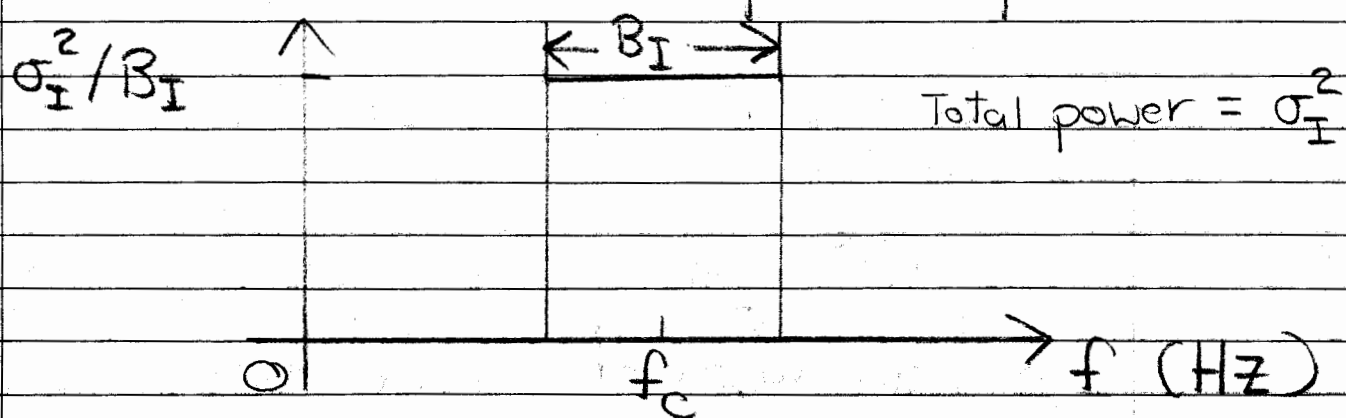


### 6.6.13 Broadband Interference Sources

Consider a 10-element linear array designed for frequency  $f_c$

A narrowband plane wave signal arrives from  $u = u_s$ .

There are two plane wave interferers from  $u_1 = 0.3$  and  $u_2 = 0.5$ . They have broadband temporal spectra:



Let  $B_{fI} = B_I / f_c$  (normalized)

- (a) Design a MVDR beamformer using the assumption that the interferers are narrowband and evaluate array gains for several values of  $B_{fI}$ .

Solution:

I need to determine the actual  $S_n(\omega)$  in order to solve the problem.

For a single broadband source, it can be shown that the  $(n, m)$ th element of the array covariance matrix is:

$$[S_x(\omega)]_{n,m} =$$

$$\sigma_I^2 \exp(j\pi(n-m)u_I) \operatorname{sinc}\left(\frac{\pi(n-m)B_{FI}u_I}{2}\right)$$

If  $B_{FI} = 0$ ,  $\operatorname{sinc}(0) = 1$  and  $S_x(\omega)$  collapses to that of a rank one narrowband interference source

To solve part (a), I did the following:

(1) Generated an assumed  $S_n(\omega)$  making the assumption that the interferers are narrowband, rank-one sources

(2) Built  $\underline{w} = S_n^{-1} \underline{v}_s / (\underline{v}_s^H S_n^{-1} \underline{v}_s)$

assuming  $u_g = 0$  and using the narrowband  $S_n(\omega)$

(3) Evaluated array gain with the expression

$$AG = |\underline{w}^H \underline{v}_s|^2 / (\underline{w}^H \underline{p}_n \underline{w})$$

where  $\underline{p}_n$  is derived from the actual  $S_n(\omega)$

- (a) The beam pattern is the same for each value of  $B_I$  because the narrowband assumption was made in each case. It is attached.

