

Laplace Transforms

- Important analytical method for solving *linear* ordinary differential equations.
 - Application to nonlinear ODEs? Must linearize first.
- Laplace transforms play a key role in important process control concepts and techniques.
 - Examples:
 - Transfer functions
 - Frequency response
 - Control system design
 - Stability analysis

Definition

The Laplace transform of a function, $f(t)$, is defined as

$$F(s) = \mathcal{L}[f(t)] = \int_0^{\infty} f(t) e^{-st} dt \quad (3-1)$$

where $F(s)$ is the symbol for the Laplace transform, \mathcal{L} is the Laplace transform operator, and $f(t)$ is some function of time, t .

Note: The \mathcal{L} operator transforms a time domain function $f(t)$ into an s domain function, $F(s)$. s is a *complex variable*:

$$s = a + bj, \quad j \doteq \sqrt{-1}$$

Inverse Laplace Transform, \mathcal{L}^{-1} :

By definition, the inverse Laplace transform operator, \mathcal{L}^{-1} , converts an s -domain function back to the corresponding time domain function:

$$f(t) = \mathcal{L}^{-1}[F(s)]$$

Important Properties:

Both \mathcal{L} and \mathcal{L}^{-1} are *linear operators*. Thus,

$$\begin{aligned}\mathcal{L}[ax(t) + by(t)] &= a\mathcal{L}[x(t)] + b\mathcal{L}[y(t)] \\ &= aX(s) + bY(s)\end{aligned}\quad (3-3)$$

where:

- $x(t)$ and $y(t)$ are arbitrary functions
- a and b are constants
- $X(s) \triangleq \mathcal{L}[x(t)]$ and $Y(s) \triangleq \mathcal{L}[y(t)]$

Similarly,

$$\mathcal{L}^{-1}[aX(s) + bY(s)] = ax(t) + by(t)$$

Laplace Transforms of Common Functions

1. Constant Function

Let $f(t) = a$ (a constant). Then from the definition of the Laplace transform in (3-1),

$$\mathcal{L}(a) = \int_0^{\infty} a e^{-st} dt = -\frac{a}{s} e^{-st} \bigg|_0^{\infty} = 0 - \left(-\frac{a}{s} \right) = \boxed{\frac{a}{s}} \quad (3-4)$$

2. Step Function

The unit step function is widely used in the analysis of process control problems. It is defined as:

$$S(t) \triangleq \begin{cases} 0 & \text{for } t < 0 \\ 1 & \text{for } t \geq 0 \end{cases} \quad (3-5)$$

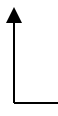
Because the step function is a special case of a “constant”, it follows from (3-4) that

$$\mathcal{L}[S(t)] = \frac{1}{s} \quad (3-6)$$

3. Derivatives

This is a very important transform because derivatives appear in the ODEs we wish to solve.

$$\mathcal{L}\left[\frac{df}{dt}\right] = sF(s) - f(0) \quad (3-9)$$


 initial condition at $t = 0$

Similarly, for higher order derivatives:

$$\begin{aligned} \mathcal{L}\left[\frac{d^n f}{dt^n}\right] = & s^n F(s) - s^{n-1} f(0) - s^{n-2} f^{(1)}(0) - \\ & - \dots - s f^{(n-2)}(0) - f^{(n-1)}(0) \end{aligned} \quad (3-14)$$

where:

- n is an arbitrary positive integer

- $f^{(k)}(0) \triangleq \left. \frac{d^k f}{dt^k} \right|_{t=0}$

Special Case: All Initial Conditions are Zero

Suppose $f(0) = f^{(1)}(0) = \dots = f^{(n-1)}(0)$. Then

$$\mathcal{L}\left[\frac{d^n f}{dt^n}\right] = s^n F(s)$$

In process control problems, we usually assume zero initial conditions. *Reason:* This corresponds to the nominal steady state when “deviation variables” are used, as shown in Ch. 4.

