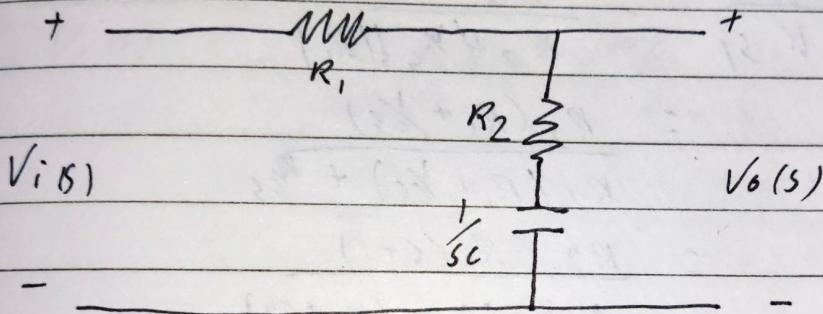


## Experiment one

→ Design and study of Lead, Lag and Lag-lead compensator network

- 1) Compute the transfer function and Pole-Zero plot of the following.
- as Lag compensator



$$\rightarrow V_o(s) = \frac{V_i(s)(R_2 + \frac{1}{sC})}{(R_1 + R_2 + \frac{1}{sC})}$$

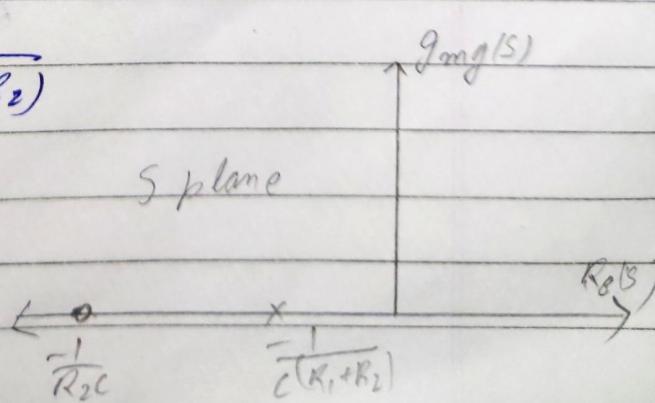
$$T(s) \Rightarrow \frac{V_o(s)}{V_i(s)}$$

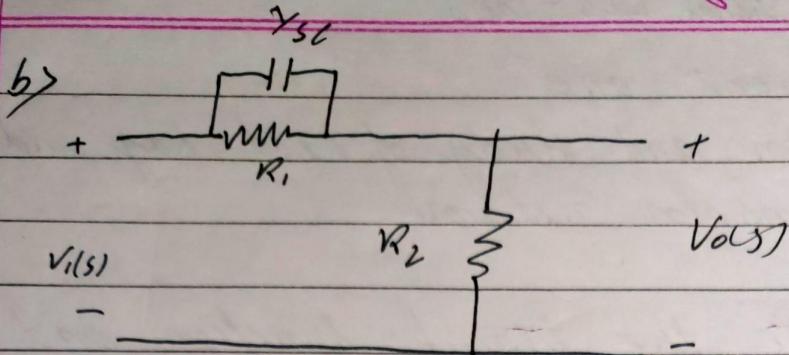
$$T(s) = \frac{1 + R_2 s C}{((R_1 + R_2)s + 1)}$$

