

Lecture 04 – 05

Single-Stage Amplifiers

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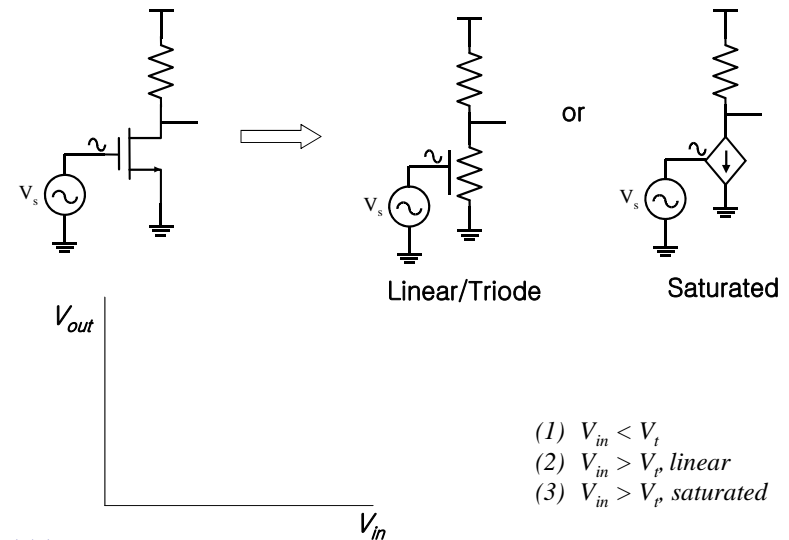
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1

Single Stage Amplifier

- Intuitive large signal analysis



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3

Introduction & Outline

- Why do we need an amplifier? Linear amplifier?
- Analyzing gain of amplifiers
- Common source amplifier
- Resistance of MOS
- Common gate amplifier
- Source Follower
- Cascode
- Current Mirrors

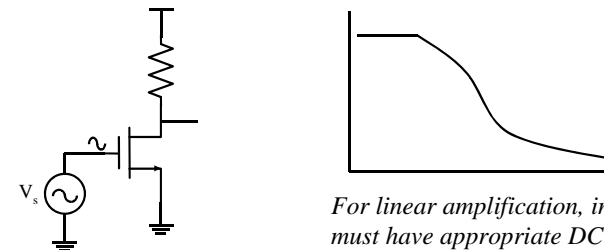
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2

Common Source Amplifier

- Quantitative large signal analysis



For linear amplification, input must be “small” and must have appropriate DC voltage (i.e. the transistor must be biased at a certain point) → Must be operated in the **saturated region!**

$$V_o = V_{DD} - R_D I_D$$

- (1) $V_{in} < V_t : I_D = 0$
 (2) $V_{in} > V_p : V_o > V_{in} - V_t : I_D =$
 (3) $V_{in} > V_p : V_o < V_{in} - V_t : I_D =$

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4

Obtaining A_v for *small enough* inputs

1. $V_o = V_{DD} - I_D(V_i)R_D \rightarrow \text{Obtain } v_o = f(v_i)$

2. $V_o = V_{DD} - I_D(V_i)R_D \rightarrow \text{Solve for } dv_o/dv_i$

3. Use Small Signal Model

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5

2. $A_v = dv_o/dv_i$

$$V_o = V_{DD} - I_D R_D$$

$$\begin{aligned} \frac{dV_o}{dV_i} &= -R_D \frac{dI_D}{dV_i} & g_m &= ? \\ &= -R_D g_m \end{aligned}$$

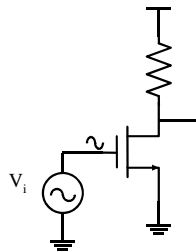
- Linearized analysis!
 \rightarrow Only works for "small" input signals under appropriate conditions

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7

1. Obtaining $v_o = f(v_i)$



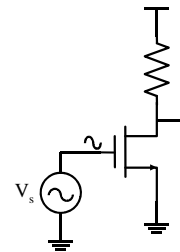
$$\begin{aligned} V_o &= V_{DD} - I_D R_D \\ &= V_{DD} - R_D \cdot k(V_i - V_{TH})^2 \\ &= V_{DD} - R_D \cdot k(V_{iDC} + v_i - V_{TH})^2 \end{aligned}$$

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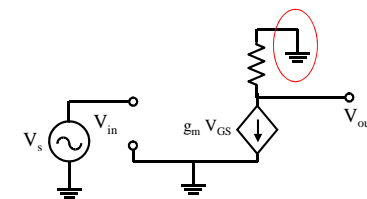
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6

3. Small Signal Model



Note: how about current source?