4. Exponential Functions

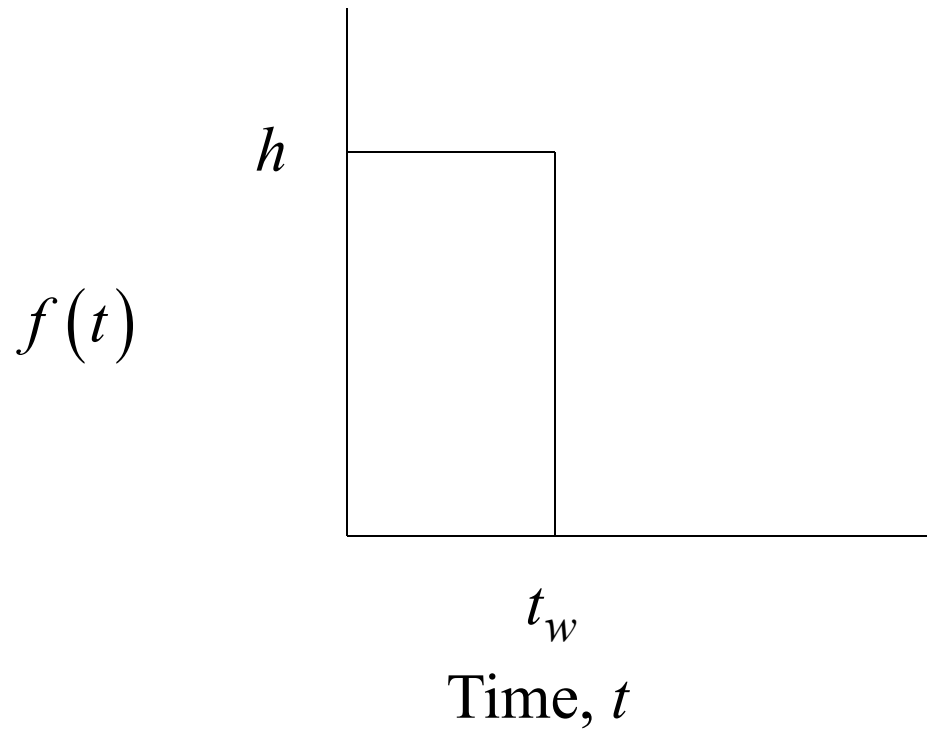
Consider $f(t) = e^{-bt}$ where $b > 0$. Then,

$$\begin{aligned}\mathcal{L}[e^{-bt}] &= \int_0^{\infty} e^{-bt} e^{-st} dt = \int_0^{\infty} e^{-(b+s)t} dt \\ &= \frac{1}{b+s} \left[-e^{-(b+s)t} \right]_0^{\infty} = \boxed{\frac{1}{s+b}}\end{aligned}\quad (3-16)$$

5. Rectangular Pulse Function

It is defined by:

$$f(t) = \begin{cases} 0 & \text{for } t < 0 \\ h & \text{for } 0 \leq t < t_w \\ 0 & \text{for } t \geq t_w \end{cases} \quad (3-20)$$



The Laplace transform of the rectangular pulse is given by

$$F(s) = \frac{h}{s} \left(1 - e^{-t_w s} \right) \quad (3-22)$$

6. Impulse Function (or Dirac Delta Function)

The impulse function is obtained by taking the limit of the rectangular pulse as its width, t_w , goes to zero but holding the area under the pulse constant at one. (i.e., let $h = \frac{1}{t_w}$)

