-1}$$



$\frac{1}{2}$

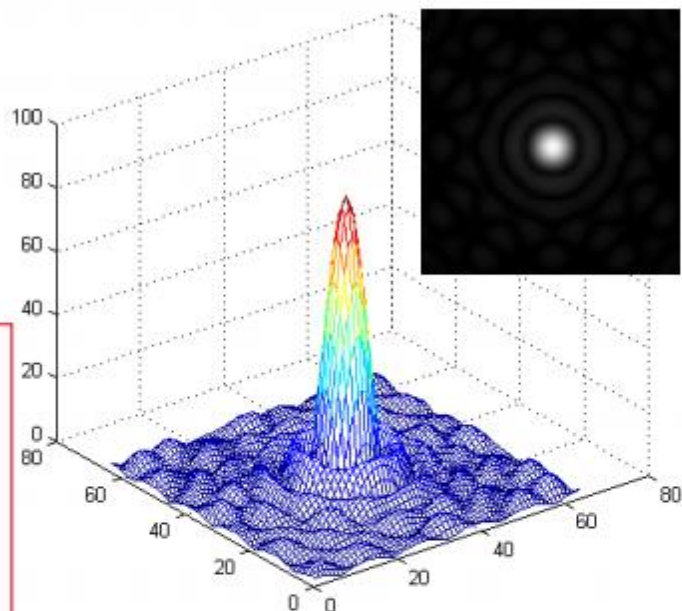
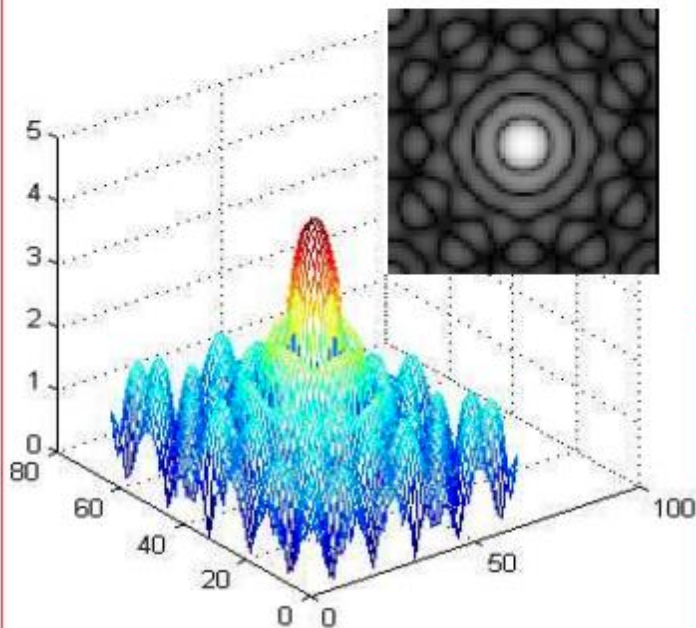
$\frac{1}{2}$





$f(x,y)$

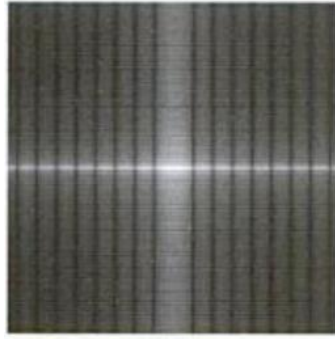
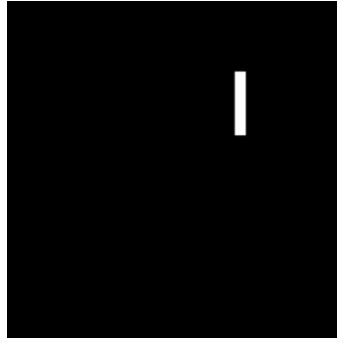
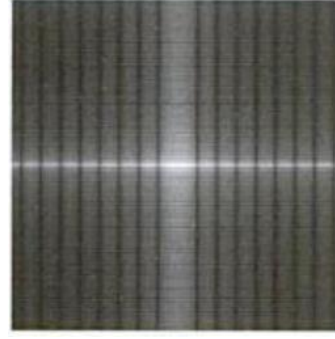
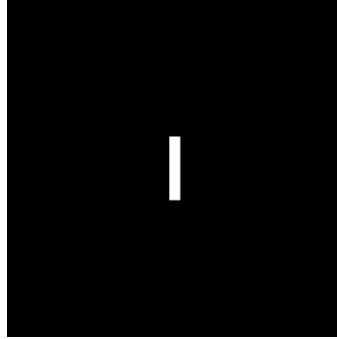
(64x64)



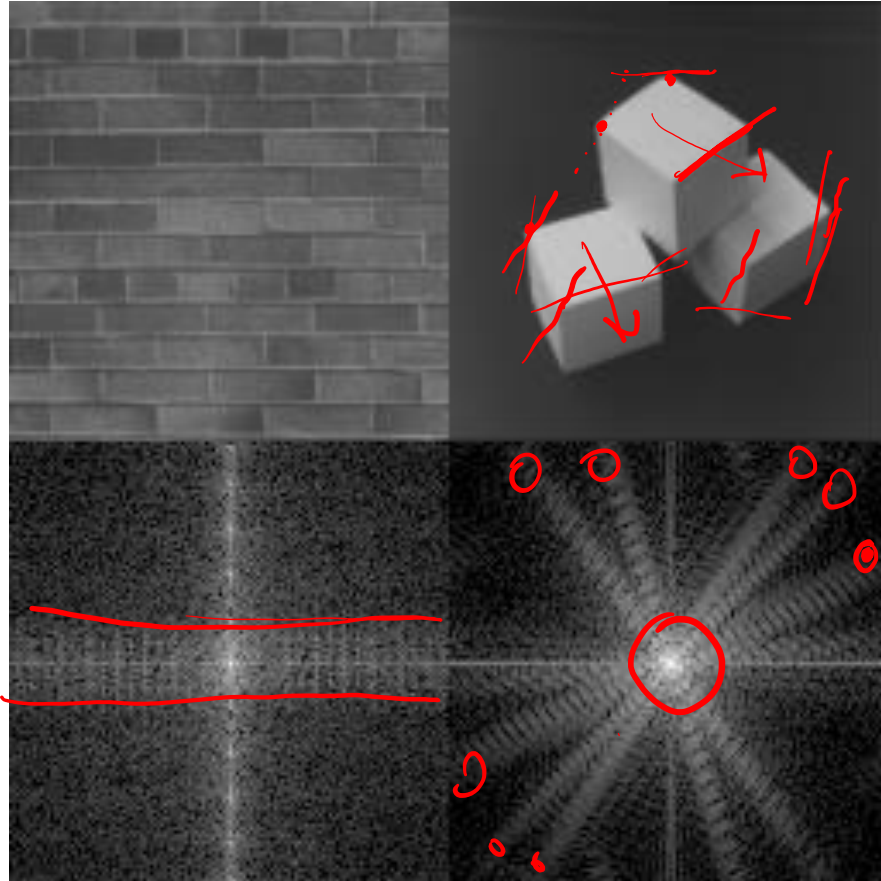
$|F(u,v)|$

$\log(1+|F(u,v)|)$

DFT Example (Translation - Magnitude)



