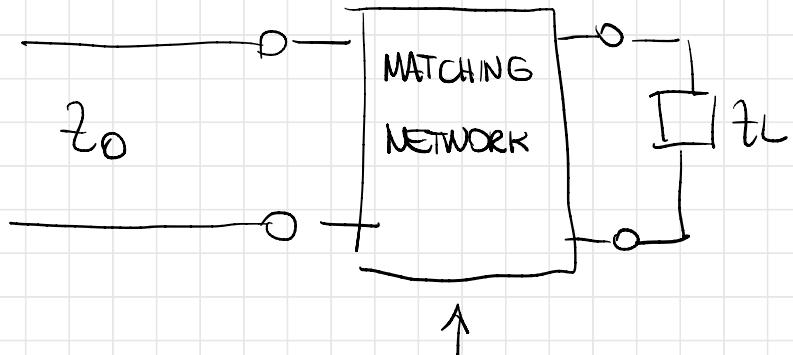


# MICROWAVE ENGINEERING

Lecture 16:  
Impedance  
Matching and  
Tuning



Typically lossless

- Max power is delivered to the load (assuming the generator is matched)
- Power loss is minimized

- Improves signal-to-noise ratio
- Reduces amplitude and phase errors in power distribution networks.

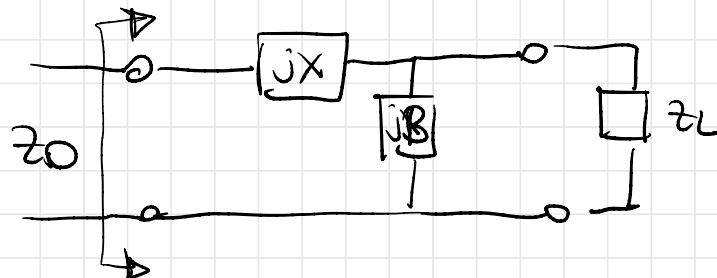
Matching network should be chosen to:

- MINIMIZE COMPLEXITY
- MATCHING REQUIRED BANDWIDTH
- IMPLEMENTATION REQUIREMENTS
- ADJUSTABILITY TO VARIABLE LOADS

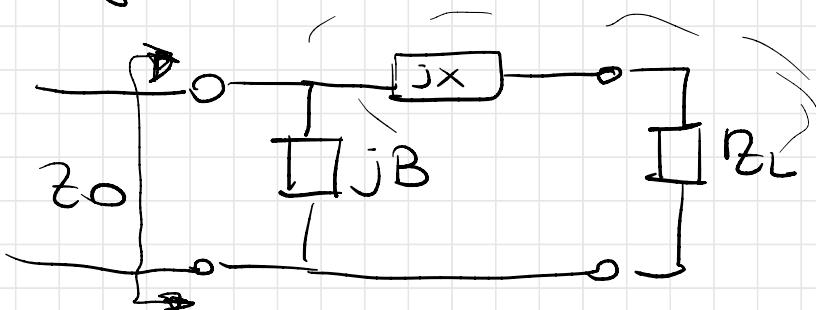
## MATCHING WITH WRAPPED ELEMENTS

The simplest approach consists of an L-section

- ① If  $z_0 = \frac{Z_L}{Z_0}$  falls inside the  $1+jx$  circle on the Smith's chart then the network is:



② If  $z_e = \frac{z_L}{z_0}$  falls outside the  $1+jX$  circle on Smith's chart ( $\frac{R_e}{z_0} < 1$ ) then the matching network is:-



LIMIT:  $jB$  and  $jX$  are actual inductors and capacitors.

Let's find the values of  $B$  and  $X$ :

If our load is  $Z_L = R_L + jX_L$        $z_L = \frac{Z_L}{Z_0}$

and if  $\boxed{R_L > Z_0}$

$$Z_0 = jX + \frac{1}{jB + \frac{1}{R_L + jX_L}}$$

after rearranging

$$B = \frac{X_L \pm \sqrt{R_L/Z_0} \sqrt{-\underline{R_L^2 + X_L^2 - Z_0 R_L}}}{R_L^2 + X_L^2} \rightarrow \text{always } > 0$$

$$X = \underline{B} + \frac{X_L Z_0}{R_L} - \frac{Z_0}{B R_L}$$

Two possible solutions can be found for  $X$  and  $B$ :

