Control Systems

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Abstract—**This manual is an introduction to control systems based on GATE problems.Links to sample Python codes are available in the text.**

Download python codes using

svn co https://github.com/gadepall/school/trunk/
   control/codes

1 SIGNAL FLOW GRAPH

1.1 Mason's Gain Formula

1.2 Matrix Formula

2 BODE PLOT

2.1 Introduction

2.2 Example

3 SECOND ORDER SYSTEM

3.1 Daming

3.2 Example

3.1. In the Feedback System given below

$$G(s) = \frac{1}{s^2 + 2s} \qquad (3.1.1)$$

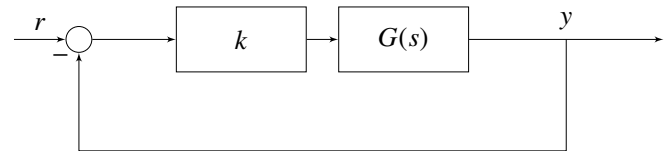The step response of the closed-loop system should have minimum settling time and have no overshoot

Fig. 3.1: Block Diagram of given question

3.2. The required value of gain 'k' to achieve this is

Solution:

Settlingtime : The time required for the transient's damped oscillations to reach and stay within 2% of the steady-state value.

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Overshoot : The amount that the waveform overshoots the steady state, or final, value at the peak time, expressed as a percentage of the steady-state value.

The Transfer function of the negative unity feedback system is given by

$$H(s) = \frac{G(s)}{1 + G(s)H(s)} \qquad (3.2.1)$$

Where G(s) is the open loop gain of the feedback system and H(s) is the gain of feedback system
In our case

$$G(s) = k * G(s) = kG(s) \qquad (3.2.2)$$
$$H(s) = 1 \qquad (3.2.3)$$

From (3.2.1) , (3.2.2) and (3.2.3)
Transfer function of Whole feedback system becomes

$$H(s) = \frac{kG(s)}{1 + kG(s)} \qquad (3.2.4)$$

From (3.1.1) Substituting G(s) in (3.2.4)

Our Transfer Function becomes

$$H(s) = \frac{k * \frac{1}{s^2+2s}}{1 + k * \frac{1}{s^2+2s}} \qquad (3.2.5)$$

On further simplifying we get,

$$H(s) = \frac{k}{s^2 + 2s + k} \qquad (3.2.6)$$

Depending upon the value of 'k' the system can be :

- Undamped System
- Underdamped System
- Critical Damped System
- Over Damped System

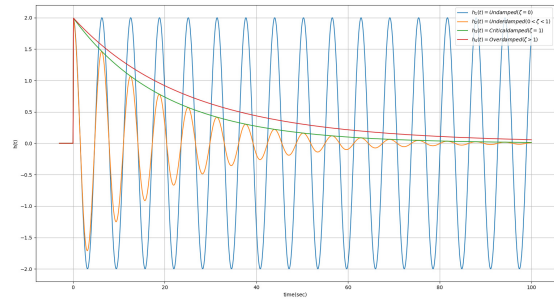Plot of various types of systems in time domain:

Fig. 3.2: Different Types of systems

Python code for the above plot is

codes/ee18btech11035_1.py

Now, graphically observing the system which has minimum settling time and also doesn't overshoot.

From Figure:(3.2)
The system with minimum settling is critical damped system also this system doesn't overshoot.

So,when unit step is given as an input for critical damped system the output of the system has minimum settling time also the output doesn't overshoot.

So, our transfer function H(s) should be a critical damped system.
The general form of a second order system is

$$H(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \qquad (3.2.7)$$

where ζ is the damping factor

- $\zeta = 0$ for a undamped system
- $0 < \zeta < 1$ for a underdamped system
- $\zeta = 1$ for a critical damped system
- $\zeta > 1$ for a overdamped system

In our case the damping factor for H(s) is 1 as it should be critical damped system