Some examples of images and spectra

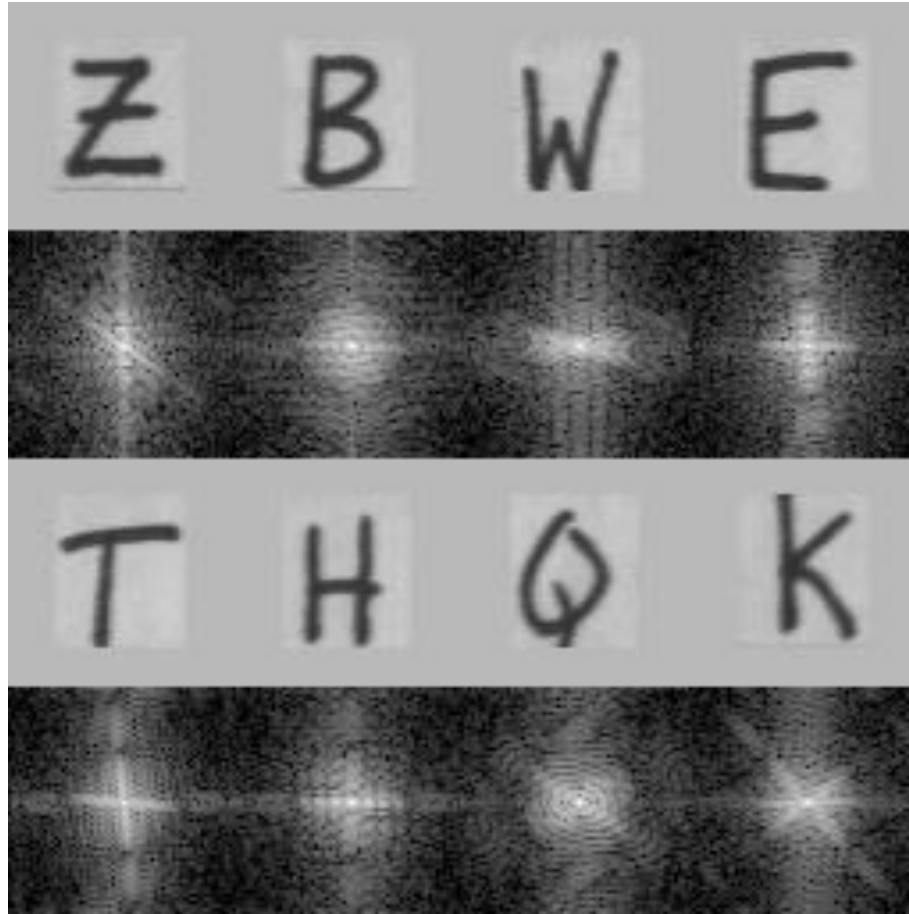


Handwritten red annotations and a formula:

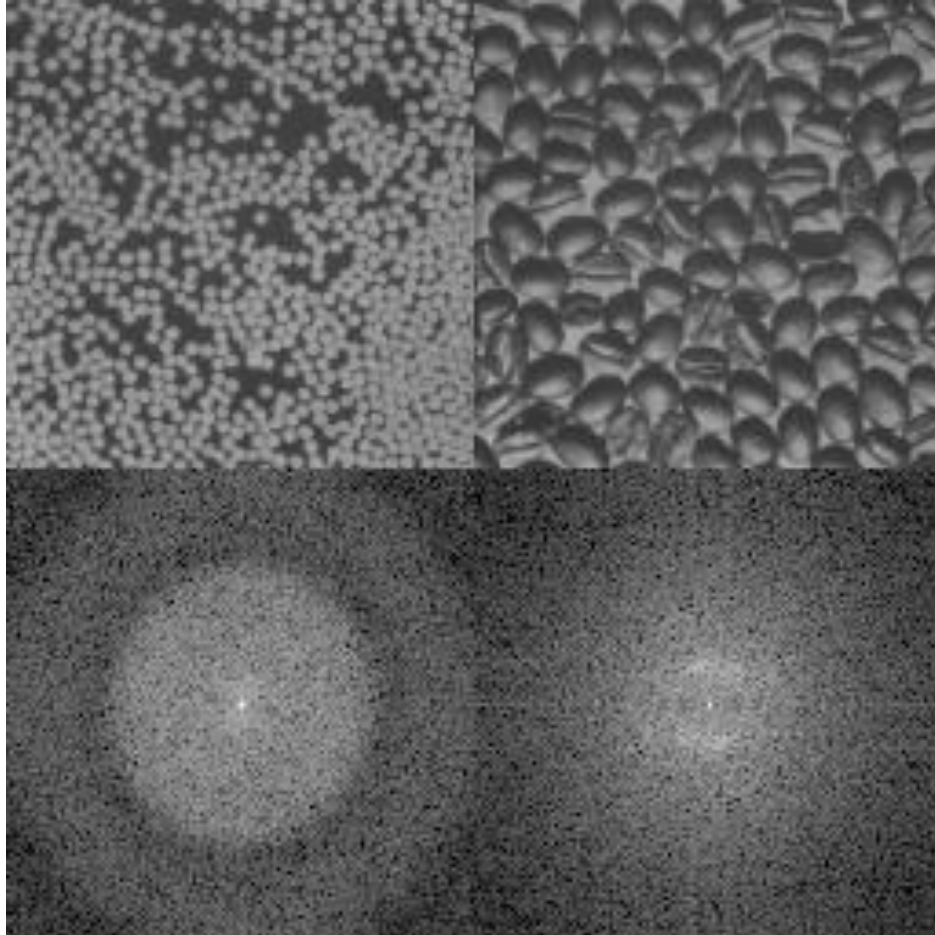
$$F[m,n] = \frac{1}{f} \int \int f(x,y) e^{-j2\pi(mx+ny)} dx dy$$

The formula is written in red ink, with the original text partially obscured by the handwritten annotations. The annotations include a wavy line with an arrow pointing right, and a curly brace with an arrow pointing up and right.

