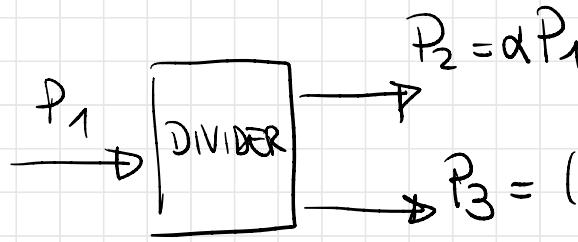


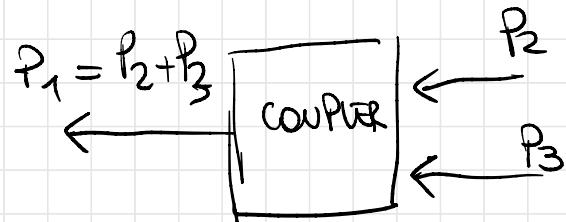
# MICROWAVE ENGINEERING

Lecture 26:  
Power dividers  
and couplers





Either 3 or 4 ports



3 ports  $\rightarrow$  T-junctions

4 ports  $\rightarrow$  Directional couplers  
or hybrids

## BASIC PROPERTIES

### 3 PORTS NETWORKS (T-JUNCTIONS)

If it's passive and there is no anisotropy  $\rightarrow$  [S] is symmetrical

NETWORK IS RECIPROCAL

NOTE : We cannot have a reciprocal, lossless and matched T-junction.

Demonstration : if  $S_{ii} = 0$  (all ports matched) and  $S_{ij} = S_{ji}$

$$[S] = \begin{bmatrix} 0 & S_{12} & S_{13} \\ S_{12} & 0 & S_{23} \\ S_{13} & S_{23} & 0 \end{bmatrix}$$

If the network is also lossless  $\rightarrow [S]$  must be unitary

$$\begin{cases} |S_{12}|^2 + |S_{13}|^2 = 1 \\ |S_{12}|^2 + |S_{23}|^2 = 1 \end{cases}$$

$$\left| S_{13} \right|^2 + \left| S_{23} \right|^2 = 1$$

and

$$\begin{aligned} S_{13}^* S_{23} &= 0 \\ S_{23}^* S_{12} &= 0 \\ S_{12}^* S_{13} &= 0 \end{aligned}$$

Both sets of equations cannot be satisfied

Example 1

### A CIRCULATOR

NON RECIPROCAL  
MATCHED  
LOSSLESS

$$[S] = \begin{bmatrix} 0 & S_{12} & S_{13} \\ S_{21} & 0 & S_{23} \\ S_{31} & S_{32} & 0 \end{bmatrix}$$

If lossless  $[S]$  is unitary

$$S_{31}^* S_{32} = 0$$

$$S_{21}^* S_{23} = 0$$

$$S_{12}^* S_{13} = 0$$

$$|S_{12}|^2 + |S_{13}|^2 = 1$$

$$|S_{21}|^2 + |S_{23}|^2 = 1$$

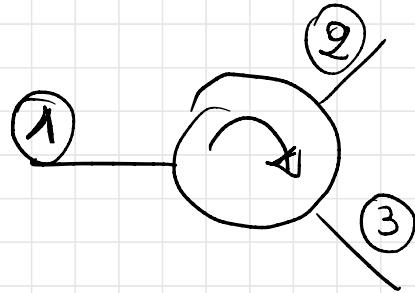
$$|S_{31}|^2 + |S_{32}|^2 = 1$$

$$\textcircled{1} \rightarrow S_{12} = S_{23} = S_{31} = 0 \rightarrow |S_{21}|^2 = |S_{32}|^2 = |S_{13}|^2 = 1$$

$$\textcircled{2} \rightarrow S_{21} = S_{32} = S_{13} = 0 \rightarrow |S_{12}|^2 = |S_{23}|^2 = |S_{31}|^2 = 1$$