Let, $\delta(t) \triangleq$ impulse function

Then, $\mathcal{L}[\delta(t)] = 1$

Some Commonly Used Laplace Transforms

$$\text{Unit Step} \Leftrightarrow 1/s$$

$$f(t - \theta) \Leftrightarrow F(s) e^{-\theta s}$$

$$t^n \Leftrightarrow \frac{n!}{s^{n+1}}$$

$$\frac{d f(t)}{dt} \Leftrightarrow s F(s) - f(0)$$

$$e^{-at} \Leftrightarrow \frac{1}{s + a}$$

$$\frac{d^2 f(t)}{dt^2} \Leftrightarrow s^2 F(s) - s f(0) - f'(0)$$

$$\sin(\omega t) \Leftrightarrow \frac{\omega}{s^2 + \omega^2}$$

$$e^{-at} \sin(\omega t) \Leftrightarrow \frac{\omega}{(s + a)^2 + \omega^2}$$

Table 3.1. Laplace Transforms

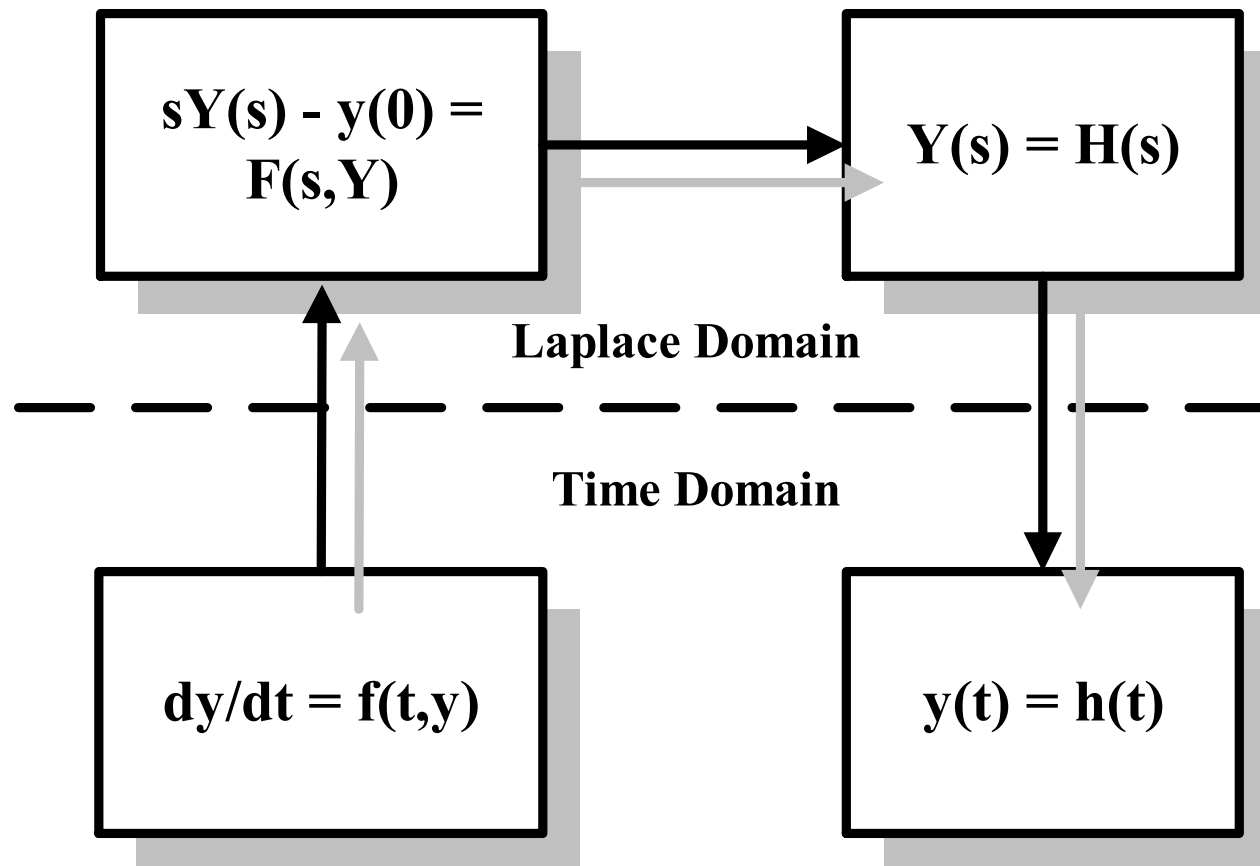
Table 3.1 Laplace Transforms for Various Time-Domain Functions*

