

# MICROWAVE ENGINEERING

Lecture 17:  
Impedance  
Matching port  
II

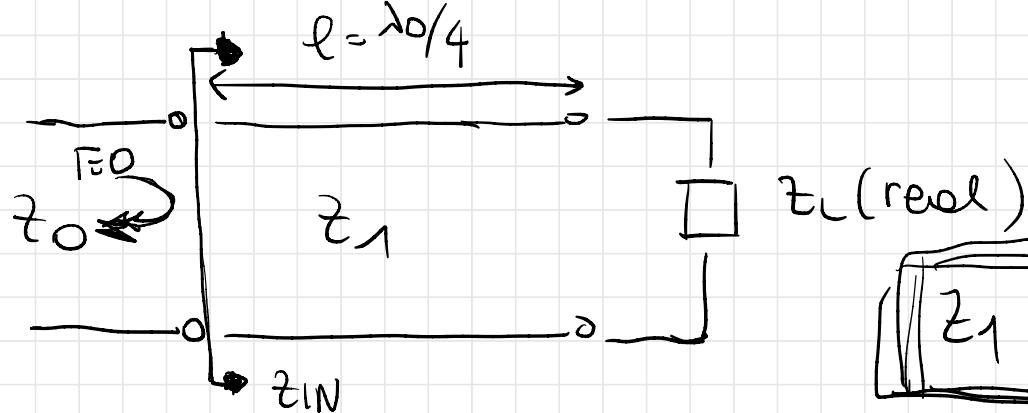
# QUARTER-WAVE TRANSFORMER

## PROS

- Simple Design
- Can be modified to multi-section

## CONS

- Matching for real loads
- Single frequency design



$$z_1 = \sqrt{z_0 z_L}$$

$$Z_{IN} = Z_1 \frac{Z_L + j Z_1 \tan \beta L}{Z_1 + j Z_L \tan \beta L}$$

$$\beta L = \frac{2\pi}{\lambda_0}, \frac{\lambda_0}{4} = \frac{\pi}{2} - \Theta$$

$$\Gamma = \frac{Z_{IN} - Z_0}{Z_{IN} + Z_0} = \frac{Z_1 (Z_L - Z_0) + j \tan \beta L (Z_1^2 - Z_0 Z_L)}{Z_1 (Z_L + Z_0) + j \tan \beta L (Z_1^2 + Z_0 Z_L)}$$

$$Z_1^2 = Z_0 Z_L \quad \text{if } \Theta = \frac{\pi}{2} \Rightarrow \Gamma = 0$$

$$\Gamma = \frac{z_L - z_0}{z_L + z_0 + 2j \tan \phi (\sqrt{z_0 z_L})}$$

The magnitude of  $\Gamma$

$$|\Gamma| = \frac{|z_L - z_0|}{\sqrt{(z_L + z_0)^2 + 4 z_0 z_L \tan^2 \phi}} = b = \tan \phi$$

$$= \frac{1}{\sqrt{\left(\frac{(z_L + z_0)^2}{(z_L - z_0)^2} + 4 t^2 \frac{z_0 z_L}{(z_L - z_0)^2}\right)^\frac{1}{2}}} =$$

1

-

=

$$\left\{ 1 + \frac{4z_0 z_L}{(z_L - z_0)^2} + \frac{4t^2 z_0 z_L}{(z_L - z_0)^2} \right\}^{\frac{1}{2}}$$

1

=

$$\left\{ 1 + \frac{4z_0 z_L}{(z_L - z_0)^2} \sec^2 \theta \right\}^{\frac{1}{2}}$$