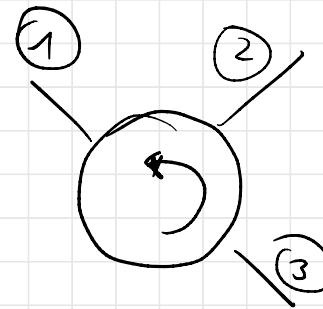
We can have 2 types of circulators:

$$[S] = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$$



or

$$[S] = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 1 & 0 & 0 \end{bmatrix}$$



Example 2: Lossless, reciprocal, with only 2 ports matched  
Example 3: lossy, nonreciprocal, Matched at all ports (RESISTIVE PROVIDER)

## FOUR PORTS NETWORKS (DIRECTIONAL COUPLES)

If reciprocal  
and  
matched

$$[S] = \begin{bmatrix} 0 & S_{12} & S_{13} & S_{14} \\ S_{12} & 0 & S_{23} & S_{24} \\ S_{13} & S_{23} & 0 & S_{34} \\ S_{14} & S_{24} & S_{34} & 0 \end{bmatrix}$$

$$\left\{ \begin{array}{l} S_{13}^* S_{23} + S_{14}^* S_{24} = 0 \quad \leftarrow \text{Multiply by } S_{24}^* \\ S_{14}^* S_{13} + S_{24}^* S_{23} = 0 \quad \leftarrow \text{Multiply by } S_{13}^* \end{array} \right.$$

Subtracting the first eq. from the second

$$- S_{2u}^* S_3^* S_{23} + S_{1u}^* |S_{2u}|^2 = 0$$

$$S_{1u}^* |S_{13}|^2 + S_{2u}^* S_{23} S_{13}^* = 0$$

$$\boxed{S_{1u}^* (|S_{13}|^2 - |S_{2u}|^2) = 0} \quad (1)$$

Then we have

$$S_{12}^* S_{23} + S_{1u}^* S_{3u} = 0 \quad \leftarrow \text{ multiply by } S_{12}$$

$$S_{1u}^* S_{12} + S_{3u}^* S_{23} = 0 \quad \leftarrow \quad " \quad " \quad S_{3u}$$

$$|S_{12}|^2 S_{23} + S_{12} S_{14}^* S_{34} = 0$$

$$- S_{14}^* S_{34} S_{12} + |S_{34}|^2 S_{23} = 0$$

$$\boxed{S_{23} (|S_{12}|^2 - |S_{34}|^2) = 0} \quad ②$$

If ① and ② have to be satisfied  $\Rightarrow S_{14} = S_{23} = 0$

From which result is :

$$[S] = \begin{bmatrix} 0 & S_{12} & S_{13} & 0 \\ S_{12} & 0 & 0 & S_{24} \\ S_{13} & 0 & 0 & S_{34} \\ 0 & S_{24} & S_{34} & 0 \end{bmatrix}$$

DIRECTIONAL COUPLER

For the coupler to be lossless

$$\left. \begin{array}{l} |S_{12}|^2 + |S_{13}|^2 = 1 \\ |S_{12}|^2 + |S_{24}|^2 = 1 \\ |S_{13}|^2 + |S_{34}|^2 = 1 \\ |S_{24}|^2 + |S_{34}|^2 = 1 \end{array} \right\} \Rightarrow \begin{array}{l} |S_{13}| = |S_{24}| \\ & \& \\ |S_{12}| = |S_{34}| \end{array}$$

If we choose a phase reference for 3 of the 4 ports:

$$\left\{ \begin{array}{l} S_{12} = S_{34} = \alpha \\ S_{13} = \beta e^{j\theta} \\ S_{24} = \beta e^{j\phi} \end{array} \right. \quad \begin{array}{l} \alpha, \beta \text{ real} \\ \theta, \phi \text{ constants} \end{array}$$

$$[S] = \begin{bmatrix} 0 & \alpha & \beta e^{j\theta} & 0 \\ \alpha & 0 & 0 & \beta e^{j\phi} \\ \beta e^{j\theta} & 0 & 0 & \alpha \\ 0 & \beta e^{j\phi} & \alpha & 0 \end{bmatrix}$$

$$S_{12}^* S_{13} + S_{23}^* S_{31} = 0$$

$$\cancel{\alpha} \beta e^{j\theta} + \cancel{\beta} e^{j\phi} \cancel{\alpha} = 0$$

$$\boxed{0 + \phi = \pi \pm 2n\pi}$$

So we can have the following couplers:

