

The flesh of polar codes

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Easy channels

Among all channels, there are two classes for which it is easy to communicate optimally:

- The **perfect channels**: the output Y determines the input X .
- The **useless channels**: the output Y is independent of the input X .

In a perfect world all channels would be of one of those extremal types.

Arikan's **polarization** is a technique to convert any binary-input channel to a mixture of binary-input **extremal** channels.

- The technique is **information lossless**, and of **low complexity**.

Polar transform

Given two copies of a binary input channel $W: \mathbb{F}_2 \rightarrow \mathcal{Y}$

- Set

$$X_1 = U_1 \oplus U_2$$

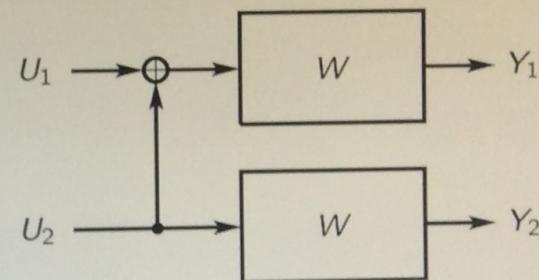
$$X_2 = U_2$$

with (U_1, U_2) uniform on \mathbb{F}_2^2 .

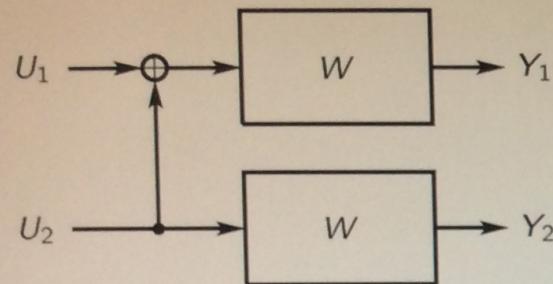
- As (X_1, X_2) is also uniform on \mathbb{F}_2^2 ,

$$\begin{aligned} 2I(W) &= I(X_1 X_2; Y_1 Y_2) = I(U_1 U_2; Y_1 Y_2) = I(U_1; Y_1 Y_2) + I(U_2; Y_1 Y_2 | U_1) \\ &= I(U_1; Y_1 Y_2) + I(U_2; Y_1 Y_2 | U_1) \\ &= I(W^-) + I(W^+) \end{aligned}$$

- $I(W^-) \leq I(W) \leq I(W^+)$.



Are the synthetic channels real?



The students need to be convinced they are not being fooled.

- The channel $W^- : U_1 \rightarrow Y_1 Y_2$ is a bonafide channel: U_1 is controlled by the transmitter, $Y_1 Y_2$ is observed by the receiver.
- The channel $W^+ : U_2 \rightarrow Y_1 Y_2 \textcolor{red}{U}_1$ poses problems: U_1 is not observed by the receiver.
- We can **synthesize** W^+ by first processing the output of W^- to estimate U_1 . Even then, we only have \hat{U}_1 . Is it legitimate to pretend that we have U_1 instead of \hat{U}_1 ?

Genie-aided receiver:

$$\tilde{U}_1 = \phi_1(Y^2)$$

$$\tilde{U}_2 = \phi_2(Y^2 \textcolor{red}{U}_1)$$

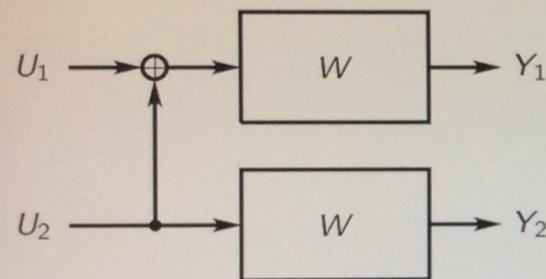
Implementable receiver:

$$\hat{U}_1 = \phi_1(Y^2)$$

$$\hat{U}_2 = \phi_2(Y^2 \hat{U}_1)$$

- If the genie-aided receiver makes no errors, then, the implementable receiver also makes no errors. Indeed, the **block** error events $\{\tilde{U}^2 \neq U^2\}$ and $\{\hat{U}^2 \neq U^2\}$ are identical.
- We can choose to analyze the genie-aided receiver instead of the implementable one. For the purposes of this analysis the synthetic channels are real. The implementation will use \hat{U}_1 instead.