| $f(t)$ | $F(s)$ |
|---|----------------------------|
| 1. $\delta(t)$ (unit impulse) | 1 |
| 2. $S(t)$ (unit step) | $\frac{1}{s}$ |
| 3. t (ramp) | $\frac{1}{s^2}$ |
| 4. t^{n-1} | $\frac{(n-1)!}{s^n}$ |
| 5. e^{-bt} | $\frac{1}{s+b}$ |
| 6. $\frac{1}{\tau}e^{-t/\tau}$ | $\frac{1}{\tau s + 1}$ |
| 7. $\frac{t^{n-1}e^{-bt}}{(n-1)!}$ ($n > 0$) | $\frac{1}{(s+b)^n}$ |
| 8. $\frac{1}{\tau^n(n-1)!}t^{n-1}e^{-t/\tau}$ | $\frac{1}{(\tau s + 1)^n}$ |
| 9. $\frac{1}{b_1 - b_2}(e^{-b_2t} - e^{-b_1t})$ | $\frac{1}{(s+b_1)(s+b_2)}$ |

Table 3.1. Laplace Transforms

| | |
|---|---|
| 10. $\frac{1}{\tau_1 - \tau_2} (e^{-t/\tau_1} - e^{-t/\tau_2})$ | $\frac{1}{(\tau_1 s + 1)(\tau_2 s + 1)}$ |
| 11. $\frac{b_3 - b_1}{b_2 - b_1} e^{-b_1 t} + \frac{b_3 - b_2}{b_1 - b_2} e^{-b_2 t}$ | $\frac{s + b_3}{(s + b_1)(s + b_2)}$ |
| 12. $\frac{1}{\tau_1} \frac{\tau_1 - \tau_3}{\tau_1 - \tau_2} e^{-t/\tau_1} + \frac{1}{\tau_2} \frac{\tau_2 - \tau_3}{\tau_2 - \tau_1} e^{-t/\tau_2}$ | $\frac{\tau_3 s + 1}{(\tau_1 s + 1)(\tau_2 s + 1)}$ |
| 13. $1 - e^{-t/\tau}$ | $\frac{1}{s(\tau s + 1)}$ |
| 14. $\sin \omega t$ | $\frac{\omega}{s^2 + \omega^2}$ |
| 15. $\cos \omega t$ | $\frac{s}{s^2 + \omega^2}$ |
| 16. $\sin(\omega t + \phi)$ | $\frac{\omega \cos \phi + s \sin \phi}{s^2 + \omega^2}$ |
| 17. $e^{-bt} \sin \omega t$ | $\left. \begin{array}{l} \frac{\omega}{(s + b)^2 + \omega^2} \\ \frac{s + b}{(s + b)^2 + \omega^2} \end{array} \right\} \quad b, \omega \text{ real}$ |
| 18. $e^{-bt} \sin \omega t$ | |
| 19. $\frac{1}{\tau \sqrt{1 - \zeta^2}} e^{-\zeta t/\tau} \sin(\sqrt{1 - \zeta^2} t/\tau)$ ($0 \leq \zeta < 1$) | $\frac{1}{\tau^2 s^2 + 2\zeta \tau s + 1}$ |
| 20. $1 + \frac{1}{\tau_2 - \tau_1} (\tau_1 e^{-t/\tau_1} - \tau_2 e^{-t/\tau_2})$ ($\tau_1 \neq \tau_2$) | $\frac{1}{s(\tau_1 s + 1)(\tau_2 s + 1)}$ |
| 21. $1 - \frac{1}{\tau \sqrt{1 - \zeta^2}} e^{-\zeta t/\tau} \sin[\sqrt{1 - \zeta^2} t/\tau + \psi]$ | $\frac{1}{\tau^2 s^2 + 2\zeta \tau s + 1}$ |

Method for Solving Linear ODE's using Laplace Transforms



Solution of ODEs by Laplace Transforms

Procedure:

1. Take the \mathcal{L} of both sides of the ODE.
2. Rearrange the resulting algebraic equation in the s domain to solve for the \mathcal{L} of the output variable, e.g., $Y(s)$.
3. Perform a partial fraction expansion.
4. Use the \mathcal{L}^{-1} to find $y(t)$ from the expression for $Y(s)$.