$$\theta = \beta L$$

$$\sec^2 \theta = 1 + \tan^2 \theta$$

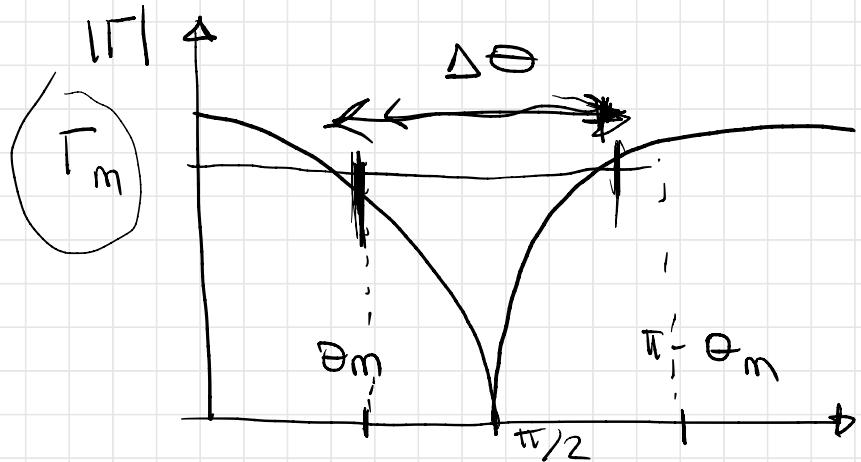
Assuming  $f \approx f_0$        $\ell \approx \frac{\lambda_0}{f}$        $\theta \approx \frac{\pi}{2}$



$$\sec^2 \theta \gg 1$$

then —

$$|\Gamma| = \left| \frac{z_L - z_0}{2\sqrt{z_0 z_L}} \right| \cos \theta \quad \text{for } \theta \approx \frac{\pi}{2}$$



$$\theta = \beta e$$

We can define a max deviation we can accept that is  $\Gamma_m$  that defines the bandwidth of the transformer

$$\Delta\theta = 2 \left( \frac{\pi}{2} - \theta_m \right)$$

$$\text{if } \Gamma = \Gamma_m \quad \theta = \theta_m \quad \vartheta = \pi - \theta_m$$

$$\frac{1}{\Gamma_m^2} = 1 + \left( \frac{\frac{2\sqrt{z_0 z_L}}{z_L - z_0} \sec \theta_m}{2} \right)^2$$

or

$$\underline{\omega_m} = \frac{\Gamma_m}{\sqrt{1 - \Gamma_m^2}} \frac{2\sqrt{Z_0 Z_L}}{|Z_L - Z_0|} \quad \Leftarrow$$

If we assume to have a TEM line

$$\theta = \beta z = \frac{2\pi f}{V_p}, \frac{V_p}{4f\beta} = \frac{\pi f}{2f_0}$$

at  $\theta = \theta_m$

$$f_m = \frac{2\theta_m f_0}{\pi}$$

The fractional bandwidth

$$BW = \frac{\Delta f}{f_0} = \frac{2(f_0 - f_m)}{f_0} = 2 - \frac{2f_m}{f_0}$$

$$= 2 - 4 \frac{\Theta_m}{\pi} =$$

$$= 2 - \frac{4}{\pi} \cos^{-1} \left[ \frac{\Gamma_m}{\sqrt{1 - \Gamma_m^2}} \frac{\frac{2}{\sqrt{z_0 z_L}}}{|z_L - z_0|} \right]$$