$$\text{Pole} \Rightarrow \sigma = -\frac{1}{C(R_1 + R_2)}$$

$$\text{Zero} \Rightarrow \sigma = -\frac{1}{R_2 C}$$

S plane





Lead compensator

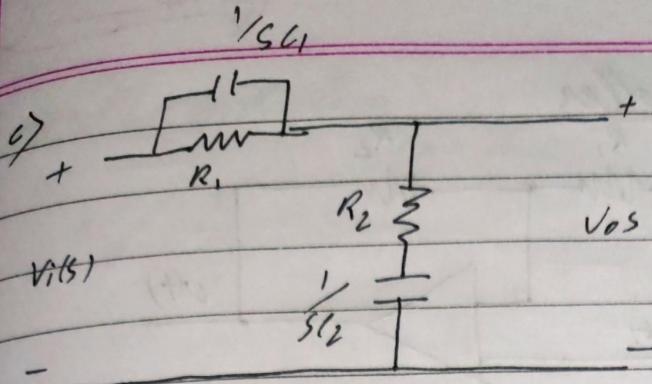


$$\begin{aligned}\frac{V_{o(s)}}{V_{i(s)}} &= \frac{R_2}{R_2 + Y_{S1}(R_1)} \\ &= \frac{R_2(R_1 + Y_{S1})}{R_2(R_1 + Y_{S1}) + R_1 Y_{S1}} \\ &= \frac{R_2(R_1(s+1))}{R_1 R_2(s + (R_1 + R_2))}\end{aligned}$$

$$B = \frac{R_2}{R_1 + R_2}$$

$$T = R_1 C$$

$$\begin{aligned}\frac{V_{o(s)}}{V_{i(s)}} &= \frac{R_2(R_1 C s + 1)}{(R_1 + R_2) \left[ \left( \frac{R_1 R_2 C s}{R_1 + R_2} \right) + 1 \right]} \\ &= B \left( \frac{sT + 1}{B s T + 1} \right)\end{aligned}$$

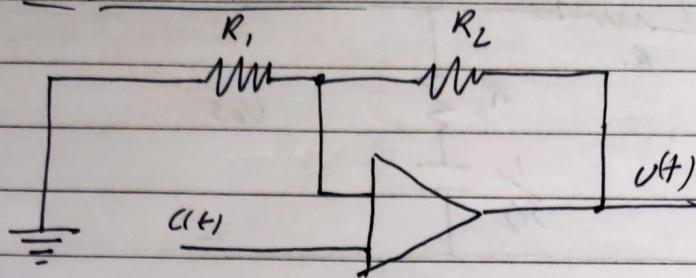


Lag-Lead compensator

$$\begin{aligned}
 \rightarrow \frac{V_{OS1}}{V_{(S)}} &= \frac{R_2 + \frac{1}{SC_2}}{R_2 + SC_2 + \left( \frac{R_1 SC_1}{R_1 + SC_1} \right)} \\
 &= \frac{(SC_2 R_2 + 1)(1 + SC_1 R_1)}{(1 + SC_1 R_2)(1 + SC_1 R_1) + (R_1 S C_2)} \\
 &= \frac{R_1 C_1 \left( \frac{1}{R_1 C_1} + s \right) R_2 C_2 \left( \frac{1}{R_2 C_2} + s \right)}{R_1 C_1 R_2 C_2 \left[ s^2 + s \left( \frac{1}{R_1 C_1} + \frac{1}{R_2 C_2} + \frac{1}{R_1 R_2 C_1 C_2} \right) + \frac{1}{R_1 R_2 C_1 C_2} \right]} \\
 &= \frac{(s + \frac{1}{T_1})(s + \frac{1}{T_2})}{(s + \beta/\tau_1)(s + \beta/\tau_2)}
 \end{aligned}$$

$$\beta = \frac{R_2}{R_1 + R_2}, \quad T_1 = R_1 C_2, \quad T_2 = R_2 C_1$$

$\Rightarrow$  P-controller



$\therefore$  It is a Non-Inverting opam

By virtual ground condition

$$v^+ = v^-$$

so Node 1 voltage is ' $v(s)$ '

$\rightarrow$  By nodal analysis

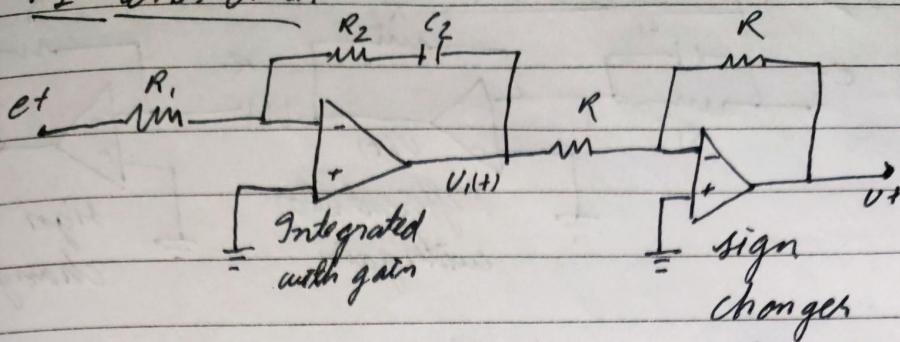
$$\frac{C(s) - 0}{R_1} + \frac{C(s) - v(s)}{R_2} = 0$$

