7.5 Mathematical Models of Discrete-Time Systems

- 7.5.1 Linear Time-Invariant Difference Equations
 - (1) Definition of difference
- (1) Forward
- 2 Backward

- (2) The difference equation and its solving method
- **1** Iteration
- 2 Z-transformation

- 7.5.2 Impulse-Transfer Function
 - (1) Definition (2) Properties
- (3) Limitation

- 7.5.3 Impulse Transfer Function of Open-Loop Systems
- (1) Switch between factors
 (2) No switch between factors
 (3) With ZOH
- 7.5.4 Impulse Transfer Function of **Closed-Loop Systems**
- (1) General Method(2) Mason's formula

Chapter 7 Analysis and Design of Linear Discrete-Time System (Sampled-data System)

- 7.1 Introduction
- 7.2 The Sampling Process and Sampling Theorem
- 7.3 Signal Recovery and Zero-Order Hold
- 7.4 Z-Transform and Inverse Z Transform
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7.6 Performance Analysis of Discrete-Time Systems

> Stability

- 貌
- > Dynamic Performance 🖖
- > Steady-state Errors



7.6.1 Stability of Discrete systems

1. Preliminaries

Stability is the most important performance of a system.

When we sampled a continuous systems, we still have a "continuous" system → the same properties hold as before:

A necessary and sufficient condition for a feedback system to be stable is that <u>all the poles of the system transfer function</u> have negative real parts.

Now, we introduced the variable $z=e^{Ts}$, how does stability look in the new variable?

First, we must understand the relationship between s-domain and z-domain.

2. s-Domain to z-Domain Mapping

Because
$$z = e^{sT}$$
, let $s = \sigma + j\omega$ then $|z| = e^{\sigma T}$

$$\angle z = \omega T$$
, thus
$$\begin{cases} \sigma > 0 & |z| > 1 \\ \sigma < 0 & |z| < 1 \\ \sigma = 0 & |z| = 1 \end{cases}$$

The mapping relationship can be described as in the following figures.

$$z = e^{Ts} = e^{T(\sigma + j\omega)} = e^{T\sigma} e^{j\omega T}$$

$$\sigma < 0 \quad |z| = \frac{1}{e^{-T\sigma}} < 1$$

$$\sigma = 0 \quad |z| = 1$$

$$\sigma > 0 \quad |z| = e^{T\sigma} > 1$$

$$Re$$
Re
$$Re$$

For a C.L.discrete-time system with unit feedback, the impulse transfer function is: C(z)

$$\Phi(z) = \frac{G(z)}{1 + G(z)}$$

Its characteristic function is: 1+G(z)=0

Necessary and Sufficient Condition for Stability of Linear Discrete-Time Systems

-All poles of $\Phi(z)$ lie in the unit circle of z plane.

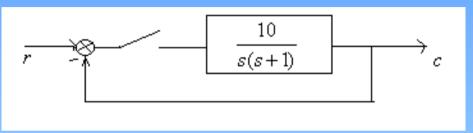
$$\Phi(z) = \frac{M(z)}{D(z)} = \frac{\prod_{i=1}^{m} (z - \alpha_i)}{\prod_{j=1}^{n} (z - \beta_j)} = \sum_{j=1}^{n} \frac{C_j z}{z - \beta_j} = K(z)$$

$$c(k) = \sum_{j=1}^{n} C_{j} \beta_{j}^{k} \stackrel{k \to \infty}{=} 0 \qquad \qquad \left| \beta_{j} \right| < 1 \qquad \qquad -- \text{Necessity}$$

$$c(k) = \sum_{j=1}^{n} C_{j} \beta_{j}^{k} \stackrel{k \to \infty}{=} 0 \qquad |\beta_{j}| < 1 \qquad -\text{Necessity}$$

$$c^{*}(t) = \sum_{k=0}^{\infty} \left(\sum_{j=1}^{n} C_{j} \beta_{j}^{k} \right) \cdot \delta(t - kT) \qquad -\text{Sufficiency}$$

Example The discrete-time system is shown as the following figure, suppose T=1, is the system stable?