$X > 0 \rightarrow$  inductor       $B > 0 \rightarrow$  capacitor

$X < 0 \rightarrow$  capacitor       $B < 0 \rightarrow$  inductor

If  $\boxed{R_L < Z_0}$

$$\frac{1}{Z_0} = jB + \frac{1}{R_L + j(X+X_L)}$$

Solving for  $B$  and  $X$

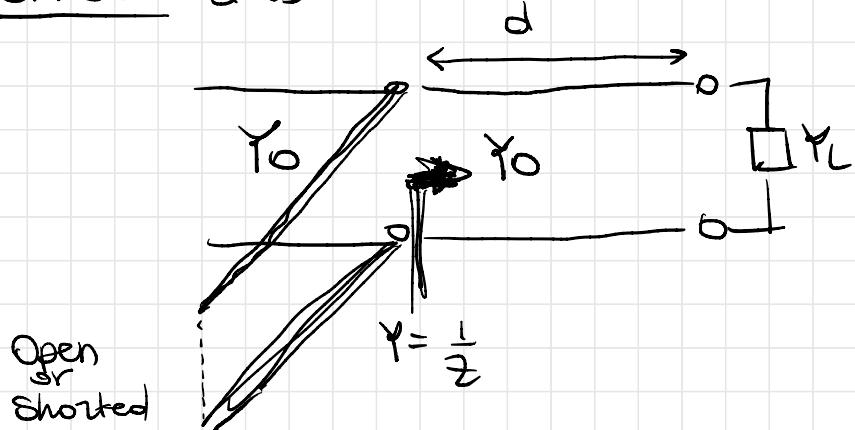
$$X = \pm \sqrt{R_L(Z_0 - R_L)} - X_L \quad \begin{array}{l} Z_0 > R_L \\ \sqrt{\phantom{x}} > 0 \end{array}$$

$$B = \pm \sqrt{\frac{(Z_0 - R_L)}{R_L}} \cdot \frac{1}{Z_0}$$

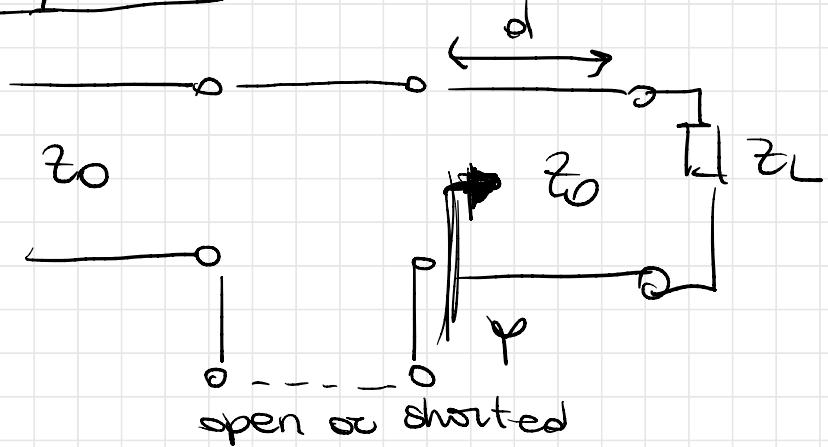
## SINGLE STUB TUNING

Very convenient and easy  
to fabricate.

### series stub



## parallel / shunt stub



## ADJUSTABLE PARAMETERS

- distance  $d$
- the value of susceptance or reactance of the stub.

- shunt-stub =  $\begin{cases} \text{open} \\ \text{shorted} \end{cases}$

] we select  $d$  so that the admittance  $\Upsilon = Y_0 + jB$  where seen from a distance  $d$

- series - stub  We select  $d$  so that the impedance seen is  $Z = Z_0 + jX$

The stub reactance is chosen to cancel the imaginary part  $\rightarrow -jB$  or  $-jX$