Example 3.1

Solve the ODE,

$$5\frac{dy}{dt} + 4y = 2 \qquad y(0) = 1 \qquad (3-26)$$

First, take \mathcal{L} of both sides of (3-26),

$$5(sY(s) - 1) + 4Y(s) = \frac{2}{s}$$

Rearrange,

$$Y(s) = \frac{5s + 2}{s(5s + 4)} \qquad (3-34)$$

Take \mathcal{L}^{-1} ,

$$y(t) = \mathcal{L}^{-1} \left[\frac{5s + 2}{s(5s + 4)} \right]$$

From Table 3.1,

$$\boxed{y(t) = 0.5 + 0.5e^{-0.8t}} \qquad (3-37)$$

Example of Solution of an ODE

$$\frac{d^2 y}{dt^2} + 6 \frac{dy}{dt} + 8y = 2 \quad y(0) = y'(0) = 0$$

ODE w/initial conditions

$$s^2 Y(s) + 6s Y(s) + 8Y(s) = 2/s$$

$$Y(s) = \frac{2}{s(s+2)(s+4)}$$

Apply Laplace transform to each term

$$Y(s) = \frac{1}{4s} + \frac{-1}{2(s+2)} + \frac{1}{4(s+4)}$$

Solve for Y(s)

$$y(t) = \frac{1}{4} - \frac{e^{-2t}}{2} + \frac{e^{-4t}}{4}$$

Apply partial fraction expansions w/Heaviside

Partial Fraction Expansions

Basic idea: Expand a complex expression for $Y(s)$ into simpler terms, each of which appears in the Laplace Transform table. Then you can take the \mathcal{L}^{-1} of both sides of the equation to obtain $y(t)$.

Example:

$$Y(s) = \frac{s+5}{(s+1)(s+4)} \quad (3-41)$$

Perform a partial fraction expansion (PFE)

$$\frac{s+5}{(s+1)(s+4)} = \frac{\alpha_1}{s+1} + \frac{\alpha_2}{s+4} \quad (3-42)$$

where coefficients α_1 and α_2 have to be determined.

To find α_1 : Multiply both sides by $s + 1$ and let $s = -1$

$$\therefore \alpha_1 = \left. \frac{s+5}{s+4} \right|_{s=-1} = \frac{4}{3}$$

To find α_2 : Multiply both sides by $s + 4$ and let $s = -4$

$$\therefore \alpha_2 = \left. \frac{s+5}{s+1} \right|_{s=-4} = -\frac{1}{3}$$

A General PFE

Consider a general expression,

$$Y(s) = \frac{N(s)}{D(s)} = \frac{N(s)}{\prod_{i=1}^n (s + b_i)} \quad (3-46a)$$

Here $D(s)$ is an n -th order polynomial with the roots ($s = -b_i$) all being *real* numbers which are *distinct* so there are no repeated roots.

The PFE is:

$$Y(s) = \frac{N(s)}{\prod_{i=1}^n (s + b_i)} = \sum_{i=1}^n \frac{\alpha_i}{s + b_i} \quad (3-46b)$$

Note: $D(s)$ is called the “characteristic polynomial”.