By comparing the equations (3.2.6) and (3.2.7)
We get,

$$\omega_n^2 = k \qquad (3.2.8)$$

$$\omega_n = \sqrt{k} \qquad (3.2.9)$$

$$2\zeta\omega_n s = 2s \qquad (3.2.10)$$

$$\zeta\omega_n = 1 \qquad (3.2.11)$$

From (3.2.9) and (3.2.11)

$$\zeta\sqrt{k} = 1 \qquad (3.2.12)$$

As $\zeta = 1$ (3.2.12) becomes

$$\sqrt{k} = 1 \qquad (3.2.13)$$

$$k = 1 \qquad (3.2.14)$$

Therefore, Transfer function is

$$H(s) = \frac{1}{s^2 + 2s + 1} \qquad (3.2.15)$$
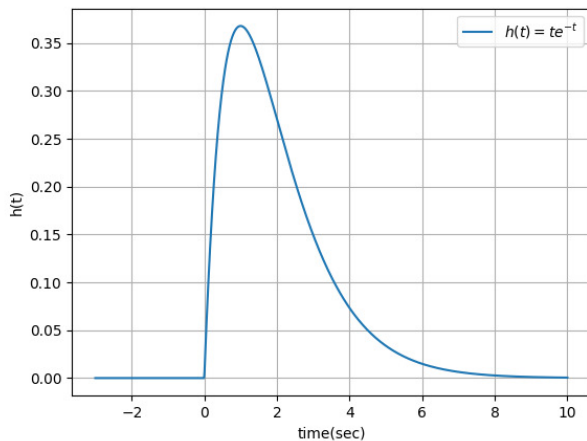
3.3. Plot of Transfer function.
Solution:

Fig. 3.3: Plot of h(t)

Python code for the above plot h(t) is

codes/ee18btech11035_2.py

$$H(s) = \frac{1}{s^2 + 2s + 1} \qquad (3.3.1)$$

$$\implies H(s) = \frac{1}{(s + 1)^2} \qquad (3.3.2)$$

Calculating the time domain equivalent for the transfer function

$$\frac{1}{s} = u(t) \qquad (3.3.3)$$

$$\frac{1}{s + 1} = e^{-t}u(t) \qquad (3.3.4)$$

$$\frac{1}{(s + 1)^2} = te^{-t}u(t) \qquad (3.3.5)$$

$$\qquad (3.3.6)$$

Therefore,

$$h(t) = te^{-t}u(t) \qquad (3.3.7)$$
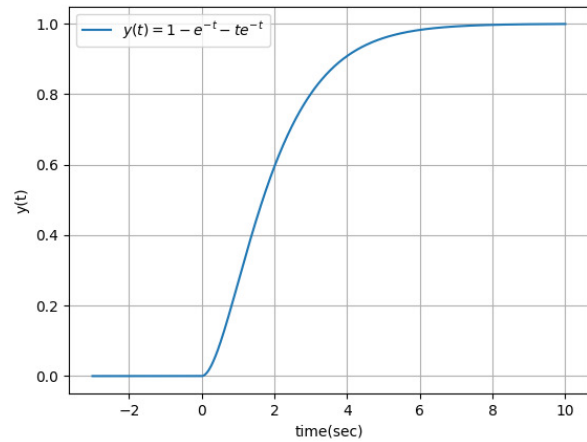
3.4. Plot of Output when input is unit step.
Solution:

Fig. 3.4: Plot of y(t)

Python code for the above plot y(t) is

codes/ee18btech11035_3.py

We know that

$$x(t) = u(t) \qquad (3.4.1)$$

Converting x(t) in s-domain

$$X(s) = \frac{1}{s} \qquad (3.4.2)$$

$$Y(s) = H(s) * X(s) \qquad (3.4.3)$$

$$Y(s) = \frac{1}{(s + 1)^2} * \frac{1}{s} \qquad (3.4.4)$$

Converting Y(s) into partial fraction
We get,

$$Y(s) = \frac{1}{s} - \frac{1}{s + 1} - \frac{1}{(s + 1)^2} \qquad (3.4.5)$$

Converting Y(s) into y(t)
As mentioned in (3.3.3), (3.3.4) and (3.3.5)

$$y(t) = (1 - e^{-t} - te^{-t})u(t) \qquad (3.4.6)$$