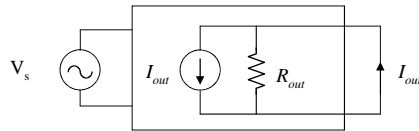
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8

$G_m R_{out}$

- In a linear circuit, the voltage gain is equal to $-G_m R_{out}$ where G_m is the transconductance of the circuit when the output is shorted to ground and R_{out} is the output resistance of the circuit when the input voltage is zero.

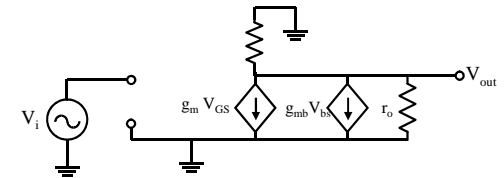
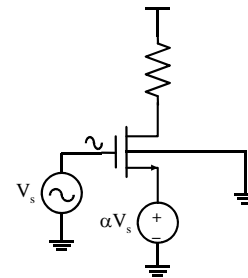


*Note $A_v = \frac{\sum \text{Output Resistance}}{\sum \text{Source Resistance}}$

First Order Approximation.
→ Good for first glance analysis only.
(Needs accuracy improvement.)

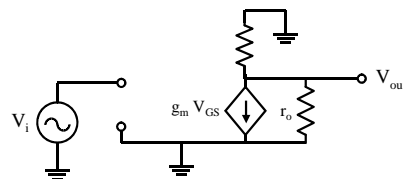
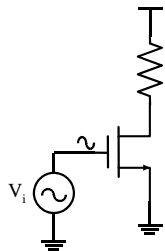
Non-Ideality : g_{mb}

- Body-effect : threshold voltage variation



Non-Ideality : g_{ds}

- Channel length modulation



Summary

- Transistor is a voltage-controlled current source
- For small enough signals at appropriate DC levels, transistor can be linearized.
- Non-idealities from channel length modulation and body-effect can decrease the gain

Biasing (Section 4.5 of Sedra/Smith)

1. Biasing with fixed V_{GS}
2. AC coupling capacitors with resistors
3. Current Sources

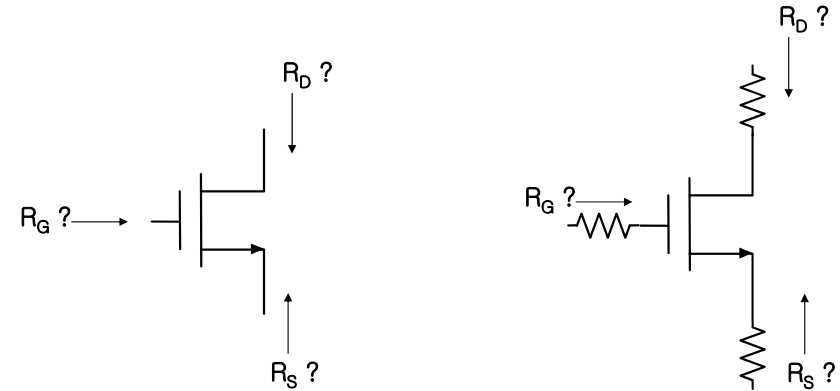
Note: Fig 4.33(a) of Sedra/Smith

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13

Small Signal Equivalent R in MOS



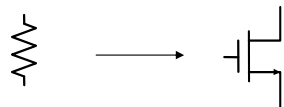
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15

How to replace the “Resistor”

- Why replace resistor?
 - Resistor is expensive and very inaccurate in CMOS process
 - Achieve higher gain (small signal)
- Replace with what?
 - Transistors are FREE!
 - Not a linear resistor



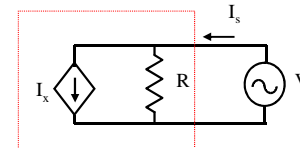
Which resistor
(transconductance) do I use?