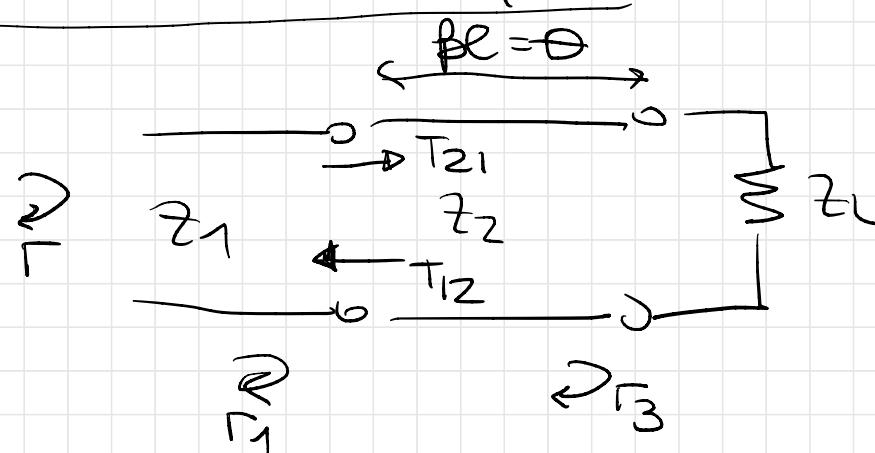
usually given in percentage

NOTE

- Formulas are valid for TEM lines where there is no dispersion

## THEORY OF SMALL REFLECTIONS

- Single section transformer

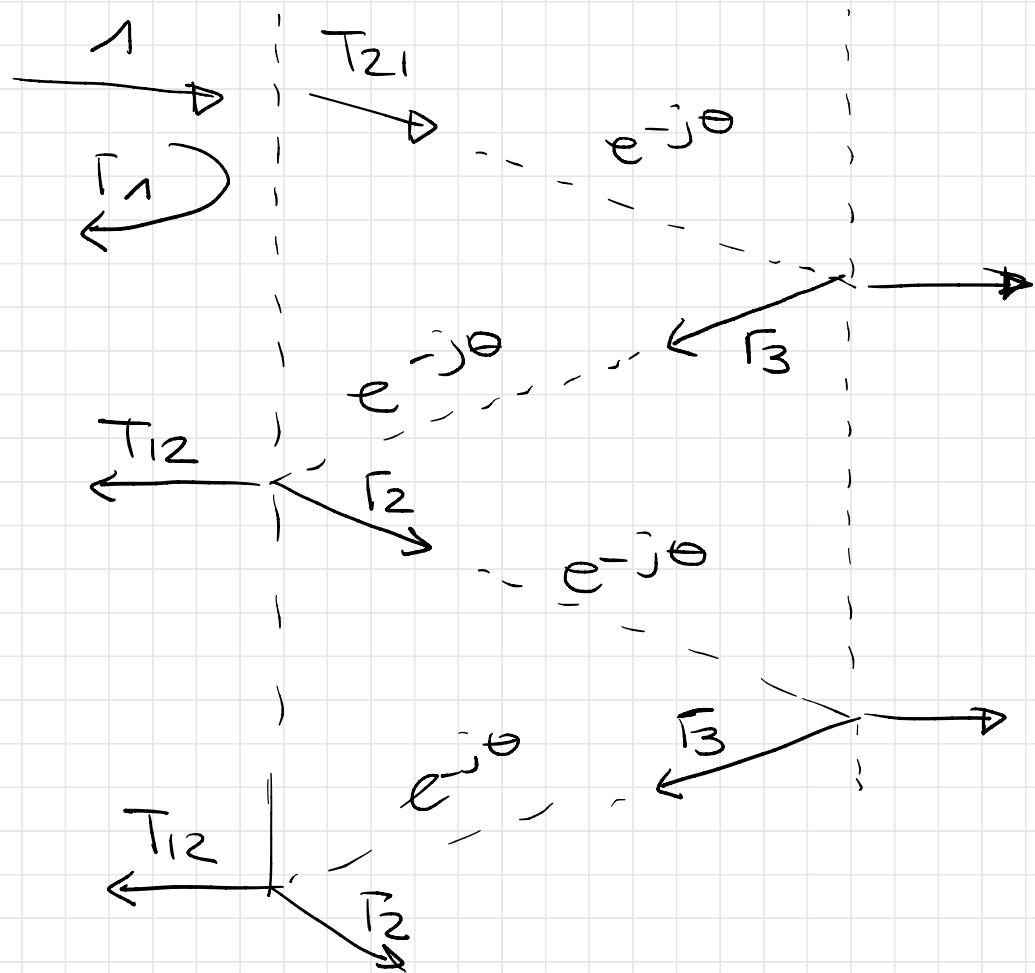


$$\Gamma_1 = \frac{z_2 - z_1}{z_2 + z_1} \quad \Gamma_2 = -\Gamma_1$$

$$\Gamma_3 = \frac{z_L - z_2}{z_L + z_2}$$

$$T_{21} = 1 + \Gamma_1 = \frac{2z_2}{z_1 + z_2}$$

$$T_{12} = 1 + \Gamma_2 = \frac{2z_1}{z_1 + z_2}$$



$$\Gamma = \Gamma_1 + \underbrace{\Gamma_{12} \Gamma_{21} \Gamma_3 e^{-2j\theta}}_{\dots} + \Gamma_{12} \Gamma_{21} \Gamma_3^2 \Gamma_2 e^{-4j\theta} +$$

