

# Semester S1 Foundations of electromagnetic wave propagation

## **EXAMINATION**

## PART S. VERDEYME



# E(rasmus) Mundus on Innovative Microwave Electronics and Optics Master



Consider the waveguide presented figure 1. The material embedded in this support is characterized by its relative permeability  $\mu_r = 1$  and its relative permittivity  $\varepsilon_r = 1$ . It is bounded by perfect magnetic walls (PMW), placed in x=0, x=a=22.86 mm, and by perfect electric walls (PEW), placed at y=0 and y=b=10.16 mm. We will work in the frequency domain.

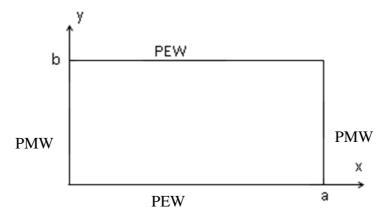


Figure 1

#### 1) TEM Mode

- a) Can you consider that the propagation along this support (figure 1) is quite similar to the propagation along a microstrip line? Explain why.
- b) We give for this TEM mode:

$$\overrightarrow{H(x,y)} = \frac{V}{bZ_0} \overrightarrow{e_x}$$

$$\overrightarrow{E(x,y)} = -\frac{v}{b}\overrightarrow{e_y}$$

where:

Z<sub>0</sub> is the plane wave impedance

V is the voltage on the conductor placed in the plane y=b (the other conductor is grounded)

Compute the expression of:

- ➤ the waveguide metallic losses per unit length on the TEM mode, for conductors characterized by their surface resistance R<sub>s</sub>
- > the power transmitted through the line
- > the attenuation coefficient of the waveguide.

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#### 2) TE Mode

This waveguide is now excited on TE modes. On a PMW, the  $H_z$  component cancels (in the planes defined by x=0 and x=a), and you know the condition for the magnetic field on the PEW.

- a. From the propagation equation, considering the waveguide lossless,:
  - $\triangleright$  compute the expressions of the first TE mode cutoff frequency and the  $H_z$  component of this mode.
  - From the following expressions, compute all the E and H field components of the first TE mode.

$$\left( \gamma^{2} + k_{o}^{2} \right) \overrightarrow{E_{t}}(\xi, \eta) = -\gamma \overrightarrow{\nabla_{t}} E_{z}(\xi, \eta) + j \omega \mu \overrightarrow{u} \wedge \overrightarrow{\nabla_{t}} H_{z}(\xi, \eta)$$

$$\left( \gamma^{2} + k_{O}^{2} \right) \overrightarrow{H_{t}}(\xi, \eta) = -\gamma \overrightarrow{\nabla_{t}} H_{z}(\xi, \eta) - j \omega \varepsilon \overrightarrow{u} \wedge \overrightarrow{\nabla_{t}} E_{z}(\xi, \eta)$$

With  $\vec{u}$  the unitary vector in the z direction.

- b. The metallic enclosure placed in the plane y=0 and y=b is now again a real conductor, characterized by its conductivity  $\sigma$ .
  - ➤ Compute the waveguide metallic losses on the first TE mode
  - ➤ Compute the power transmitted through the waveguide
  - > Compute the attenuation coefficient of the waveguide.
- 3) Compare now the attenuation per unit length of this support on the TE and TEM modes:
  - ➤ At the cutoff frequency of the TE mode
  - $\triangleright$  For  $\beta >> K_c$

Conclude