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1 Introduction

2 Moving average and filters

A lot of papers describes the definition and the properties of moving average and linear filters (see for example Ladiray (2018)). Here we summarize some of the main results.

Let p et f two integers, a moving average M_θ or M is defined by a set of coefficients $\theta = (\theta_{-p}, \dots, \theta_f)'$ such as:

$$M_\theta(X_t) = \sum_{k=-p}^{+f} \theta_k X_{t+k}$$

- $p + f + 1$ is called the *moving average order*.
- When $p = f$ the moving average is said to be *centered*. If we also have $\forall k : \theta_{-k} = \theta_k$, the moving average M_θ is said to be *symmetric*. In this case, the quantity $h = p = f$ is called the *bandwidth*.

2.1 Gain and phase shift functions

Let $X_t = e^{i\omega t}$, the result of the moving average M_θ in X_t is:

$$Y_t = M_\theta X_t = \sum_{k=-p}^{+f} \theta_k e^{i\omega(t+k)} = \left(\sum_{k=-p}^{+f} \theta_k e^{i\omega k} \right) \cdot X_t.$$

The function $\Gamma_\theta(\omega) = \sum_{k=-p}^{+f} \theta_k e^{i\omega k}$ is called the *transfer function*. It can be rewritten as:

$$\Gamma_\theta(\omega) = G_\theta(\omega) e^{-i\Phi_\theta(\omega)}$$

where $G_\theta(\omega) = |\Gamma_\theta(\omega)|$ is the *gain* or *amplitude* function and $\Phi_\theta(\omega)$ is the *phase shift* or *time shift* function¹. For all symmetric moving average we have $\Phi_\theta(\omega) = 0$.

To sum up, applying a moving average to an harmonic times series affects in in two different ways:

- by multiplying it by an amplitude coefficient $G_\theta(\omega)$;
- by “shifting” it in time by $\Phi_\theta(\omega)/\omega$, which directly affects the detection of turning points².

¹This function is sometimes represented as $\phi_\theta(\omega) = \frac{\Phi_\theta(\omega)}{\omega}$ to measure the phase shift in number of periods.

²When $\Phi_\theta(\omega)/\omega > 0$ the time shift is positive: a turning point is detected with delay.

Example: with $M_{\theta_0}X_t = \frac{1}{2}X_{t-1} + \frac{1}{2}X_t$ we have:

$$\Gamma_{\theta_0}(\omega) = \frac{1}{2} + \frac{1}{2}e^{-i\omega} = |\cos(\omega/2)|e^{-i\frac{\omega}{2}}$$

The figure 1 illustrates the gain and the phase shift for $\omega = \pi/2$ and $X_t = \sin(\omega t)$.

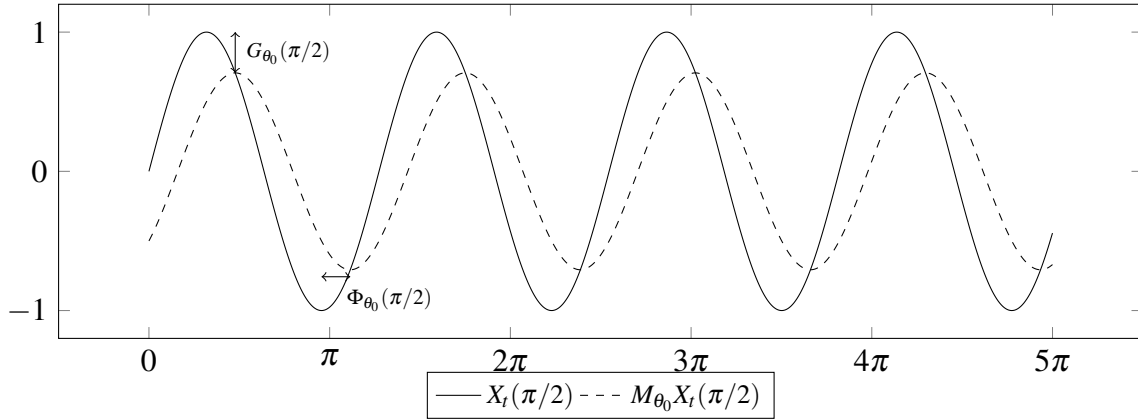


Figure 1: Smoothing of the time series $X_t = \sin(\omega t)$ by the moving average $M_{\theta_0}X_t = \frac{1}{2}X_{t-1} + \frac{1}{2}X_t$ for $\omega = \pi/2$.

