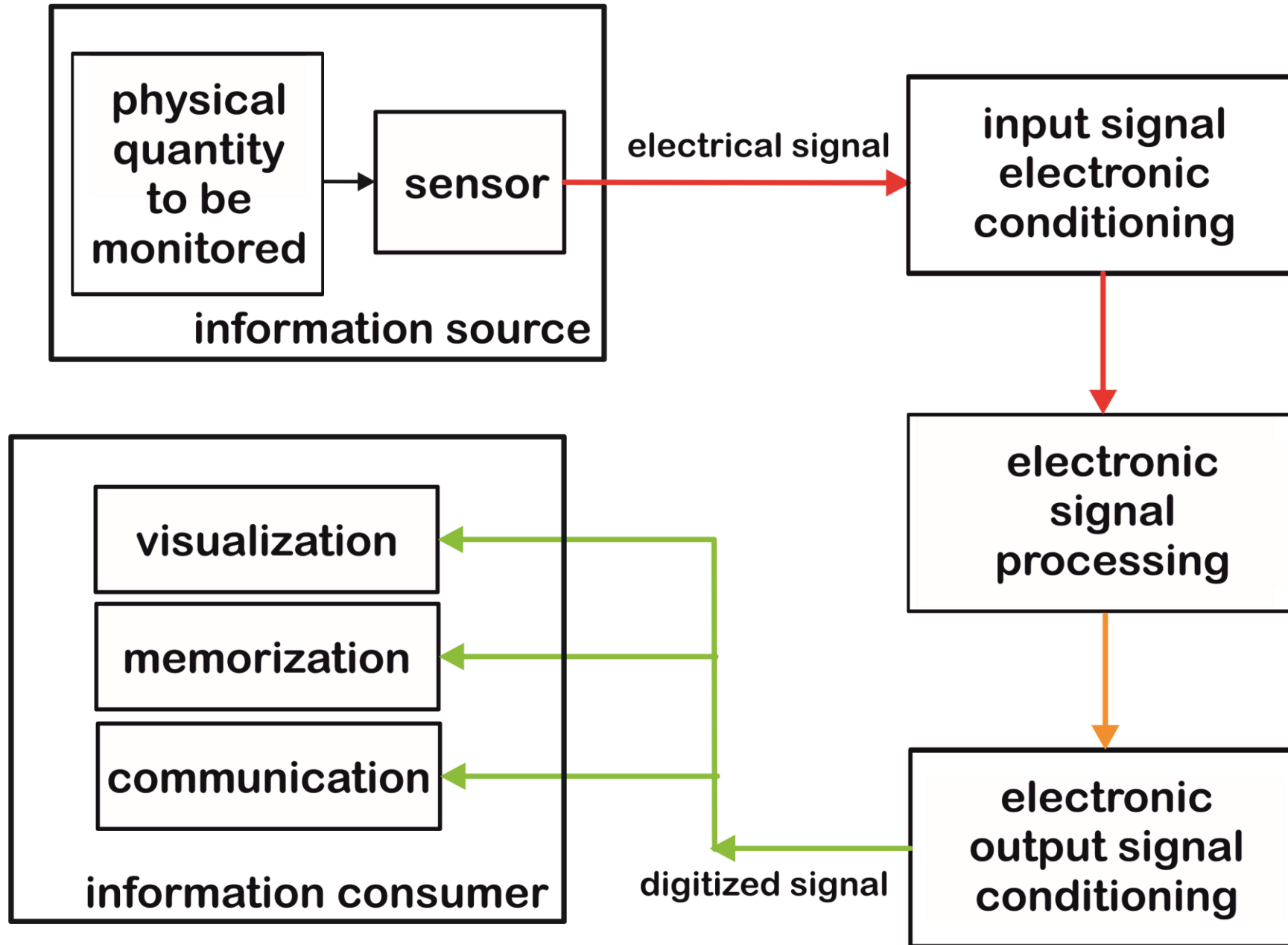
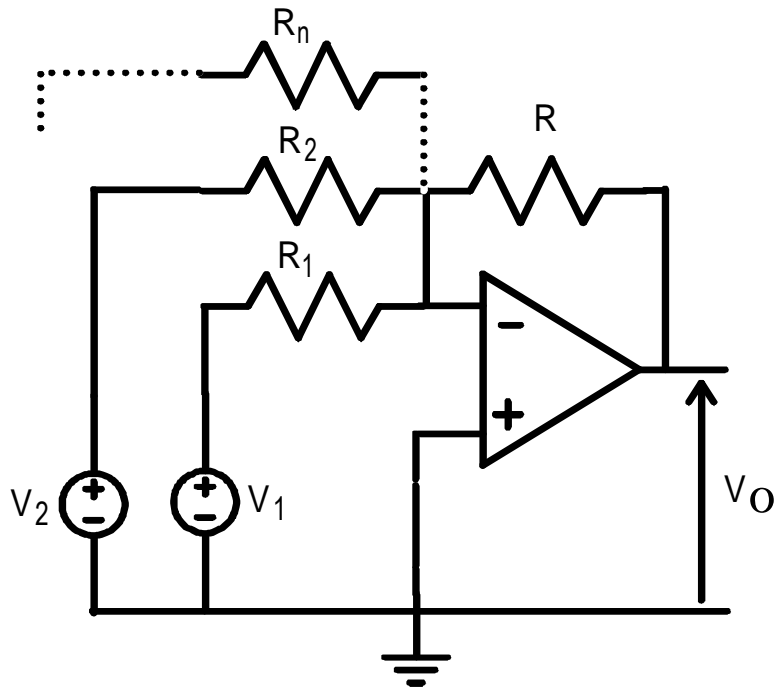


