

6.S.5

- D plane wave signals plus white noise

$$x = V f + n \quad (N \times 1)$$

where

$$V = [v_1 \dots v_D] \quad (N \times D)$$

$$\underline{f} = [f_1 \dots f_D]^T \quad (D \times 1)$$

- $S_x = V S_f V^* + S_n$

- MMSE estimate of all D signals

$$\hat{\underline{f}} = \begin{bmatrix} \hat{f}_1 \\ \vdots \\ \hat{f}_D \end{bmatrix} = H_0 x$$

where  $H_0$  is the  $D \times N$  filter

$$H_0 = S_f x^* S_x^{-1}$$

- assume noise uncorrelated with signals

$$S_f x^* = E\{f x^*\} = E\{f f^*\} V^* = S_f V^*$$

$$\Rightarrow H_0 = S_f V^* S_x^{-1} = (I + S_f V^* S_n^{-1} V)^{-1} S_f V^* S_n^{-1}$$

- assuming  $S_f^{-1}$  exists, this can also be written as

$$H_0 = \underbrace{(I + S_f V^* S_n^{-1} V)^{-1} (S_f^{-1})^{-1}}_{A^{-1} B^{-1} = (BA)^{-1}} V^* S_n^{-1}$$

$$= (S_f^{-1} + V^* S_n^{-1} V)^{-1} V^* S_n^{-1} = S_f \underbrace{S_f^{-1} (S_f^{-1} + V^* S_n^{-1} V)^{-1} V^*}_{A^{-1} B^{-1} = (BA)^{-1}} S_n^{-1}$$

$$H_0 = S_f (I + V^* S_n^{-1} V S_f)^{-1} V^* S_n^{-1}$$

- when  $S_n = \sigma_w^2 I \Rightarrow \boxed{H_0 = \frac{S_f}{\sigma_w^2} (I + V^* V \frac{S_f}{\sigma_w^2})^{-1} V^*}$

- MMSE estimate of 1<sup>st</sup> signal only

$$\hat{f}_1 = H_1 x \quad \text{where } H_1 \text{ is the } 1 \times N \text{ filter}$$

$$H_1 = S_{f_1 x}^* S_x^{-1}$$

$$S_{f_1 x}^* = E\{f_1 x^*\} = E\{f_1 f^*\} V^* = \Sigma_1 V^*$$

$$\text{where } \Sigma_1 \text{ is first row of } S_f = \begin{bmatrix} \Sigma_1 & \dots \\ \vdots & \vdots \\ \Sigma_f & \dots \end{bmatrix}$$

$$\Rightarrow \underline{\underline{H_1 = \Sigma_1 V^* S_x^{-1}}}$$

$$\begin{aligned} \text{comparing to } H_0 &= S_f V^* S_x^{-1} = \begin{bmatrix} \Sigma_1 \\ \vdots \\ \Sigma_f \end{bmatrix} V^* S_x^{-1} \\ &= \begin{bmatrix} \Sigma_1 V^* S_x^{-1} \\ \vdots \\ \Sigma_f V^* S_x^{-1} \end{bmatrix} \end{aligned}$$

we see that  $H_1$  is first row of  $H_0$ , thus estimating  $f_1$  as one of  $D$  desired signals, or by itself with remaining signals as interferers results in the same processor.

$$\begin{aligned} \text{for white noise } H_0 &= \frac{S_f}{\sigma_{w^2}} (I + V^* V \frac{S_f}{\sigma_{w^2}})^{-1} V^* \\ &= \frac{1}{\sigma_{w^2}} \begin{bmatrix} \Sigma_1 (I + V^* V S_f / \sigma_{w^2})^{-1} V^* \\ \vdots \\ \Sigma_f (I + V^* V S_f / \sigma_{w^2})^{-1} V^* \end{bmatrix} \end{aligned}$$

$$\Rightarrow \boxed{H_1 = \frac{\Sigma_1}{\sigma_{w^2}} (I + V^* V \frac{S_f}{\sigma_{w^2}})^{-1} V^*, \quad \Sigma_1 = \text{first row of } S_f}$$

The result  $H_1 = \frac{\sum_i (I + V^\# V \frac{S_f}{\sigma_w^2})^{-1} V^\#}{\sigma_w^2}$  holds for

arbitrary  $D < N-1$  and does not require  $S_f$  diagonal

expanding this for  $D=2$  does not lead to any enlightening expressions for the general case.