$$= \Gamma_1 + \Gamma_{12} \Gamma_{21} \Gamma_3 e^{-j2\theta} \sum_{n=0}^{\infty} \Gamma_2^n \Gamma_3^n e^{-jn\theta}$$

$$\sum_{n=0}^{\infty} x^n = \frac{1}{1-x} \quad |x| < 1$$

expanding  $\Gamma$  we get  $\Gamma = \Gamma_1 + \frac{\Gamma_{12} \Gamma_{21} \Gamma_3 e^{-2j\theta}}{1 - \Gamma_2 \Gamma_3 e^{-2j\theta}}$

$$\text{Replacing } \Gamma_2 = -\Gamma_1 \quad \bar{\Gamma}_{21} = 1 + \Gamma_1$$

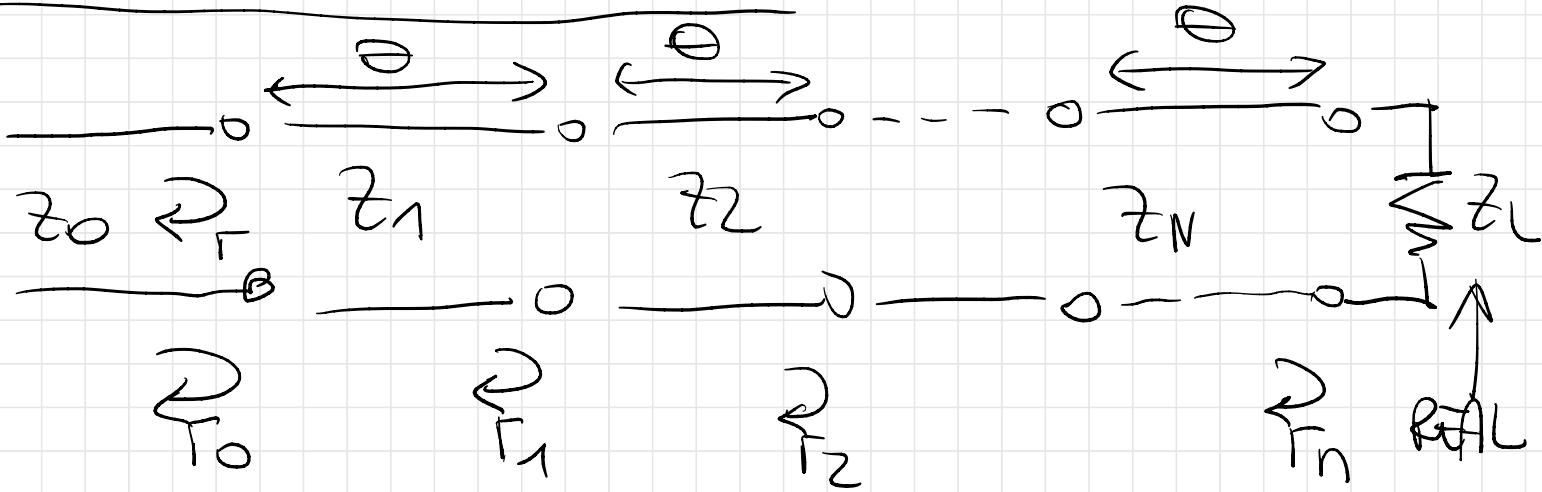
$$\bar{\Gamma}_{12} = 1 + \bar{\Gamma}_2 = 1 - \bar{\Gamma}_1$$

$$\boxed{\Gamma = \frac{\Gamma_1 + \Gamma_3 e^{-2j\theta}}{1 + \Gamma_1 \Gamma_3 e^{-2j\theta}}}$$

If the discontinuities are small ( $|z_1 - z_2|$  and  $|z_2 - z_L|$  are small) then

$$|\Gamma_1 \Gamma_3| \ll 1 \Rightarrow \boxed{\Gamma \approx \Gamma_1 + \Gamma_3 e^{-2j\theta}}$$

