DIGITAL SIGNAL PROCESSING: COSC390

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1. **Introduction:** Types of filters (reading: ch. 3, ch. 5)
    - Butterworth
    - Bessel
    - Chebyshev
2. LTI systems and their properties (reading: ch. 5)
3. Convolution (reading: ch. 7)
    - Implementation with FFT
    - Impulse and step response

Future lectures will cover:

1. SNR of filtered signals: SNR
2. Common filter kernels (moving average, windows)
3. Recursive filters
4. FIR and IIR definitions

INTRODUCTION: TYPES OF FILTERS

Chapter 3 lists three types of anti-aliasing filters: Butterworth, Bessel, and Chebyshev. Filters are examples of linear, time-invariant (LTI) devices

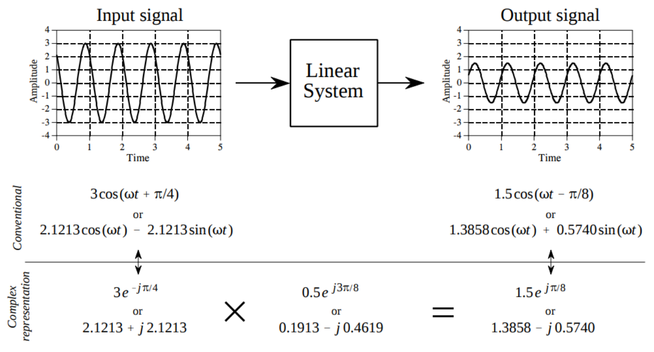

Figure 1: A linear, time-invariant system has special properties encapsulated by the *convolution operation.*

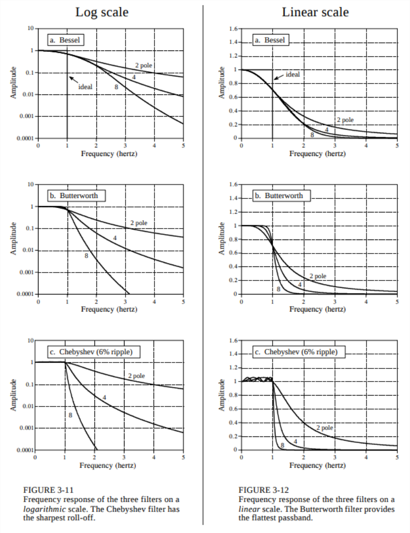

Figure 2: Comparison of transfer function magnitudes.

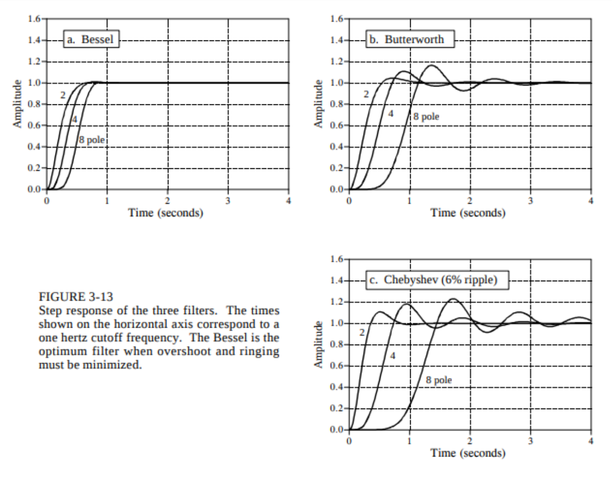

FIGURE 3-13
Step response of the three filters. The times shown on the horizontal axis correspond to a one hertz cutoff frequency. The Bessel is the optimum filter when overshoot and ringing must be minimized.

Figure 3: Comparison of transfer function step responses.

