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Lesson 6  
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## II Definition of the characteristic impedance $Z_c(p)$ and of the reflection coefficient $\gamma(p)$

### 2.1 $Z_c(p)$

$$\left| \begin{array}{l} V(z,p) = V_L(p) e^{-\gamma(p)z} + V_R(p) e^{\gamma(p)z} \\ I(z,r) = I_L(p) e^{-\gamma(p)z} + I_R(p) e^{\gamma(p)z} \end{array} \right.$$

$$|| Z_c(p) = \frac{V_L(p)}{I_L(p)}$$

$$\text{Impedance at point } z_0 \text{ of the line } Z(p) = \frac{V(z_0,p)}{I(z_0,p)}$$

$$-\frac{\partial V(z,t)}{\partial z} = R e(z,t) + L \frac{\partial I(z,t)}{\partial t}$$

$$- \frac{\partial V(z,p)}{\partial z} = (R + Lp) I(z,p)$$

$$\begin{aligned} \gamma(p) V_L(p) e^{-\gamma(p)z} - \gamma(p) V_R(p) e^{\gamma(p)z} &= \\ (\eta + Lp) (I_L(p) e^{-\gamma(p)z} + I_R(p) e^{\gamma(p)z}) & \end{aligned}$$

$$\gamma(p) V_L(p) = (\eta + Lp) I_L(p)$$

$$- \gamma(p) V_R(p) = (\eta + Lp) I_R(p)$$

$$\frac{V_L(p)}{I_L(p)} = \frac{R + Lp}{\gamma(p)} = \frac{R + Lp}{\sqrt{(R + Lp)(G + Cp)}} = \sqrt{\frac{R + Lp}{G + Cp}}$$

$$|| Z_c(p) = \sqrt{\frac{R + Lp}{G + Cp}}$$

$$\frac{V_R(p)}{I_R(p)} = - Z_c(p)$$

$$|| I(z,p) = \frac{1}{Z_c(p)} (V_L(p) e^{-\gamma(p)z} - V_R(p) e^{\gamma(p)z})$$

$$|| V(z,p) = V_L(p) e^{-\gamma(p)z} + V_R(p) e^{\gamma(p)z}$$

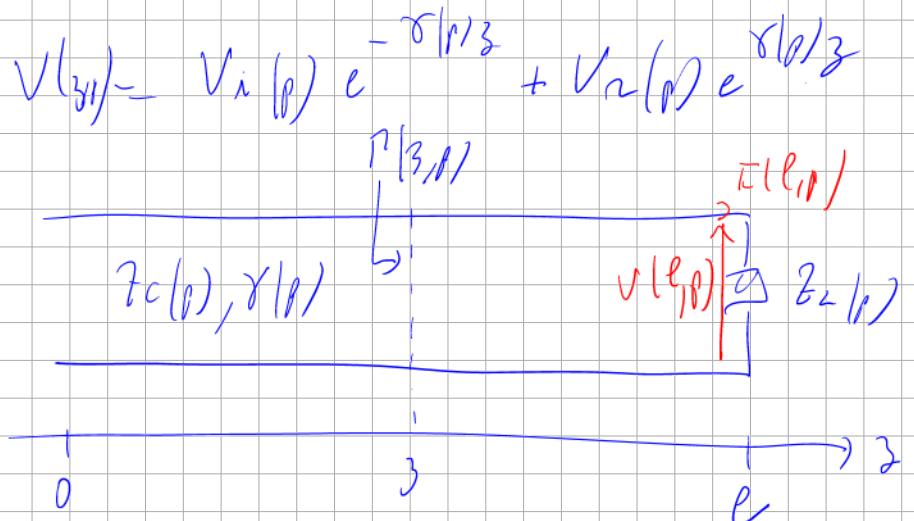
$$\text{lossless line : } Z_c(p) = \sqrt{\frac{L}{C}} \quad n = G = 0$$

$$\text{harmonic domain : } p = j \omega$$

$$|| Z_c = \sqrt{\frac{\eta + jL\omega}{G + jC\omega}}$$

2-2-  $\Gamma(p)$

$$\Gamma_{(3,p)} = \frac{V_{u(p)} e^{-\gamma(p)z}}{V_u(p) e^{-\gamma(p)z}} = \frac{V_u(p)}{V_i(p)} e^{2\gamma(p)z}$$



$$V(l,p) = Z_L(p) I(l,p)$$

$$V_u(p) e^{-\gamma(p)l} + V_l(p) e^{\gamma(p)l} = \frac{Z_L(p)}{Z_0} (V_u(p) e^{-\gamma(p)l} - V_l(p) e^{\gamma(p)l})$$

$$V_u(p) e^{-\gamma(p)l} \left[ 1 - \frac{Z_L(p)}{Z_0} \right] = -V_l(p) e^{\gamma(p)l} \left[ 1 + \frac{Z_L(p)}{Z_0} \right]$$

