Lesson 13

Digital Logic

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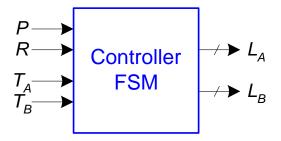
Factoring State Machines

- Break complex FSMs into smaller interacting FSMs
- Example: Modify traffic light controller to have Parade Mode.
 - Two more inputs: P, R
 - When P = 1, enter Parade Mode & Bravado Blvd light stays green
 - When R = 1, leave Parade Mode

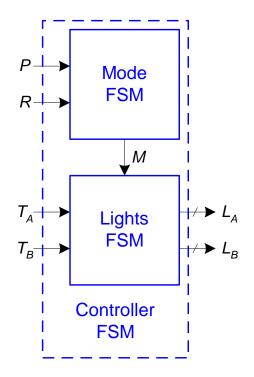


Parade FSM

Unfactored FSM



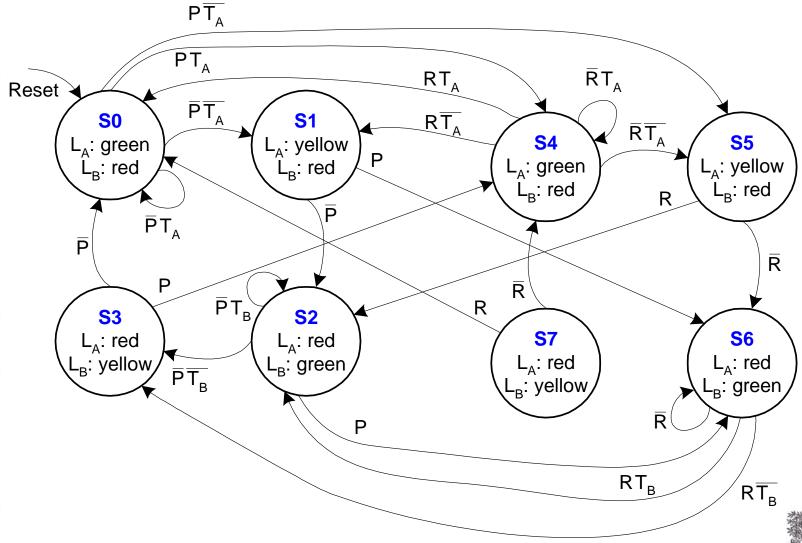
Factored FSM



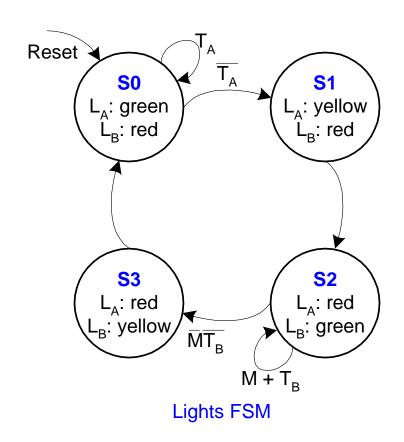


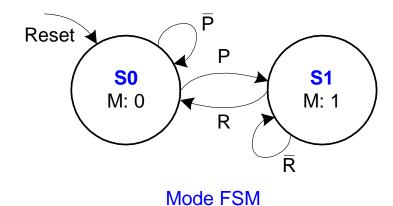
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Unfactored FSM



Factored FSM







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FSM Design Procedure

- 1. Identify inputs and outputs
- 2. Sketch state transition diagram
- 3. Write state transition table
- 4. Select state encodings
- 5. For Moore machine:
 - 1. Rewrite state transition table with state encodings
 - 2. Write output table
- 6. For a Mealy machine:
 - 1. Rewrite combined state transition and output table with state encodings
- 7. Write Boolean equations for next state and output logic
- 8. Sketch the circuit schematic