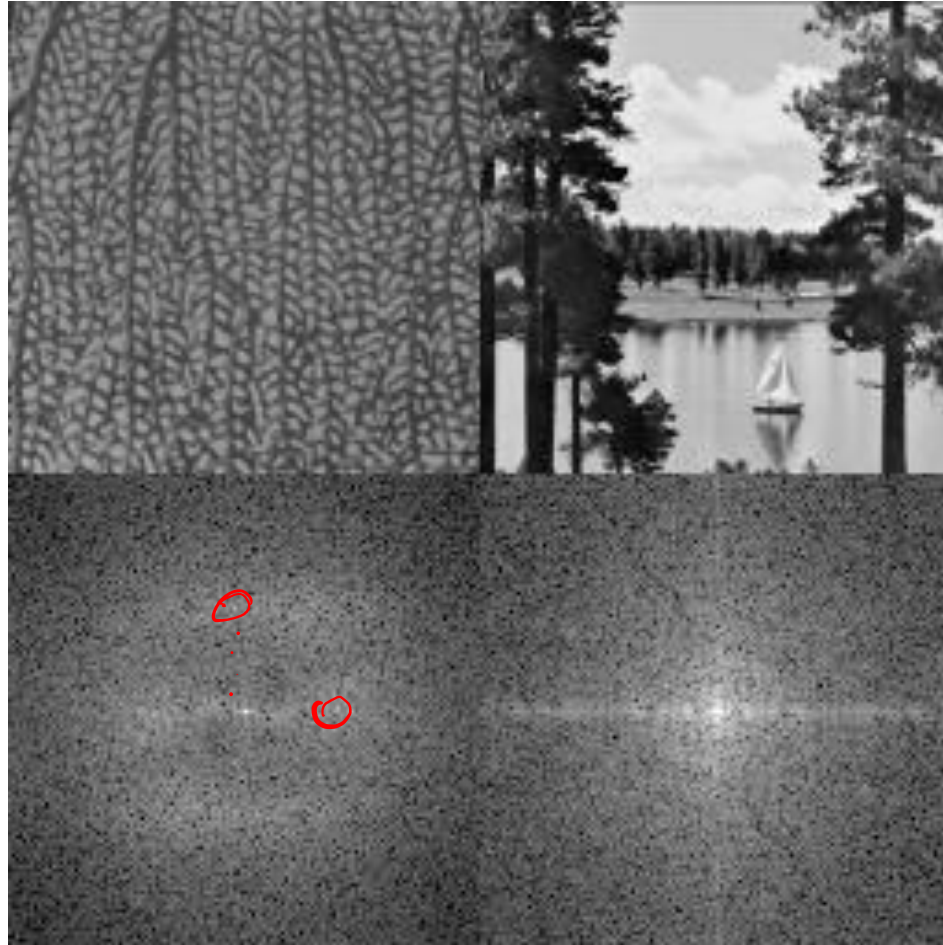
Some examples of images and spectra



Some examples of images and spectra



Some examples of images and spectra



Important Terms

- Magnitude spectrum

$$|F(\omega)| = \left[R^2(\omega) + I^2(\omega) \right]^{1/2}$$

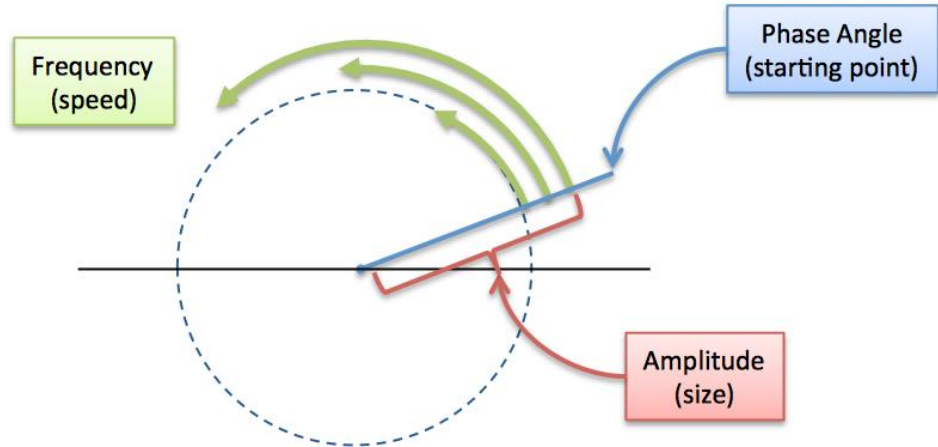
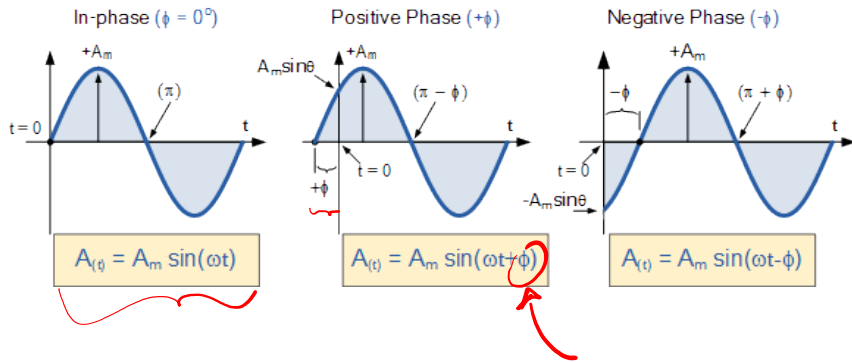
- Phase Spectrum

$$\phi(\omega) = \tan^{-1} \left[\frac{I(\omega)}{R(\omega)} \right]$$

- Power Spectrum

$$P(\omega) = |F(\omega)|^2$$

Phase



Magnitude and Phase Spectra



Figure 4a
Original

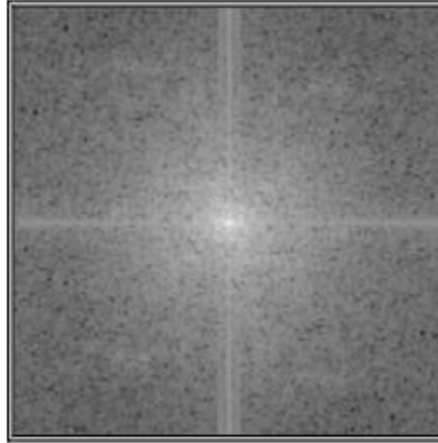


Figure 4b
 $\log(|A(\Omega, \Psi)|)$

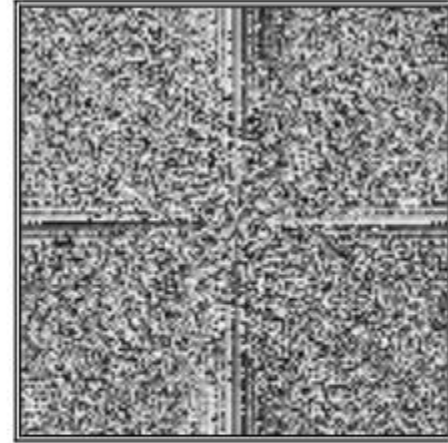


Figure 4c
 $\phi(\Omega, \Psi)$

$$\tan^{-1}\left(\frac{\text{Im}}{\text{Re}}\right)$$

We generally do not display PHASE images because most people who see them shortly thereafter succumb to hallucinogenics or end up in a Tibetan monastery – John Brayer