Special Situations:

Two other types of situations commonly occur when $D(s)$ has:

- i) Complex roots: e.g., $b_i = 3 \pm 4j$ ($j \triangleq \sqrt{-1}$)
- ii) Repeated roots (e.g., $b_1 = b_2 = -3$)

Example 3.2 (continued)

Recall that the ODE, $\ddot{y} + 6\dot{y} + 11y = 1$ with zero initial conditions resulted in the expression

$$Y(s) = \frac{1}{s(s^3 + 6s^2 + 11s + 6)} \quad (3-40)$$

The denominator can be factored as

$$s(s^3 + 6s^2 + 11s + 6) = s(s+1)(s+2)(s+3) \quad (3-50)$$

Note: Normally, numerical techniques are required in order to calculate the roots.

The PFE for (3-40) is

$$Y(s) = \frac{1}{s(s+1)(s+2)(s+3)} = \frac{\alpha_1}{s} + \frac{\alpha_2}{s+1} + \frac{\alpha_3}{s+2} + \frac{\alpha_4}{s+3} \quad (3-51)$$

Solve for coefficients to get

$$\alpha_1 = \frac{1}{6}, \quad \alpha_2 = -\frac{1}{2}, \quad \alpha_3 = \frac{1}{2}, \quad \alpha_4 = -\frac{1}{6}$$

(For example, find α , by multiplying both sides by s and then setting $s = 0$.)

Substitute numerical values into (3-51):

$$Y(s) = \frac{1/6}{s} - \frac{1/2}{s+1} + \frac{1/2}{s+2} + \frac{1/6}{s+3}$$

Take \mathcal{L}^{-1} of both sides:

$$\mathcal{L}^{-1}[Y(s)] = \mathcal{L}^{-1}\left[\frac{1/6}{s}\right] - \mathcal{L}^{-1}\left[\frac{1/2}{s+1}\right] + \mathcal{L}^{-1}\left[\frac{1/2}{s+2}\right] + \mathcal{L}^{-1}\left[\frac{1/6}{s+3}\right]$$

From Table 3.1,

$$y(t) = \frac{1}{6} - \frac{1}{2}e^{-t} + \frac{1}{2}e^{-2t} - \frac{1}{6}e^{-3t} \quad (3-52)$$

Important Properties of Laplace Transforms

1. *Final Value Theorem*

It can be used to find the steady-state value of a system (providing that a steady-state value exists.)

Statement of FVT:

$$\lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} [sY(s)]$$

providing that the limit exists (is finite) for all $\text{Re}(s) \geq 0$, where $\text{Re}(s)$ denotes the real part of complex variable, s .

2. *Initial Value Theorem*

Statement of IVT:

$$\lim_{t \rightarrow 0} y(t) = \lim_{s \rightarrow \infty} [sY(s)]$$

Important Properties of Laplace Transforms

$$\lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} [sY(s)]$$

$$\lim_{t \rightarrow 0} y(t) = \lim_{s \rightarrow \infty} [sY(s)]$$

Proof:

$$\mathcal{L}\left[\frac{df}{dt}\right] = sF(s) - f(0) \quad (3-9)$$

$$\int_0^{\infty} f'(t)e^{-st} dt = sF(s) - f(0)$$

when $s \rightarrow 0$

$$\int_0^{\infty} f'(t) dt = \lim_{s \rightarrow 0} sF(s) - f(0)$$

$$f(\infty) - f(0) = \lim_{s \rightarrow 0} sF(s) - f(0)$$

when $s \rightarrow \infty$

$$0 = \lim_{s \rightarrow \infty} sF(s) - f(0)$$

Example:

Suppose,

$$Y(s) = \frac{5s + 2}{s(5s + 4)} \quad (3-34)$$

Then,

$$y(\infty) = \lim_{t \rightarrow \infty} y(t) = \lim_{s \rightarrow 0} \left[\frac{5s + 2}{5s + 4} \right] = 0.5$$

3 . *Time Delay*

Time delays occur due to fluid flow, time required to do an analysis (e.g., gas chromatograph). The delayed signal can be represented as

$$y(t - \theta) \quad \theta = \text{time delay}$$

Also,

$$\mathcal{L}[y(t - \theta)] = e^{-\theta s} Y(s)$$