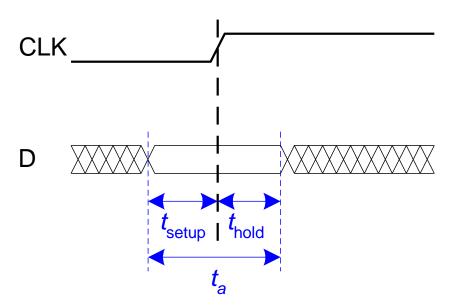
Timing

- Flip-flop samples D at clock edge
- D must be stable when sampled
- Similar to a photograph, D must be stable around clock edge
- If not, metastability can occur



Input Timing Constraints

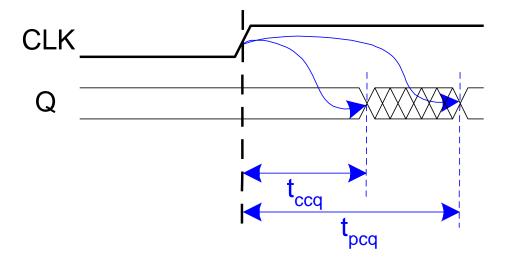
- Setup time: $t_{\text{setup}} = \text{time } before \text{ clock edge data must be stable (i.e. not changing)}$
- Hold time: t_{hold} = time after clock edge data must be stable
- Aperture time: t_a = time *around* clock edge data must be stable ($t_a = t_{\text{setup}} + t_{\text{hold}}$)





Output Timing Constraints

- Propagation delay: t_{pcq} = time after clock edge that the output Q is guaranteed to be stable (i.e., to stop changing)
- Contamination delay: t_{ccq} = time after clock edge that Q might be unstable (i.e., start changing)





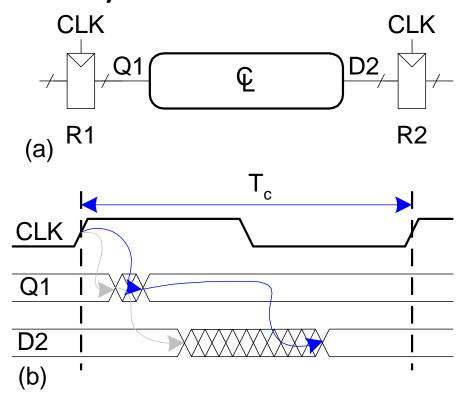
Dynamic Discipline

- Synchronous sequential circuit inputs must be stable during aperture (setup and hold) time around clock edge
- Specifically, inputs must be stable
 - at least t_{setup} before the clock edge
 - at least until t_{hold} after the clock edge



Dynamic Discipline

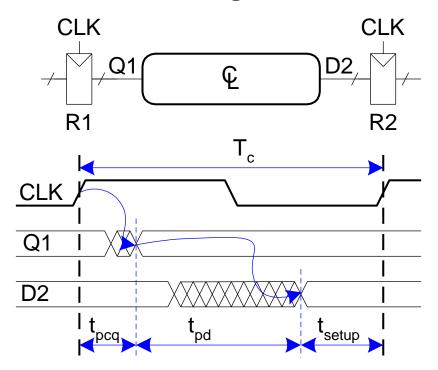
 The delay between registers has a minimum and maximum delay, dependent on the delays of the circuit elements





Setup Time Constraint

- Depends on the maximum delay from register R1 through combinational logic to R2
- The input to register R2 must be stable at least $t_{\rm setup}$ before clock edge

