

# Dual-Beam DFRC Precoding for Joint Communications & Sensing

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## Abstract

This report presents a minimal, reproducible demonstrator of dual-function radar-communications (DFRC) beamforming on a uniform linear array (ULA). We co-design a transmit weight vector to maintain downlink communications SINR towards a user equipment (UE) while meeting a sensing mainlobe constraint towards a look-angle and suppressing sidelobes elsewhere. A ridge-regularized pattern-fitting approach is used to approximate constrained designs and to sweep sensing gain targets, yielding a Pareto curve of “UE SINR vs sensing gain”. The code is lightweight (Python) and intended as an entry point for 6G ISAC studies.

## 1 Objectives

- Form a UE beam and a sensing beam with one digital precoder on a ULA.
- Quantify the trade-off between communications SINR and sensing mainlobe gain.
- Enforce sidelobe suppression over a broad angular sector.
- Provide a compact, reproducible codebase and figures.

## 2 System Model

Consider an  $M$ -element ULA with inter-element spacing  $d = \lambda/2$ . The steering vector at angle  $\theta$  (broadside at 0) is

$$\mathbf{a}(\theta) = \left[ 1 \quad e^{j2\pi \frac{d}{\lambda} \sin \theta} \quad \dots \quad e^{j2\pi \frac{d}{\lambda} (M-1) \sin \theta} \right]^\top. \quad (1)$$

Let  $\mathbf{w} \in \mathbb{C}^M$  be the transmit weights. The array response towards  $\theta$  is  $B(\theta) = \mathbf{a}(\theta)^H \mathbf{w}$  and the normalized power pattern is  $|B(\theta)|^2$ .

The UE effective channel is modeled as

$$\mathbf{h}_{\text{UE}} = \frac{\mathbf{a}(\theta_{\text{UE}}) + \kappa \mathbf{h}_{\text{NLoS}}}{\|\mathbf{a}(\theta_{\text{UE}}) + \kappa \mathbf{h}_{\text{NLoS}}\|_2}, \quad (2)$$

with small Rayleigh perturbation  $\mathbf{h}_{\text{NLoS}}$  and  $\kappa \ll 1$ . The single-user downlink SINR (noise variance  $\sigma^2$ ) is

$$\text{SINR}_{\text{UE}} = \frac{|\mathbf{h}_{\text{UE}}^H \mathbf{w}|^2}{\sigma^2}. \quad (3)$$

## 3 Design Formulation

The idealized constrained problem is

$$\max_{\mathbf{w}} \quad |\mathbf{h}_{\text{UE}}^H \mathbf{w}|^2 \quad (4)$$

$$\text{s.t.} \quad |\mathbf{a}(\theta_s)^H \mathbf{w}|^2 \geq G_{\min}, \quad (5)$$

$$|\mathbf{a}(\theta)^H \mathbf{w}|^2 \leq S_{\max}, \quad \forall \theta \in \Theta_{\text{SL}}, \quad (6)$$

$$\|\mathbf{w}\|_2^2 \leq P_{\max}.$$

To obtain a simple, robust baseline and sweep the trade-off efficiently, we use *ridge-regularized pattern fitting*: enforce complex gains at a small set of target angles and penalize sidelobes elsewhere,

$$\min_{\mathbf{w}} \|\mathbf{Aw} - \mathbf{b}\|_2^2 + \lambda \|\mathbf{w}\|_2^2, \quad \Rightarrow \quad \mathbf{w} = (\mathbf{A}^H \mathbf{A} + \lambda \mathbf{I})^{-1} \mathbf{A}^H \mathbf{b}. \quad (7)$$

Rows of  $\mathbf{A}$  stack  $\mathbf{a}(\theta_{\text{UE}})$ ,  $\mathbf{a}(\theta_s)$  with desired complex gains  $b_{\text{UE}} = 1$  and  $b_s = \sqrt{G_{\text{targ}}}$ , and a dense set of sidelobe angles with target 0 weighted by  $w_{\text{SL}}$ . The solution is normalized to meet a transmit power constraint.

## 4 Methodology

**Angles and Grid.** Choose  $\theta_{\text{UE}}$  and sensing angle  $\theta_s$ . Build a grid  $\Theta_{\text{SL}}$  covering  $[-90^\circ, 90^\circ]$  while excluding small windows around the targets.

**Sweep.** For gain targets  $G_{\text{targ}}$  (in dB on voltage, i.e.,  $20 \log_{10}$ ), solve the ridge system and compute the pair  $(G_{\text{targ}}, \text{SINR}_{\text{UE}})$  to form a Pareto curve.