Example: Erasure channel



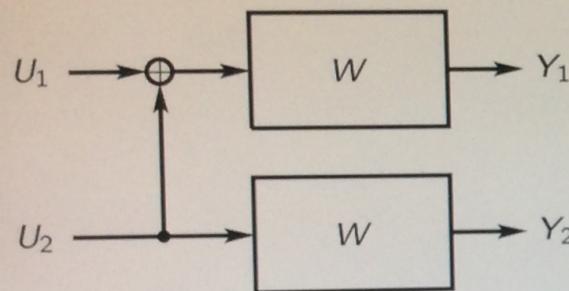
Suppose W is a BEC(p), i.e.,

$$Y = \begin{cases} X & \text{with probability } 1 - p, \\ ? & \text{with probability } p \end{cases}$$

- W^- has input U_1 , output

$$(Y_1, Y_2) = \begin{cases} (U_1 \oplus U_2, U_2) & \text{w.p. } (1-p)^2 \\ (?, U_2) & \text{w.p. } p(1-p) \\ (U_1 \oplus U_2, ?) & \text{w.p. } (1-p)p \\ (?, ?) & \text{w.p. } p^2 \end{cases}$$

Example: Erasure channel



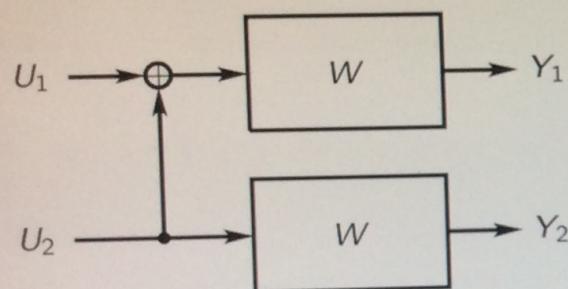
Suppose W is a BEC(p), i.e.,

$$Y = \begin{cases} X & \text{with probability } 1 - p, \\ ? & \text{with probability } p \end{cases}$$

- W^- is a BEC(p^-); $p^- = 2p - p^2$.
- W^+ has input U_2 , output

$$(Y_1, Y_2, U_1) = \begin{cases} (U_1 \oplus U_2, U_2, U_1) & \text{w.p. } (1-p)^2 \\ (?, U_2, U_1) & \text{w.p. } p(1-p) \\ (U_1 \oplus U_2, ?, U_1) & \text{w.p. } (1-p)p \\ (?, ?, U_1) & \text{w.p. } p^2 \end{cases}$$

Example: Erasure channel



Suppose W is a BEC(p), i.e.,

$$Y = \begin{cases} X & \text{with probability } 1 - p, \\ ? & \text{with probability } p \end{cases}$$

- W^- is a BEC(p^-); $p^- = 2p - p^2$.
- W^+ is a BEC(p^+); $p^+ = p^2$.

- We already begin to see some extremalization: W^+ is better than W , while W^- is worse.

Polarization construction

What we can do once, we can do many times. Given $W : \mathbb{F}_2 \rightarrow \mathcal{Y}$,

- Duplicate $W : X \rightarrow Y$ and obtain
 $W^- : U_1 \rightarrow Y_1 Y_2$ and $W^+ : U_2 \rightarrow Y_1 Y_2 U_1$.
- Duplicate W^- (and also W^+)
- and obtain

$$W^{--} : V_1 \rightarrow Y_1 Y_2 Y_3 Y_4$$

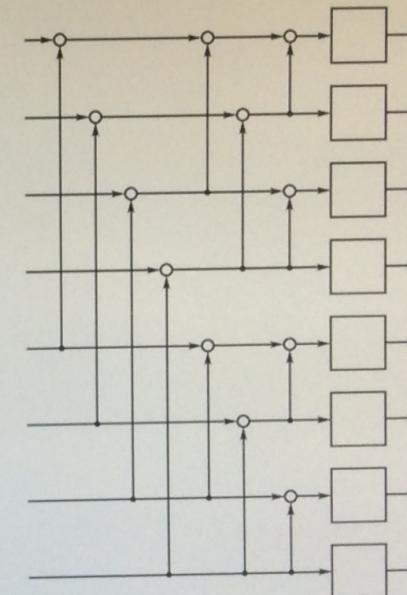
$$W^{-+} : V_2 \rightarrow Y_1 Y_2 Y_3 Y_4 V_1$$

