

① Linear Settling.

↓ For a LTI (Linear time invariant) Systems (Linear systems)

Scaling the input will scales the output

∴ Settling time $\neq f(V_{in})$

2 Time Constant τ of the System does not depends on input level

∴ We need to settle to a higher Voltage in the same time

→ need larger slope → need larger current

∴ Linear settling: scaling the input scales the output slope and output current

3 $V_{out} = V_{step} (1 - e^{-\frac{t}{\tau}})$

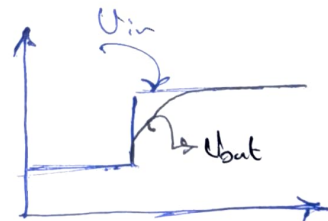
$$\frac{dV_{out}}{dt} = V_{step} \left(\frac{1}{\tau} e^{-\frac{t}{\tau}} \right)$$

∴ $\frac{dV_{out}}{dt} \propto V_{step}$

∴ $I_{out} = C \frac{dV_{out}}{dt} \rightarrow I_{out} \propto V_{step}$

4 If I_{out} reaches the max value

∴ Settling changes from linear settling to non-linear settling 😞



② The non-Linear Settling.

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1 non-linear settling: Output slope and current NOT \propto input

\therefore Output is linear ramp \rightarrow finite constant slope

2 SR (Slow rate) is the max possible slope of the op-amp output

3 Linear ramp \rightarrow Constant charging a cap

$$I_{out} = C_L \frac{dV_{out}}{dt} \rightarrow SR = \left(\frac{dV_{out}}{dt} \right)_{max} = \frac{I_{out,max}}{C_L}$$

4 Slope independent of the input level

\therefore non-linear behaviour \rightarrow non-linear system

③ Example.

1 Small input step, Linear settling

1 output voltage has exponential settling behaviour

2 When the output settles $\rightarrow \Delta V_{id} = V_{in} - V_{out} \approx 0$

$$\therefore I_{out} = g_m \Delta V_{id}$$

2 Large input step, Slewing

1 M2 off $\rightarrow I_{SS}$ is fully steered to M1 \rightarrow M3 \rightarrow M4 \rightarrow then to C_L

2 Non-linear settling: Output slope and charging current \propto input

\therefore Linear ramp \rightarrow Constant current (I_{SS}) charging a capacitor

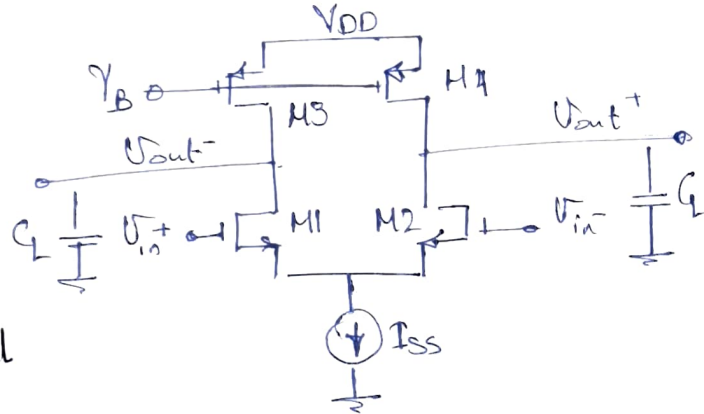
3 As V_{out} approaches V_{in} the circuit resumes linear settling (exp)