$$\frac{C(s)}{R_1} + \frac{C(s)}{R_2} = \frac{U(s)}{R_2}$$

$$\frac{R_2}{R_1} + 1 = T(s)$$

$$T(s) = \frac{R_2}{R_1} + 1$$

### Q) PI-controller



→ By Laplace transform

$$\frac{O - e(s)}{R_1} + \frac{O - M(s)}{R_2 + 1/C_2 s} = 0$$

$$V_1(s) \Rightarrow -e(s) \frac{(1 + R_2 C_2 s)}{R_1 C_2 s} = 0$$

Now,

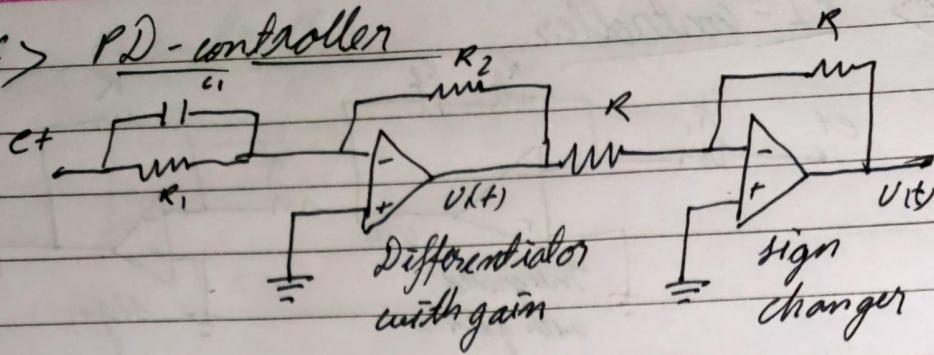
$$V+ = -V_1(+)$$
 [inverting opam]

$$V(s) = -M(s)$$

using ① and ②

~~$$\frac{V(s)}{e(s)} = T(s) = -\frac{(1 + R_2 C_2 s)}{R_1 C_2 s}$$~~

$\Rightarrow$  PD-controller



$$\rightarrow \frac{U_1(s)}{C(s)} = -\frac{R_2}{R_1 \frac{1}{C_1 s}} \frac{R_1 C_1 s + 1}{C_1 s}$$

