

Homework 3, Unit 0: Foundations and Fundamentals

Prof. Jordan C. Hanson

February 19, 2025

1 Memory Bank

- **Homogeneous system:** Let k be a constant, and let $s_{\text{in}}(t)$ and $s_{\text{out}}(t)$ be the input and output signals to a system S , respectively. S is *homogeneous* if:

$$s_{\text{out}}(t) = S[s_{\text{in}}(t)] \quad (1)$$

$$ks_{\text{out}}(t) = S[ks_{\text{in}}(t)] \quad (2)$$

- **Additive system:** Let $s_1(t)$ and $s_2(t)$ be two input signals to a system S , with outputs $s'_1(t)$ and $s'_2(t)$. S is *additive* if:

$$s'_1(t) = S[s_1(t)] \quad (3)$$

$$s'_2(t) = S[s_2(t)] \quad (4)$$

$$s'_1(t) + s'_2(t) = S[s_1(t) + s_2(t)] \quad (5)$$

- **Shift-invariant system:** Let $s_{\text{in}}(t)$ and $s_{\text{out}}(t)$ be input and output signals to a system S , and let t_0 be a constant. S is *shift invariant* if:

$$s_{\text{out}}(t) = S[s_{\text{in}}(t)] \quad (6)$$

$$s_{\text{out}}(t - t_0) = S[s_{\text{in}}(t - t_0)] \quad (7)$$

- $F(f) = \mathcal{F}\{f(t)\} = \int_{-\infty}^{\infty} f(t)e^{-2\pi jft} dt \dots$ The Fourier Transform.
- $\mathcal{F}^{-1}\{F(f)\} = \int_{-\infty}^{\infty} F(f)e^{2\pi jft} df \dots$ The Inverse Fourier Transform.
- The **Dirac δ -function** is a distribution defined by the following property:

$$f(t_0) = \int_{-\infty}^{\infty} f(t)\delta(t - t_0)dt \quad (8)$$

In words, the integral of a δ -function times a function f is the value of the function at t_0 .

- **Convolution:** this is an operation that characterizes the response $h[n]$ of a linear system.

$$y[i] = h[n] * x[n] = \sum_{j=0}^{M-1} h[j]x[i - j] \quad (9)$$

In words, the output at sample i is equal to the produce of the system response h and the input signal x , summed over the proceeding M samples (from $j = 0$ to $j = M - 1$).

2 Linear Systems

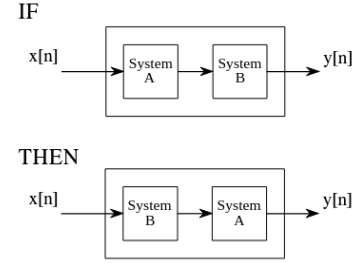


Figure 1: Linear systems **commute**.

1. Consider Fig. 1, which depicts two linear systems A and B. Symbolically, systems A and B **commute** if $A\{B\{x[n]\}\} = B\{A\{x[n]\}\}$. (a) Let $A\{x[n]\} = 2x[n] - 1$, and $B\{x[n]\} = 0.5x[n]$. Which system, A or B, is a linear system? For the system that is not linear, which linear property does it break? (b) Modify the non-linear system to make it linear, and show that A and B commute.

2. Consider Eq. 8 in the Memory Bank. Let $f(t) = a_1 \cos(2\pi f_1 t) + a_2 \cos(2\pi f_2 t)$, with $T_1 = 1/f_1$, $T_2 = 1/f_2$, and $f_2 = 2f_1$. Evaluate the following:

- $\int_{-\infty}^{\infty} f(t)\delta(t - T_1)dt$
- $\int_{-\infty}^{\infty} f(t)\delta(t - T_2)dt$

3. Let $f(t) = a\delta(t - t_0)$. (a) Show that the magnitude of the **Fourier transform** of this impulse is a . (b) Show that the phase angle, ϕ , is $-2\pi ft_0$. (c) Show that the group delay, $\tau_g = -d\phi/d\omega$ is t_0 .
2. For the output spectra of the previous exercise, low-pass and high-pass, calculate the group delays.²
3. (a) Show that the inverse Fourier transform of $S(f) = (a/2)(\delta(f - f_0) + \delta(f + f_0))$ is a cosine function. (b) Show that the inverse Fourier transform of $S(f) = (a/2j)(\delta(f - f_0) - \delta(f + f_0))$ is a sine function.
4. Let $\delta[n]$ represent a digital impulse: $[1000\ 0000]$ ¹. (a) If $y[n] = S[x[n]] = 0.5x[n - 2]$, what is $S[\delta[n]]$? (b) $y[n]$ is the *impulse response* of S . What is the *step response*, if the step input is $s[n] = [01111111]$?