$B_{fI} = B_I/f_c$	Array Gain (dB)
0	32.74 (upper bound)
0.1	31.27
0.2	28.70
0.4	24.76

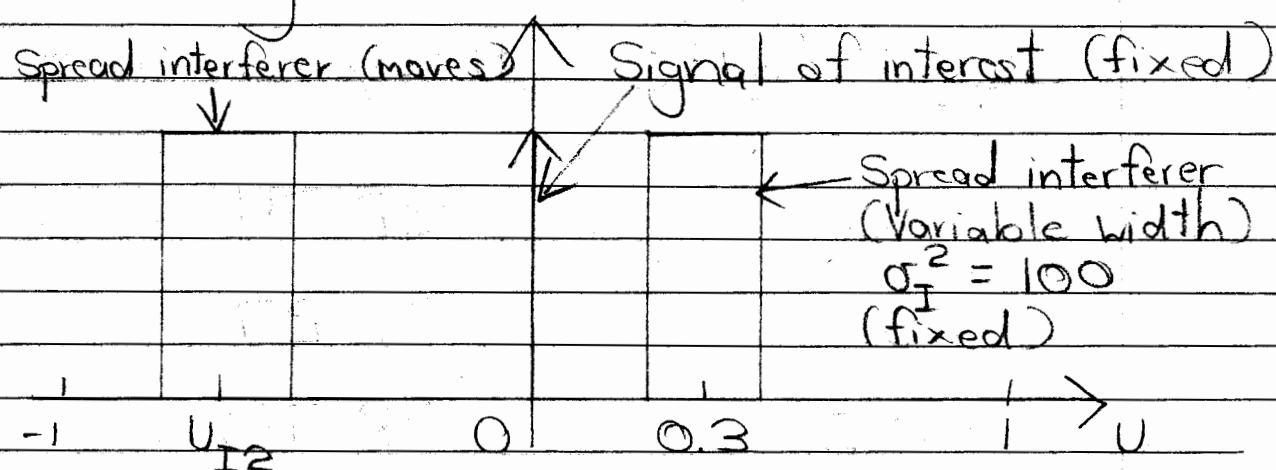
The trend is expected. As the interferer bandwidth increases, the mismatch between actual and assumed  $S_n(\omega)$  increases and the processor is mismatched to the environment. That results in performance degradation.

- (b) Let the DOA and INR of the second interferer be variables.

Plot array gain vs.  $U_{I2}$  for several values of INR.

Solution:

I wrote a script to implement the following scenario:



(a) The second spread interferer has variable width, but same width as the first one.

(b)  $-1 \leq u_{I2} \leq 1$

(c) Signal of interest from  $u_s = 0$

(d) The INR of the second interferer varies

(e) The white noise floor is set to one.

The beamformer is designed assuming narrowband interferers.

INR of interferer #1 = 20 dB

INR of second interferer is variable

The DOF of the second interferer varies  
 $(-1 \leq u_{I2} \leq 1)$

I have plotted array gain vs  $u_{I2}$   
 for  $INR's = -10, 0, 10$  and  $20$  dB  
 and  $B_{fI} = 0$  and  $0.1$ .

Plots are attached.

(c) Design a MVDR beamformer using the  
 actual  $S_n(\omega)$ . Plot beam patterns and  
 compute array gains.

Solution:

I included this in the code I used to  
 solve part (a). Beam patterns are attached  
 and array gains tabulated below.