① The symmetrical coupler:  $\Theta = \phi = \frac{\pi}{2}$  (equal phase)

$$[S] = \begin{bmatrix} 0 & \alpha & j\beta & 0 \\ \alpha & 0 & 0 & j\beta \\ j\beta & 0 & 0 & \alpha \\ 0 & j\beta & \alpha & 0 \end{bmatrix}$$

② The Antisymmetrical coupler :  $\Theta = 0 \quad \Phi = \pi$

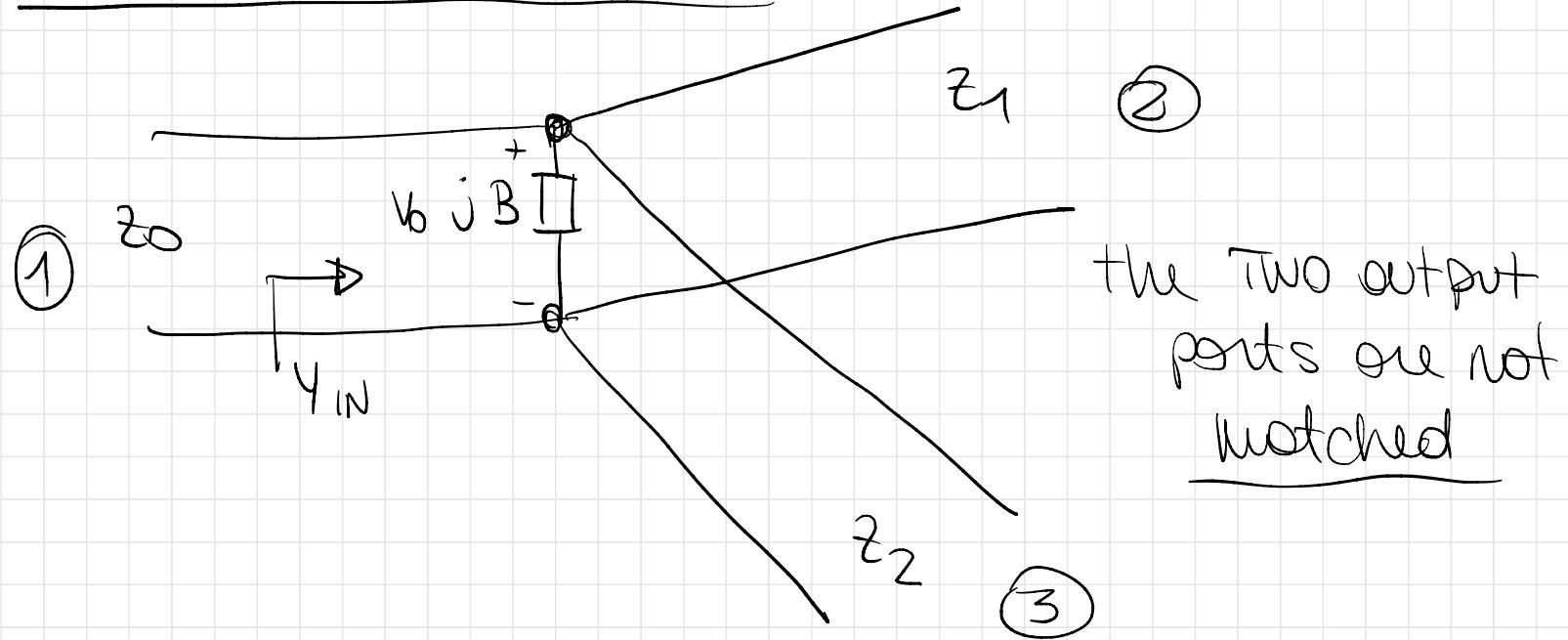
$$[S] = \begin{bmatrix} 0 & \alpha & \beta & 0 \\ \alpha & 0 & 0 & -\beta \\ \beta & 0 & 0 & \alpha \\ 0 & -\beta & \alpha & 0 \end{bmatrix}$$