3 Fourier Transforms and Basic Filters

1. Suppose we pass a signal $s(t)$ into a low-pass filter. The signal as a function of frequency is $S(f)$, the Fourier transform of $s(t)$. The output of the low-pass filter will be $S(f)$ times $1/(1 + j\omega\tau)$, where $\omega = 2\pi f$, and $\tau = RC$. That is, the output will be $S(f)/(1 + j\omega\tau)$. (a) Calculate the Fourier transform $S(f)$, if $s(t) = a\delta(t - t_0)$ (as we did in class). (b) Suppose we pass our impulse $s(t)$ into a low-pass filter. What is the magnitude of the output, as a function of frequency? (c) Repeat this exercise, but with a high-pass filter response: $j\omega\tau/(1 + j\omega\tau)$.

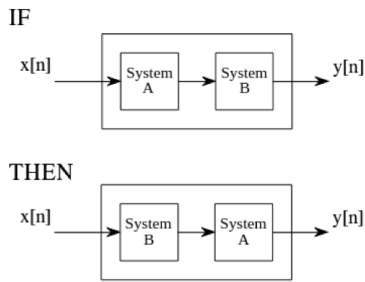
4 Convolution and Octave Code

1. For the following exercises, use Eq. 9. Let the digital impulse be $\delta[n]$ which is 1 for $n = 0$, and 0 if $n \neq 0$. For example, $\delta[n - 5]$ is 1 when $n = 5$. (a) Show that if $x[n] = \delta[n]$, $y[n] = h[n] * x[n] = h[n]$. That is, if the input data is an impulse, the output is the system response. (b) Show that if the input impulse is shifted ($x[n] = \delta[n - n_0]$), the output is $h[n]$, shifted by the same amount.
2. In **octave**, use the `conv` function to convolve a 440 Hz sine wave with a $\delta[n - n_0]$ impulse. Shift the phase of the sine output by varying n_0 .

¹Let the index for data in this list of numbers start with $n = 0$.

²Hint: multiply the numerator and denominator of ratios by the complex conjugate of the denominator, to aid in splitting the complex expression into real and imaginary parts.

2.

Figure 1: Linear systems **commute**.

1. Consider Fig. 1, which depicts two linear systems A and B. Symbolically, systems A and B **commute** if $A\{B\{x[n]\}\} = B\{A\{x[n]\}\}$. (a) Let $A\{x[n]\} = 2x[n] - 1$, and $B\{x[n]\} = 0.5x[n]$. Which system, A or B, is a linear system? For the system that is not linear, which linear property does it break? (b) Modify the non-linear system to make it linear, and show that A and B commute.

a). To be linear system, k is constant

$$\textcircled{1}. S\{kx[n]\} = kS\{x[n]\}$$

$$\textcircled{2}. S\{x_1[n] + x_2[n]\} = S\{x_1[n]\} + S\{x_2[n]\}$$

consider $A\{x[n]\} = 2x[n] - 1$

$$\textcircled{1}. A\{kx[n]\} = 2(kx[n]) - 1$$

$$= 2kx[n] - 1$$

$$kA\{x[n]\} = k(2x[n] - 1)$$

$$= 2kx[n] - k$$

\neq

Thus $A\{x[n]\}$ not linear.

Fails Homogeneity property

consider $B\{x[n]\} = 0.5x[n]$

$$\textcircled{1}. B\{kx[n]\} = 0.5kx[n]$$

$$kB\{x[n]\} = 0.5kx[n]$$

$$\textcircled{2}. B\{x_1[n] + x_2[n]\} = 0.5(x_1[n] + x_2[n])$$

$$= 0.5x_1[n] + 0.5x_2[n]$$

$$B\{x_1[n]\} + B\{x_2[n]\} = 0.5x_1[n] + 0.5x_2[n]$$

Thus. System B is linear

b). Change $A\{x[n]\} = 2x[n]$

①. $A\{kx[n]\} = 2kx[n]$

$= kA\{x[n]\} = 2kx[n]$

②. $A\{x_1[n] + x_2[n]\} = 2x_1[n] + 2x_2[n]$

$A\{x_1[n]\} + A\{x_2[n]\} = 2x_1[n] + 2x_2[n]$

Now A is linear system.