### SHUNT STUB MATCHING

$$Z_L = \frac{1}{Y_L} = R_L + jX_L$$

$$Z = Z_0 \frac{(R_L + jX_L) + jZ_0 \tan \beta d}{Z_0 + j(R_L + jX_L) \tan \beta d}$$

$$\gamma = \frac{1}{Z} = G + jB$$

$$t = \tan \beta d$$

$$\textcircled{1} \quad G = \frac{R_L (1 + t^2)}{R_L^2 + (X_L + Z_0 t)^2}$$

$$\textcircled{2} \quad B = \frac{R_L^2 t - (Z_0 - X_L t)(X_L + Z_0 t)}{Z_0 [R_L^2 + (X_L + Z_0 t)^2]}$$

d has to be chosen so that  $G = \gamma_0 = \frac{1}{Z_0}$

$$Z_0 (R_L - Z_0) t^2 - 2X_L Z_0 t + (R_L Z_0 - R_L^2 - X_L^2) = 0$$

Solving for  $t$ :

$$t = \frac{X_L \pm \sqrt{R_L [(Z_0 - R_L)^2 + X_L^2] / Z_0}}{R_L - Z_0}$$

$$\boxed{\text{If } R_L \neq Z_0}$$

$\hookrightarrow d$  is calculated

$$\text{If } \boxed{R_L = Z_0} \quad t = -\frac{X_L}{2Z_0}$$

The principal solutions are:

$$\boxed{\frac{d}{\lambda} = \begin{cases} \frac{1}{2\pi} \tan^{-1} t & t \geq 0 \\ \frac{1}{2\pi} (\pi + \tan^{-1} t) & t < 0 \end{cases}}$$

The stub susceptance

$$\underline{B_s = -B}$$

If the stub is open-circuit:

$$\frac{l_0}{\lambda} = \frac{1}{2\pi} \tan^{-1} \left( \frac{B_s}{Y_0} \right) = -\frac{1}{2\pi} \tan^{-1} \left( \frac{B}{Y_0} \right)$$

If the stub is short-circuited

$$\frac{l_{s0}}{\lambda} = \frac{-1}{2\pi} \tan^{-1} \left( \frac{Y_0}{B_s} \right) = \frac{1}{2\pi} \tan^{-1} \left( \frac{Y_0}{B} \right)$$

If  $l_0$  and  $l_s$  are negative  $\rightarrow$  we add  $\frac{\lambda}{2}$  line to make the distance or length positive

## SERIES-STUB

If load admittance is  $\gamma_L = \frac{1}{Z_L} = G_L + jB_L$

then after a distance d

$$\gamma = \gamma_0 \frac{(G_L + jB_L) ; \text{to tancd}}{\gamma_0 + j(G_L + jB_L) \tan \beta d} \quad t = \tan \beta d$$

$$z = R + jX = \frac{1}{\gamma}$$

$$R = \frac{G_L (1+t^2)}{G_L^2 + (B_L + \gamma_0 t)^2} \leftarrow$$

$$X = \frac{G_L^2 t - (\gamma_0 - t B_L)(B_L + t \gamma_0)}{\gamma_0 [G_L^2 + (B_L + \gamma_0 t)^2]}$$

d is chosen so that  $R = Z_0 = \frac{1}{\gamma_0}$

it follows :

$$\rightarrow \gamma_0 (G_L - \gamma_0) t^2 - 2 B_L \gamma_0 t + (G_L \gamma_0 - G_L^2 - B_L^2) = 0$$

$$t = \frac{B_L \pm \sqrt{G_L [(\gamma_0 - G_L)^2 + B_L^2]}}{G_L - \gamma_0} \quad G_L \neq \gamma_0$$

$$\text{If } G = \infty \Rightarrow t = -\frac{B_L}{2Z_0}$$

the principal solutions are

$$\frac{d}{\lambda} = \begin{cases} \frac{1}{2\pi} \tan^{-1} t & t \geq 0 \\ \frac{1}{2\pi} (\pi + \tan^{-1} t) & t < 0 \end{cases}$$

Imposing the susceptance of the step cancels  $X$  then  
 $(X_S = -X)$

short-circuit

$$\left| \frac{ls}{\lambda} = \frac{1}{2\pi} \tan^{-1} \left( \frac{X_S}{Z_0} \right) = \frac{1}{2\pi} \tan^{-1} \left( \frac{X}{Z_0} \right) \right.$$

open-area

$$\frac{lo}{\lambda} = -\frac{1}{2\pi} \tan^{-1} \left( \frac{z_0}{x_s} \right) = \frac{1}{2\pi} \tan^{-1} \left( \frac{z_0}{x} \right)$$