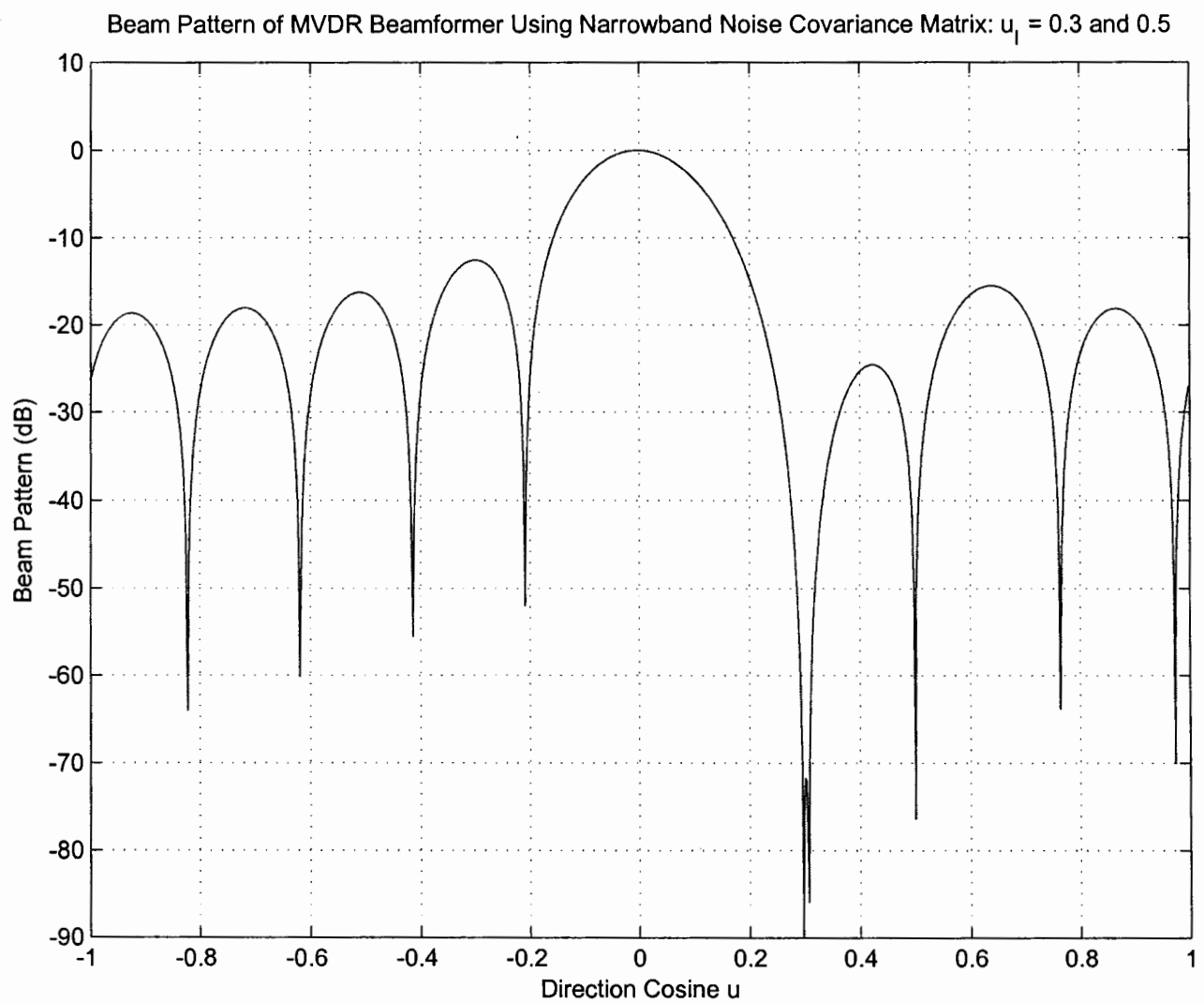
$B_{fI} = B_I / f_c$	Gain w/Correct $S_n(\omega)$ (dB)
0	32.74 (32.74)
0.1	32.37 (31.27)
0.2	32.12 (28.7)
0.4	31.61 (24.7)

The numbers in parantheses are the gains  
 provided by the MVDR processor that  
 assumes  $B_{fI} = 0$  for both interferers.