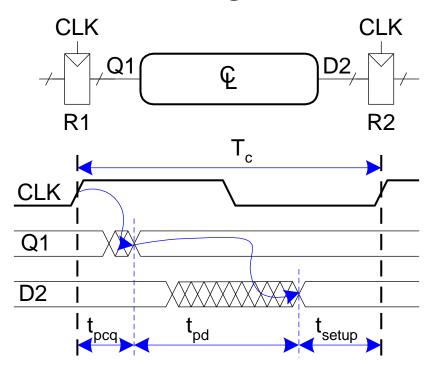






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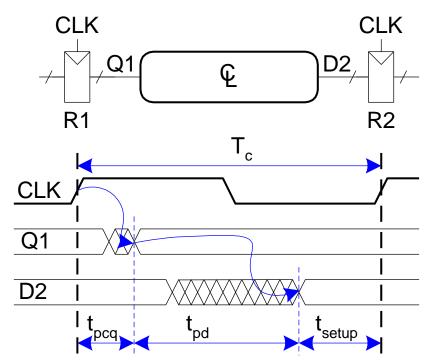
$$T_c \ge t_{pcq} + t_{pd} + t_{\text{setup}}$$

$$t_{pd} \le$$



Setup Time Constraint

- Depends on the maximum delay from register R1 through combinational logic to R2
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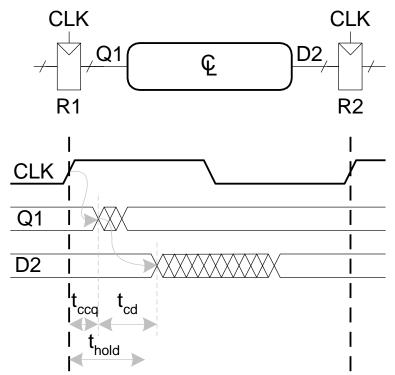
$$T_c \ge t_{pcq} + t_{pd} + t_{\text{setup}}$$
$$t_{pd} \le T_c - (t_{pcq} + t_{\text{setup}})$$

 $(t_{pcq} + t_{setup})$: sequencing overhead



Hold Time Constraint

- Depends on the minimum delay from register R1 through the combinational logic to R2
- The input to register R2 must be stable for at least t_{hold} after the clock edge

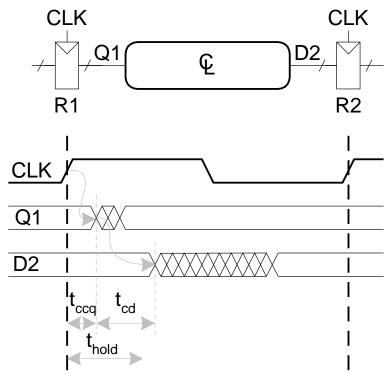






Hold Time Constraint

- Depends on the minimum delay from register R1 through the combinational logic to R2
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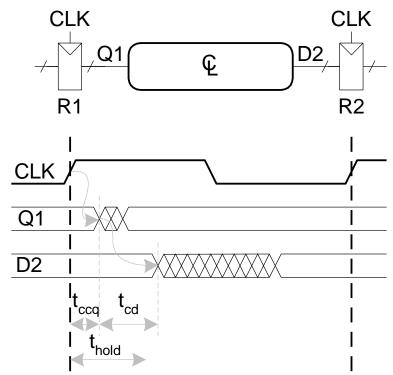
$$t_{\rm hold} < t_{ccq} + t_{cd}$$

$$t_{cd} >$$



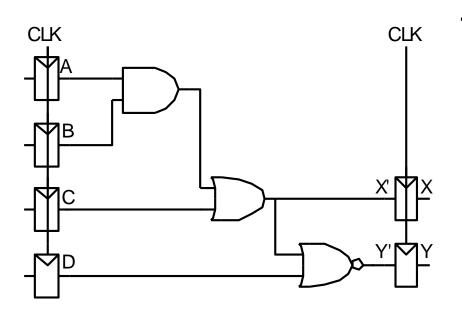
Hold Time Constraint

- Depends on the minimum delay from register R1 through the combinational logic to R2
- The input to register R2 must be stable for at least $t_{\rm hold}$ after the clock edge