2.2 Desirable properties of a moving average

The moving average are often constructed under some specific constraints. In the report we will focus on two constraints:

- the preservation of certain kind of trends;
- the variance reduction.

2.2.1 Trend preservation

Is is often desirable for a moving average to conserve certain kind of trends. A moving average M_θ conserve a function of the time $f(t)$ if $\forall t : M_\theta f(t) = f(t)$.

We have the following properties for the moving average M_θ :

- To conserve a constant series $X_t = a$ we need

$$\forall t : M_\theta(X_t) = \sum_{k=-p}^{+f} \theta_k X_{t+k} = \sum_{k=-p}^{+f} \theta_k a = a \sum_{k=-p}^{+f} \theta_k = a$$

the sum of the coefficients of the moving average $\sum_{k=-p}^{+f} \theta_k$ must then be equal to 1.

- To conserve a linear trend $X_t = at + b$ we need:

$$\forall t : M_\theta(X_t) = \sum_{k=-p}^{+f} \theta_k X_{t+k} = \sum_{k=-p}^{+f} \theta_k [a(t+k) + b] = at \sum_{k=-p}^{+f} k \theta_k + b \sum_{k=-p}^{+f} \theta_k = at + b$$

which is equivalent to:

$$\sum_{k=-p}^{+f} \theta_k = 1 \quad \text{and} \quad \sum_{k=-p}^{+f} k \theta_k = 0$$

- In general, it can be shown that M_θ conserve a polynomial of degree d if and only if:

$$\sum_{k=-p}^{+f} \theta_k = 1 \quad \text{and} \quad \forall j \in \llbracket 1, d \rrbracket : \sum_{k=-p}^{+f} k^j \theta_k = 0$$

2.2.2 Variance reduction

All time series are affected by noise that can blur the signal extraction. Hence, we seek to reduce the variance of the noise. The sum of the sum of the squares of the coefficients $\sum_{k=-p}^{+f} \theta_k^2$ is the *variance reduction* ratio.

Indeed, let $\{\varepsilon_t\}$ a sequence of independent random variables with $\mathbb{E}[\varepsilon_t] = 0$, $\mathbb{V}[\varepsilon_t] = \sigma^2$.

$$\mathbb{V}[M_\theta \varepsilon_t] = \mathbb{V}\left[\sum_{k=-p}^{+f} \theta_k \varepsilon_{t+k}\right] = \sum_{k=-p}^{+f} \theta_k^2 \mathbb{V}[\varepsilon_{t+k}] = \sigma^2 \sum_{k=-p}^{+f} \theta_k^2$$

2.3 Real-time estimation and asymmetric moving average

For symmetric filters, the phase shift function is equal to zero. Therefore, there is no delay in any frequency: that's why they are preferred to the asymmetric one. However, they cannot be used in the beginning and in the end of the time series because no past/future value can be used. Thus, for real-time estimation, it is needed to build asymmetric moving average that approximate the symmetric moving average.

Grun-Rehonne, Guggemos, and Ladiray (2018) Wildi (2018)

à compléter

2.3.1 Timeliness (gugemos)

timeliness

$$\int_{\omega_1}^{\omega_2} f(\rho_\theta(\omega), \varphi_\theta(\omega)) d\omega$$

2.4 Quality indicators (? trouver un autre nom)

To compare the different moving average / filters the following indicators are used:

$$\begin{cases} \text{constant bias} = \text{bias0} & = \sum_{k=-p}^{+p} \theta_k \\ \text{linear bias} = \text{bias1} & = \sum_{k=-p}^{+p} k \theta_k \\ \text{quadratic bias} = \text{bias2} & = \sum_{k=-p}^{+p} k^2 \theta_k \\ \text{variance reduction} & = \sum_{k=-p}^{+p} \theta_k^2 \end{cases}$$

à compléter, mettre valeurs de référence

3 Local polynomial filters

In this section we detail the filters that arised from fitting a local polynomial to our time series, as described by Proietti and Luati (2008).