and also

$$W^{+-} : V_3 \rightarrow Y_1 Y_2 Y_3 Y_4 V_1 V_2$$

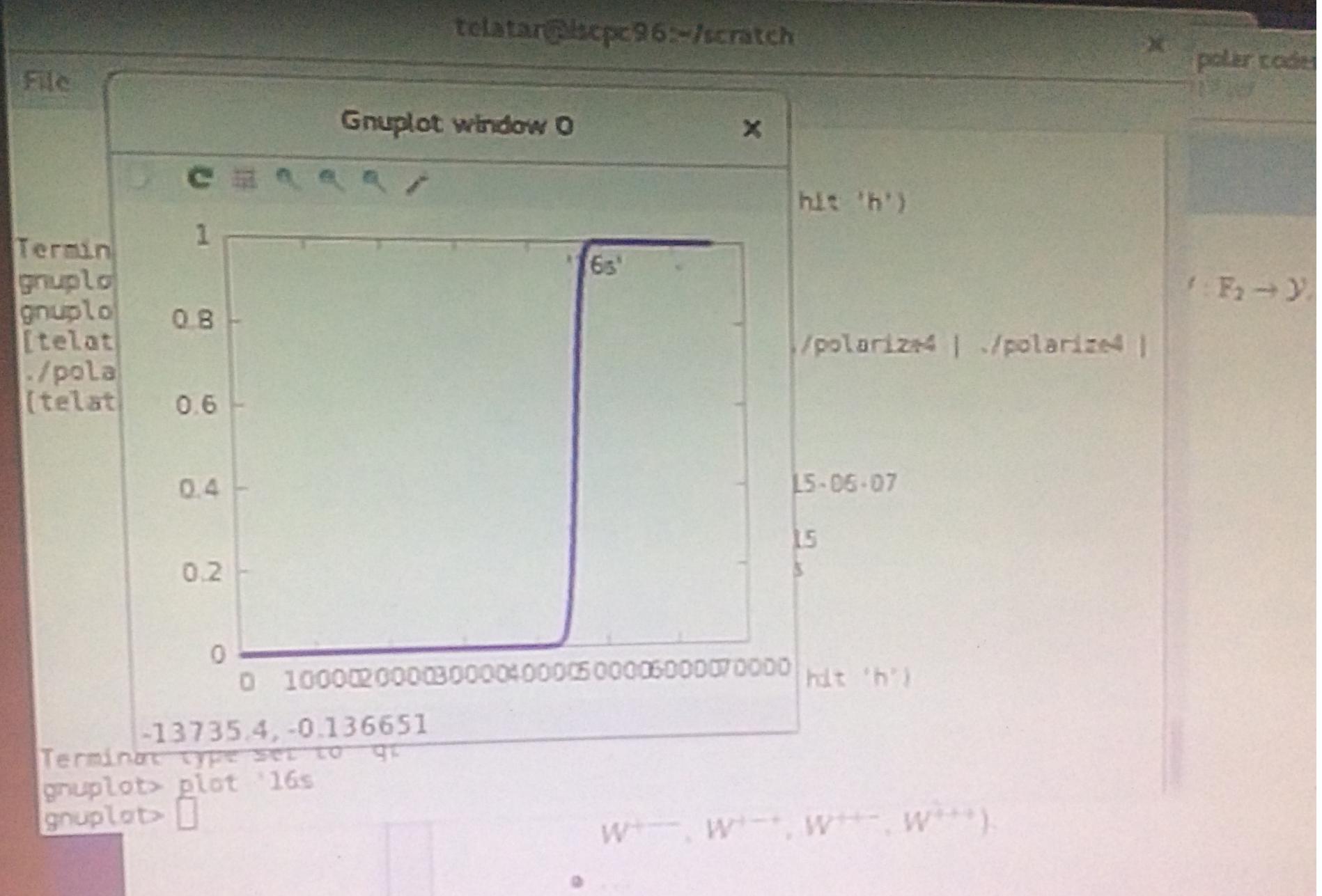
$$W^{++} : V_4 \rightarrow Y_1 Y_2 Y_3 Y_4 V_1 V_2 V_3$$

- Duplicate W^{--} (and W^{-+} , W^{+-} , W^{++}) and obtain W^{---} and W^{--+} (and W^{--+} , W^{---} , W^{+--} , W^{--+} , W^{++-} , W^{+++}).