$$t_{\text{hold}} < t_{ccq} + t_{cd}$$
 $t_{cd} > t_{\text{hold}} - t_{ccq}$





Timing Characteristics

$$t_{ccq}$$
 = 30 ps

$$t_{pcq} = 50 \text{ ps}$$

$$t_{\text{setup}} = 60 \text{ ps}$$

$$t_{\text{hold}} = 70 \text{ ps}$$

$$t_{pd} =$$

$$t_{cd} =$$

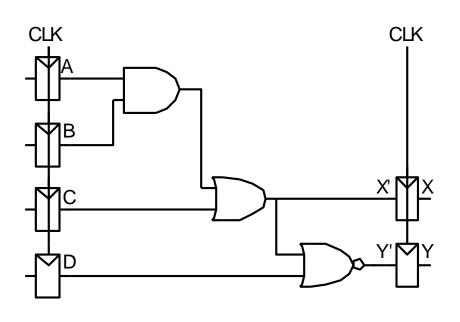
Setup time constraint:

$$T_c \ge$$

$$f_c =$$

$$t_{ccq} + t_{cd} > t_{hold}$$
?





Timing Characteristics

$$t_{ccq}$$
 = 30 ps

$$t_{pcq} = 50 \text{ ps}$$

$$t_{\text{setup}} = 60 \text{ ps}$$

$$t_{\text{hold}} = 70 \text{ ps}$$

$$t_{pd}$$
 = 3 x 35 ps = 105 ps

$$t_{cd} = 25 \text{ ps}$$

Setup time constraint:

$$T_c \ge (50 + 105 + 60) \text{ ps} = 215 \text{ ps}$$

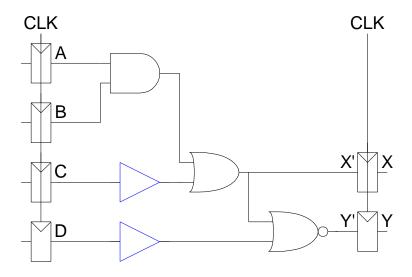
$$f_c = 1/T_c = 4.65 \text{ GHz}$$

$$t_{cca} + t_{cd} > t_{hold}$$
?

$$(30 + 25) ps > 70 ps ? No!$$



Add buffers to the short paths:



$$t_{pd} =$$

$$t_{cd} =$$

Setup time constraint:

$$T_c \ge$$

$$f_c =$$

Timing Characteristics

$$t_{cca} = 30 \text{ ps}$$

$$t_{pcq}$$
 = 50 ps

$$t_{\text{setup}} = 60 \text{ ps}$$

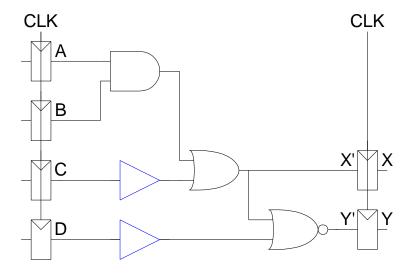
$$t_{\text{hold}} = 70 \text{ ps}$$

$$\begin{bmatrix} t_{pd} & = 35 \text{ ps} \\ t_{cd} & = 25 \text{ ps} \end{bmatrix}$$

$$t_{\text{ccq}} + t_{cd} > t_{\text{hold}}$$
?



Add buffers to the short paths:



$$t_{pd}$$
 = 3 x 35 ps = 105 ps

$$t_{cd}$$
 = 2 x 25 ps = 50 ps

Setup time constraint:

$$T_c \ge (50 + 105 + 60) \text{ ps} = 215 \text{ ps}$$

$$f_c = 1/T_c = 4.65 \text{ GHz}$$

Timing Characteristics

$$t_{cca} = 30 \text{ ps}$$

$$t_{pcq} = 50 \text{ ps}$$

$$t_{\text{setup}} = 60 \text{ ps}$$

$$t_{\text{hold}} = 70 \text{ ps}$$

$$\begin{bmatrix} t_{pd} & = 35 \text{ ps} \\ t_{cd} & = 25 \text{ ps} \end{bmatrix}$$

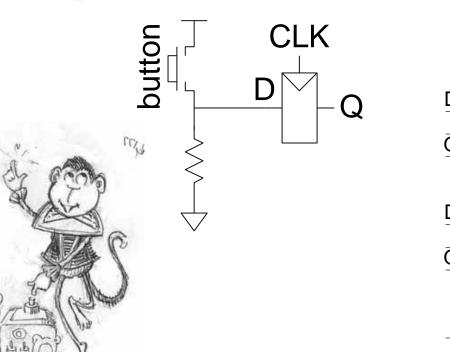
$$t_{ccq} + t_{cd} > t_{hold}$$
?