We assume that our time series y_t can be decomposed as

$$y_t = \mu_t + \varepsilon_t$$

where μ_t is the signal (trend) and $\varepsilon_t \stackrel{i.i.d}{\sim} \mathcal{N}(0, \sigma^2)$ is the noise. We assume that μ_t can be locally approximated by a polynomial of degree d of the time t between y_t and the neighboring observations $(y_{t+j})_{j \in \llbracket -h, h \rrbracket}$. Then $\mu_t \simeq m_t$ with:

$$\forall j \in \llbracket -h, h \rrbracket : y_{t+j} = m_{t+j} + \varepsilon_{t+j}, \quad m_{t+j} = \sum_{i=0}^d \beta_i j^i$$

This signal extraction problem is then equivalent to the estimation of $m_t = \beta_0$. In matrix notation we can write:

$$\underbrace{\begin{pmatrix} y_{t-h} \\ y_{t-(h-1)} \\ \vdots \\ y_t \\ \vdots \\ y_{t+(h-1)} \\ y_{t+h} \end{pmatrix}}_y = \underbrace{\begin{pmatrix} 1 & -h & h^2 & \dots & (-h)^d \\ 1 & -(h-1) & (h-1)^2 & \dots & -(h-1)^d \\ \vdots & \vdots & \vdots & \dots & \vdots \\ 1 & 0 & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \dots & \vdots \\ 1 & h-1 & (h-1)^2 & \dots & (h-1)^d \\ 1 & h & h^2 & \dots & h^d \end{pmatrix}}_X \underbrace{\begin{pmatrix} \beta_0 \\ \beta_1 \\ \vdots \\ \vdots \\ \vdots \\ \beta_d \end{pmatrix}}_{\beta} + \underbrace{\begin{pmatrix} \varepsilon_{t-h} \\ \varepsilon_{t-(h-1)} \\ \vdots \\ \varepsilon_t \\ \vdots \\ \varepsilon_{t+(h-1)} \\ \varepsilon_{t+h} \end{pmatrix}}_{\varepsilon}$$

Two parameters are crucial in determining the accuracy of the approximation:

- the degree d of the polynomial;
- the number of neighbored $H = 2h + 1$ (or the *bandwidth* h).

In order to estimate β we need $H \geq d + 1$ and the estimation is done by the weighted least squares (WLS), which consists of minimizing the following objective function:

$$S(\hat{\beta}_0, \dots, \hat{\beta}_d) = \sum_{j=-h}^h \kappa_j (y_{t+j} - \hat{\beta}_0 - \hat{\beta}_1 j - \dots - \hat{\beta}_d j^d)^2$$

where κ_j is a set of weights called *kernel*. We have $\kappa_j \geq 0$, $\kappa_{-j} = \kappa_j$ and with $K = \text{diag}(\kappa_{-h}, \dots, \kappa_h)$, the estimate of β can be written as $\hat{\beta} = (X' K X)^{-1} X' K y$. With $e_1 = \begin{pmatrix} 1 & 0 & \dots & 0 \end{pmatrix}'$, the estimate of the trend is:

$$\hat{m}_t = e_1 \hat{\beta} = w' y = \sum_{j=-h}^h w_j y_{t-j} \text{ with } w = K X (X' K X)^{-1} e_1$$

To conclude, the estimate of the trend \hat{m}_t can be obtained applying the symmetric filter w to y_t ³. Moreover, $X' w = e_1$ so:

$$\sum_{j=-h}^h w_j = 1, \quad \forall r \in \llbracket 1, d \rrbracket : \sum_{j=-h}^h j^r w_j = 0$$

Hence, the filter w preserve deterministic polynomial of order d .