Magnitude and Phase Spectra

Both matter for reconstruction



Figure 4a
Original

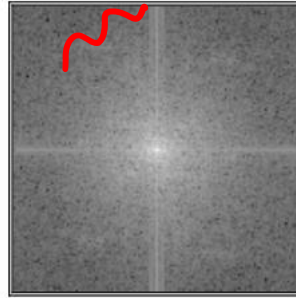


Figure 4b
 $\log(|A(\Omega, \Psi)|)$

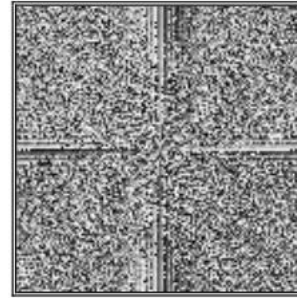


Figure 4c
 $\phi(\Omega, \Psi)$

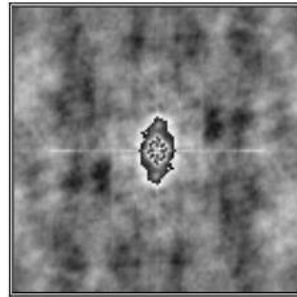


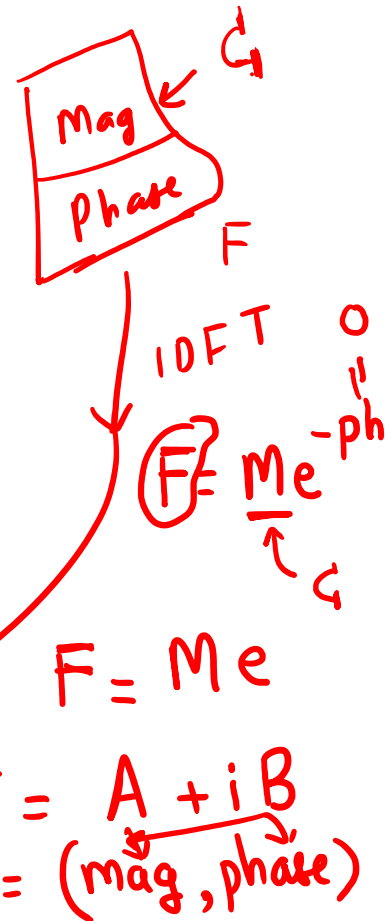
Figure 5a

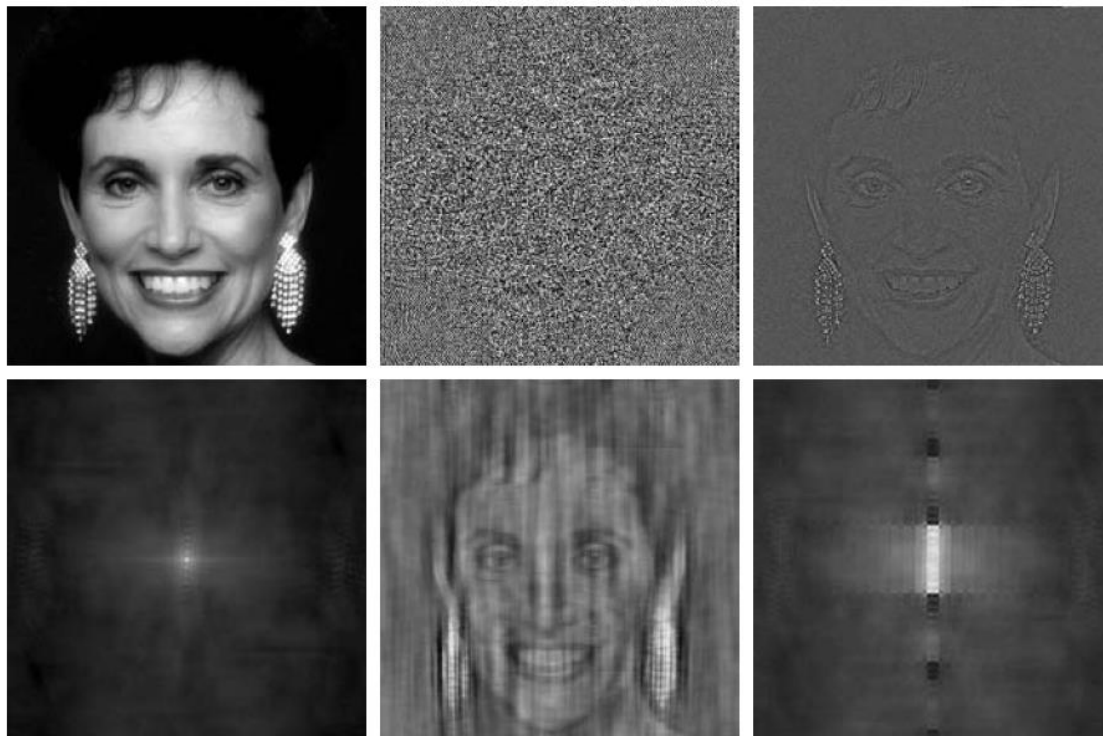
$\phi(\Omega, \Psi) = 0$



Figure 5b

$|A(\Omega, \Psi)| = \text{constant}$



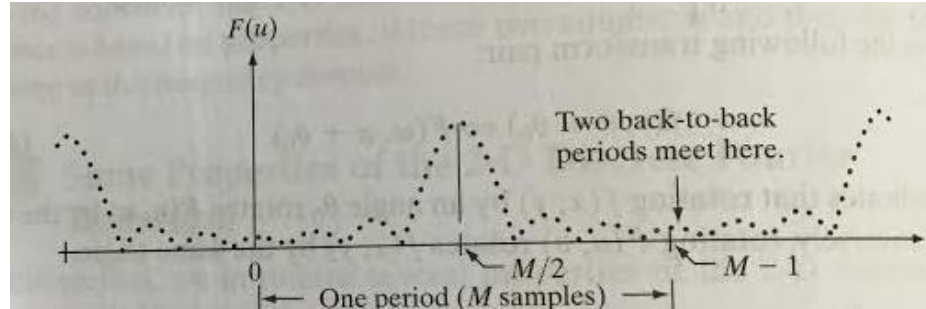
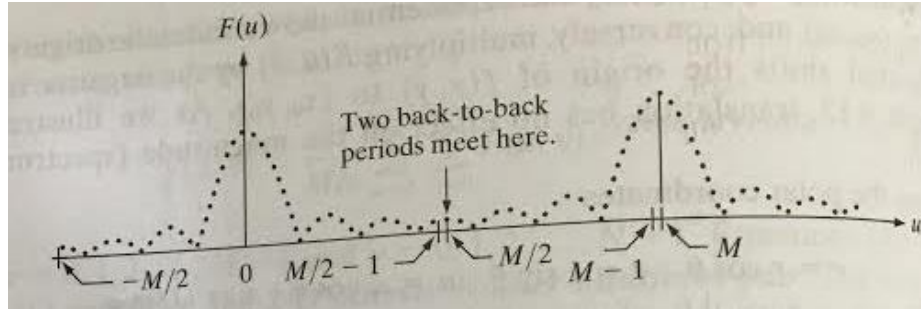


a	b	c
d	e	f

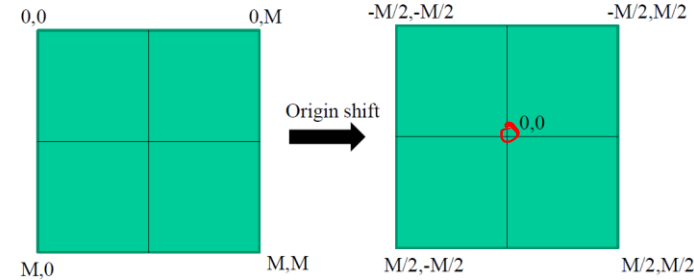
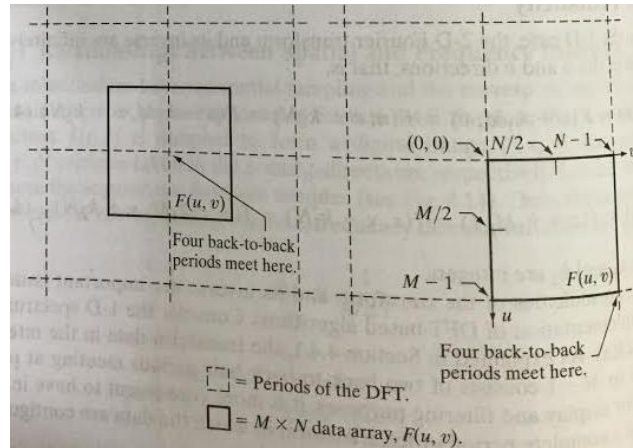
FIGURE 4.27 (a) Woman. (b) Phase angle. (c) Woman reconstructed using only the phase angle. (d) Woman reconstructed using only the spectrum. (e) Reconstruction using the phase angle corresponding to the woman and the spectrum corresponding to the rectangle in Fig. 4.24(a). (f) Reconstruction using the phase of the rectangle and the spectrum of the woman.

Shifting origin

1-D

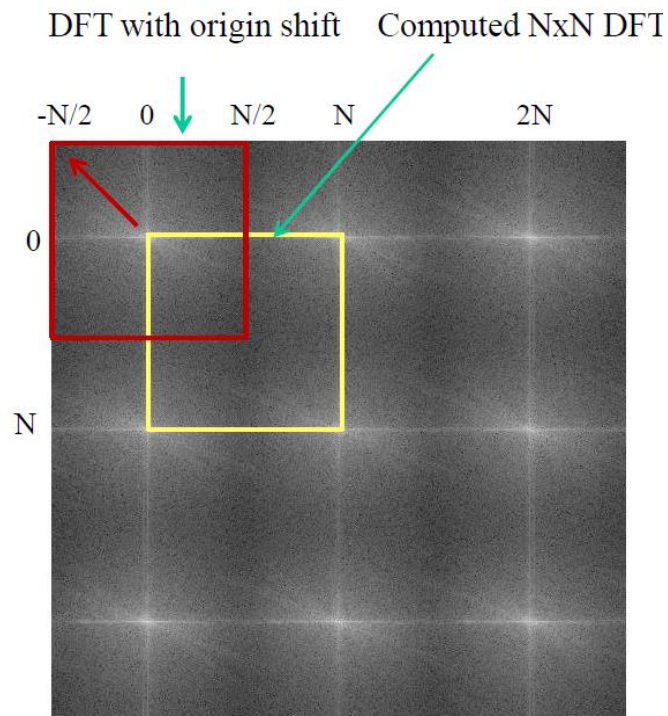
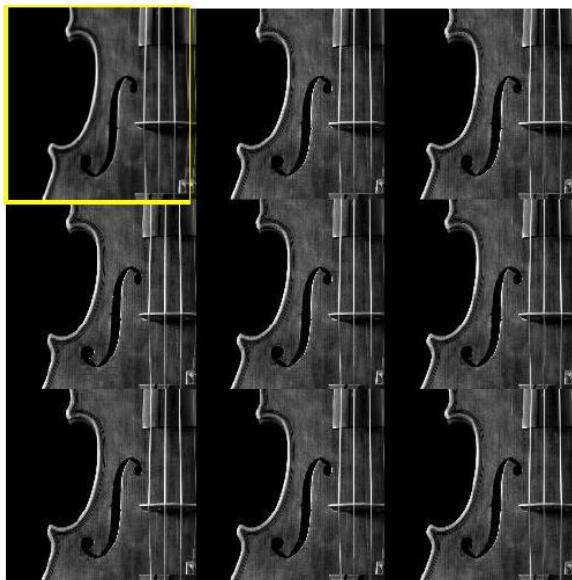


2-D

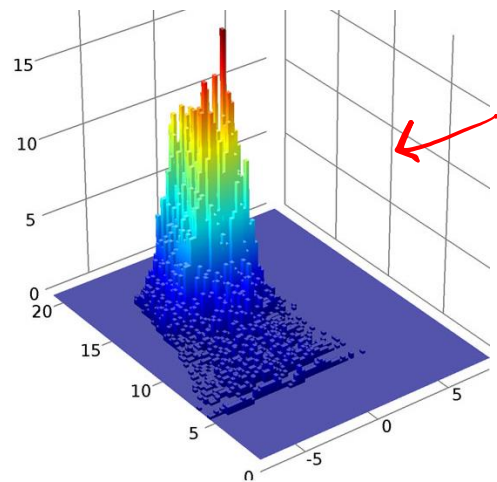


$$\underline{f[x, y]e^{j2\pi(\frac{u_0x}{M} + \frac{v_0y}{M})}} \leftrightarrow F(u - u_0, v - v_0)$$