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14

Resistance Multiplication



$$R_{\text{eff}} = R \left(1 - \frac{dI_x}{dI_s} \right)$$

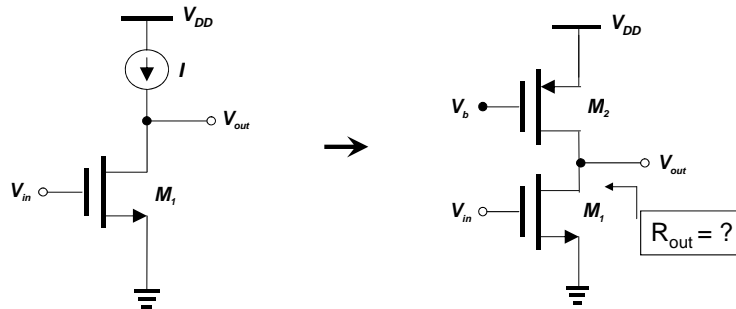
e.g.) $R = 1$
 $V_s = 1V \rightarrow 1.1V$
 $I_x = 1A \rightarrow 0.91A$

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16

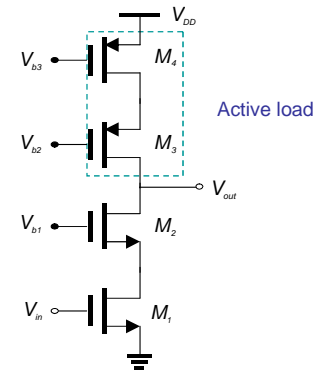
Active Load



CS stage with Active (current source) load

$A_v = ?$

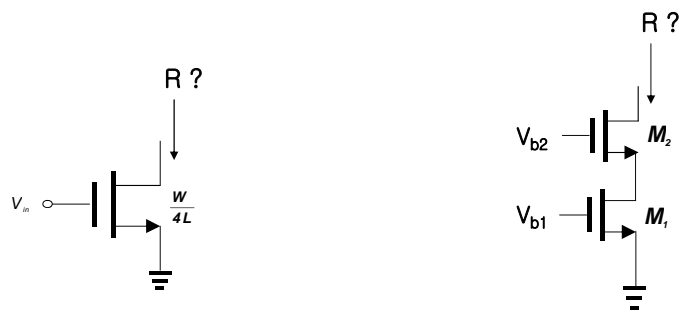
The MOS Cascode Amplifier



Active load

Increasing the output resistance

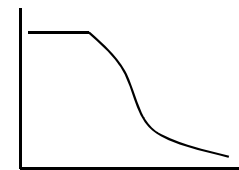
- Can we increase the resistance even higher?



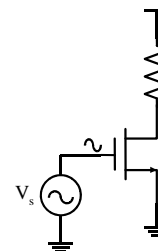
Increasing the channel length

Cascode Stage

Problem of Common-Source Amp

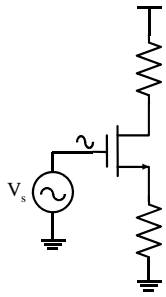


- Problem
 - limited linear range
 - input dependant gain
- Solution
 - reduce g_m
 - reduce sensitivity



Intuition: V_s rise $\rightarrow I_D$ rise $\rightarrow V_s$ rise $\rightarrow V_{gs}$ fall $\rightarrow I_d$ fall

Solution: Degenerate Resistor



$$A_v =$$

$$R_{out} =$$

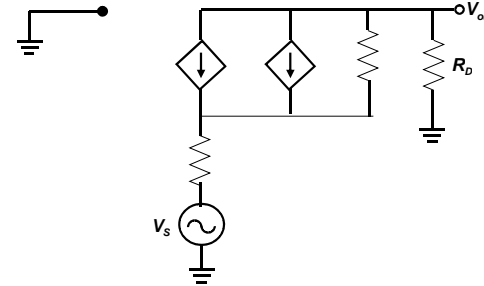
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21

Common Gate Amplifier

- small-signal model

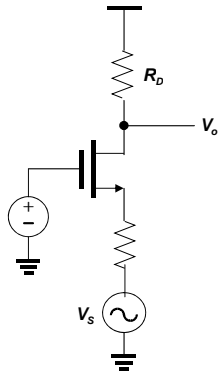


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23

Common Gate Amplifier



$$A_v = ?$$

$$R_{out} = ?$$

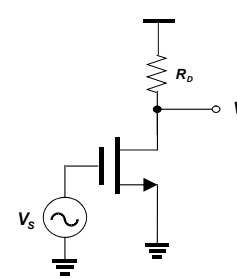
$$R_{in} = ?$$

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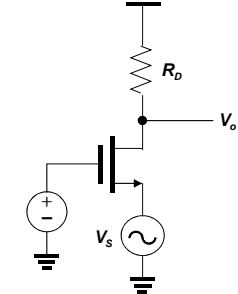
22