3.1 Different kernels

In signal extraction, we generally look for weighting observations according to their distance from time t : this is the role of the kernel function. For that, we introduce a kernel function κ_j , $j = 0, \pm 1, \dots, \pm h$ with $\kappa_j \geq 0$ and $\kappa_j = \kappa_{-j}$. An important class of kernels is the Beta kernels. In the discrete, up to a proportional factor (so that $\sum_{j=-h}^h \kappa_j = 1$):

$$\kappa_j = \left(1 - \left| \frac{j}{h+1} \right|^r \right)^s$$

with $r > 0$, $s \geq 0$. The following kernels are considered in this report:

- $r = 1, s = 0$ uniform kernel:
- $r = s = 1$ triangle kernel:

$$\kappa_j^U = 1$$

$$\kappa_j^T = \left(1 - \left| \frac{j}{h+1} \right| \right)$$

³ w is symmetric due to the symmetry of the kernel weights κ_j .

- $r = 2, s = 1$ Epanechnikov (or Parabolic) kernel:

$$\kappa_j^E = \left(1 - \left|\frac{j}{h+1}\right|^2\right)$$

- $r = s = 2$ biweight kernel:

$$\kappa_j^{BW} = \left(1 - \left|\frac{j}{h+1}\right|^2\right)^2$$

- $r = 2, s = 3$ triweight kernel:

$$\kappa_j^{TW} = \left(1 - \left|\frac{j}{h+1}\right|^2\right)^3$$

- $r = s = 3$ tricube kernel:

$$\kappa_j^{TC} = \left(1 - \left|\frac{j}{h+1}\right|^3\right)^3$$

- Gaussian kernel⁴:

$$\kappa_j^G = \exp\left(-\frac{j^2}{4h}\right)$$

- Henderson kernel (see section 3.1.1 for its construction):

$$\kappa_j = \left[1 - \frac{j^2}{(h+1)^2}\right] \left[1 - \frac{j^2}{(h+2)^2}\right] \left[1 - \frac{j^2}{(h+3)^2}\right]$$

- Trapezoidal kernel:

$$\kappa_j^{TP} = \begin{cases} \frac{1}{3(2h-1)} & \text{if } j = \pm h \\ \frac{2}{3(2h-1)} & \text{if } j = \pm(h-1) \\ \frac{1}{2h-1} & \text{otherwise} \end{cases}$$

The figure 2 summarises the coefficients of the different kernels. Analysing the coefficients we can already anticipate some properties of the associated filters:

- For h small the triweight kernel has the narrowest distribution, for h high ($h \geq 15$) the gaussian kernel become narrowest. The narrowest a distribution is, the smallest the weights of furthest neighbors are: the associated filter should have a high weight in the current observation (t).
- For h high the Henderson kernel is equivalent to the triweight kernel (since $h+1 \sim h+2 \sim h+3$, $\kappa_j^H \sim \kappa_j^{TW}$), the associated filter should also be equivalent.

3.1.1 Specific symmetric filters

When $p = 0$ (local adjustment by a constant) we obtain the **Nadaraya-Watson's** estimator.

With the uniform kernel we obtain the **Macauley filter**. When $p = 0, 1$, this is the arithmetic moving average: $w_j = w = \frac{1}{2h+1}$.

The **Epanechnikov** kernel is often recommended as the optimal kernel that minimize the mean square error of the estimation by local polynomial.

⁴The gaussian kernel is generally defined as $e^{-\frac{x^2}{2\sigma^2}}$. Here we take $\sigma^2 = 2/h$ with the following idea: the biggest h is, the narrowest the filter is.

Figure 2: Coefficients of the different kernels for h from 2 to 30.

Note: to see the animation the PDF must be open with Acrobat Reader, KDE Okular, PDF-XChange or Foxit Reader. Otherwise you will only be able to see the results for $h = 2$.