$$\Rightarrow U_1(s) = -\frac{R_2 (R_1 C_1 s + 1) C(s)}{R_1 s^2}$$

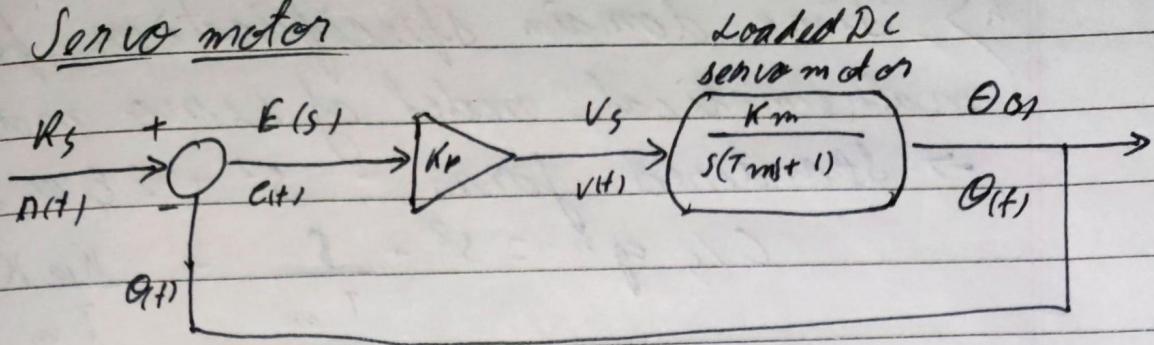
$$U(t) = -u_1(t)$$

$$U(s) = -u_1(s)$$

$$-U(s) = -\frac{R_2 (R_1 C_1 s + 1) \cdot C(s)}{R_1 s^2}$$

$$\frac{U(s)}{C(s)} = \frac{R_2 (R_1 C_1 s + 1)}{R_1 s^2}$$

### g) Servo motor



$$\rightarrow \frac{R_s s_1}{\theta(s)} = \frac{\frac{K_p K_m}{s(T_m s + 1)}}{1 + \frac{K_p K_m}{s(T_m s + 1)}}$$

$$= \frac{K_p K_m}{s(T_m s + 1) + K_p K_m}$$

$$\text{Char. eqn. : } H(s) = 0$$

$$T_m s^2 + s + K_p K_m = 0$$

$$s_{1,2} = \frac{-1 \pm \sqrt{1 - 4 T_m K_p K_m}}{2 T_m}$$

Poles of transfer function depend on the values of  $K_p$  &  $K_m$

↳ Time domain specification for mathematical model of servo motor.

→ Standard form :  $s^2 + 2\xi\omega_n s + \omega_n^2$

$$C/s \text{ eqn} = s^2 + \frac{s}{T_m} + \frac{K_p K_m}{T_m}$$

$$\text{natural frequency } \omega_n = \sqrt{\frac{K_p K_m}{T_m}}$$

$$2\xi\omega_n = \frac{1}{T_m}$$

$$\xi = \frac{1}{2T_2} \sqrt{\frac{T_m}{K_p K_m}}$$

$$\xi = \frac{1}{2\sqrt{T_m K_p K_m}}$$

$$\text{raise time } t_r = \frac{\pi - \theta}{\omega_d}$$

$$\omega_d = \omega_n \sqrt{1 - \xi^2}$$

$$= \sqrt{\frac{4 K_p K_m T_m - 1}{2 T_m}}$$

$$\theta = \left[ (\cos^{-1} \xi) \frac{\pi}{180} \right]$$

$$t_r = \frac{2 T_2 \left[ \pi - \frac{\pi}{180} \cos^{-1} \xi \right]}{\sqrt{4 T_m K_p K_m - 1}}$$

$$\text{peak time } t_p = \frac{n\pi}{\text{Cod}}$$

$$n = 1$$

$$t_p = \frac{2\pi T_m}{\sqrt{4 T_m K_p K_m - 1}}$$

$$\text{peak overshoot } M_o = e^{-\frac{\pi}{\sqrt{1 - \xi^2}}}$$

$$= e^{-\pi} \frac{\sqrt{2} \sqrt{T_m K_p K_m}}{\sqrt{1 - \frac{1}{n T_m K_p K_m}}}$$

$$M_o = e^{-\frac{\pi}{\sqrt{4 K_p K_m T_m - 1}}}$$

settling time

2% EB

$$t_{sett} = \frac{4}{\xi \omega_n}$$

$$= \frac{4}{\sqrt{2} \sqrt{T_m}} = 8 T_m$$

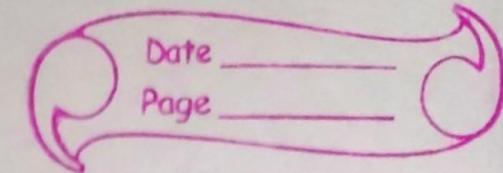
5% EB

$$t_{sett} = \frac{3}{\xi \omega_n} = 6 T_m$$

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### 1<sup>st</sup> order system

- 1 independent energy storage element
- it does not exhibit ripples, irrespective of position of poles
- Time constant plays a major role



### 2<sup>nd</sup> order system

- 2 independent energy storage element
- it may or may not exhibit oscillatory behavior
- Natural frequency and damping ratio plays a major role.