Check A . B commutativity.

$A\{B\{x[n]\}\} = B\{A\{x[n]\}\}$

$= A\{0.5x[n]\} \quad | \quad = B\{2x[n]\}$

$= 2 \cdot (0.5x[n]) \quad | \quad = 0.5(2x[n])$

$= \underline{\underline{x[n]}} \quad | \quad = \underline{\underline{x[n]}}$

LHS = RHS, Thus Commutative

2.2.

2. Consider Eq. 8 in the Memory Bank. Let $f(t) = a_1 \cos(2\pi f_1 t) + a_2 \cos(2\pi f_2 t)$, with $T_1 = 1/f_1$, $T_2 = 1/f_2$, and $f_2 = 2f_1$. Evaluate the following:

- $\int_{-\infty}^{\infty} f(t) \delta(t - T_1) dt$
- $\int_{-\infty}^{\infty} f(t) \delta(t - T_2) dt$

$$\int_{-\infty}^{\infty} f(t) \delta(t - t_0) dt = f(t_0)$$

$$T_1 = \frac{1}{f_1}$$

$$T_2 = \frac{1}{f_2} \Rightarrow T_2 = \frac{1}{2f_1} = \frac{T_1}{2}$$

$$\textcircled{1} \int_{-\infty}^{\infty} f(t) \delta(t - T_1) dt = f(T_1)$$

$$= a_1 \cos(2\pi f_1 T_1) + a_2 \cos(2\pi f_2 T_1)$$

$$= a_1 \cos(2\pi f_1 \cdot \frac{1}{f_1}) + a_2 \cos(2\pi 2f_1 \cdot \frac{1}{f_1})$$

$$= a_1 \underbrace{\cos(2\pi)}_{=1} + a_2 \underbrace{\cos(4\pi)}_{=1}$$

$$= a_1 + a_2$$

$$\textcircled{2} \int_{-\infty}^{\infty} f(t) \delta(t - T_2) dt = f(T_2)$$

$$= a_1 \cos(2\pi f_1 T_2) + a_2 \cos(2\pi f_2 T_2)$$

$$= a_1 \cos(2\pi f_1 \cdot \frac{1}{2f_1}) + a_2 \cos(2\pi 2f_1 \cdot \frac{1}{2f_1})$$

$$= \underbrace{a_1 \cos(\pi)}_{=-1} + \underbrace{a_2 \cos(2\pi)}_{=1}$$

$$= a_2 - a_1$$

3. Let $f(t) = a\delta(t - t_0)$. (a) Show that the magnitude of the **Fourier transform** of this impulse is a . (b) Show that the phase angle, ϕ , is $-2\pi f t_0$. (c) Show that the group delay, $\tau_g = -d\phi/d\omega$ is t_0 .

$$\begin{aligned}
 a). \quad F(f) &= \int_{-\infty}^{\infty} f(t) e^{-2\pi j f t} dt = \int_{-\infty}^{\infty} a \delta(t - t_0) e^{-2\pi j f t} dt \\
 &= a \int_{-\infty}^{\infty} \delta(t - t_0) e^{-2\pi j f t} dt \\
 &= a e^{-2\pi j f t_0} \\
 \left| a e^{-2\pi j f t} \right| &= \sqrt{a e^{-2\pi j f t_0} a e^{2\pi j f t_0}} \\
 &= \sqrt{a^2 e^0} \\
 &= a
 \end{aligned}$$

$$\begin{aligned}
 b). \quad \phi &= \text{Im} (a e^{-2\pi j f t_0}) \\
 &= \underline{-2\pi f t_0}
 \end{aligned}$$

$$\begin{aligned}
 c). \quad T_g &= - \frac{d\phi}{d\omega} \quad \omega = 2\pi f \\
 \phi &= -\omega t_0
 \end{aligned}$$

$$- \left(\frac{d}{d\omega} (-\omega t_0) \right) = t_0$$

2.4.