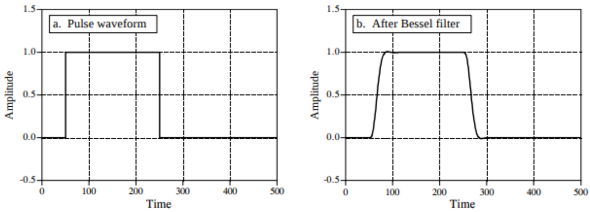

FIGURE 3-14
Pulse response of the Bessel and Chebyshev filters. A key property of the Bessel filter is that the rising and falling edges in the filter's output looking similar. In the jargon of the field, this is called *linear phase*. Figure (b) shows the result of passing the pulse waveform in (a) through a 4 pole Bessel filter. Both edges are smoothed in a similar manner. Figure (c) shows the result of passing (a) through a 4 pole Chebyshev filter. The left edge overshoots on the *top*, while the right edge overshoots on the *bottom*. Many applications cannot tolerate this distortion.

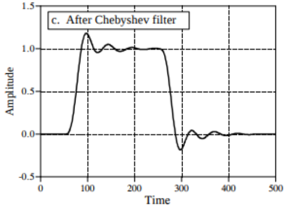

Figure 4: Comparison of transfer function pulse responses.

The single-pole Butterworth transfer functions are derived from the single-pole RC filter circuit:

$$H_{LP}(\omega) = \frac{\omega_0}{\omega_0 + j\omega} \tag{1}$$

$$H_{HP}(\omega) = \frac{\omega}{\omega - j\omega_0} \tag{2}$$

- What frequency causes a singularity in the transfer functions?
- What is the phase and group delay of this filter?

General expression for the transfer function of Butterworth filter (low-pass):

$$|H_{LP}(\omega)| = \frac{G_0}{\sqrt{1+\left(\frac{j\omega}{\omega_0}\right)^{2n}}} \tag{3}$$

The integer n is the number of poles. G_0 is the *gain*, and ω_0 is the corner or cutoff frequency.

- Can we plot the poles in the complex plane?
- What is the phase and group delay of this filter?

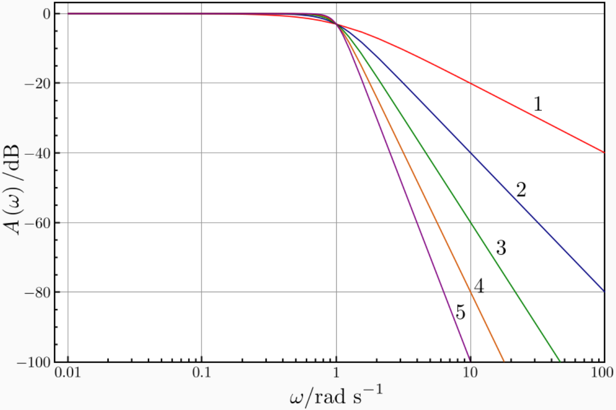

Figure 5: Gain of butterworth filters with *n* poles.

The n-th order low-pass Bessel filter transfer function is a ratio of **reverse Bessel polynomials:**

$$H(\omega) = \frac{\theta_n(0)}{\theta_n(j\omega/\omega_0)} \tag{4}$$

where the reverse Bessel polynomials $\theta_n(x)$ are given by

$$\theta_n(x) = \sum_{k=0}^{n} \frac{(n+k)!}{(n-k)!k!} \frac{x^{n-k}}{2^k} \tag{5}$$

- What is θ_3?
- How do we turn this into a high-pass filter?
- What are the pole locations of the 3rd-order Bessel filter?

The n-th order low-pass Chebyshev filter transfer function is

$$|H(\omega)| = \frac{1}{\sqrt{1 + \epsilon^2 T_n^2(\omega/\omega_0)}} \tag{6}$$

where the Chebyshev polynomials $T_n(x)$ are given by

$$T_n(x) = \cos(n \cos^{-1}(x)) \quad |x| < 1 \tag{7}$$

$$T_n(x) = \cosh(n \cosh^{-1}(x)) \quad x \geq 1 \tag{8}$$

$$T_n(x) = (-1)^n \cosh(n \cosh^{-1}(-x)) \quad x \leq 1 \tag{9}$$

- Can we plot $T_2(x)$ in Octave?
- Pole locations are interesting (next slide).