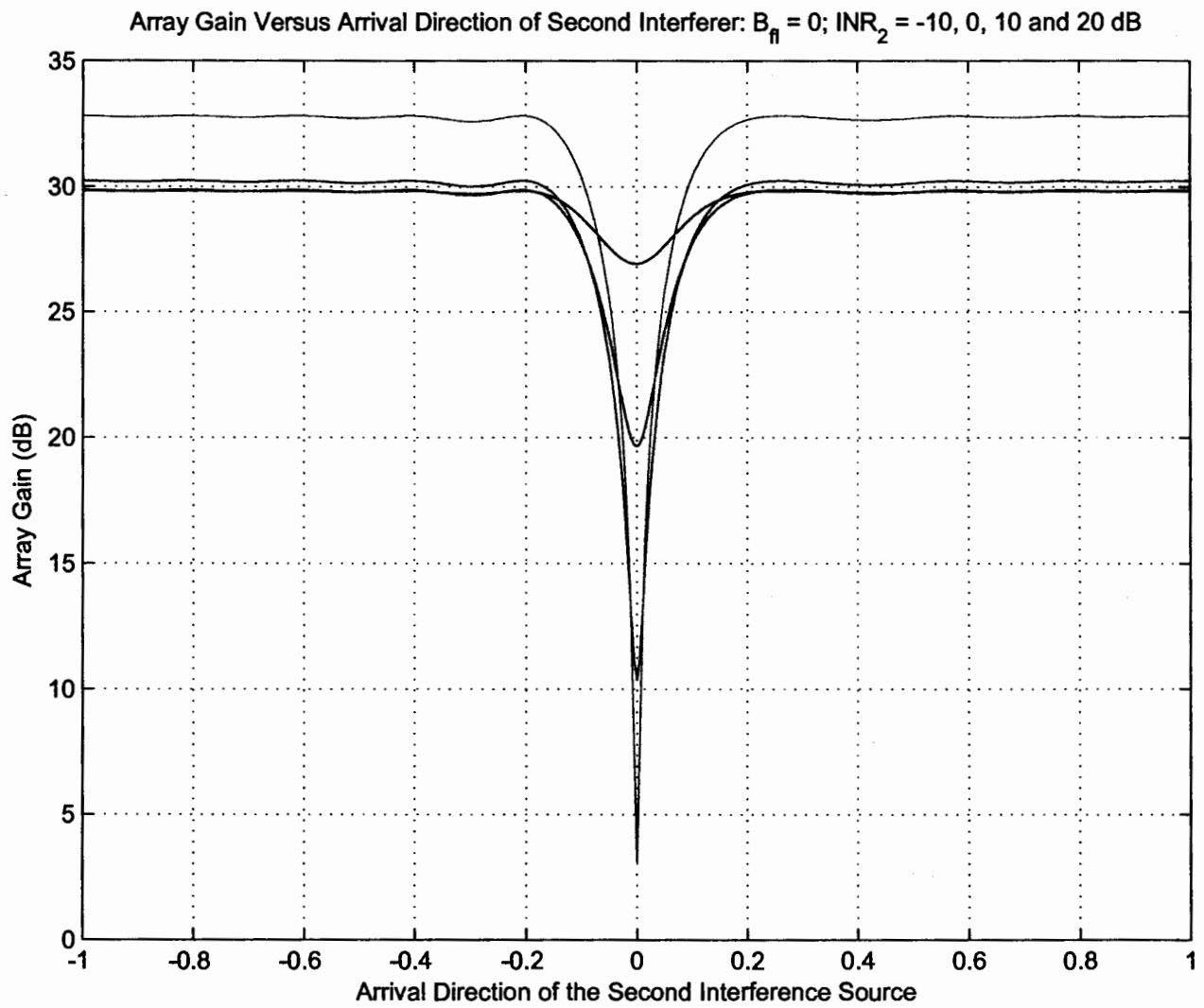
As the bandwidth increases, the performance degradation becomes more apparent.

Beam pattern plots follow.

