HW-13

Q1. Determine (a) the transfer function of the network shown in fig P12.2 and (b) its pole and zeros.

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$$\frac{V_0}{V_{SS}} = \frac{R_2}{R_2 + l_1 II \frac{I}{SC_1}}$$

$$\frac{V_0}{||P_1||} + \frac{V_0}{||P_2||} = \frac{R_2}{||P_2||} = \frac{R_2}{||P$$

$$= \frac{R_2}{R_2 + \frac{R_1}{1 + SC_1 R_1}} = \frac{h_2(1 + SR_1 C_1)}{R_1 + R_2 + R_1 \cdot R_2 C_1 S} = \frac{R_2}{l_1 + R_2} \cdot \frac{1 + \frac{S}{1 + R_1 C_1}}{l_1 + \frac{S}{1 + R_2 C_1 R_2}}$$

Bode Plot

1201Bldec

odB

1/C1.(P.11P2) -, pole

Q2. Determine the pole and zero quality factors Qp and Q2, (b) the Pole and zero resonant frequencies wp and wz, C the pole factor βp , and D the pole angle Δp . The transfer function has the general form as given by $H(s) = \frac{5s^2 + 15s + 100}{s^2 + 20s + 200}$

H(s) = A.
$$5^{2} + 28z wo_{1}25 + wo_{1}2$$
 $5^{2} + 28pwo_{1}pS + wo_{1}p^{2}$

Bw₁p= $28pwo_{1}p - Q_{1} = \frac{wo_{1}p}{28pwo_{1}p} = \frac{\sqrt{200}}{20} = \frac{\sqrt{2}}{2} = \frac{Q_{1}p}{2}$

Bw₁p= $28z wo_{1}z = Q_{1} = \frac{wo_{1}p}{28zwo_{1}z} = \frac{\sqrt{20}}{3} = \frac{\sqrt{2}}{3} = \frac{Q_{2}p}{3}$

Bw₁z= $\frac{\sqrt{2}}{3} = \frac{\sqrt{2}}{3} = \frac{2}{3} = \frac{\sqrt{2}}{3} = \frac{\sqrt{2}}{3} = \frac{\sqrt{2}}{3} = \frac{\sqrt{2}$

Xp+Bp= Wop2=200

2ap=2fwgp=20/dp=10

 $P_{1} = -\alpha p - y_{3}p = -10 - 10j$ $P_{2} = -\alpha p + \beta_{3}p = -10 + 10j$ $P_{2} = -\alpha p + \beta_{3}p = -10 + 10j$ $Q_{p} = \alpha s^{-1} \left(\frac{\alpha p}{\sqrt{\alpha p^{2} + \beta_{n}^{2}}}\right) = \alpha s^{-1} \left(\frac{10}{\sqrt{200}}\right) = \left(\cos^{-1}\left(\frac{1}{\sqrt{2}}\right) = \frac{1}{45}\right)^{2}$

(2)

Q3 A coil having an inductance of 10MH is intended for applications around 1-MHz frequency. Its Q is specified to be 200. Find the equivalent parallel resistence Rp. What is the value of the capacitor required to produce resonance at 1MHz? What additional parallel resistance is requiered to produce 2 3-dB bandwith of lower

3

$$Q = \frac{Wo}{BW} = \frac{1/\sqrt{LC'}}{1/2C} = 2\sqrt{\frac{C'}{C'}}$$

$$R = \frac{1}{1/2C} = 2\sqrt{\frac{C'}{C'}}$$

$$R = \frac{1}{1/2C} = 2\sqrt{\frac{C'}{C'}}$$

277 R-1+00 Qiain ideal deger olsa de bu durunda BW=0 Olduquadan kullarilamamaktadır. Trade-off

$$\frac{\partial}{\partial t} = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{LC}} = \frac{2\pi \text{ IMHz}}{\sqrt{LC}} = \frac{1}{\sqrt{LC}} = \frac{2\pi \text{ IMHz}}{\sqrt{LC}} = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{LC}$$

Qy. Derive the response of an RC low-pass filter to a sinusoidal Wave form

$$\frac{2}{\sqrt{-v_{in}}} + \frac{2}{\sqrt{-v_{in}}} + \frac{2}{\sqrt{-v_{$$

$$= \frac{A}{Rc} \frac{W}{S_{4}^{2}W^{2}} \longrightarrow \left(S + \frac{1}{Rc}\right)V(S) = V(O) \neq \frac{A}{Rc} \frac{W}{S_{4}^{2}W^{2}}$$

$$\sqrt{6}) = \frac{\sqrt{(0^{-1})}}{5 + \frac{1}{lc}} + \frac{A}{lc} \cdot \frac{W}{(s^{2} + w^{2})(s + \frac{1}{lc})}$$

$$\frac{1}{(S_{+}^{2}\omega^{2}).(s_{+}^{2})} = \frac{F_{s}+G}{s_{+}^{2}\omega^{2}} + \frac{H}{s_{+}^{2}\omega^{2}} - (F_{s}+G).(s_{+}^{2}\omega^{2}) + H(s_{+}^{2}\omega^{2}) = 1$$

$$F_{5}^{2} + \frac{G}{2c} + G_{5} + \frac{F}{2c} +$$

$$H \left[1+(RCW)^{2}\right] = (RC)^{2}$$
 $H = \frac{(RC)^{2}}{1+(RCW)^{2}}$ $G = \frac{H}{RC} = \frac{RC}{1+(RCW)^{2}}$

$$G = \frac{H}{RC} = \frac{RC}{1 + (RCW)^2}$$

$$F = -\frac{(Rc)^2}{1+(Rcw)^2}$$

$$V_{Per}(s) = \frac{AW}{Rc} \left[\frac{FS+G}{s^2+w^2} \right] = \frac{AFw}{Rc}, \frac{s}{S^2+w^2} + \frac{AG}{Rc}, \frac{w}{S^2+w^2}$$

$$V_{Per}(+) = \frac{AFw}{Rc}, \alpha s (w + \frac{AG}{Rc}) + \frac{AG}{Rc} \sin w + \frac{AG}{Rc}$$

$$V_{L}(s) = \frac{AFw}{Rc}$$

$$V_{L}(s) = \frac{AFw}{$$