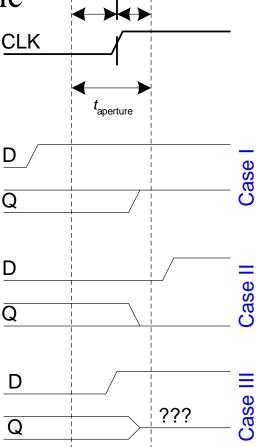
$$(30 + 50) ps > 70 ps ? Yes!$$



Violating Dynamic Discipline

Asynchronous (for example, user) inputs might violate the dynamic discipline CLK

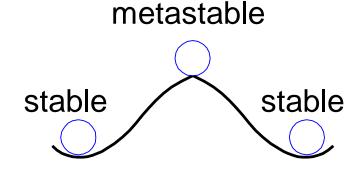






Metastability

- **Bistable devices:** two stable states, and a metastable state between them
- **Flip-flop:** two stable states (1 and 0) and one metastable state
- If a flip-flop lands in metastable state, could stay there for an undetermined amount of time







Flip-Flop Internals

Flip-flop has **feedback**: if Q is somewhere between 1 and 0, cross-coupled gates drive output to either rail (1 or 0)

N1

N2

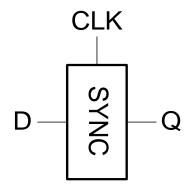


• If flip-flop input changes at random time (flip-flop samples metastable input), if you wait long enough time, the **metastable** output will resolve to 1 or 0 with **high probability**



Synchronizers

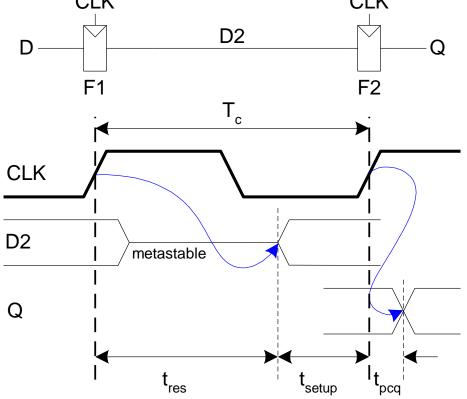
- Asynchronous inputs are inevitable (user interfaces, systems with different clocks interacting, etc.)
- Synchronizer goal: make the probability of failure (the output Q still being metastable) low
- Synchronizer cannot make the probability of failure 0





Synchronizer Internals

- Synchronizer: built with two back-to-back flip-flops
- Suppose D is transitioning when sampled by F1
- Internal signal D2 has $(T_c t_{\text{setup}})$ time to resolve to 1 or 0





Parallelism

Two types of parallelism:

- Spatial parallelism
 - duplicate hardware performs multiple tasks at once
- Temporal parallelism
 - task is broken into multiple stages
 - also called pipelining
 - for example, an assembly line



Parallelism Definitions

- Token: Group of inputs processed to produce group of outputs
- Latency: Time for one token to pass from start to end
- Throughput: Number of tokens produced per unit time

Parallelism increases throughput



Parallelism Example

- Ben Bitdiddle bakes cookies to celebrate traffic light controller installation
- 5 minutes to roll cookies
- 15 minutes to bake
- What is the latency and throughput without parallelism?



Parallelism Example

- Ben Bitdiddle bakes cookies to celebrate traffic light controller installation
- 5 minutes to roll cookies
- 15 minutes to bake
- What is the latency and throughput without parallelism?