Solution:
$$G(z) = Z\left[\frac{10}{s(s+1)}\right] = \frac{6.32z}{(z-1)(z-0.368)}$$

$$1+G(z)=0 \Rightarrow z^2+4.952z+0.368=0$$

$$\Rightarrow z_1=-0.076 \quad z_2=-4.876$$

$$\therefore |z_2|>1 \quad \text{So the system is unstable.}$$

3. The Stability Criterion of Discrete-Time Systems

- For continuous-time systems, we can use Routh
 criterion to determine the stability of the system, where
 the stable area is on LHP (left-hand-plane) of [s]-domain.
- Unfortunately, for discrete-time systems, the stable area is unit circle, not LHP of [z]-domain, we cannot directly apply the Routh criterion as we have to test on something else than LHP.

w-transformation and Routh criterion in w-domain

We find a transformation that maps the unit circle back onto the LHP while maintaining the algebraic structure of rational functions.

A particular transformation that will accomplish this would be the bilinear transformation:

$$z = \frac{w+1}{w-1}$$
, $w = \frac{z+1}{z-1}$

Suppose

$$z = x + j y \qquad w = u + j v$$

Then

$$w = \frac{z+1}{z-1} = \frac{(x+1)+jy}{(x-1)+jy}$$

$$u + jv = \frac{(x^2 + y^2) - 1}{(x - 1)^2 + y^2} = \frac{2y}{(x - 1)^2 + y^2}$$

$$\therefore u = \frac{x^2 + y^2 - 1}{(x - 1)^2 + y^2}$$

$$w = \frac{z+1}{z-1} = \frac{x+1+jy}{x-1+jv} = \frac{x^2-1+y^2-j2y}{(x-1)^2+v^2} = u+jv$$

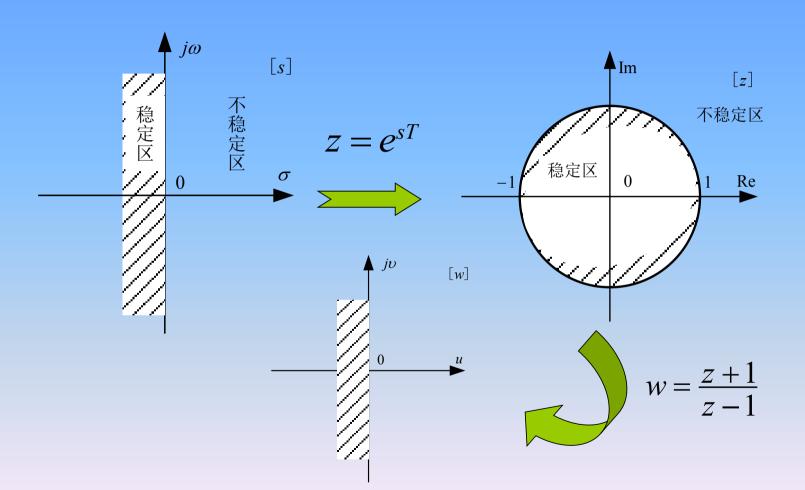
[w] imaginary axis
$$u = 0 = \frac{x^2 + y^2 - 1}{(x - 1)^2 + y^2} = 0$$

$$\Rightarrow x^2 + y^2 = 1$$
 [z] unit circle

$$x^{2} + y^{2} \begin{cases} < 1 & \text{in z-plane} \\ > 1 \end{cases}$$

$$u < 0 & \text{in w-plane} \\ u > 0$$

$$\begin{cases} u < 0 \\ u > 0 \end{cases}$$
 in w-plane



Example 1 Determine the stability from the characteristic equation of a discrete system.

$$D(z) = 45z^{3} - 117z^{2} + 119z - 39 = 0$$

$$\int z = (w+1)/(w-1)$$

$$= 45(\frac{w+1}{w-1})^{3} - 117(\frac{w+1}{w-1})^{2} + 119(\frac{w+1}{w-1}) - 39 = 0$$

$$D(w) = 45(w+1)^{3} - 117(w+1)^{2}(w-1) + 119(w+1)(w-1)^{2} - 39(w-1)^{3} = 0$$

$$D(w) = w^{3} + 2w^{2} + 2w + 40 = 0$$

$$Routh \qquad w^{3} \qquad 1 \qquad 2$$

$$w^{2} \qquad 2 \qquad 40$$

$$w^{1} \qquad -18 \qquad \qquad Unstable!$$

$$w^{0} \qquad 40$$

Example 2 Consider the discrete-time system as shown in the figure, if T=0.1, determine the stability of the system.

$$\begin{array}{c|c}
\hline
 & 1-e^{-rs} \\
\hline
 & s \\
\hline
 & s(0.1s+1)(0.05s+1)
\end{array}$$