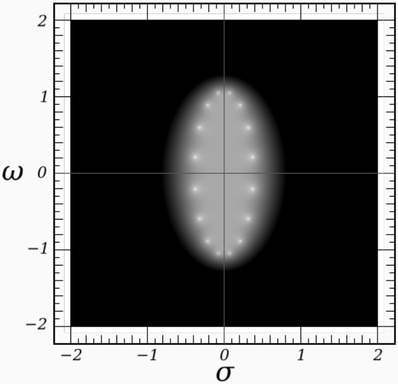

Figure 6: Eight-pole Chebyshev filter in the complex plane. The poles form an ellipse, due to the trigonometric nature of the definition of Chebyshev polynomials.

In the octave signal package, we can access the transfer functions of these filters:

```
pkg load signal;
[b1,a1] = butter(n,omega); %(e.g. include ”high”)
[b2,a2] = besself(n,omega);
[b3,a3] = cheby1(n,rp,omega); %rp pass-band ripple
x = (...); %data
y = filter(b1,a1,x);
```

Use **help** function on these for more information. The **filter** function is using the pole-zero information stored in the coefficients *a* and *b* to apply the transfer function to the data (more later).

INTRODUCTION: TYPES OF FILTERS, OCTAVE PROGRAMMING EXAMPLE

If you have the **signal** package, you can specify a n-th order butterworth filter with **butter** as above. Otherwise, you can write a small function using Eq. 3 for the digital response.

1. Create a white-noise sample (**randn**) of 10^4 samples. You can also specify a time vector of the same size.
2. Plot the spectrum of this noise using **fft**[1].).
3. **Filter** the noise with a 2nd-order low-pass butterworth filter, at a cutoff frequency of 0.2 f_s. Either use the **filter** function and fft, or multiply in the Fourier domain.
4. **Filter** the noise with a 2nd-order high-pass butterworth filter, at a cutoff frequency of 0.4 f_s.
5. Now do (3) then (4), and plot the spectrum. Repeat for (4) then (3). Do you see the same spectrum?

[1] See the **Aliasing.m** script in Unit 1

LTI SYSTEMS

Filters are LTI systems. Let's review the properties of LTI systems:

1. Continuous vs. discrete
2. Scaling property
3. Distributive property
4. Time-invariance (also causality)
5. Commutative property
6. Combination of properties

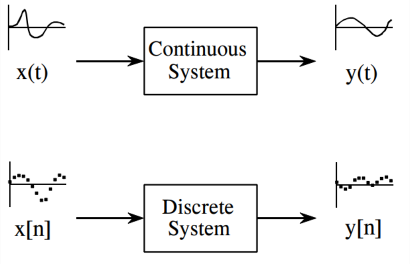

Figure 7: A continuous LTI system response to continuous data, and a discrete LTI system response to discrete data. All properties should hold both cases.

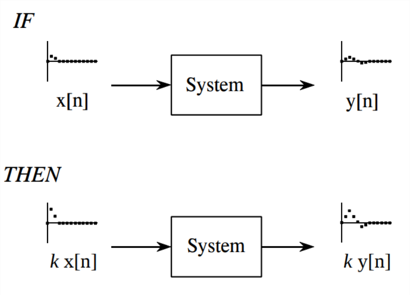

Figure 8: Scaling: If the data is scaled by a real constant, the output should be scaled by a real constant.

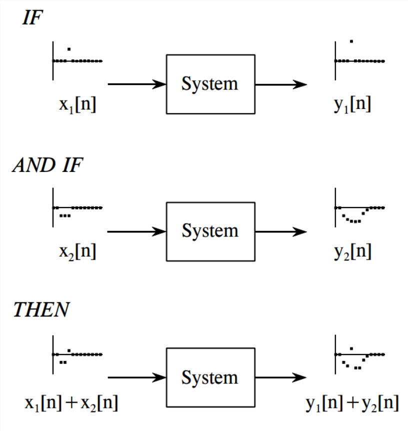

Figure 9: Distributive property: the LTI system should respond to each signal separately.