Measuring chain



Summing amplifier



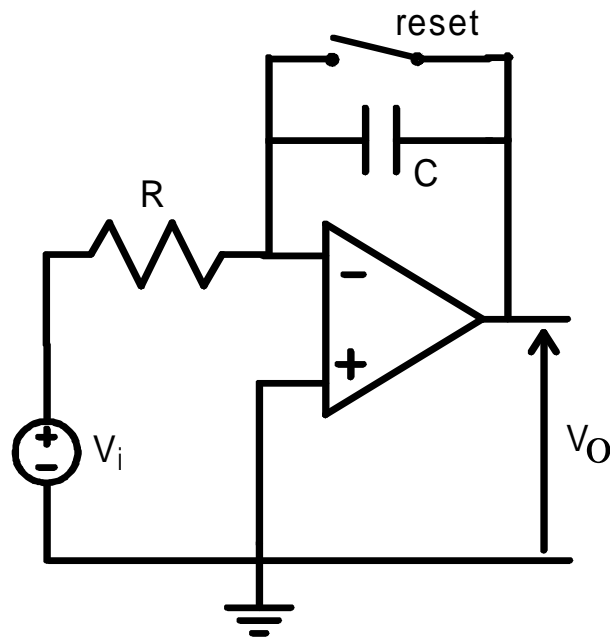
$$\square \quad V_o = -R \cdot \left(\frac{V_1}{R_1} + \frac{V_2}{R_2} + \dots + \frac{V_n}{R_n} \right)$$

setting $R_1 = R_2 = R_3 = R_n$

$$\square \quad V_o = -\frac{R}{R_1} \cdot (V_1 + V_2 + \dots + V_n)$$

- ☐ input impedance relatively small
- ☐ the gain can be different for each input
- ☐ to get an accurate sum it is necessary to use resistors with small tolerances

Integrator



○ the reset switch is opened at the instant $t=0$:

○
$$V_o(s) = -V_i(s) \cdot \frac{1/sC}{R} = -\frac{1}{s} \cdot \frac{V_i(s)}{RC}$$

⇒
$$V_o(s) = -V_i(s) \cdot \frac{1/sC}{R} = -\frac{1}{s} \cdot \frac{V_i(s)}{RC}$$

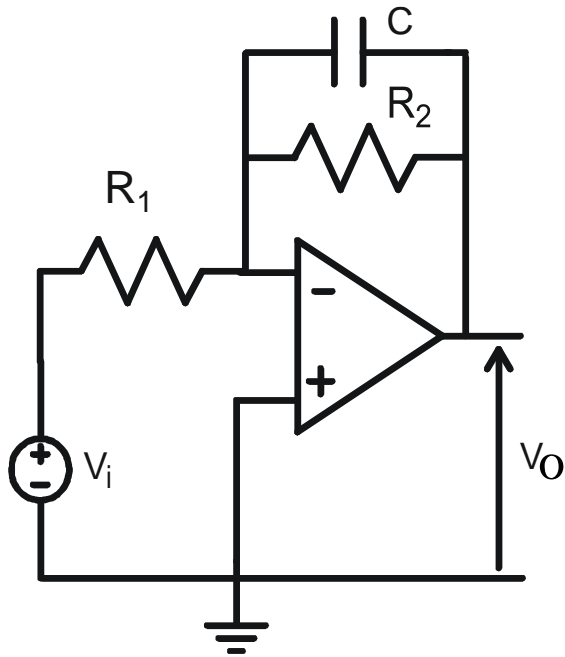
□ if there is a continuous component of the input signal then the circuit will saturate