Part (a)

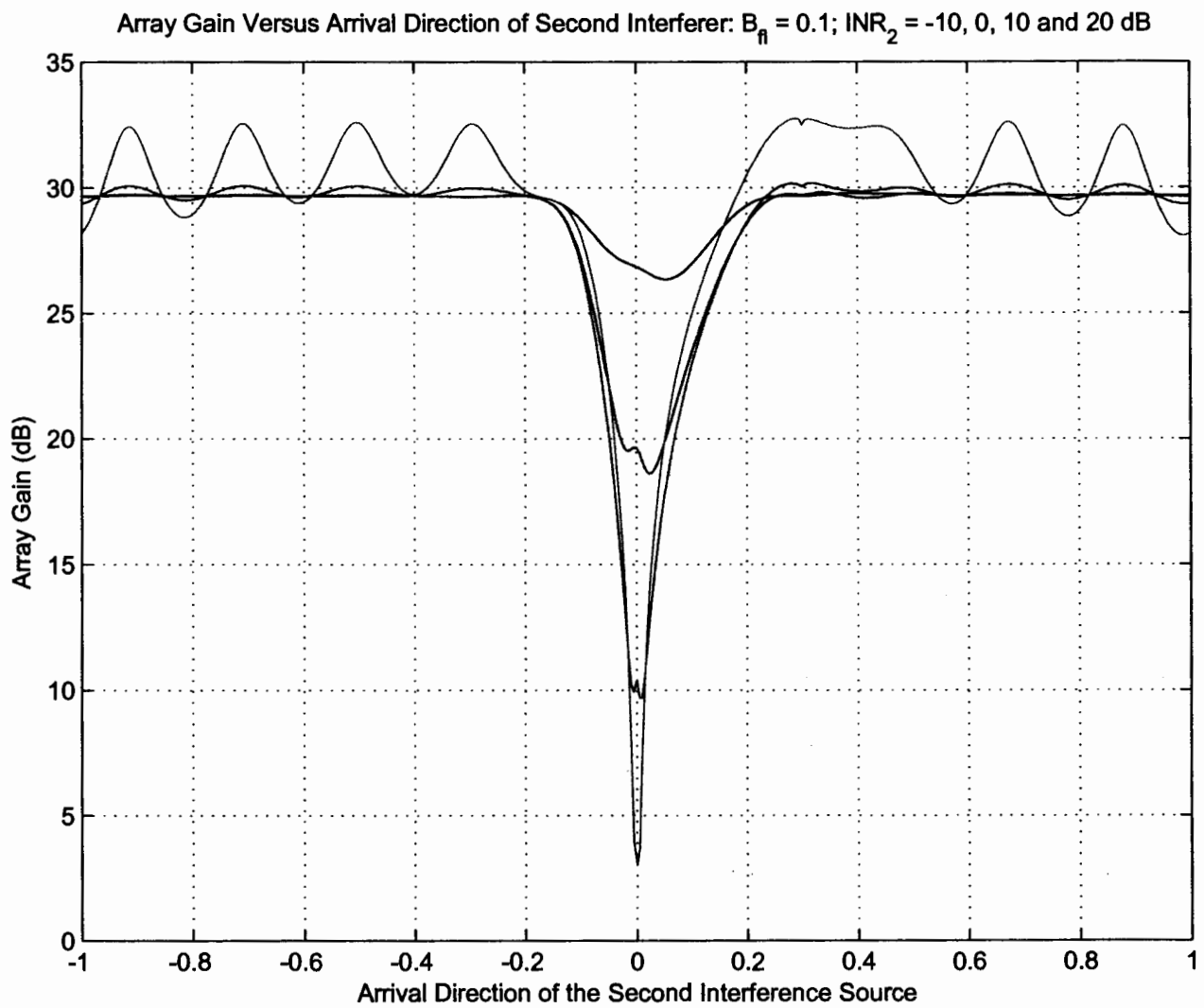


Part (b)