Solution:

$$G(z) = \frac{z - 1}{z} Z \left[\frac{2}{s^2 (0.1s + 1)(0.05s + 1)} \right]$$

$$= \frac{z - 1}{z} \left[-\frac{0.3z}{z - 1} + \frac{0.4z}{(z - 1)^2} + \frac{0.4z}{z - e^{-10T}} - \frac{0.1z}{z - e^{-20T}} \right]$$

$$1 + G(z) = 0$$
 $z^3 - 1.001z^2 + 0.3356z + 0.0535 = 0$

$$z = \frac{w+1}{w-1}$$
2.33w³ + 3.68w² + 1.65w + 0.34 = 0
$$w^{3} \quad 2.33 \quad 1.65$$

$$w^{2} \quad 3.68 \quad 0.34$$
The element of the system of the system

The elements in the first column are all positive, the system is stable.

Example 3 Consider the range of K.

system shown in the figure (T=1). Determine the stable range of K.

Consider the
$$e^*$$
 e^*
 e^*

Routh criterion in w domain

$$G(z) = Z \left[\frac{1 - e^{-Ts}}{s} \cdot \frac{K}{s(s+1)} \right] = (1 - z^{-1})K \cdot Z \left[\frac{1}{s^2(s+1)} \right]$$

$$= \frac{(z-1)K}{z} \cdot Z \left[\frac{1}{s^2} - \frac{1}{s} + \frac{1}{s+1} \right] = \frac{(z-1)K}{z} \cdot \left[\frac{Tz}{(z-1)^2} - \frac{z}{z-1} + \frac{z}{z-e^{-T}} \right]$$

$$= K \left[\frac{(T-1+e^{-T})z + (1-e^{-T} - Te^{-T})}{(z-1)(z-e^{-T})} \right] \stackrel{T=1}{=} \frac{0.368K(z+0.718)}{(z-1)(z-0.368)}$$

$$\Phi(z) = \frac{G(z)}{1 + G(z)} = \frac{0.368K(z + 0.718)}{z^2 + (0.368K - 1.368)z + (0.264K + 0.368)}$$

$$\Phi(z) = \frac{G(z)}{1+G(z)} = \frac{0.368K(z+0.718)}{z^2+(0.368K-1.368)z+(0.264K+0.368)}$$

$$D(z) = z^2+(0.368K-1.368)z+(0.264K+0.368) = 0$$

$$\begin{vmatrix} z = \frac{w+1}{w-1} \\ = (\frac{w+1}{w-1})^2 + (0.368K-1.368)(\frac{w+1}{w-1}) + (0.264K+0.368) = 0 \end{vmatrix}$$

$$(w+1)^2+(0.368K-1.368)(w+1)(w-1)+(0.264K+0.368)(w-1)^2 = 0$$

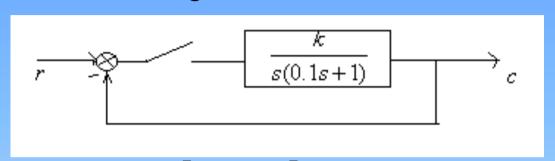
$$D(w) = 0.632Kw^2+(1.264-0.528K)w+(2.736-0.104K) = 0$$

$$\begin{cases} K>0 \\ 1.264-0.528K>0 \\ 2.736-0.104K>0 \end{cases}$$

$$K > 0$$

$$K < 2.394 \\ 2.736-0.104K>0$$

Exercise Consider the system shown in the figure (T=0.1). Determine the stable range of K. If it is a continuous-time system, Determine the stable range of K



Thin:

$$G(z) = Z \left[\frac{k}{s(0.1s+1)} \right] = \frac{0.632kz}{z^2 - 1.368z + 0.368}$$

$$1 + G(z) = 0 \qquad z^2 + (0.632k - 1.368)z + 0.368 = 0$$

$$z = \frac{w+1}{w-1} \qquad 0.632kw^2 + 1.264w + 2.736 - 0.632k = 0$$

For continuous-time system, 0 < k, then system is stable; For discrete-time system, 0 < k < 4.33, stable.

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Summary

s-Domain to z-Domain Mapping

Necessary and Sufficient Condition for Stability of Linear Discrete-Time Systems

— All poles of $\Phi(z)$ lie in the unit circle of z plane Routh criterion in w domain (Generalized Routh Criterion)