4. Let $\delta[n]$ represent a digital impulse: $[1000\ 0000]^1$. (a) If $y[n] = S[x[n]] = 0.5x[n-2]$, what is $S[\delta[n]]$? (b) $y[n]$ is the *impulse response* of S . What is the *step response*, if the step input is $s[n] = [01111111]$?

$$a). \quad \delta[n-2] = \begin{cases} 1 & n=2 \\ 0 & n \neq 2 \end{cases}$$

$$S[\delta[n]] = 0.5 \delta[n-2]$$

$$= [0, 0, 0.5, 0, 0, 0, 0, 0]$$

$$b). \quad h[n-2] = \begin{cases} 1 & n \geq 2 \\ 0 & n < 2 \end{cases}$$

$$S[h[n]] = [0, 0, 0.5, 0.5, 0.5, 0.5, 0.5, 0.5]$$

3. 1. Suppose we pass a signal $s(t)$ into a low-pass filter. The signal as a function of frequency is $S(f)$, the Fourier transform of $s(t)$. The output of the low-pass filter will be $S(f)$ times $1/(1 + j\omega\tau)$, where $\omega = 2\pi f$, and $\tau = RC$. That is, the output will be $S(f)/(1 + j\omega\tau)$. (a) Calculate the Fourier transform $S(f)$, if $s(t) = a\delta(t - t_0)$ (as we did in class). (b) Suppose we pass our impulse $s(t)$ into a low-pass filter. What is the magnitude of the output, as a function of frequency? (c) Repeat this exercise, but with a high-pass filter response: $j\omega\tau/(1 + j\omega\tau)$.

$$a). S(f) = F\{s(t)\} = \int_{-\infty}^{\infty} s(t) e^{-j2\pi ft} dt$$

$$s(t) = a\delta(t - t_0)$$

$$\begin{aligned} F\{s(t)\} &= a \int_{-\infty}^{\infty} \delta(t - t_0) e^{-j2\pi ft} dt \\ &= a e^{-j2\pi ft_0} \end{aligned}$$

$$b). H_{lp} = \frac{1}{1 + j\omega\tau} \quad \omega = 2\pi f$$

$$\begin{aligned} \text{output} &= S(f) \cdot H_{lp} \\ &= a e^{-j2\pi ft_0} \cdot \frac{1}{1 + j2\pi f\tau} \end{aligned}$$

$$|\text{output}| = \left| a e^{-j2\pi ft_0} \right| \left| \frac{1}{1 + j2\pi f\tau} \right|$$

$$= |a| \cdot \frac{1}{\sqrt{1 + (2\pi f\tau)^2}}$$

$$= \frac{a}{\sqrt{1 + (2\pi f\tau)^2}}$$

$$c). H_{hp} = \frac{j\omega\tau}{1 + j\omega\tau} = \frac{j2\pi f\tau}{1 + j2\pi f\tau}$$

$$\text{output} = a e^{-j2\pi ft_0} \cdot \frac{j2\pi f\tau}{1 + j2\pi f\tau}$$

$$|\text{output}| = |a| \cdot \left| \frac{j2\pi f\tau}{1 + j2\pi f\tau} \right|$$

$$= a \cdot \frac{2\pi f\tau}{\sqrt{1 + (2\pi f\tau)^2}}$$

$$= \frac{2\pi f\tau}{\sqrt{1 + (2\pi f\tau)^2}}$$

3.2

2. For the output spectra of the previous exercise, low-pass and high-pass, calculate the group delays.²

①. LPF :

$$\begin{aligned}
 H_{lp} &= \frac{1}{1+j\omega\tau} \\
 &= \frac{1}{1+j\omega\tau} \times \frac{1-j\omega\tau}{1-j\omega\tau} \\
 &= \frac{1-j\omega\tau}{1+(\omega\tau)^2}
 \end{aligned}$$

$$\phi_{lp}(\omega) = \text{Im}(H_{lp}) = \tan^{-1}(-\omega\tau)$$

$$\begin{aligned}
 \tau_g &= -\frac{d}{d\omega} \tan^{-1}(-\omega\tau) \\
 &= -\frac{-\tau}{1+(\omega\tau)^2} \\
 &= \frac{\tau}{1+(\omega\tau)^2} = \frac{\tau}{1+(2\pi f\tau)^2}
 \end{aligned}$$