Loess is a locally weighted polynomial regression that use tricube kernel.

The **Henderson filter** is a specific case of a local cubic fit ($p = 3$). It is often used Henderson for trend estimation (for example it's the filter used in the seasonal adjustment). For a fixed bandwidth, Henderson found the kernel that gave the smoothest estimates of the trend. He showed that the three following problems were equivalent:

1. minimize the variance of third difference of the series by the application of the moving average;

2. minimize the sum of squares of third difference of the coefficients of the filter, it's the *smoothness criteria*: $S = \sum_j (\nabla^3 \theta_j)^2$;
3. fit a local cubic polynomial by weighted least squares, where the weights are chose to minimize the sum of squares of the resulting filter.

Resolving the last problem leads to the kernel presented in section 3.1.

3.1.2 Analysis of symmetric filters

In this section, all the filters are computed by local polynomial of degree $d = 3$. The figure 3 plots the coefficients of the filters for the differents kernels presented in different kernels presented in section 3.1 and for different bandwidth h . The table 1 shows the variance reduction of the different filters. We find the similar results than in section 2:

- For h small the triweight kernel gives the filter with the narrowest distribution, for h high ($h \geq 15$) the gaussian kernel become narrowest. The narrowest a distribution is, the higher the variance reduction should be. Indeed, the distribution of the coefficients of the filter can be interpreted as the output signal of an additive outlier. As a result, with a wide distribution, an additive outlier will be more persistent than with a narrow distribution. Therefore, it's the triweight that has the higher variance reduction for all $h \leq 30$.
- For h small, the trapezoidal filter seems to produce similar results than the Epanechnikov one.
- For h high the Henderson filter is equivalent to the one computed by the triweight kernel.

Moreover, we find that for all the filters, the coefficients decrease, when the distance to the central observation increases, until a negative value and then increase towards 0 (except for the uniform kernel). Negative coefficients might be disturbing but they arise from the cubic polynomial constraints. Indeed to preserve polynomial of degree 2 (and so 3) we need $\sum_{j=-h}^h j^2 \theta_i = 0$, which constraint some coefficients to be negative. However, those negative coefficients are negligible compare to the central higher coefficients (they are more 80% smaller than the central coefficient for all kernels, except for uniform and trapezoidal with high bandwidth).

3.1.3 Gain functions

Figure 4 plots the gain functions of the different filters. Gain functions are usually plotted between 0 and π . However, locally weighted polynomial regression are low-pass filters: they leave almost unchanged low frequency components (such as the trend) and attenuate high frequency fluctuations (noise). That's why we are only interested in low frequencies. For a monthly data, a cycle of 3 years correspond to the frequency $2\pi/36$ and a cycle of 7 years to the frequency $2\pi/84$.

Figure 3: Coefficients of symmetric filters computed by local polynomial of degree 3, according to the different kernels and for h from 2 to 30.

Note: to see the animation the PDF must be open with Acrobat Reader, KDE Okular, PDF-XChange or Foxit Reader. Otherwise you will only be able to see the results for $h = 2$.

When the bandwidth h increases, the gain function decreases for low frequencies: short business cycles will then be attenuated. For a fixed value of h , gaussian, Henderson and triweight filters will preserve more short business cycles than the others filters (especially uniform, trapezoidal and Epanechnikov).

tracer et interpréter gain après $2\pi/12$ pour voir à partir de quelles fréquences sont atténuées ?
ou bien c'est résumé avec le pouvoir de réduction de la variance ?