Shifting origin







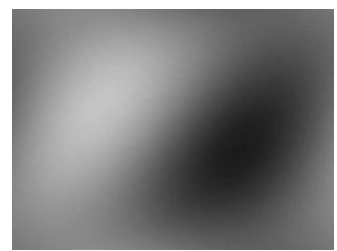
$$F[m,n] = \sum_{y=0}^{y=(N-1)} \sum_{x=0}^{x=(M-1)} f[x,y] e^{-2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$

$$f[x,y] = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} F[m,n] e^{2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$

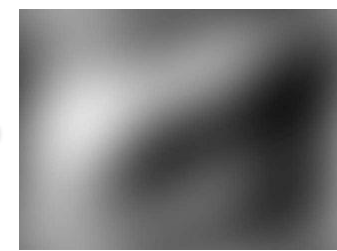
$$f[x,y] = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} F[m,n] e^{2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$



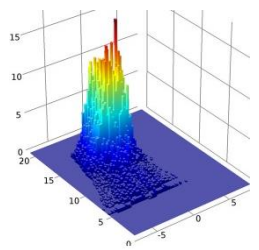
$F[0,0]$



$F[1,0]$



From 50% of the lowest frequencies



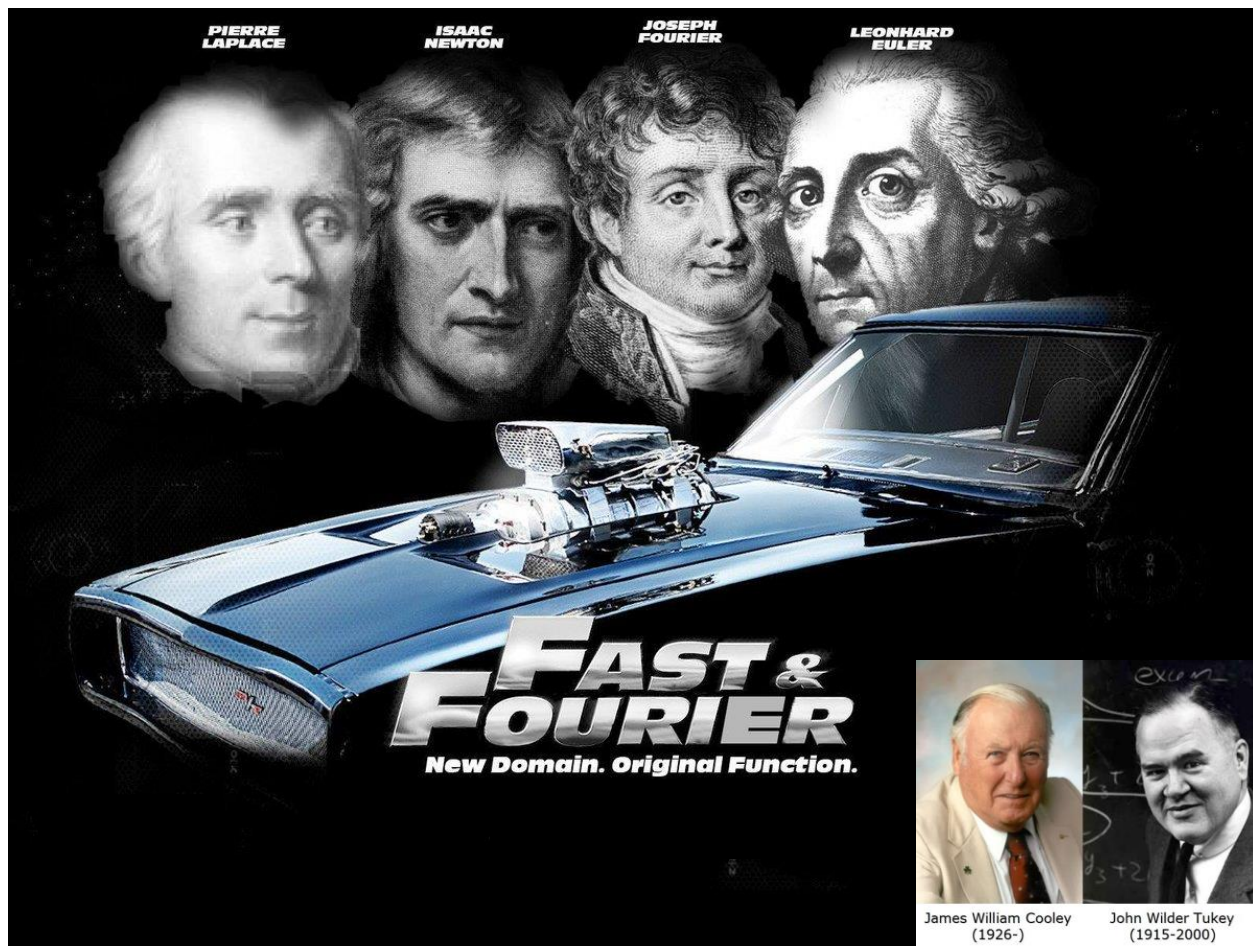
Adding up to 50% lowest frequencies



(Some) Properties of FT

	Name:	Condition:	Property:
1	Amplitude scaling	$f(t) \leftrightarrow F(\omega)$, constant K	$Kf(t) \leftrightarrow KF(\omega)$
2	Addition	$f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$, \dots	$f(t) + g(t) + \dots \leftrightarrow F(\omega) + G(\omega) + \dots$
3	Hermitian	Real $f(t) \leftrightarrow F(\omega)$	$F(-\omega) = F^*(\omega)$
4	Even	Real and even $f(t)$	Real and even $F(\omega)$
5	Odd	Real and odd $f(t)$	Imaginary and odd $F(\omega)$
6	Symmetry	$f(t) \leftrightarrow F(\omega)$	$F(t) \leftrightarrow 2\pi f(-\omega)$
7	Time scaling	$f(t) \leftrightarrow F(\omega)$, real s	$f(st) \leftrightarrow \frac{1}{ s } F(\frac{\omega}{s})$
8	Time shift	$f(t) \leftrightarrow F(\omega)$	$f(t - t_o) \leftrightarrow F(\omega)e^{-j\omega t_o}$
9	Frequency shift	$f(t) \leftrightarrow F(\omega)$	$f(t)e^{j\omega_o t} \leftrightarrow F(\omega - \omega_o)$
10	Modulation	$f(t) \leftrightarrow F(\omega)$	$f(t) \cos(\omega_o t) \leftrightarrow \frac{1}{2}F(\omega - \omega_o) + \frac{1}{2}F(\omega + \omega_o)$
11	Time derivative	Differentiable $f(t) \leftrightarrow F(\omega)$	$\frac{df}{dt} \leftrightarrow j\omega F(\omega)$
12	Freq derivative	$f(t) \leftrightarrow F(\omega)$	$-jtf(t) \leftrightarrow \frac{d}{d\omega} F(\omega)$
13	Time convolution	$f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$	$f(t) * g(t) \leftrightarrow F(\omega)G(\omega)$
14	Freq convolution	$f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$	$f(t)g(t) \leftrightarrow \frac{1}{2\pi} F(\omega) * G(\omega)$
15	Compact form	Real $f(t)$	$f(t) = \frac{1}{2\pi} \int_0^\infty 2 F(\omega) \cos(\omega t + \angle F(\omega)) d\omega$
16	Parseval, Energy W	$f(t) \leftrightarrow F(\omega)$	$W \equiv \int_{-\infty}^\infty f(t) ^2 dt = \frac{1}{2\pi} \int_{-\infty}^\infty F(\omega) ^2 d\omega$

- DFT has an efficient version called FFT (Fast Fourier Transform)



DFT vs FFT computation times

$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$

$O(N^2)$

n	$N = 2^n$	N^2	$N \log N$
10	1 024	1 048 576	10 240
12	4 096	16 777 216	49 152
14	16 384	268 435 456	229 376
16	65 536	4 294 967 296	1 048 576