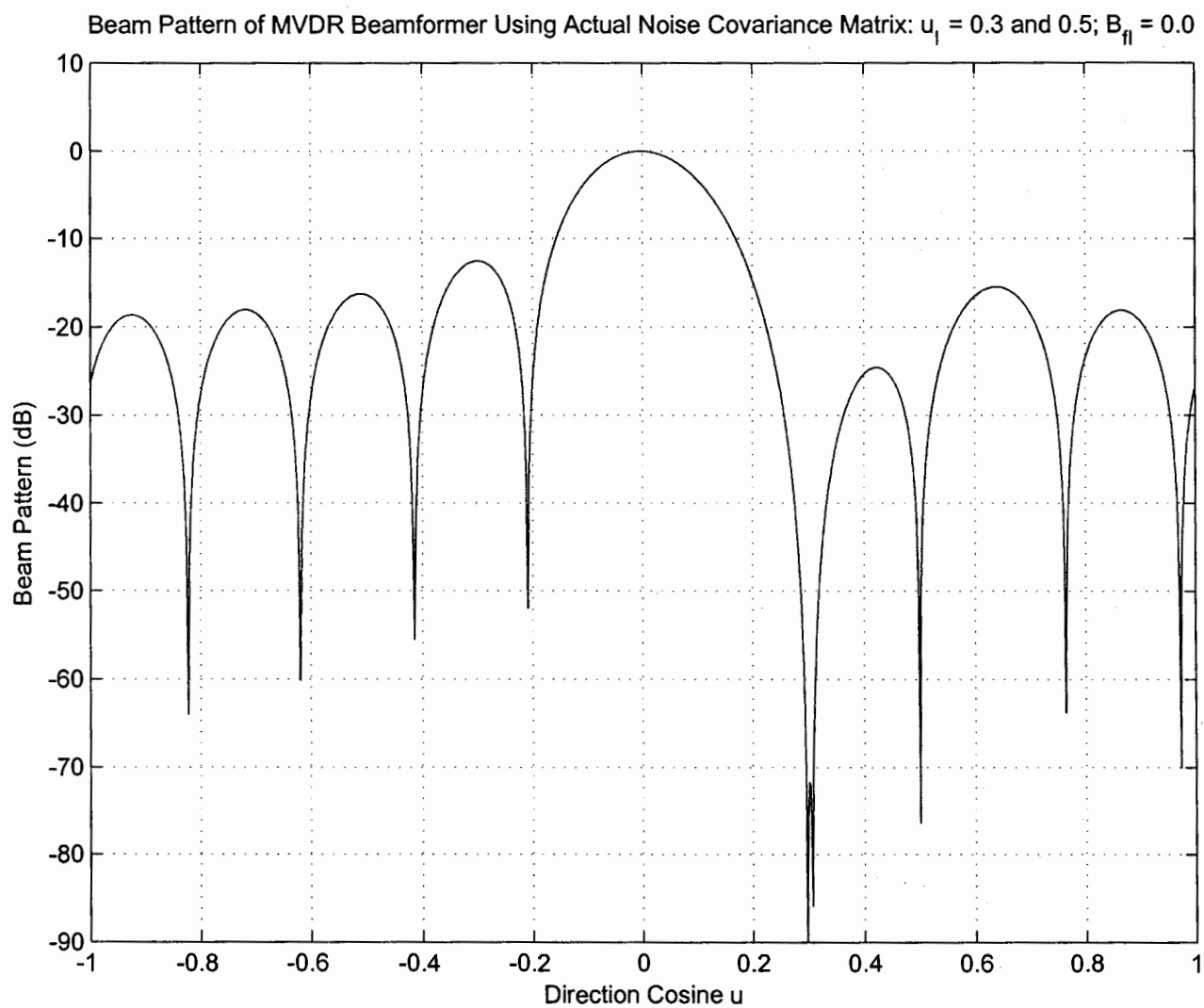




Part (b)