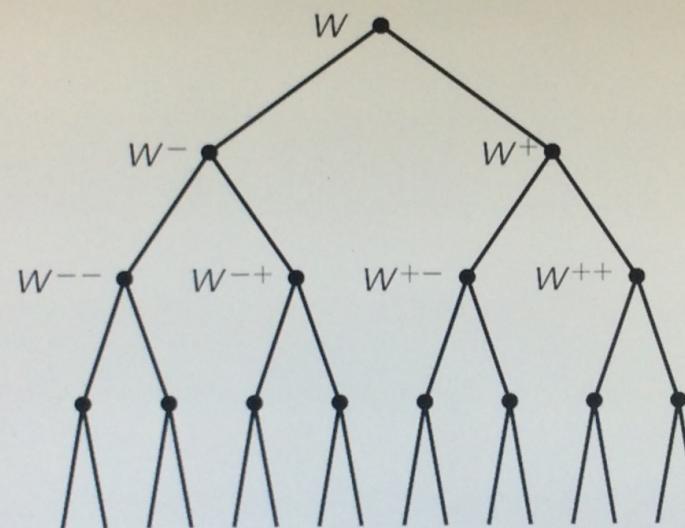


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Family tree of channels



Polarization happens

Theorem

For any $\epsilon > 0$ and any W , the fraction of ϵ -mediocre channels vanishes as we repeatedly apply the polar transform:

$$\mu_t(\epsilon) := \frac{1}{2^t} \sum_{s^t \in \{+,-\}^t} \mathbb{1}\{I(W^{s^t}) \in (\epsilon, 1-\epsilon)\}, \quad \lim_{t \rightarrow \infty} \mu_t(\epsilon) = 0.$$

Theorem (Restricted to BEC)

For any $\epsilon > 0$, and any $W = \text{BEC}(p)$, the fraction of ϵ -mediocre channels vanishes as we repeatedly apply the polar transform:

$$\mu_t(\epsilon) := \frac{1}{2^t} \sum_{s^t \in \{+,-\}^t} \mathbb{1}\{p(s^t) \in (\epsilon, 1-\epsilon)\}, \quad \lim_{t \rightarrow \infty} \mu_t(\epsilon) = 0.$$



Proof of polarization for BECs

Theorem (Restricted to BEC)

For any $\epsilon > 0$, and any $W = \text{BEC}(p)$, the fraction of ϵ -mediocre channels vanishes as we repeatedly apply the polar transform:

$$\mu_t(\epsilon) := \frac{1}{2^t} \sum_{s^t \in \{+,-\}^t} \mathbb{1}\{p(s^t) \in (\epsilon, 1-\epsilon)\}, \quad \lim_{t \rightarrow \infty} \mu_t(\epsilon) = 0.$$

- For a channel $P = \text{BEC}(q)$ define its **ugliness**, $\text{ugly}(P) = \sqrt{4q(1-q)}$.
- The children of P , the channels P^+ and P^- , have ugliness

$$\begin{aligned} \text{ugly}(P^+) &= \sqrt{4q^2(1-q^2)} & \text{ugly}(P^-) &= \sqrt{4q(2-q)(1-q)^2} \\ &= \text{ugly}(P) \sqrt{q(1+q)} & &= \text{ugly}(P) \sqrt{(2-q)(1-q)} \end{aligned}$$

- $\frac{1}{2} \text{ugly}(P^+) + \frac{1}{2} \text{ugly}(P^-) = \text{ugly}(P) \frac{1}{2} [\sqrt{q(1+q)} + \sqrt{(2-q)(1-q)}] \leq \text{ugly}(P) \sqrt{\frac{3}{4}}$.
- (observation first made by B. Hajek)



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Polarization proof for BECs

Theorem (Restricted to BEC)