• However if  $S_f = \begin{bmatrix} \sigma_1^2 & 0 \\ 0 & S_I \end{bmatrix}$  (signal 1 uncorrelated with rest)

$$\Rightarrow \Sigma_1 = \begin{bmatrix} \sigma_1^2 & 0 \end{bmatrix}$$

we can get an interesting expression. Note that

$$\frac{\Sigma_1}{\sigma_w^2} (I + V^\# V \frac{S_f}{\sigma_w^2})^{-1} = \frac{\sigma_1^2}{\sigma_w^2} \cdot \text{first row of } (I + V^\# V \frac{S_f}{\sigma_w^2})^{-1}$$

$$\text{partition } V = [V_1 : V_I] \Rightarrow V^\# V = \begin{bmatrix} V_1^\# V_1 & V_1^\# V_I \\ V_I^\# V_1 & V_I^\# V_I \end{bmatrix}$$

$$\text{define } \rho_{1I} = \frac{1}{N} V_1^\# V_I \Rightarrow V^\# V = \begin{bmatrix} N & N \rho_{1I} \\ N \rho_{1I}^\# & V_I^\# V_I \end{bmatrix}$$

$$I + V^\# V \frac{S_f}{\sigma_w^2} = \begin{bmatrix} 1 + N \sigma_1^2 / \sigma_w^2 & N \rho_{1I} S_I / \sigma_w^2 \\ N \rho_{1I}^\# \sigma_1^2 / \sigma_w^2 & I + V_I^\# V_I S_I / \sigma_w^2 \end{bmatrix}$$

1<sup>st</sup> row of inverse of Partitioned Matrix

$$\begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix}^{-1} = \begin{bmatrix} (A_{11} - A_{12} A_{22}^{-1} A_{21})^{-1} & - (A_{11} - A_{12} A_{22}^{-1} A_{21})^{-1} A_{12} A_{22}^{-1} \\ \cdot & \cdot \end{bmatrix}$$

$$(A_{11} - A_{12} A_{22}^{-1} A_{21})^{-1} = \left[ 1 + N \sigma_1^2 / \sigma_w^2 - N \rho_1^T S_I / \sigma_w^2 (I + V_I^H V_I S_I / \sigma_w^2)^{-1} N \rho_1^H \sigma_1^2 / \sigma_w^2 \right]^{-1}$$

$$A_{12} A_{22}^{-1} = N \rho_1^T S_I / \sigma_w^2 (I + V_I^H V_I S_I / \sigma_w^2)^{-1} = N \rho_1^T (I + \frac{S_I}{\sigma_w^2} V_I^H V_I)^{-1} \frac{S_I}{\sigma_w^2}$$

$$\Rightarrow H_1 = \frac{\sigma_1^2}{\sigma_w^2} (A_{11} - A_{12} A_{22}^{-1} A_{21})^{-1} \left[ v_1^H - A_{12} A_{22}^{-1} v_2^H \right]$$

letting  $S_n \equiv \sigma_w^2 I + V_I S_I V_I^H$ , we see that

$$(A_{11} - A_{12} A_{22}^{-1} A_{21})^{-1} = (1 + \sigma_1^2 v_1^H S_n^{-1} v_1)^{-1} = \frac{\Lambda}{\sigma_1^2 + \Lambda}$$

$$\text{where } \Lambda = (v_1^H S_n^{-1} v_1)^{-1}$$

$$\Rightarrow H_1 = \frac{\sigma_1^2 \Lambda}{\sigma_1^2 + \Lambda} \cdot \frac{N}{\sigma_w^2} \left[ \frac{v_1^H}{N} - N \rho_1^T (I + \frac{S_I}{\sigma_w^2} V_I^H V_I)^{-1} \frac{S_I}{\sigma_w^2} \cdot \frac{v_2^H}{N} \right]$$

which corresponds to the processor shown in Fig. 6.34.