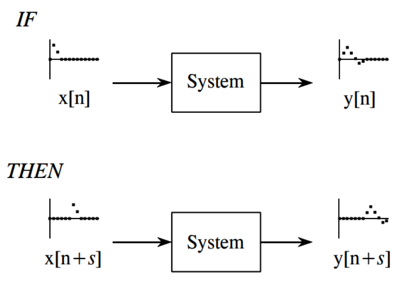

Figure 10: Time-invariance: the LTI system should respond when the signal arrives, independent of global time. Also, the filter should not respond *before* the signal arrives (causality).

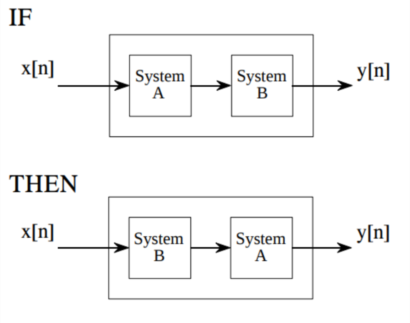

Figure 11: Commutative property: the LTI system should not depend on the existence of previous systems.

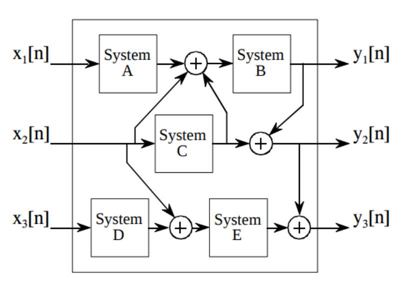

Figure 12: Combination of properties: an LTI system may be built from a combination of LTI systems.

Suppose **A(x)**, **B(x)**, **C(x)**, **D(x)**, and **E(x)** represent operators of LTI systems. What is the formula for a) y_1, b) y_2 and c) y_3? What is the the Fourier transform of the simplest output?

The **convolution** operator has all of the necessary properties of the LTI system operator. The convolution of data streams x_1 and x_2 can be implemented with the following basic algorithm:

```
x_1 = (...); %data1
x_2 = (...); %data2
y_1 = fft(x_1);
y_2 = fft(x_2);
Z_1 = y_1.*y_2;
result = real(ifft(Z_1));
```

Or, in one line:

```
result = real(ifft(fft(x_1).*fft(x_2)));
```

LTI SYSTEMS: PROGRAMMING WITH OCTAVE

Obtain the impulse response of the first-order low-pass butterworth filter. The impulse response is given by the identity:

$$x[n] \circ \delta[n] = x[n] \tag{10}$$

Recall the transfer function $h(\omega)$ for the first-order butterworth filter. Take the inverse Fourier transform of this function, to obtain the impulse response $h(t)$. What are the units of $h(t)$?

1. Plot this as a function of time in Octave.
2. Now define the delta function vector (impulse vector):

```
d = zeros(size(t));
d(1) = 1;
```

3. Convolve d with $h(t)$. What do you see?

The impulse response of a filter is the function with which the signal is convolved:

$$x[n] \circ f[n] = y[n] \tag{11}$$

1. Plot this as a function of time in Octave.
2. Now define the delta function vector (impulse vector):

```
d = zeros(size(t));
d(1) = 1;
```

3. Convolve d with $h(t)$. What do you see?

You can also use the built-in Octave function **conv** for convolution, but pay attention to sizes:

```
x_1 = (...); %data
x_2 = (...); %data
y = conv(x_1,x_2);
size(y)
size(x_1)
size(x_2)
```

Special case of **conv**:

```
y = conv(x_1,x_2,"same"); %default; "full"
size(y)
size(x_1)
size(x_2)
```

Convolve a sine-wave of a given frequency with the response of a first-order low-pass butterworth filter, and plot the result. Does the amplitude make sense? *Hint: don't forget factors of* Δt*!*

SPECIAL TOPIC: IMPULSE RESPONSE OF RF ANTENNAS

A paper on RF antenna response for UHE neutrino research:
`https://doi.org/10.1016/j.astropartphys.2014.09.002`