(phases are  
180° apart)

NOTE :  $\alpha$  and  $\beta$  are NOT INDEPENDENT

$$\boxed{\alpha^2 + \beta^2 = 1}$$

# LOSSLESS POWER DIVIDER

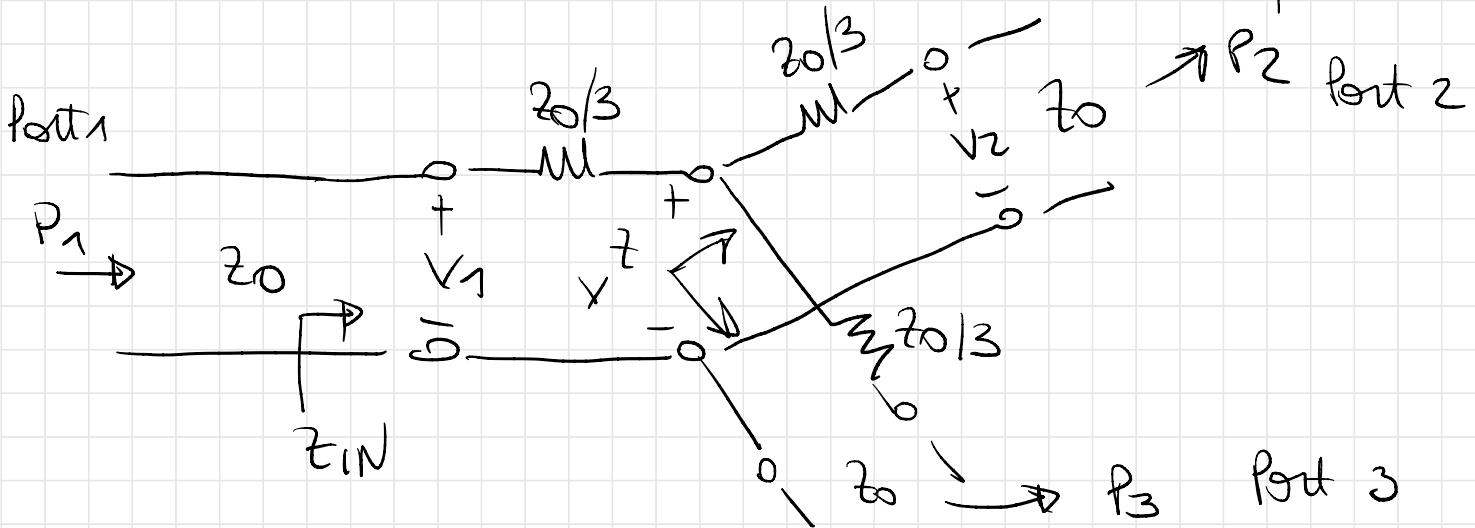


$$y_{IN} = jB + \frac{1}{z_1} + \frac{1}{z_2} = \frac{1}{z_0}$$

For example : If we assume  $Z_0 = 50 \Omega$  and we want a 3dB (equal split) power divider then (if  $B=0$ )  $\rightarrow Z_1 = Z_2 = 100 \Omega$

## RESISTIVE POWER DIVIDER

Lossless, reciprocal, matched at all ports!



$$Z = Z_0/3 + Z_0 = \frac{4}{3} Z_0$$

$$Z_{IN} = \frac{Z_0}{3} + \frac{2}{3} Z_0 = Z_0 \quad \leftarrow \text{Matched to feed the line}$$

↑  
Z/Z

↓

$$S_{11} = S_{22} = S_{33} = 0$$

We can calculate

$$V = V_1 \frac{Z_0/3}{Z_0 + \frac{2}{3} Z_0} = \frac{2}{3} V_1$$

The output voltages are

$$V_2 = V_3 = V \frac{\frac{Z_0}{Z_0 + \frac{Z_0}{3}}}{=} \frac{3}{4} V = \frac{1}{2} V_1$$