**Robustness.**  $\lambda$  regularizes against ill-conditioning and phase errors.  $w_{\text{SL}}$  controls sidelobe suppression.

## 5 Implementation

- `precoding/ula.py`: steering vector, beampattern, UE channel generator.
- `precoding/pattern_fit.py`: ridge design and SINR utility.
- `sims/dfrc_beam.py`: end-to-end sweep producing figures.

**Run:**

```

1 python - <<'PY'
2 import sys
3 sys.path.append('/mnt/data/isac')
4 from sims.dfrc_beam import run
5 run() # saves figures in data/results/
6 PY

```

## 6 Results

Figure 1 shows the normalized beampattern for a mid operating point (e.g.,  $M=12$ ,  $\theta_{\text{UE}}=10^\circ$ ,  $\theta_s=40^\circ$ ). A mainlobe is steered to  $\theta_s$  while the UE direction retains a strong response and sidelobes are suppressed across the remaining sector.

Figure 2 reports the Pareto frontier between sensing mainlobe *target voltage gain* and achieved UE SINR. As the sensing constraint tightens, the UE SINR decreases, quantifying the DFRC trade-off on a single array.

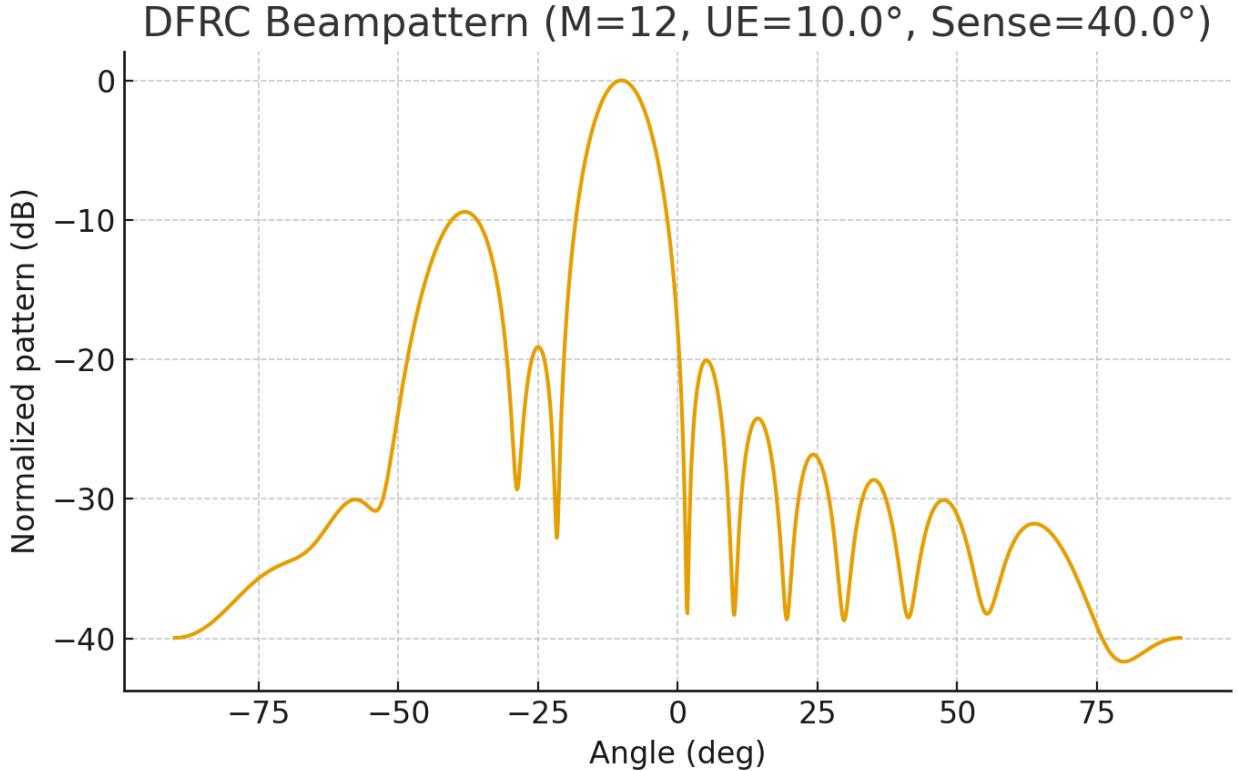


Figure 1: Normalized beampattern for a representative design.