Table 1: Variance reduction ratio ($\sum \theta_i^2$) of symmetric filters computed by local polynomial of degree 3.

h	Kernel								
	Biweight	Epanechnikov	Gaussian	Henderson	Trapezoidal	Triangular	Tricube	Triweight	Uniform
2	0.50	0.49	0.49	0.50	0.51	0.51	0.49	0.52	0.49
3	0.33	0.30	0.30	0.32	0.31	0.33	0.32	0.37	0.28
4	0.25	0.23	0.24	0.25	0.23	0.25	0.25	0.28	0.22
5	0.22	0.21	0.21	0.22	0.20	0.22	0.22	0.24	0.20
6	0.20	0.19	0.19	0.20	0.19	0.20	0.20	0.21	0.19
7	0.19	0.18	0.18	0.19	0.18	0.19	0.19	0.20	0.18
8	0.18	0.17	0.18	0.18	0.17	0.18	0.18	0.19	0.17
9	0.17	0.17	0.17	0.17	0.17	0.17	0.17	0.18	0.17
10	0.17	0.16	0.17	0.17	0.16	0.17	0.17	0.17	0.16
20	0.12	0.12	0.13	0.13	0.12	0.12	0.13	0.13	0.12
30	0.10	0.10	0.10	0.10	0.10	0.10	0.10	0.10	0.10

3.2 Asymmetric filters

3.2.1 Direct asymmetric filters (DAF)

As mentionned in section 2.3, symmetric filters cannot be used in boundary points. For real-time estimation, three different approaches can be used:

1. Build a asymmetric filter fitting local polynomial to the available observations y_t for $t \in \llbracket n-h, n \rrbracket$.
2. Apply the symmetric filter to the series extended by forecast (or backcast) $\hat{y}_{n+l|n}, l \in \llbracket 1, h \rrbracket$.
3. Build a asymmetric filter which minimize the mean square revision error subject to polynomial reproducing constraints.

Proietti and Luati (2008) show that the first two approaches are equivalent when the forecast is done by a polynomial extrapolation of order d (forecasts generated with the same polynomial model than the symmetric filter). This is called the *direct asymmetric filter* (DAF). Let q be the number of available observations in the future: q varies from 0 (real time filter) to h (symmetric filter).

Rewritting the matrix X, K y in the following way:

$$X = \begin{pmatrix} X_p \\ X_f \end{pmatrix}, \quad y = \begin{pmatrix} y_p \\ y_f \end{pmatrix}, \quad K = \begin{pmatrix} K_p & 0 \\ 0 & K_f \end{pmatrix}$$

where y_p correspond to the available data and y_f the missing data, the DAF w_a and the forecast \hat{y}_f can be written as:

$$w_a = K_p X_p (X_p' K_p X_p)^{-1} e_1, \quad \hat{y}_f = X_f (X_p' K_p X_p)^{-1} X_p' K_p y_p$$

Figure 4: Gain functions from 0 to $2\pi/12$ of symmetric filters computed by local polynomial of degree 3, according to the different kernels and for h from 2 to 30.

Note: to see the animation the PDF must be open with Acrobat Reader, KDE Okular, PDF-XChange or Foxit Reader. Otherwise you will only be able to see the results for $h = 2$.

Moreover, we have the following results with the DAF w_a :

- it satisfies the same polynomial reproduction constraints as the symmetric filter (conserve polynomial of degree d). Thus, the bias in estimating an unknown function of time has the same order of magnitude as in the interior of time support.
- w_a minimize the weighted distance (by the kernel function) between the asymmetric filter coefficients and the symmetric ones.

However, the weights $w_{a,0}$ of the DAF are highly concentrated in the current observation t with

an important change between $q = 0$ (real-time filter) and $q = h$ (see figure ??). Moreover the real-time filter doesn't have a satisfying gain functions: it is close to one for all the frequencies. Therefore, even if the real-time filter is unbiased (if the series is generated by a polynomial of degree d) it is at the expenses of a high variance.

Figure 5: Coefficients and gain function of direct asymmetric filters (DAF) computed by local polynomial of degree 3, according to the different kernels and for $h = 6$.

Note: to see the animation the PDF must be open with Acrobat Reader, KDE Okular, PDF-XChange or Foxit Reader. Otherwise you will only be able to see the results for the Henderson kernel.

For all the kernels, we find the same results as in Proietti and Luati (2008):

- For a fixed value of d , the more the data is available (q increases), the more the weight associated to the current observation $w_{a,0}$ decreases.
- For a fixed value of h and q , $w_{a,0}$ increases exponentially with the polynomial degree d (in particular, for $d = h$, $w_{a,0} = 1$).