It follows:

$$S_{21} = S_{31} = S_{23} = \frac{1}{2} \quad (-6\text{dB below the input power})$$

Because the network is also reciprocal

$$[S] = \frac{1}{2} \begin{bmatrix} 0 & 1 & 1 \\ 1 & 0 & 1 \\ 1 & 1 & 0 \end{bmatrix}$$

The power at the input:

$$P_1 = P_{IN} = \frac{1}{2} \frac{V_1^2}{Z_0}$$

The output Power is

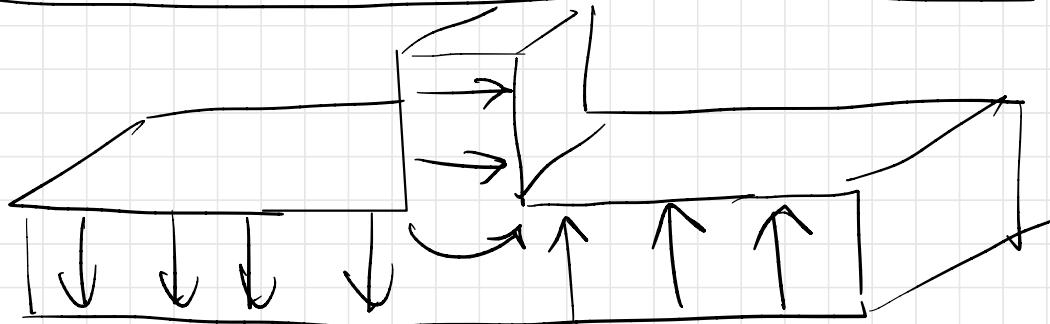
$$P_2 = P_3 = \frac{1}{2} \frac{\left(\frac{1}{2}V_1\right)^2}{Z_0} = \frac{1}{8} \frac{V_1^2}{Z_0} = \frac{1}{4} P_{IN}$$

$$P_2 + P_3 = \frac{1}{2} P_{IN}$$

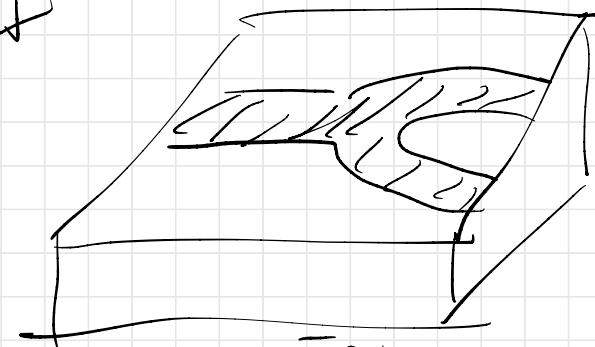
Half of the  $P_{IN}$  is lost in the resistors