Latency = 5 + 15 = 20 minutes = 1/3 hour

Throughput = 1 tray / 1/3 hour = 3 trays/hour

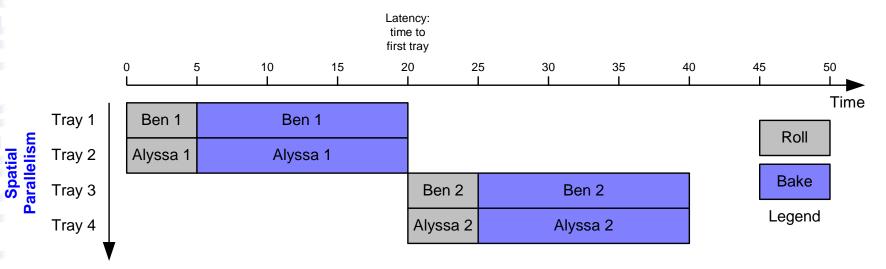


Parallelism Example

- What is the latency and throughput if Ben uses parallelism?
 - Spatial parallelism: Ben asks Allysa P. Hacker to help, using her own oven
 - Temporal parallelism:
 - two stages: rolling and baking
 - He uses two trays
 - While first batch is baking, he rolls the second batch, etc.



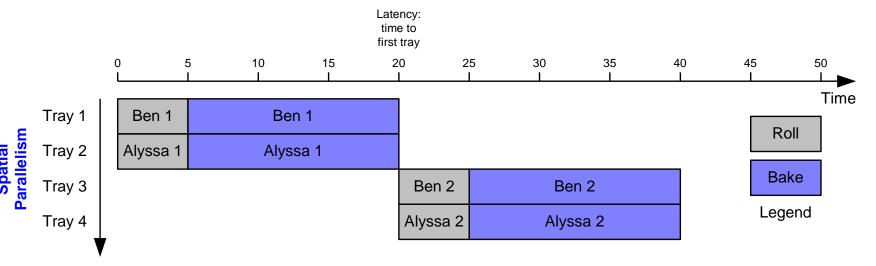
Spatial Parallelism



Latency = ?
Throughput = ?



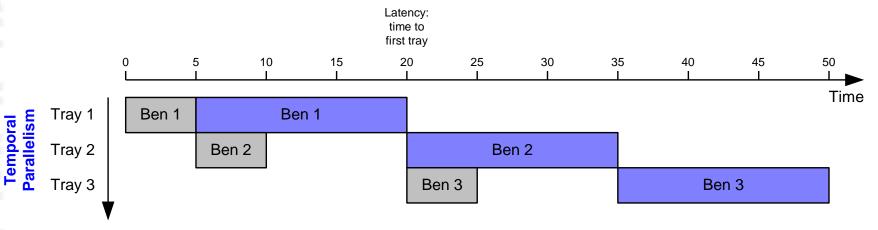
Spatial Parallelism



Latency = 5 + 15 = 20 minutes = 1/3 hour **Throughput** = 2 trays/ 1/3 hour = 6 trays/hour



Temporal Parallelism

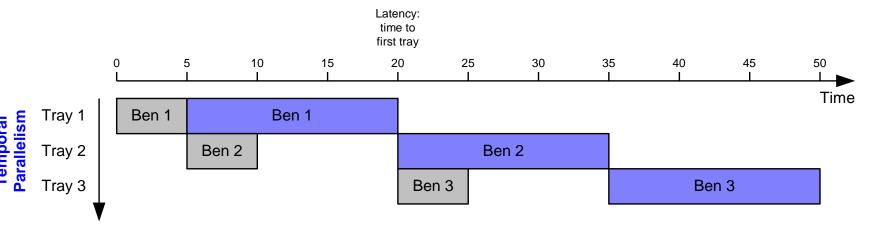


Latency = ?

Throughput = ?



Temporal Parallelism



Using both techniques, the throughput would be 8 trays/hour