♦ the solution is the approximate integrator

□ performances depend on the quality of the capacitor:

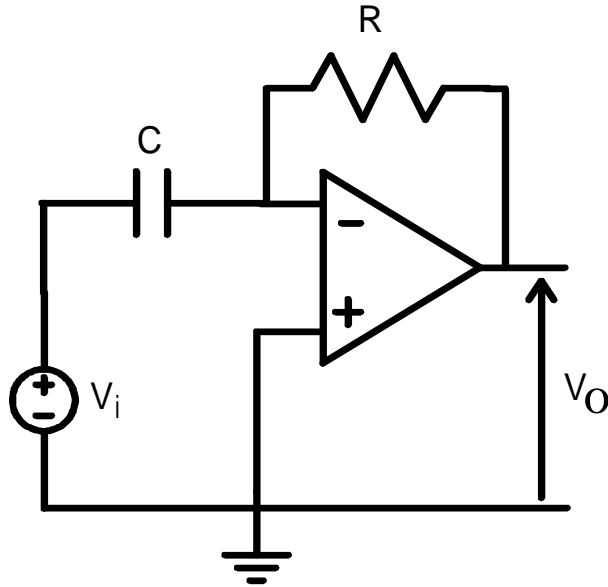
♦ polypropylene and polystyrene capacitors

Approximate integrator



$$V_o(s) = -V_i(s) \cdot \frac{1}{R_1} \frac{R_2 \cdot 1/sC}{R_2 + 1/sC} = -\frac{R_2}{R_1(1 + sR_2C)} V_i(s)$$

Differentiator



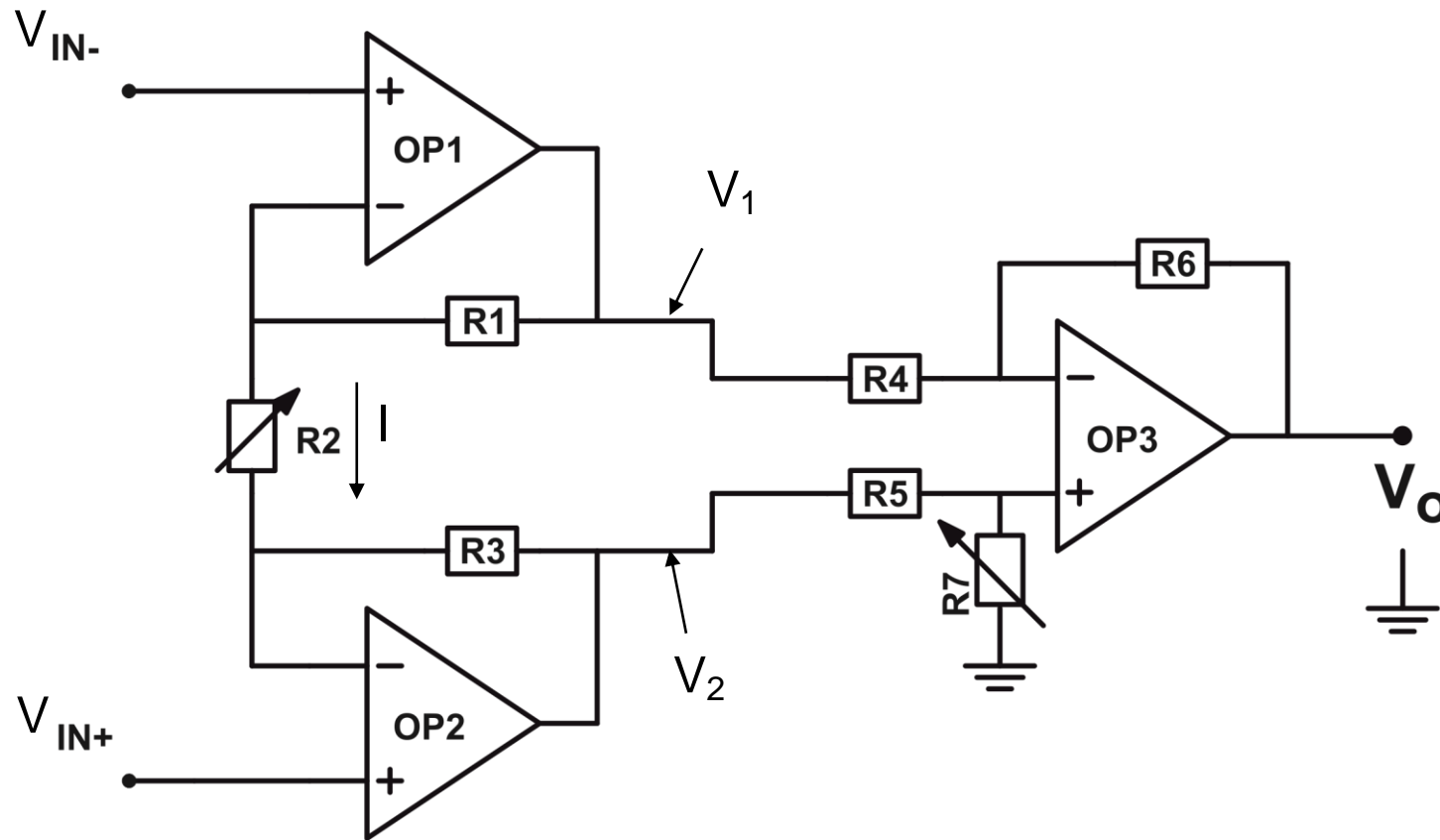
$$V_o(s) = -V_i(s) \cdot \frac{R}{1/sC} = -s \cdot RC \cdot V_i(s)$$