②. HPF :

$$\begin{aligned}
 H_{hp} &= \frac{j\omega\tau}{1+j\omega\tau} \\
 &= \frac{j\omega\tau}{1+j\omega\tau} \cdot \frac{1-j\omega\tau}{1-j\omega\tau} \\
 &= \frac{j\omega\tau(1-j\omega\tau)}{1+(\omega\tau)^2}
 \end{aligned}$$

$$\phi_{hp}(\omega) = \text{Im}(H_{hp}) = \tan^{-1}\left(\frac{\omega\tau}{1+(\omega\tau)^2}\right)$$

$$\begin{aligned}
 \tau_g &= -\frac{d}{d\omega} \tan^{-1}\left(\frac{\omega\tau}{1+(\omega\tau)^2}\right) \\
 &= \frac{\tau}{1+(\omega\tau)^2} = \frac{\tau}{1+(2\pi f\tau)^2}
 \end{aligned}$$

3. (a) Show that the inverse Fourier transform of $S(f) = (a/2)(\delta(f - f_0) + \delta(f + f_0))$ is a cosine function. (b) Show that the inverse Fourier transform of $S(f) = (a/2j)(\delta(f - f_0) - \delta(f + f_0))$ is a sine function.

$$s(t) = \mathcal{F}^{-1} \{ S(f) \} = \int_{-\infty}^{\infty} S(f) e^{j2\pi ft} df$$

a). $S(f) = \frac{a}{2} (\delta(f - f_0) + \delta(f + f_0))$

$$s(t) = \int_{-\infty}^{\infty} \frac{a}{2} (\delta(f - f_0) + \delta(f + f_0)) e^{j2\pi ft} df$$

$$= \frac{a}{2} (e^{2\pi j f_0 t} + e^{-2\pi j f_0 t})$$

$$= \frac{a}{2} (e^{j\theta} + e^{-j\theta}) \Rightarrow \theta = 2\pi f_0 t$$

$$= \frac{a}{2} \cdot 2 \cos \theta$$

$$= a \cos(2\pi f_0 t)$$

b). $S(f) = \frac{a}{2j} (\delta(f - f_0) - \delta(f + f_0))$

$$s(t) = \int_{-\infty}^{\infty} \frac{a}{2j} (\delta(f - f_0) - \delta(f + f_0)) e^{j2\pi ft} df$$

$$= \frac{a}{2j} (e^{2\pi j f_0 t} - e^{-2\pi j f_0 t})$$

$$= \frac{a}{2j} \cdot (2j \sin \theta) \Rightarrow \theta = 2\pi f_0 t$$

$$s(t) = a \sin(2\pi f_0 t)$$

4 Convolution and Octave Code

1. For the following exercises, use Eq. 9. Let the digital impulse be $\delta[n]$ which is 1 for $n = 0$, and 0 if $n \neq 0$. For example, $\delta[n - 5]$ is 1 when $n = 5$. (a) Show that if $x[n] = \delta[n]$, $y[n] = h[n] * x[n] = h[n]$. That is, if the input data is an impulse, the output is the system response. (b) Show that if the input impulse is shifted ($x[n] = \delta[n - n_0]$), the output is $h[n]$, shifted by the same amount.

$$y[n] = h[n] \cdot x[n] = \sum_{j=0}^{M-1} h[j] x[n-j]$$

$$\textcircled{1} x[n] = \delta[n]$$

$$\delta[n] = \begin{cases} 1 & n=0 \\ 0 & n \neq 0 \end{cases}$$

$$y[n] = \sum_{j=0}^{M-1} h[j] \delta[n-j]$$

$$\sum_j g[j] \delta[n-j] = g[n]$$

$$\text{Thus } y[n] = h[n] \text{ when } j=n, \text{ else } = 0$$

$$\textcircled{2} x[n] = \delta[n - n_0]$$

$$y[n] = \sum_{j=0}^{M-1} h[j] \delta[n - n_0 - j] \Rightarrow \sum_{j=0}^{M-1} h[j] \delta[(n - n_0) - j]$$

$$y[n] = h[n - n_0] \text{ when } j = n - n_0, \text{ else } = 0$$

2. In octave, use the `conv` function to convolve a 440 Hz sine wave with a $\delta[n - n_0]$ impulse. Shift the phase of the sine output by varying n_0 .