FFT(n , [a_0, a_1, \dots, a_{n-1}]):

if $n=1$: return a_0

$F_{\text{even}} = \text{FFT}(n/2, [a_0, a_2, \dots, a_{n-2}])$

$F_{\text{odd}} = \text{FFT}(n/2, [a_1, a_3, \dots, a_{n-1}])$

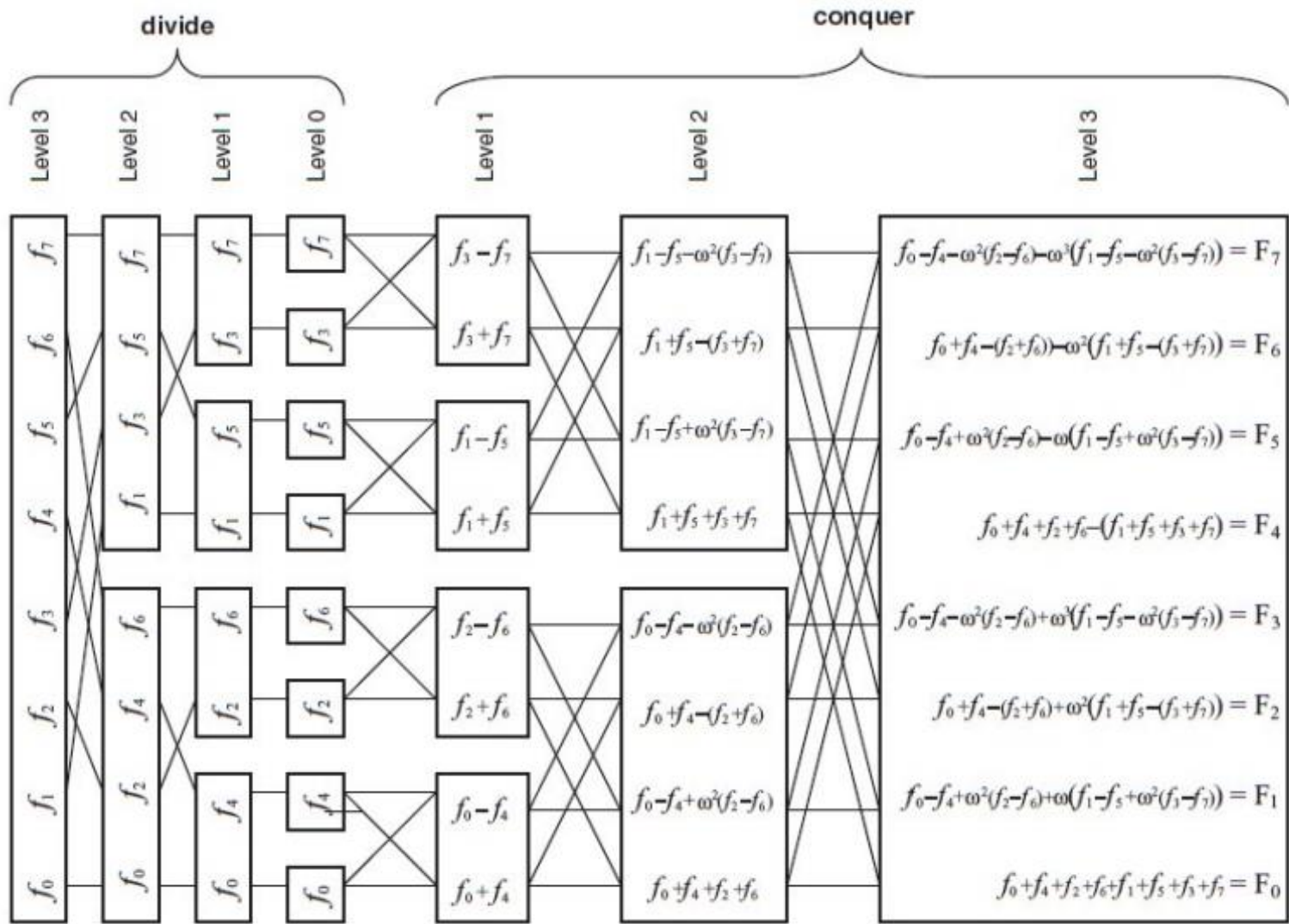
for $k = 0$ to $n/2 - 1$:

$\omega^k = e^{2\pi i k/n}$

$y^k = F_{\text{even } k} + \omega^k F_{\text{odd } k}$

$y^{k+n/2} = F_{\text{even } k} - \omega^k F_{\text{odd } k}$

return [y_0, y_1, \dots, y_{n-1}]



References

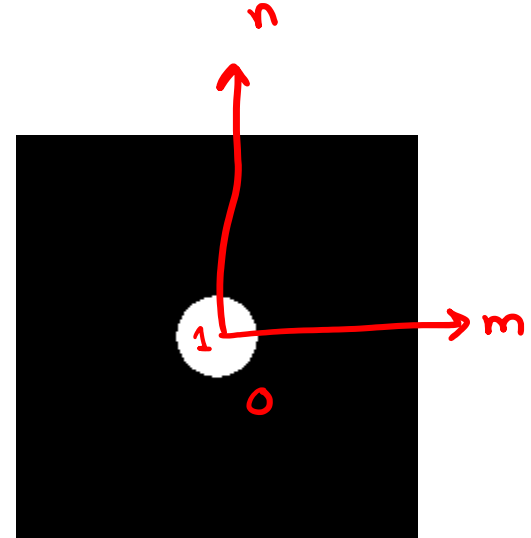
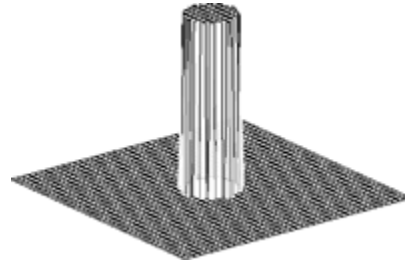
- <http://cns-alumni.bu.edu/~slehar/fourier/fourier.html>
- <https://slideplayer.com/slide/5665338/>
- <https://2e.mindsmachine.com/asf07.02.html>
- <https://radiologykey.com/a-walk-through-the-spatial-frequency-domain/>
- <https://blogs.mathworks.com/steve/2009/12/04/fourier-transform-visualization-using-windowing/>
- <https://photo.stackexchange.com/questions/40401/what-does-frequency-mean-in-an-image>
- <http://homepages.inf.ed.ac.uk/rbf/HIPR2/fourier.htm>
- <http://paulbourke.net/miscellaneous/imagefilter>
- <https://www.cs.unm.edu/~brayer/vision/fourier.html>

Image Enhancement and Filtering in Frequency Domain

Ideal Low Pass Filters

$$I \rightarrow \underbrace{F \cdot H}_{\text{Filter}} \rightarrow I$$

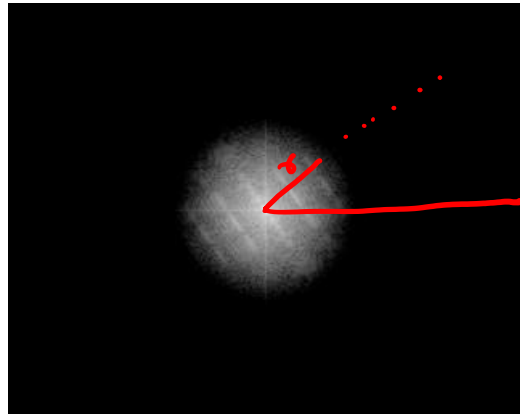
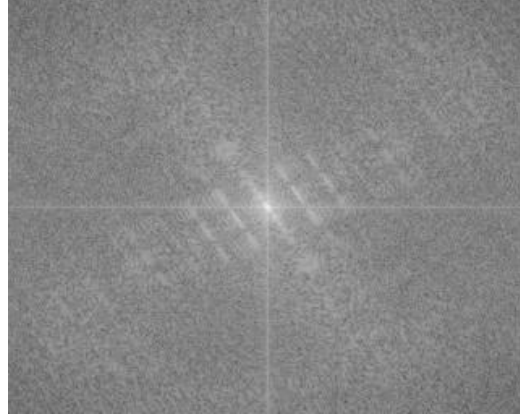
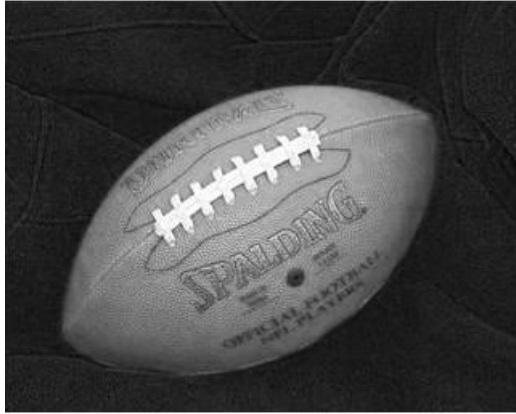
$$H(u, v) = \begin{cases} 1 & \text{if } D(u, v) \leq D_0 \\ 0 & \text{if } D(u, v) > D_0 \end{cases}$$