For any $\epsilon > 0$, and any $W = \text{BEC}(p)$, the fraction of ϵ -mediocre channels vanishes as we repeatedly apply the polar transform:

$$\mu_t(\epsilon) := \frac{1}{2^t} \sum_{s^t \in \{+,-\}^t} \mathbb{1}\{p(s^t) \in (\epsilon, 1-\epsilon)\}, \quad \lim_{t \rightarrow \infty} \mu_t(\epsilon) = 0.$$

- Recall: $\text{ugly}(\text{BEC}(q)) := \sqrt{4q(1-q)}$.
- Recall: $\frac{1}{2}[\text{ugly}(P^+) + \text{ugly}(P^-)] \leq \sqrt{\frac{3}{4}} \text{ ugly}(P)$.
- Thus: $\frac{1}{2^t} \sum_{s \in \{+,-\}^t} \text{ugly}(W^s) \leq \left(\frac{3}{4}\right)^{t/2} \text{ugly}(W)$.
- Note: $\mathbb{1}\{p(s) \in (\epsilon, 1-\epsilon)\} \leq \text{ugly}(W^s) / \sqrt{4\epsilon(1-\epsilon)}$
- Hence: $\mu_t(\epsilon) \leq \left(\frac{3}{4}\right)^{t/2} / \sqrt{4\epsilon(1-\epsilon)}$
- QED.



Corollary

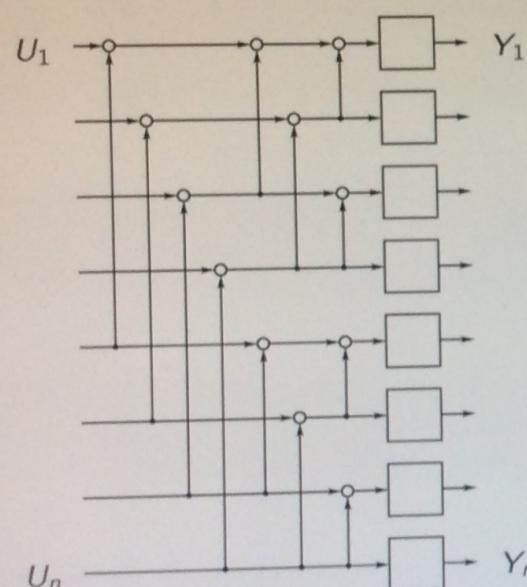
For any $\epsilon \in (0, 1)$, for any $W = \text{BEC}(p)$, the fraction of ϵ -good channels approaches $1 - p$:

$$\gamma_t(\epsilon) := \frac{1}{2^t} \sum_{s \in \{+,-\}^t} \mathbb{1}\{p(s) \leq \epsilon\}; \quad \lim_{t \rightarrow \infty} \gamma_t(\epsilon) = 1 - p.$$



Polar coding

This suggests the following coding scheme: given $W = \text{BEC}(p)$, $R < 1 - p$,



- Start with $n = 2^t$ copies of W . (I.e., blocklength is 2^t .)
- Apply the polar transform for t generations to synthesize the n channels $W^{++\cdots+}, \dots, W^{-\cdots-}$.
- Set the inputs of the best nR channels to uncoded data. Freeze the inputs of the remaining channels to 0.
- Number of ϵ -good channels is $\approx (1 - p)n$, so all the nR channels are good.
- At the receiver, successively decode U_1, \dots, U_n .
- The block error probability is upper bounded by

$$\sum_{s: p(s) \leq \epsilon} p(s) \leq n\epsilon$$

- Hmm.

Hmmm.

The polarization theorem we just proved is not strong enough. We need to be able to choose $\epsilon = \epsilon_t \ll 1/n$, and still have polarization ($n = 2^t$).

Theorem

For any $\delta > 0$, and any $W = \text{BEC}(p)$, with $n = 2^t$, and $\epsilon_t := \exp_2(-n^{\frac{1}{2} - \delta})$, the fraction of ϵ_t -good channels approaches $1 - p$:

$$\lim_{t \rightarrow \infty} \gamma_t(\epsilon_t) = 1 - p.$$



Taintedness

To prove this stronger polarization statement we need two concepts:

- Call a channel $W^s = W^{s_1 \dots s_t}$ at generation t to be **ϵ -tainted** if it has an ϵ -mediocre ancestor $W^{s_1 \dots s_i}$ at **any** generation $i \in [\sqrt{t}, t]$.
- The fraction of ϵ -tainted channels at generation t is upper bounded by

$$\sum_{i=\sqrt{t}}^t \mu_i(\epsilon) \leq \frac{1}{\sqrt{4\epsilon(1-\epsilon)}} \sum_{i=\sqrt{t}}^t \left(\frac{3}{4}\right)^{i/2} \leq \text{const}(\epsilon) \left(\frac{3}{4}\right)^{\sqrt{t}/2}$$

which vanishes as t gets large.