3.2.2 General class of asymmetric filters

To deal with the problem of the variance of the estimates of the real-time filters, Proietti and Luati (2008) suggest a general of asymmetric filters to make a tradeoff between bias and variance.

Here we consider that the data is generated by the model:

$$y = U\gamma + Z\delta + \varepsilon, \quad \varepsilon \sim \mathcal{N}(0, D)$$

The goal is to find a filter v which minimize the mean square revision error (with the symmetric filter w) subject to some constraints. The constraints are summarized by the matrix $U = \begin{pmatrix} U_p' & U_f' \end{pmatrix}'$ (with U_p the available observations of the matrix U for the asymmetric filter): $U_p'v = U'w$. The problem is equivalent to find v that minimize:

$$\varphi(v) = \underbrace{(v - w_p)'D_p(v - w_p) + w_f'D_f w_f}_{\text{revision error variance}} + \underbrace{[\delta'(Z_p'v - Z'w)]^2}_{\text{biases}^2} + \underbrace{2l'(U_p'v - U'w)}_{\text{constraints}}$$

Mean square revision error

with l a vector of Lagrange multipliers.

When $U = X$ this is equivalent to the constraint to preserve polynomial of degree d : we find the direct asymmetric filters w_a with $D = K^{-1}$.

When $U = \begin{pmatrix} 1 & \dots & 1 \end{pmatrix}'$, $Z = \begin{pmatrix} -h & \dots & +h \end{pmatrix}'$, $\delta = \delta_1$, $D = \sigma^2 I$ and when the symmetric filter is the Henderson filter we obtain the Musgrave asymmetric filters (see [ajouter biblio](#)). With the filter we assume that the data is generated by a linear process and that the asymmetric filters preserve constant signals ($\sum v_i = \sum w_i = 1$). The asymmetric filters depends on the ratio δ_1/σ , which is related to the “I-C” ratio $R = \frac{\bar{I}}{\bar{C}} = \frac{\sum |I_t - I_{t-1}|}{\sum |C_t - C_{t-1}|}$ ($\delta_1/\sigma = 2/(R\sqrt{\pi})$), the ratio between the expected absolute difference of the irregular and of the trend-cycle. In the seasonal adjustment method, the I-C ratio⁵ is used to determine the bandwidth to used for the Henderson filter. For monthly data:

- if $R < 1$ a 9-term Henderson is used ($h = 4$);
- if $1 \leq R \leq 3.5$ a 13-term Henderson is used ($h = 6$);
- if $3.5 < R$ a 23-term Henderson is used ($h = 12$).

When U corresponds to the first $d^* + 1$ first the columns of X , $d^* < d$, the constraint is that the asymmetric filter should reproduce polynomial of degree d^* , the potential bias depends on the value of δ . This will reduce the variance at the expense of a bias: it is the idea followed by Proietti and Luati (2008) to propose three class of asymmetric filters:

1. *Linear-Constant* (LC): y_t linear ($d = 1$) and v preserve constant signals ($d^* = 0$), v depend on the ratio δ_1^2/σ^2 .
2. *Quadratic-Linear* (QL): y_t quadratic ($d = 2$) and v preserve linear signals ($d^* = 1$), v depend on the ratio δ_2^2/σ^2 .
3. *Cubic-Quadratic* (CQ): y_t cubic ($d = 3$) and v preserve linear signals ($d^* = 1$), v depend on the ratio δ_2^2/σ^2 .

⁵In order to compute the I-C ratio, a first decomposition of the seasonally adjusted series is computed using a 13-term Henderson moving average.

Figure 6: Revision error of the different asymmetric filters for $h = 6$, for the IPI-FR.
Note: to see the animation the PDF must be open with Acrobat Reader, KDE Okular, PDF-XChange or Foxit Reader. Otherwise you will only be able to see the results for the Henderson kernel.

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