## Multisection transformer



NOTE : line sections are commensurate

$$\boxed{\Theta = \beta L}$$

Partial reflections are:

$$T_0 = \frac{z_1 - z_0}{z_1 + z_0}$$

$$T_m = \frac{z_{m+1} - z_m}{z_{m+1} + z_m}$$

$$T_N = \frac{z_L - z_N}{z_L + z_N}$$

Assume

$z_L$  is real;  $z_m$  increases or decreases monotonically

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$\Gamma_n$  is Real

$$\Gamma_n > 0 \quad \text{if} \quad Z_L > Z_0$$

$$\Gamma_n < 0 \quad \text{if} \quad Z_L < Z_0$$

The overall  $\Gamma$  is

$$\boxed{\Gamma(\theta) = \Gamma_0 + \Gamma_1 e^{-j2\theta} + \Gamma_2 e^{-4j\theta} + \dots + \Gamma_N e^{-j2N\theta}}$$

If we assume the transformer is symmetrical  
 $(\Gamma_0 = \Gamma_N, \Gamma_1 = \Gamma_{N-1}, \Gamma_2 = \Gamma_{N-2} \dots)$   
 ↳ Does not imply that  $\gamma_n$  are symmetrical?

In this case

$$\begin{aligned}\Gamma(\theta) = & e^{-jN\theta} \left[ \Gamma_0 [e^{-jN\theta} + e^{jN\theta}] \right. \\ & + \Gamma_1 \left[ e^{j(N-2)\theta} + e^{-j(N-2)\theta} \right] \dots \left. \right]\end{aligned}$$

↗ Finite Fourier series of cosines

$$\left\{ \begin{array}{l} \Gamma(\theta) = 2e^{-jN\theta} [\Gamma_0 \cos N\theta + \Gamma_1 \cos(N-2)\theta \\ \quad + \dots + \Gamma_m \cos(N-2m\theta) + \dots + \frac{1}{2}\Gamma_{N/2}] \\ \qquad \qquad \qquad \text{N is even} \end{array} \right.$$

$$\left\{ \begin{array}{l} \Gamma(\theta) = 2e^{-jN\theta} [\Gamma_0 \cos N\theta + \Gamma_1 \cos(N-2)\theta \\ \quad + \dots + \Gamma_m \cos(N-2m\theta) + \dots + \Gamma_{(N-1)/2}] \\ \qquad \qquad \qquad \text{N is odd} \end{array} \right.$$

# BINOMIAL MULTISECTION TRANSFORMER

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↳ provides the flattest response  
(MAXIMALLY FLAT)

$$T(\theta) = A \left( 1 + e^{-2j\theta} \right)^N$$

We have to set to zero the first  $N-1$  derivatives of  $\Gamma(\theta)$

The magnitude

$$|\Gamma(\theta)| = |A| |e^{-j\theta}|^N |e^{j\theta} + e^{-j\theta}|^N$$

$$= 2^N |A| \cos \theta^N$$

$$\text{for } \theta = \frac{\pi}{2} \Rightarrow |\Gamma(\theta)| = 0$$

$$\text{for } \theta = \frac{\pi}{2} \Rightarrow \frac{d^n \Gamma(\theta)}{d\theta^n} = 0 \quad n = 1, 2, \dots, N$$