- For small ϵ , the transformations $q \rightarrow q^+$ and $q \rightarrow q^-$ don't jump between $[0, \epsilon]$ and $[1 - \epsilon, 1]$.
- So if ϵ is small, and if $W^{s_1 \dots s_t}$ is ϵ -untainted, then, all its ancestors $\{W^{s'_i} : i \in [\sqrt{t}, t]\}$ are the same type: either all are good (BEC(q) with $q \leq \epsilon$) or all are bad (with $q \geq 1 - \epsilon$).



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Unluckyness

The other notion we need is that of luck:

- Call a channel $W^s = W^{s_1 \dots s_t}$ at generation t to be ϵ -unlucky if $\{s_i : i \in [\sqrt{t}, t]\}$ does not have its fair share of +'s:

$$\sum_{i=\sqrt{t}}^t \mathbb{1}\{s_i = +\} < \left(\frac{1}{2} - \epsilon\right)t.$$

- The fraction of ϵ -unlucky channels at generation t vanishes as t gets large.



Consider the t 'th generation descendants of $W = \text{BEC}(p)$, pick a small $\epsilon > 0$.

- The fraction of ϵ -good, ϵ -untainted, ϵ -lucky channels approaches $1 - p$ as t gets large. Take any such channel, and trace its ancestors in generations $[\sqrt{t}, t]$. For each generational step

$$\text{either } p(s^{i+1}) = p(s^i)^2$$

$$\text{or } p(s^{i+1}) = p(s^i)(2 - p(s^i)) \leq p(s^i)2 = p(s^i)\epsilon^{-\epsilon'} \leq p(s^i)^{1-\epsilon'} \quad \text{with } \epsilon' = 1/\log_2(1/\epsilon)$$

- Taking logs twice

$$\log_2 \log_2 \frac{1}{p(s^{i+1})} \geq \log_2 \log_2 \frac{1}{p(s^i)} + \begin{cases} 1 & s_{i+1} = + \\ -\epsilon'' & s_{i+1} = - \end{cases} \quad \text{with } \epsilon'' = -\log_2(1 - \epsilon')$$

- Chaining the inequalities from $i = \sqrt{t}$ till $t - 1$,

$$\begin{aligned} \log_2 \log_2 \frac{1}{p(s^t)} &\geq \log_2 \log_2 \frac{1}{\epsilon} + (\frac{1}{2} - \epsilon)t - t\epsilon'' \geq (\frac{1}{2} - \delta)t \quad \text{for small enough } \epsilon \\ - \log_2 p(s^t) &\geq 2^{(1/2 - \delta)t} = n^{1/2 - \delta}. \end{aligned}$$



Corollary

Polar codes are capacity achieving with error probability exponentially small in the square root of the block length.

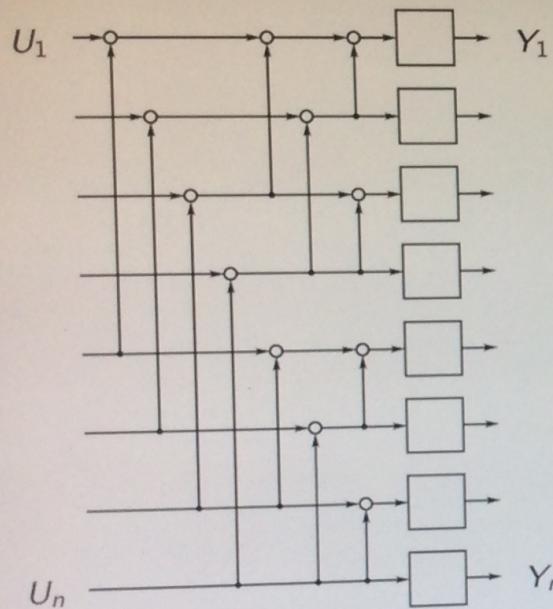
Given $\delta > 0$, $W = \text{BEC}(p)$, and $R < 1 - p$; for large $n = 2^t$ the polar code constructed by sending data on the nR best synthetic channels has error probability at most $n2^{-n^{1/2-\delta}}$.