## 7 Key Parameters

Parameter	Default
Elements $M$	12
Spacing $d/\lambda$	0.5
UE angle $\theta_{\text{UE}}$	10°
Sensing angle $\theta_s$	40°
Sidelobe mask exclude window	±8° around targets
Ridge weight $\lambda$	$10^{-3}$
Sidelobe weight $w_{\text{SL}}$	8
Noise variance $\sigma^2$	1
Gain sweep (voltage, dB)	{-10, -7, -5, -3, 0, 3, 6}

## 8 Discussion

**When ridge is enough.** For single-UE and one sensing lobe, ridge fitting with dense sidelobe samples gives clean, stable patterns and an interpretable Pareto. It is fast and differentiable.

**When to upgrade.** If hard inequality constraints are vital (e.g., guaranteed sidelobe caps), upgrade to QP/SOCP with slack variables. For multi-UE or multi-look sensing, extend  $\mathbf{A}$  and add per-direction priorities or iterative reweighting.

**Hardware non-idealities.** Phase quantization, mutual coupling, and calibration errors can be absorbed by: (i) adding perturbations in simulation, (ii) increasing  $\lambda$ , and (iii) re-fitting with measured array responses.

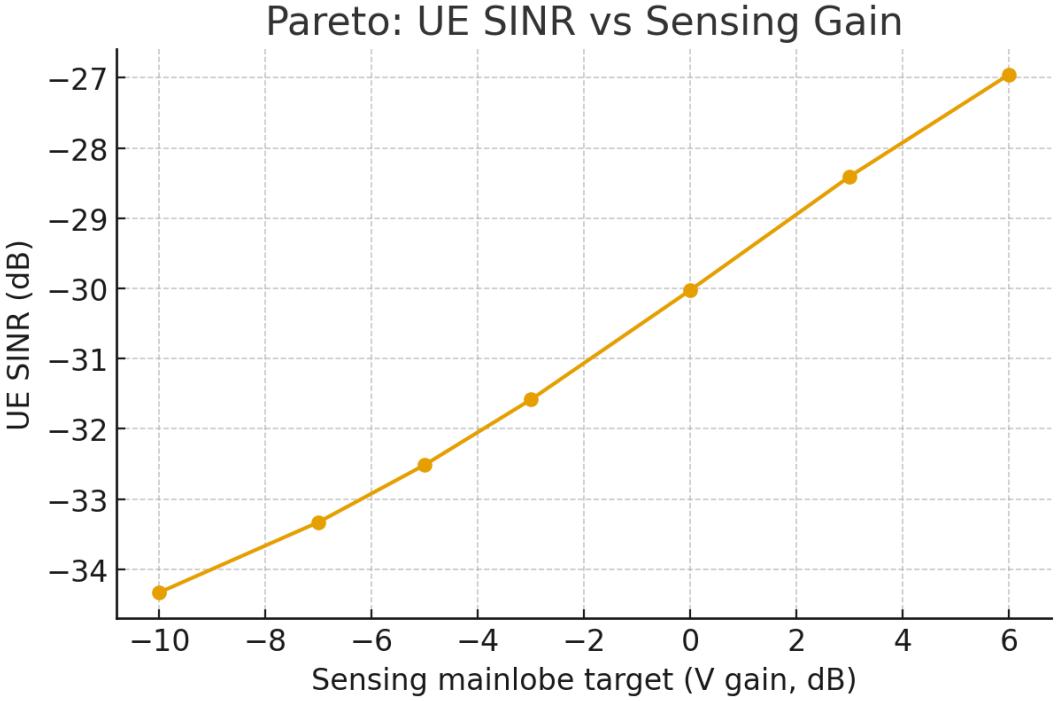


Figure 2: Pareto curve: UE SINR vs sensing mainlobe target (voltage gain, dB).

## 9 Planned Extensions

- Multi-UE DFRC with interference constraints and weighted sum-rate objective.
- Robust design under phase errors and per-antenna power constraints.
- Joint waveform-beam co-design with sensing PRB scheduling.
- Integration with Project 1 to evaluate end-to-end Pd/BLER trade-offs.

## 10 Conclusions

The DFRC beamforming demo provides a compact method to co-steer communications and sensing with a single array. The ridge approach exposes a clear SINR-sensing trade-off and serves as a fast baseline before moving to constrained convex formulations and hardware-in-the-loop tests.

**Repo:** /mnt/data/isac **Figures:** dfrc\_beampattern.png, dfrc\_pareto.png