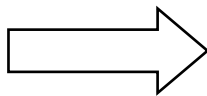
$$V_o(t) = -RC \cdot \frac{dV_i(t)}{dt}$$

- ☐ bias currents and offset voltages are not critical
- ☐ due to the non-idealities of the amplifier, the circuit may become unstable and may oscillate. A small resistor in series with C and a small capacitor in parallel with R can solve the problem.
- ☐ eventual noise at high frequency is amplified at the output

Instrumentation amplifier



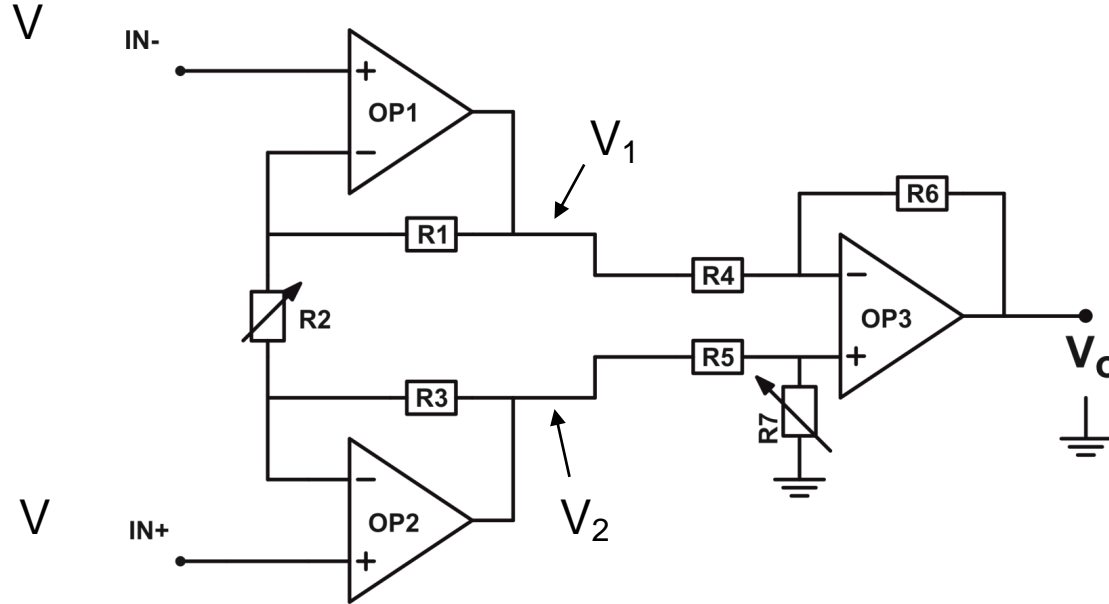
$$I = \left(\frac{V_{IN-} - V_{IN+}}{R_2} \right)$$



$$V_1 = V_{IN-} + R_1 \cdot I = V_{IN-} + \frac{R_1}{R_2} (V_{IN-} - V_{IN+})$$

$$V_2 = V_{IN+} - R_3 \cdot I = V_{IN+} - \frac{R_3}{R_2} (V_{IN-} - V_{IN+})$$

continue ...