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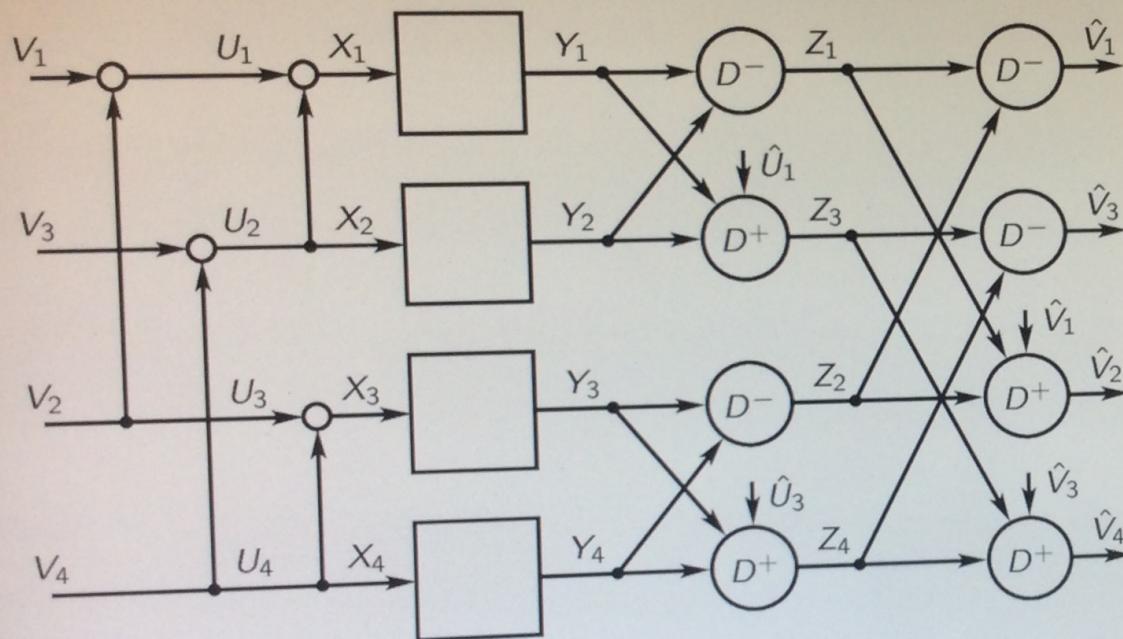
Encoding Complexity



For a polar code of blocklength $n = 2^t$,

- There are t encoding stages
- Each stage with $n/2$ exclusive-or operations.
- Encoding complexity is $n \log_2 n$.

Decoding Complexity



Receiver's first task: construct the output of the BEC W^- . Recall $(Y_1, Y_2) = (U_1 \oplus U_2, U_2)$ or useless.

$$D^-(y_1, y_2) = \begin{cases} y_1 \oplus y_2 & y_1 \neq ?, y_2 \neq ? \\ ? & \text{else} \end{cases}$$

$$D^+(z_1, z_2, v_1) = \begin{cases} z_2 & z_2 \neq ? \\ z_1 \oplus v_1 & z_1 \neq ? \\ ? & \text{else} \end{cases}$$

We have $n \log_2 n$ computation units each executing once. Decoding complexity is $n \log_2 n$.

Remarks

- Polar codes are easy to teach, and the students appear to appreciate hearing about them.
- They provide a constructive argument to prove the channel coding theorem, and allow a symmetry-in-teaching with source coding.
- See [Alsan & T., A simple proof of polarization and polarization for non-stationary memoryless channels, T-IT 2016] for a second moment method to prove the first polarization theorem for general binary input channels.
- I. Tal has an arXiv paper that also describes a simple proof of the second polarization theorem for general binary input channels.
- Do read Arikan's paper on polarization, and also his 'On the origin of polar coding' for further insight on how they came to be.



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