CS Amplifier vs CG Amplifier



Common-Source Stage

- High input resistance
- Large gain



Common-Gate Stage

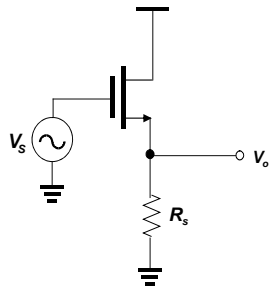
- Current buffering property
- Superior high frequency response

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24

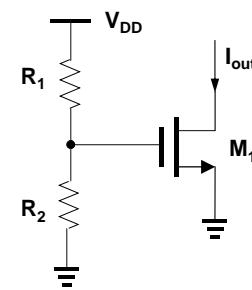
Source Follower



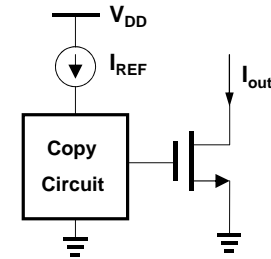
- Input resistance
- Output resistance
- Voltage gain

Problem! : I_D 's dependancy on V_s !!

How can we provide accurate current?



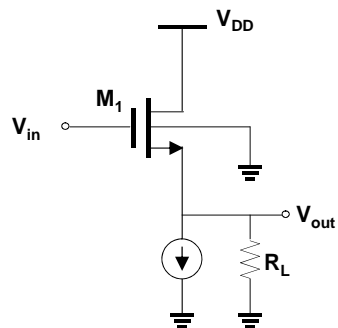
Definition of current by resistive divider



Conceptual means of copying currents

Could you set I_{out} stably?

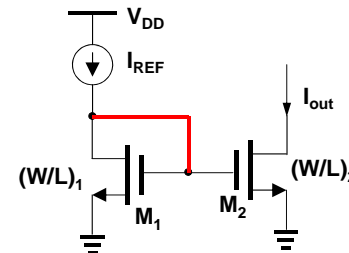
Source Follower with Current Source



- Input resistance
- Output resistance
- Voltage gain

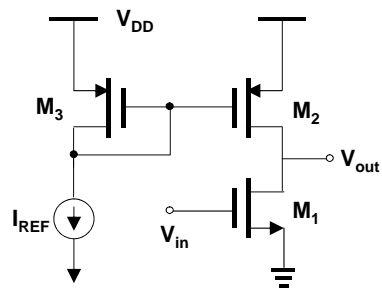
Note: Why can't this be used in common source amplifiers?

Basic Concept of Current Mirror



Basic current mirror

CS Amplifier with Active Load

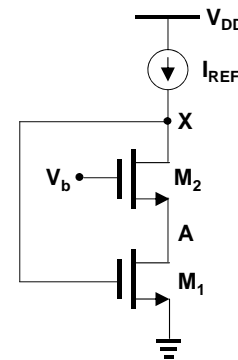


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29

*Cascode Mirror for Low-Voltage Operation

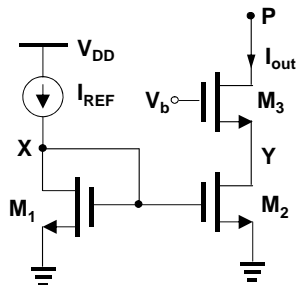


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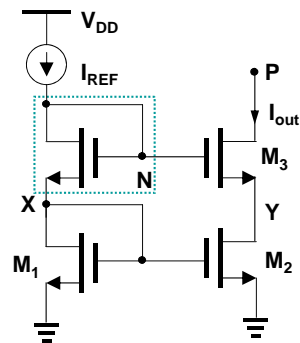
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31

Cascode Current Mirror



How can you set V_b ?



Cascode current mirror

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30

Summary

	Input resistance	Output resistance	Voltage gain
CS	∞		
CS with R_s	∞		
CG			
Source follower	∞		

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32