We can determine A by letting  $f \rightarrow 0$  or  
 $\theta \rightarrow 0$

$$\Gamma(0) = 2^N \quad A = \frac{z_L - z_0}{z_L + z_0}$$

$$A = 2^{-N} \frac{z_L - z_0}{z_L + z_0} \leftarrow$$

$$r(\theta) = A \left( 1 + e^{-j\theta} \right)^N =$$

$$= A \sum_{n=0}^N C_n^N e^{-jn\theta}$$

$$C_n^N = \frac{N!}{(N-n)! n!}$$

$$C_0^N = 1$$

$$C_1^N = N = C_{N-1}^N$$

$$C_n^N = C_{N-n}^N$$

$$\Gamma(\theta) = A \sum_{n=0}^N C_n e^{-2j n \theta} =$$

$$= \Gamma_0 + \Gamma_1 e^{-2j\theta} + \Gamma_2 e^{-4j\theta} + \dots$$

$$+ \Gamma_N e^{-2j N \theta}$$

~~if~~

$$\boxed{\Gamma_N = A C_N}$$

← Tabulated

Bandwidth calculated by imposing the  
max tolerable  $\Gamma_m$

$$\Gamma_m = 2^N |A| \cos^n \theta_m$$

where  $\theta_m < \frac{\pi}{2}$

$$\theta_m = \cos^{-1} \left[ \frac{1}{2} \left( \frac{\Gamma_m}{|A|} \right)^{\frac{1}{n}} \right]$$



Fractional bandwidth

$$\frac{\Delta f}{f_0} = \frac{2(f_0 - f_m)}{f_0} = 2 - \frac{4\Theta m}{\pi}$$

$$= 2 - \frac{4}{\pi} \cos^{-1} \left[ \frac{1}{2} \left( \frac{f_m}{|A|} \right)^{\frac{1}{2}} \right]$$

# CHEBISHER MULTISECTION TRANSFORMER

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↳ optimizes bandwidth  
at the expense of the  
passband ripples

We equate  $T(\theta)$  to the Chebyshev  
polynomial.

$$\left\{ \begin{array}{l} T_1(x) = x \\ T_2(x) = 2x^2 - 1 \end{array} \right.$$

$$\begin{cases}
 T_3(x) = 4x^3 - 3x \\
 T_4(x) = 8x^4 - 8x^2 + 1 \\
 \vdots \\
 \vdots
 \end{cases}$$

$$T_n(x) = 2x T_{n-1}(x) - T_{n-2}(x)$$

## Properties

- For  $-1 \leq x \leq 1$ ,  $|T_n(x)| \leq 1$  INSIDE  
PASSBAND

- For  $|x| > 1$ ,  $|T_n(x)| > 1 \leftarrow$  OUTSIDE PASS BAND
- For  $|x| > 1$ ,  $|T_n(x)|$  increases faster with  $x$  as  $n$  increases

If we assume  $x = \cos(\theta)$  then  $|x| < 1$

$$T_n(\cos\theta) = \cos n\theta$$

In general

$$T_m(x) = \cos(m \cos^{-1} x), \quad |x| \leq 1$$

$$T_m(x) = \cosh(m \cosh^{-1} x), \quad |x| > 1$$

For example if we have 4 sections:

$$T_1(\sec \theta m \cos \theta) = \sec \theta m \cos \theta$$

$$T_2(\sec \theta m \cos \theta) = \sec^2 \theta m (1 + \cos \theta - 1)$$

$$T_3(\sec \theta m \cos \theta) = \sec^3 \theta m [\cos(3\theta) + 3 \cos \theta] - 3 \sec \theta m \cos \theta$$

$$T_4(\sec \theta_m \cos \theta) = \sec^4 \theta_m (\cos 4\theta + \\ + 4 \cos 2\theta + 3) - 4 \sec^2 \theta_m (\cos 2\theta + 1) + 1$$

To summarize:

$$\Gamma(\theta) = A e^{-jN\theta} T_N(\sec \theta_m \cos \theta)$$

The constant  $A$  is calculated for  $\theta=0$

$$\Rightarrow \Gamma(0) = \frac{Z_L - Z_0}{Z_L + Z_0} = A T_N(\sec \theta_m)$$

$$A = \frac{Z_L - Z_0}{Z_L + Z_0} \quad \frac{1}{T_N(\sec \theta_m)}$$

If the max allowed  $\Gamma$  is  $\Gamma_m$

Since the max  $T_m(\sec \theta_m \cos \theta) = 1$

Then

$$\underline{T_N(\sec \theta_m)} = \frac{1}{\Gamma_m} \left| \frac{Z_L - Z_0}{Z_L + Z_0} \right| \approx \frac{1}{2\Gamma_m} \left| \ln \frac{Z_L}{Z_0} \right|$$

The fractional bandwidth is

$$\frac{\Delta f}{f_0} = 2 - \frac{4\Theta_m}{\pi}$$