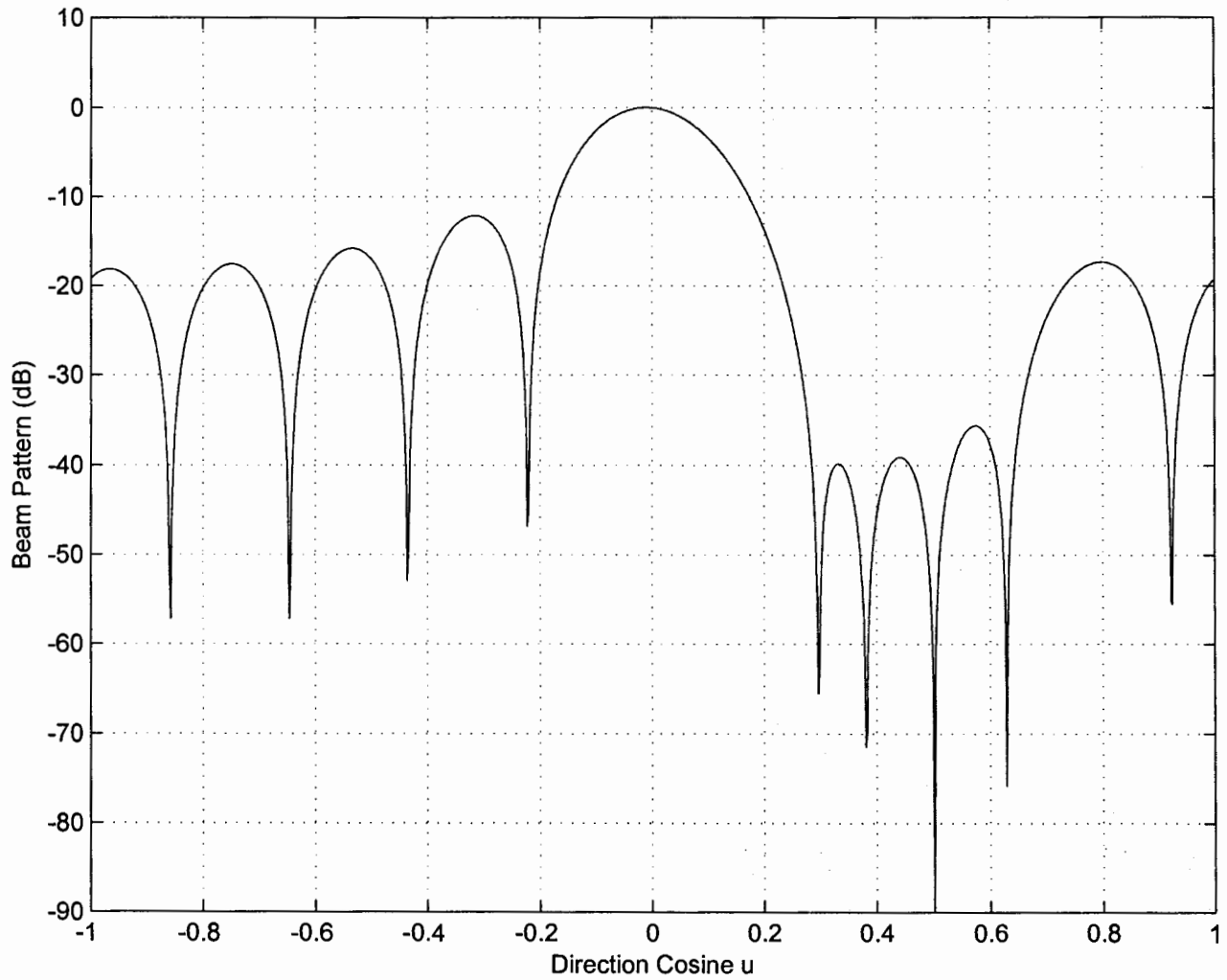


Part (c)