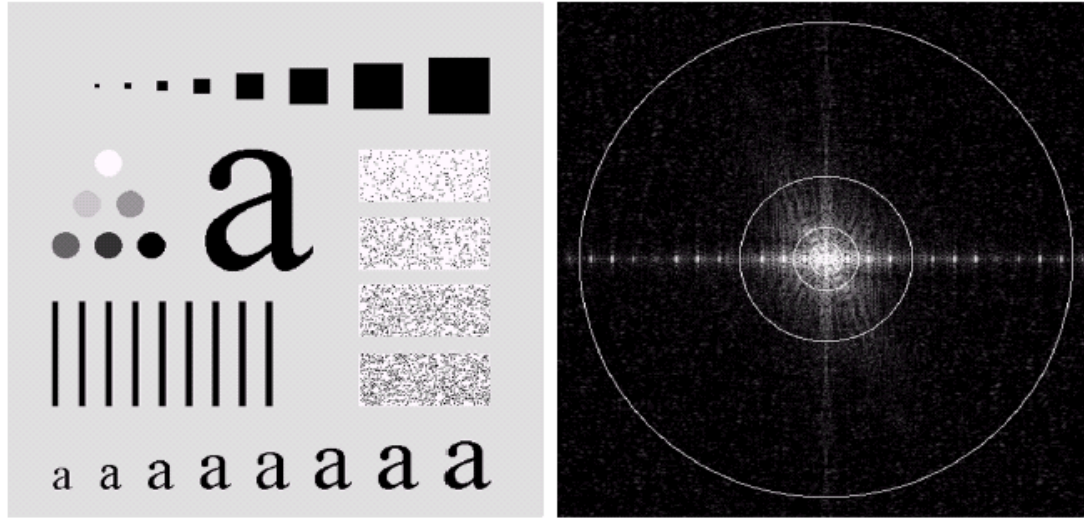
where $D(u, v) = [(u - M / 2)^2 + (v - N / 2)^2]^{1/2}$

$D_0 \rightarrow$ cut off frequency

Ideal Low Pass Filters

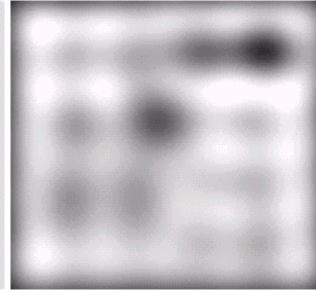
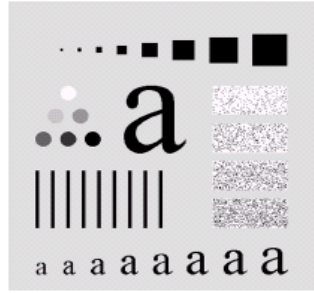


Ideal Low Pass Filters



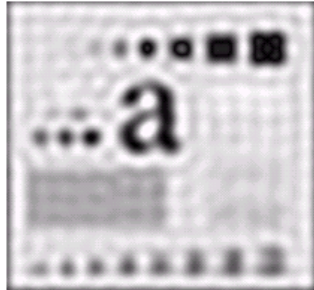
Radii 10,30,60,160 and 460 \rightarrow power 87, 93.1, 95.7, 97.8 and 99..2

Ideal Low Pass Filters



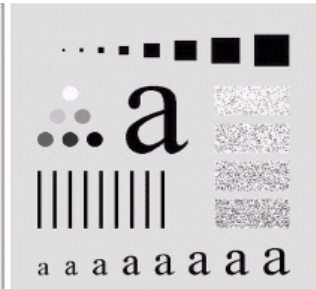
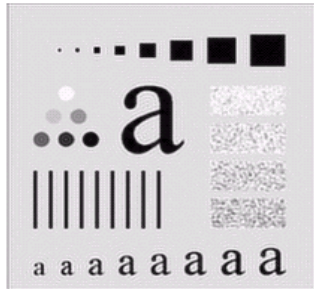
ILPF radius 10

ILPF radius 30



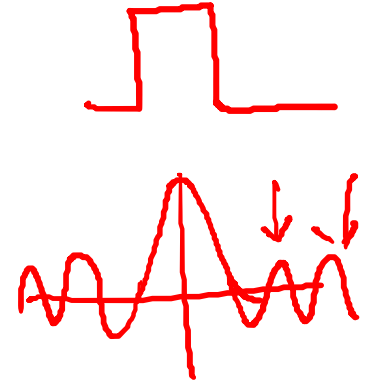
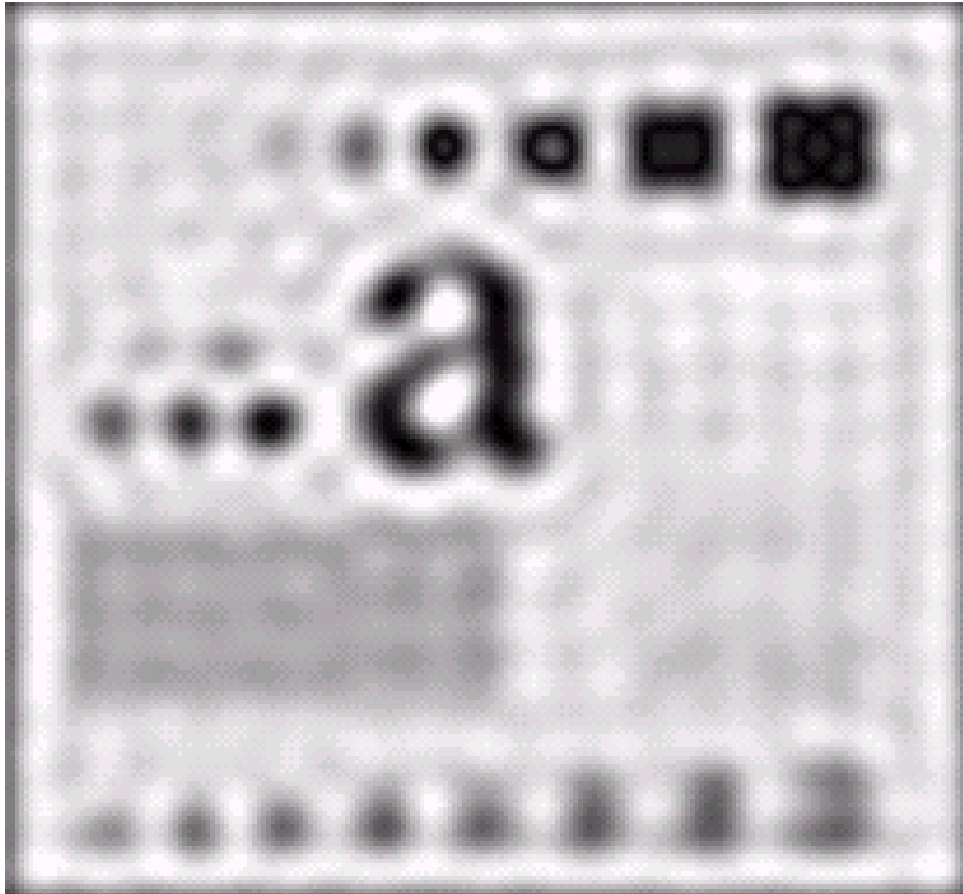
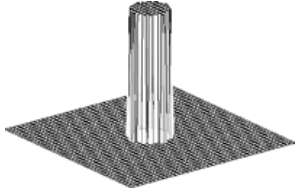
ILPF radius 60