4 $SR = \frac{I_{SS}}{C_L} \rightarrow \text{efficiency} \approx 100\%$

④ SR of FD-OTA

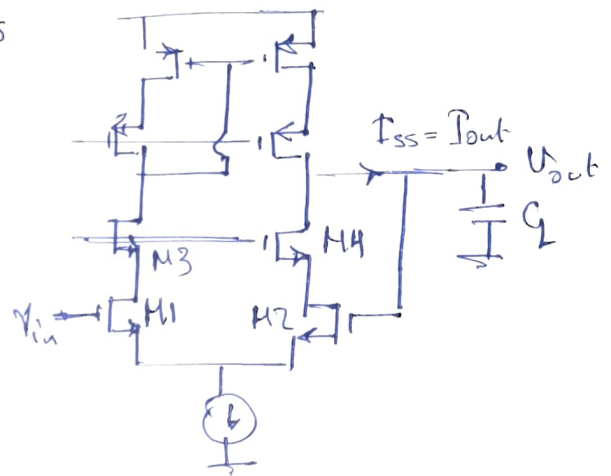
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1. $\omega_{out-} : SR = \frac{I_{SS}}{2C_L}$
 2. $\omega_{out+} : SR = \frac{I_{SS}}{2C_L}$
 3. $\omega_{outd} = \omega_{out+} - \omega_{out-} : SR = \frac{I_{SS}}{C_L}$
- Symmetrical slewing \rightarrow Necessary
to maintain constant CM level



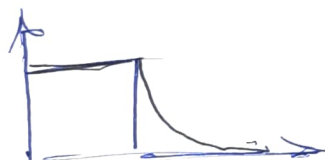
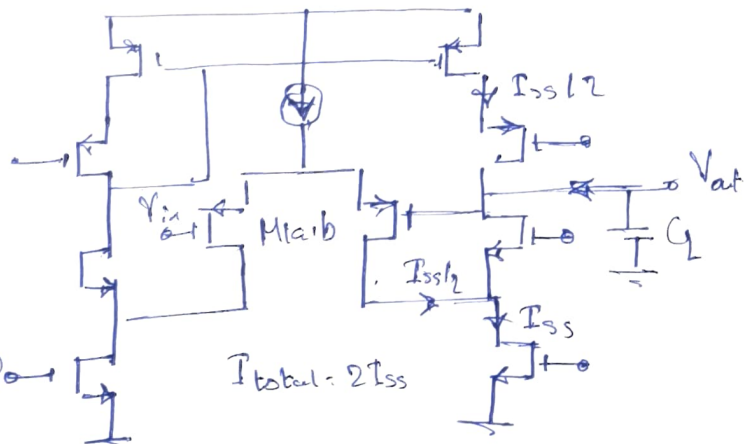
⑤ SR of Telescopic CasCode

1. applying large $V_{in} \rightarrow M_1, I_D = I_{SS}$
 $\therefore M_2$ and M_4 turned off
2. I_{SS} mirrored to the output C_L
 $\therefore I_{out\ max} = I_{SS}$
3. $SR = \frac{I_{SS}}{C_L} \rightarrow \text{efficiency} = 100\%$



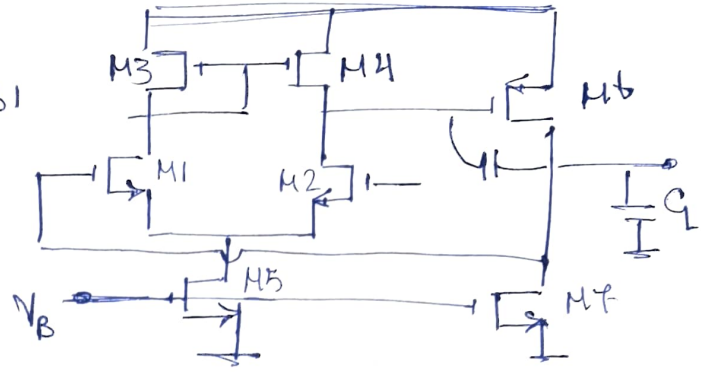
⑥ Folded CasCode OTA SR

1. usually set $I_{SS} = 2I_{B12}$
 $\therefore I_{D2, CG} = I_{D2, CS}$
2. at equilibrium
 $I_{B2} = I_{B12} + \frac{I_{SS}}{2}$
3. at slewing, $I_{B1} = I_{B2} - I_{SS} = 0$
4. $SR = \frac{I_{SS}}{C_L} \rightarrow \text{efficiency} \approx 50\%$