$$\frac{V_u(p)}{V_l(p)} = \frac{Z_L(p) - Z_0}{Z_L(p) + Z_0} e^{-2\gamma(p)l}$$

$$\Gamma(z/p) = \frac{Z_L(p) - Z_0}{Z_L(p) + Z_0} e^{2\gamma(p)z}$$

on the load  $z = l$

$$\Gamma(l,p) = \Gamma_L(p) = \frac{Z_L(p) - Z_0}{Z_L(p) + Z_0}$$

\*  $Z_L(p) = Z_0$

$\Gamma_L(p) = 0$  incident wave

the line is matched

\*  $Z_L(p) = 0$  short circuit

$$\Gamma_L(p) = -1$$

\*  $Z_L(p) \rightarrow \infty$  open circuit

$$\Gamma_L(p) = +1$$



## 2-3. harmonic study

$$\rho = f \omega$$

$$V(z) = V_1 e^{+j\beta z} + V_2 e^{-j\beta z}$$

$$I(z) = \frac{1}{Z_c} (V_1 e^{+j\beta z} - V_2 e^{-j\beta z})$$

incident wave  $V_1 e^{j\beta z}$



$\beta$   $\leftarrow$   
 $Z_c$  real loss line

$$V(z) = V_1 e^{j\beta z} (1 + \rho(z))$$

$$\rho(z) = \frac{V_2}{V_1} e^{-2j\beta z}$$

$$\rho(0) = \frac{V_2}{V_1} = \frac{Z_L - Z_c}{Z_L + Z_c}$$

$$\rho(z) = \rho(0) e^{-2j\beta z}$$

$$= \rho_L e^{-2j\beta z}$$

$$\rho_L = |\rho_L| e^{j\phi_L}$$

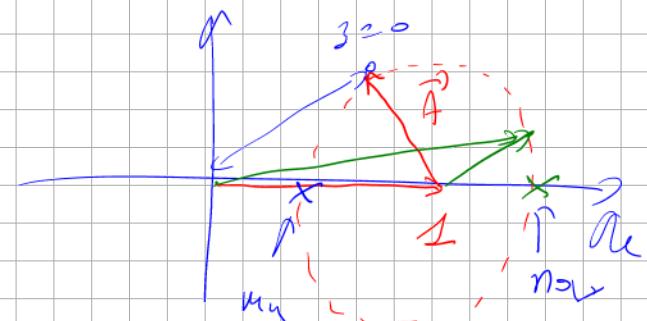
$$\rho(z) = |\rho_L| e^{j(\phi_L - 2\beta z)}$$

$$V(z) = V_1 e^{j\beta z} \left( 1 + |\rho_L| e^{j(\phi_L - 2\beta z)} \right)$$

$$|V(z)| = |V_1| \left| 1 + \frac{|\rho_L| e^{-j(\phi_L - 2\beta z)}}{1 + \frac{|\rho_L|}{|V_1|} e^{-j(\phi_L - 2\beta z)}} \right|$$

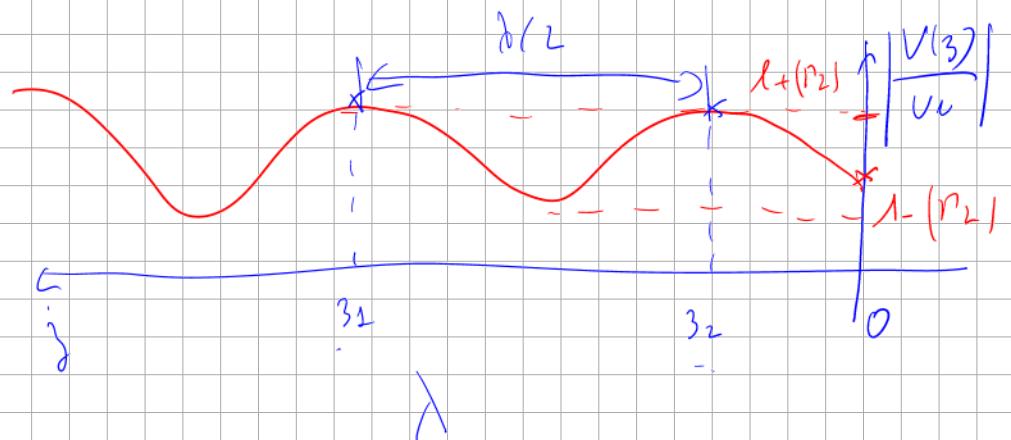
$$|V(z)|_{\max} = |V_1| \left( 1 + \frac{|\rho_L|}{|V_1|} \right)$$

$$\phi_L - 2\beta z = 2k\pi$$



$$|V(z)|_{\min} = |V_1| (1 - |\rho_L|)$$

$$\phi_L - 2\beta z = 2(k+1)\pi$$



between two max of  $|V(z)|$

$$(k_1 - 2\beta_{31}) - (k_1 - 2\beta_{32}) = 2\pi$$

$$|\chi\beta(z_2-z_1)| = \pi\pi \quad \beta = \frac{\omega}{\lambda}$$

$$|\beta_2 - \beta_1| = \frac{\lambda}{2}$$

SWR Standing wave ratio

$$\text{||} \quad S = \frac{V_{max}}{V_{min}} = \frac{1 + |\rho_L|}{1 - |\rho_L|}$$

$$\begin{aligned} Z_L &= 0 \quad \rho_L = -1 \quad V_2 = -V_1 \\ V(z) &= V_1 e^{j\beta_3 z} + V_2 e^{-j\beta_3 z} \\ &= V_1 (e^{j\beta_3 z} - e^{-j\beta_3 z}) \\ &= 2V_1 \sin(\beta_3 z) \\ N(z,t) &= \partial_z (2V_1 \sin(\beta_3 z) e^{j\omega t}) \\ &= -2V_1 \underbrace{\sin(\beta_3 z)}_{\neq \sin(\omega t - \beta_3 z)} \omega \cos(\omega t) \end{aligned}$$

$S \rightarrow \infty$  no propagation























