ILPF radius 160



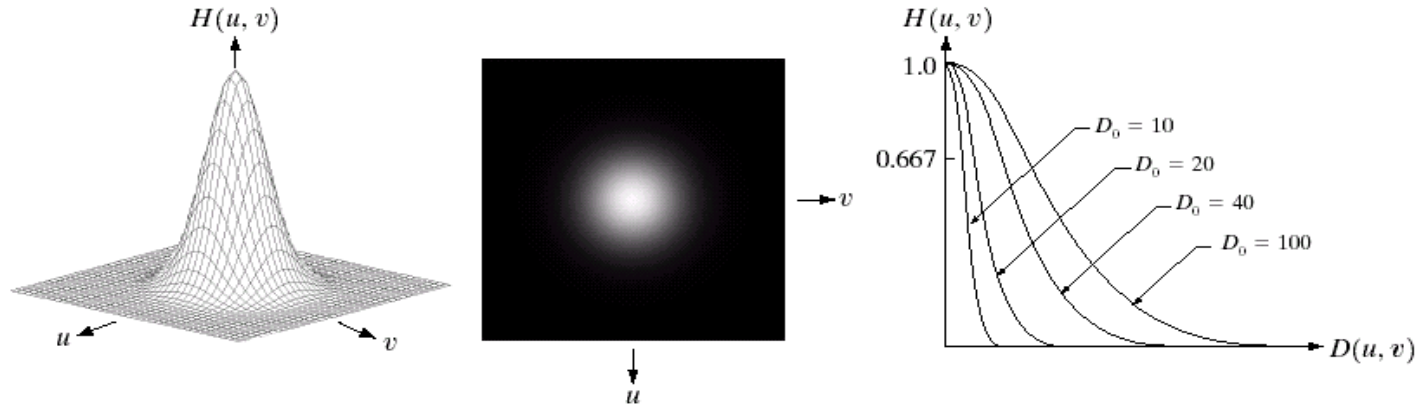
ILPF radius 460

Ideal Low Pass Filters



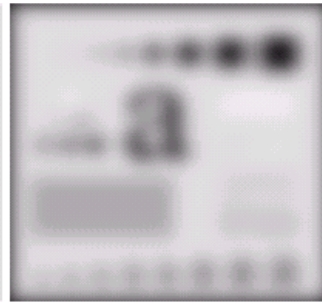
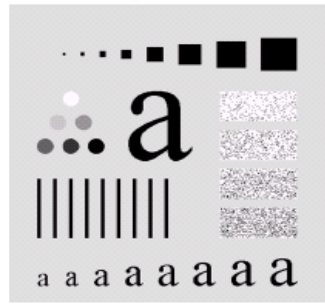
ILPF radius 30

Gaussian Low Pass Filters

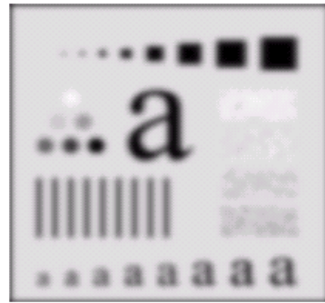


$$\underline{H(u, v) = e^{-D^2(u, v) / 2D_0^2}}$$

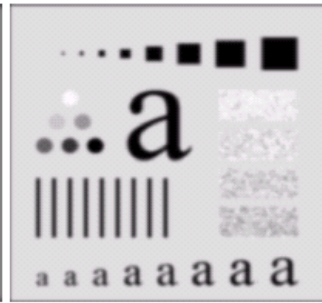
Gaussian Low Pass Filters (GLPF)



GLPF cut off
frequency 10



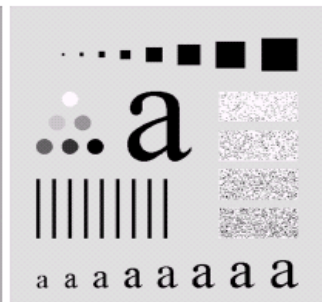
GLPF cut off
frequency 30



GLPF cut off
frequency 60

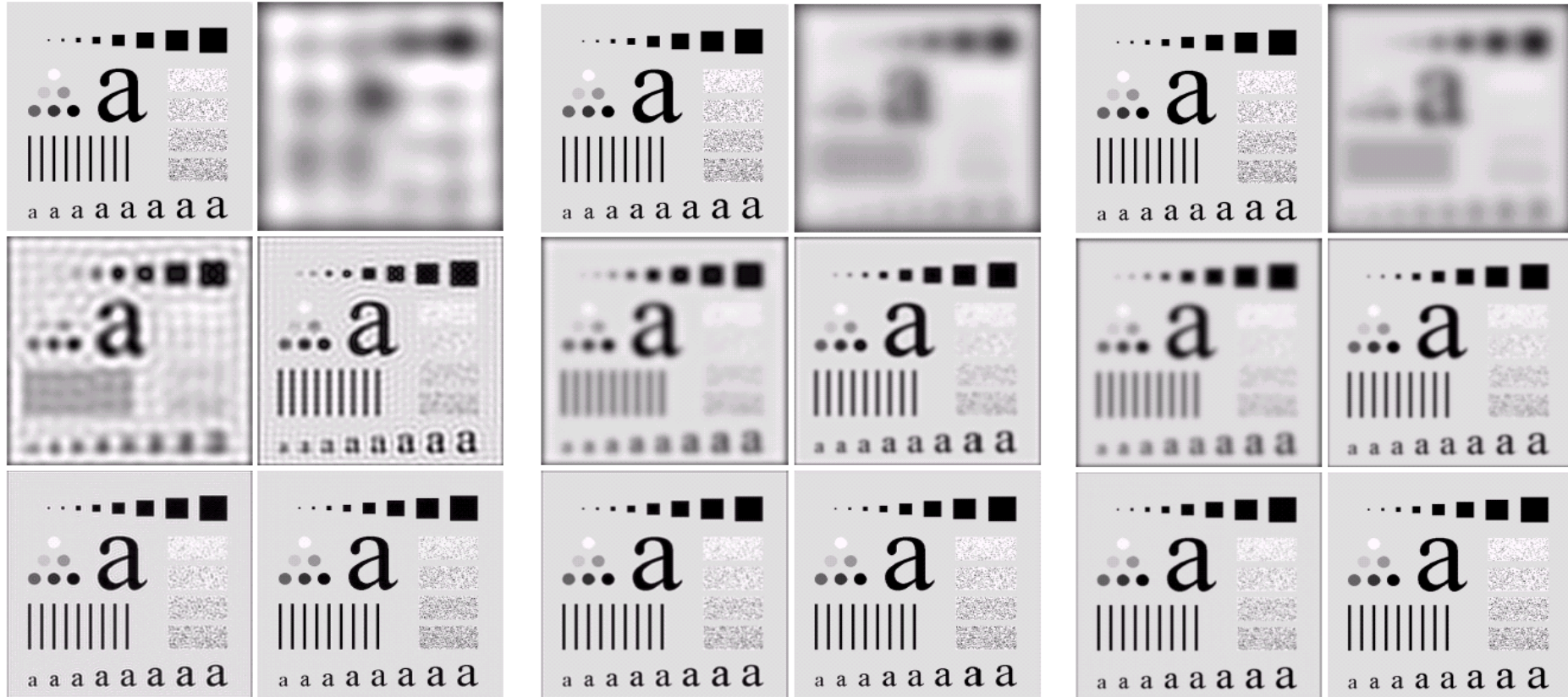


GLPF cut off
frequency 160



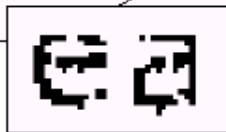
GLPF cut off
frequency 460

Comparison (ILPF, BLPF, GLPF)

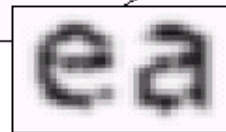


Low pass filtering application

Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.



Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.



References

- G & W (4.5.1, 4.5.2, 4.5.5, 4.6 – 4.11)
- http://mstrzel.eletel.p.lodz.pl/mstrzel/pattern_rec/fft_ang.pdf

Scribe List

2018101097
2018101099
2018101106
2018101110
2018102003
2018102005