⑦ Two stage miller OTA SR

1 I_{B2} much larger than I_{B1}
 \rightarrow SR usually limited by I_{B1}



2 Internal Slewling Limits

Output Slewling

3 Positive Slewling

$$\frac{I_{D6} - I_{B1} - I_{B2}}{C_L} > \frac{I_{B1}}{C_1 + C_C}$$

$$SR_+ \approx \frac{I_{B1}}{C_1 + C_C}$$

4 Negative Slewling

$$\frac{I_{B2} - I_{B1}}{C_L} > \frac{I_{B1}}{C_1 + C_C}$$

$$SR_- = \frac{I_{B1}}{C_1 + C_C}$$

5 Must be designed such that $SR_+ = SR_-$
 \therefore Both Limited by I_{B1}

6 Otherwise the Slewling is asymmetric
 * Causes shift in CM output
 * Slow transients in CHFB network is slow

⑧ Power supply rejection

1 Supply Lines Carry noise from regulators, digital loads, etc.
 2 We don't want V_{out} to be affected by V_{DD} noise

$$PSR = \frac{1}{V_{out}/V_{DD}}$$

$$PSRR = \frac{V_{out}/V_{in}}{V_{out}/V_{DD}} = A_{vd} \cdot PSR$$

Both PSR and PSRR usually reported in dB

⑨ PSRR Example 5T OTA

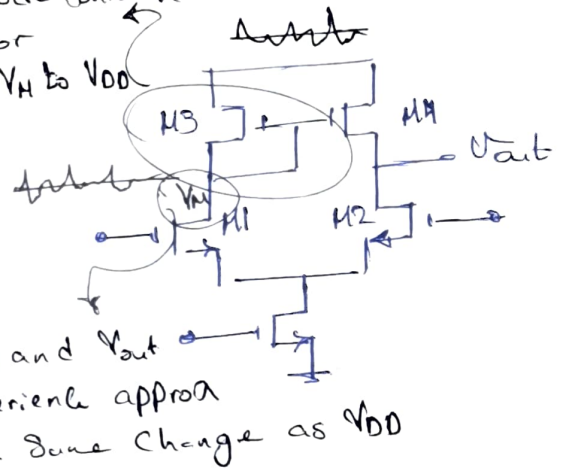
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$$PSR = \frac{1}{V_{out}/V_{DD}} = 1$$

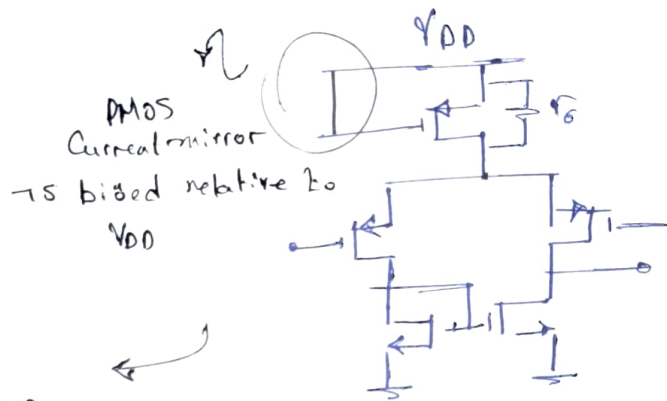
$$PSRR = A_{vd} \cdot PSR$$

$$= \frac{g_{m1,2} (r_{o2} \parallel r_{o4})}{1}$$

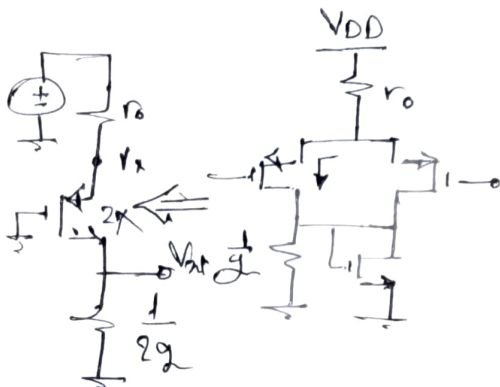
T_n diode Connected transistor clamps V_H to V_{DD}



⑩ PSRR of 5T-OT with PMOS input pair



$$PSRR = \frac{A_{vd}}{V_{out}/V_{DD}} \approx CMRR \approx (g_m r_o)^2$$



⑪ PSRR of inverter

$$\frac{V_{out}}{V_{DD}} = g_m \left(\frac{r_o}{2} \right) \text{ Very high}$$

$$\frac{V_{out}}{V_{in}} = g_m r_o$$

$$\therefore PSRR = 2 \text{ (sad face)}$$