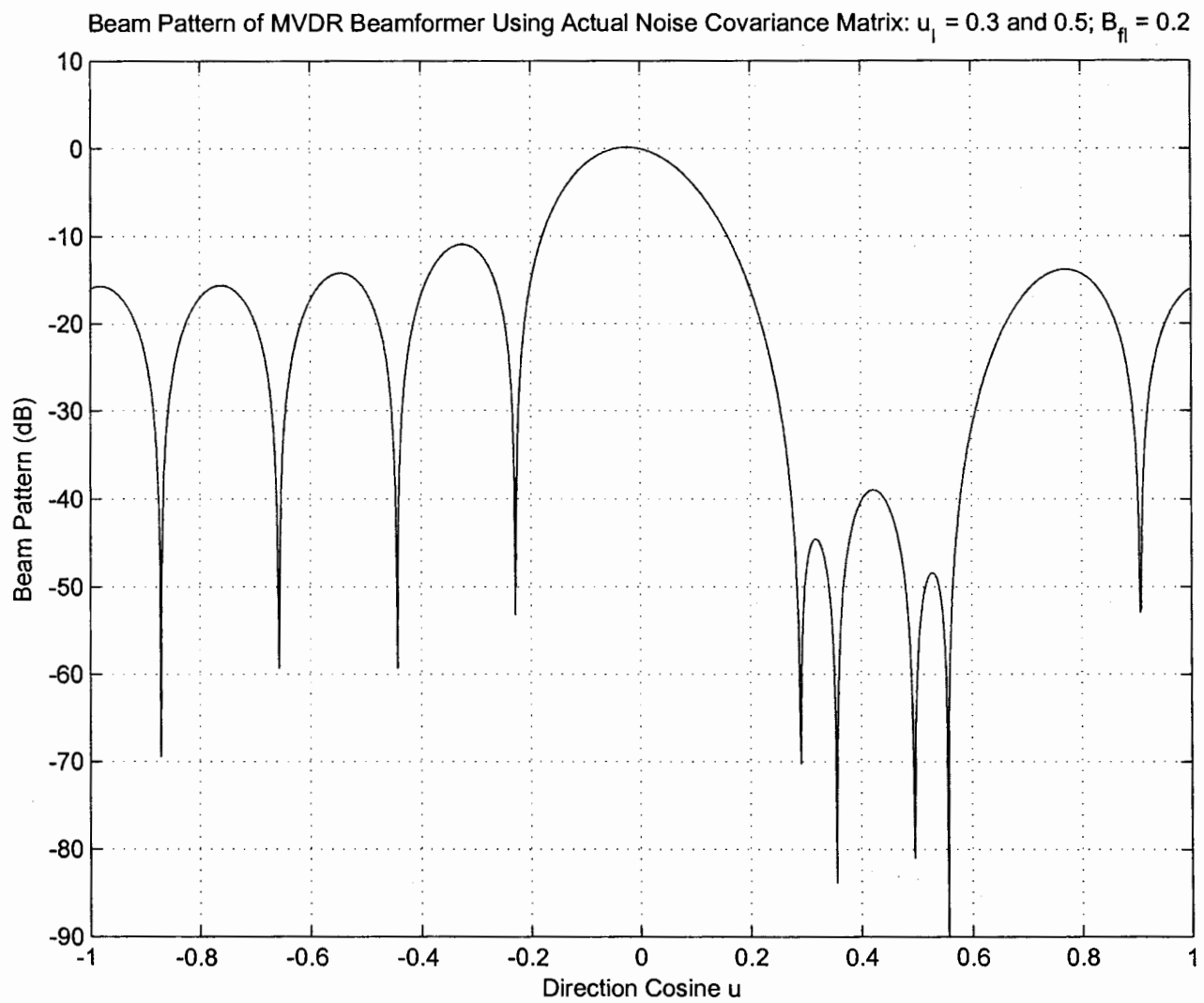


Part (c)

Beam Pattern of MVDR Beamformer Using Actual Noise Covariance Matrix:  $u_1 = 0.3$  and  $0.5$ ;  $B_{11} = 0.1$



Part (c)



Part (c)

Beam Pattern of MVDR Beamformer Using Actual Noise Covariance Matrix:  $u_1 = 0.3$  and  $0.5$ ;  $B_{fl} = 0.4$