$$V_O = V_2 \cdot \frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4}\right) - V_1 \cdot \frac{R_6}{R_4}$$

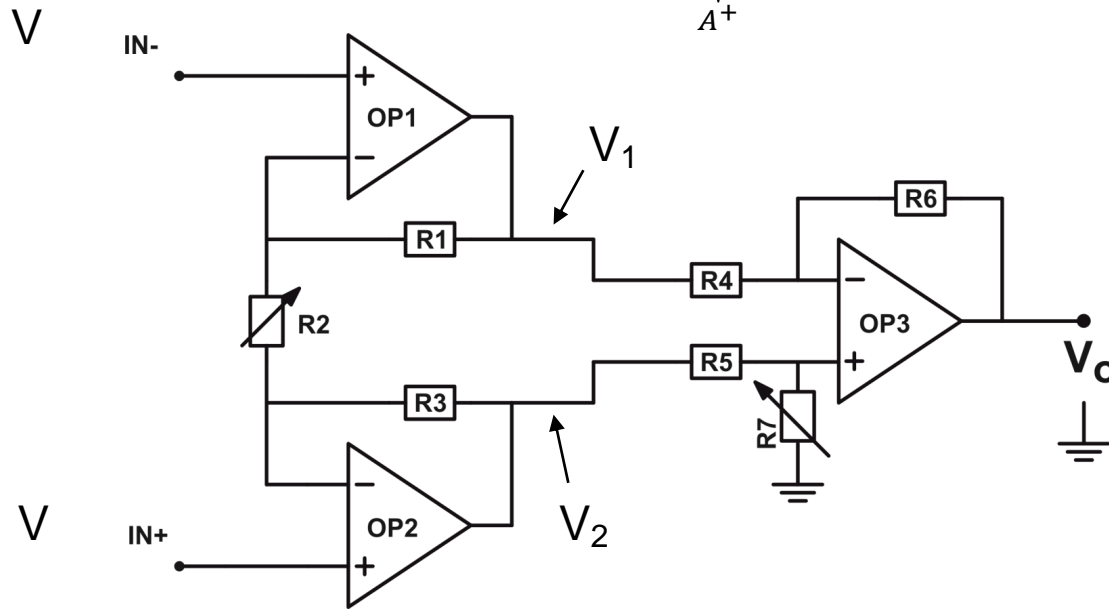
$$V_O = \left(V_{IN+} - \frac{R_3}{R_2} (V_{IN-} - V_{IN+}) \right) \cdot \frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4}\right) - \left(V_{IN-} + \frac{R_1}{R_2} (V_{IN-} - V_{IN+}) \right) \cdot \frac{R_6}{R_4}$$

$$V_O = V_{IN+} \cdot \underbrace{\left[\frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4}\right) \cdot \left(1 + \frac{R_3}{R_2}\right) + \frac{R_1}{R_2} \cdot \frac{R_6}{R_4} \right]}_{G^+} - V_{IN-} \cdot \underbrace{\left[\frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4}\right) \cdot \frac{R_3}{R_2} + \frac{R_6}{R_4} \cdot \left(1 + \frac{R_1}{R_2}\right) \right]}_{G^-}$$

continue ...

□ we want

$$\underbrace{\left[\frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4} \right) \cdot \left(1 + \frac{R_3}{R_2} \right) + \frac{R_1}{R_2} \cdot \frac{R_6}{R_4} \right]}_{A^+} = \underbrace{\left[\frac{R_7}{R_5 + R_7} \cdot \left(1 + \frac{R_6}{R_4} \right) \cdot \frac{R_3}{R_2} + \frac{R_6}{R_4} \cdot \left(1 + \frac{R_1}{R_2} \right) \right]}_{A^-}$$



□ by setting $R_1 = R_3$, $R_5 = R_4$ and $R_6 = R_7$

□ we obtain: $A^+ = A^- = \frac{R_6}{R_4} \left(1 + \frac{2R_1}{R_2} \right)$