$$P_{loss} = P_{IN} - P_2 - P_3 = \frac{1}{2} P_{IN}$$

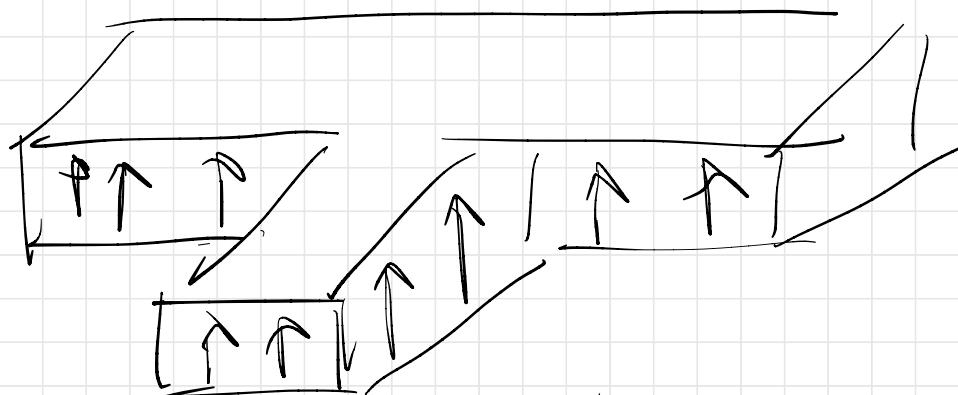
# T-Junction practical implementation



E-plane waveguide

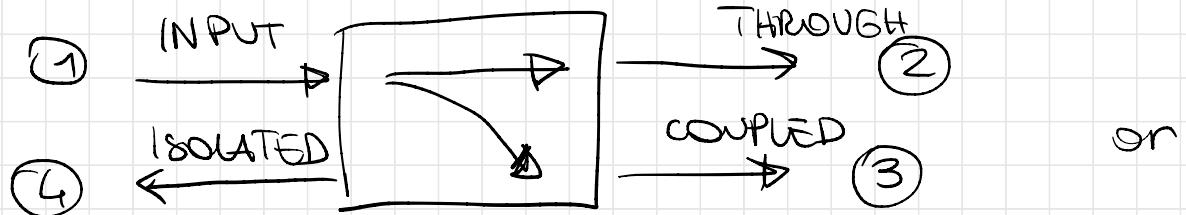


microstrip  
T-junction

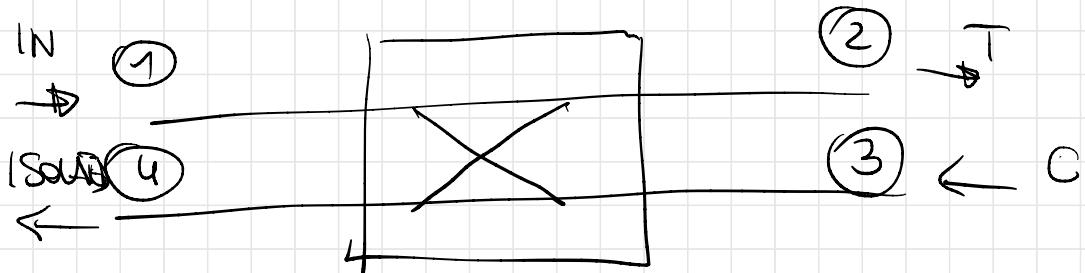


H-plane waveguide

# How does a directional coupler work?



or



Power goes in at port ① and exits at port ③ with coupling factor  $\sqrt{|S_{13}|^2} = \beta^2$

The rest of power goes to port ② with coeff.

$$|S_{12}|^2 = \alpha^2 = 1 - \beta^2$$

In an ideal directional coupler NO POWER GOES TO PORT ④

Fundamental parameters:

$$\text{COUPLING} = C = 10 \log \frac{P_1}{P_3} = -20 \log \beta \quad \text{dB}$$

$$\text{DIRECTIVITY} = D = 10 \log \frac{P_3}{P_4} = 20 \log \frac{\beta}{|S_{14}|} \quad \text{dB}$$

$$\text{ISOLATION} = I = 10 \log \frac{P_1}{P_4} = -20 \log |S_{14}| \quad \text{dB}$$

$$\bar{I} = D + C$$

The ideal coupler will have  $D = I = \infty$   
 $\alpha$  and  $\beta$  would depend on  $C$  only.

Special couplers ( $C = 3\text{dB} \rightarrow \alpha = \beta = \frac{1}{\sqrt{2}}$ )

## HYBRID COUPLERS

### ① Quadrature hybrid

SYMMETRICAL

$$[S] = \frac{1}{\sqrt{2}} \begin{bmatrix} 0 & 1 & j & 0 \\ 1 & 0 & 0 & j \\ j & 0 & 0 & -j \\ 0 & j & 1 & 0 \end{bmatrix}$$

90° phase shift between  
ports 2 and 3 when the

(input is at  
port 1).

② Magic-T Hybrid  
ANTI SYMMETRICAL

(rat-race hybrid)

180° phase shift between  
ports 2 and 3 when  
input is at port 4

$$[S] = \frac{1}{\sqrt{2}} \begin{bmatrix} 0 & 1 & 1 & 0 \\ 1 & 0 & 0 & -1 \\ 1 & 0 & 0 & 1 \\ 0 & -1 & 1 & 0 \end{bmatrix}$$