

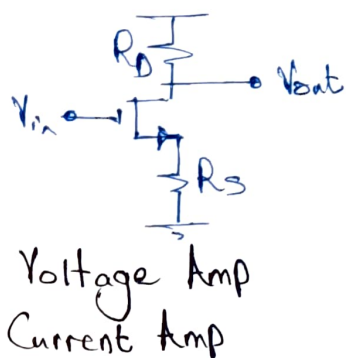
Analog Integrated Circuits. lec 06: Amplifier topologies.

Page 1

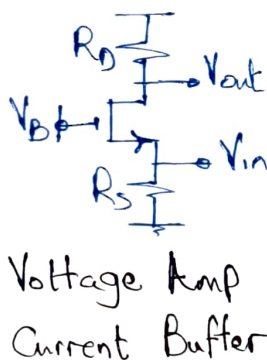
① Basic Mos topologies.

— There are 3 basic topologies of Mos amp. based on Terminal Conditions

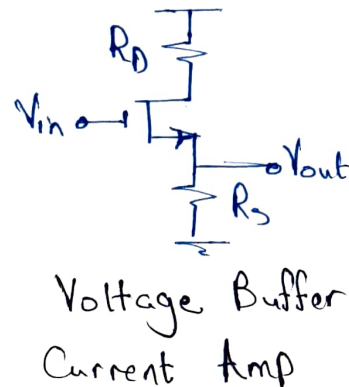
Common Source



Common Gate



Common Drain



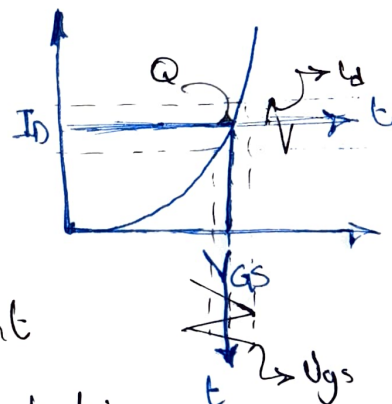
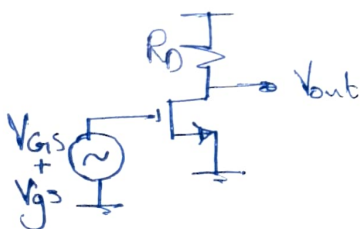
② Basic Amp operation

— The input signal is a combination of two components

1 DC Component

a certain V_{GS} gives a certain I_D @ Q point

2 AC Component → where a small perturbation of small signal around Q point will result a perturbation around I_D



③ Basic Analysis steps

1 DC analysis → Coupling and bypass caps → AC
→ Calculate Q point and define region

2 Calculate small signal parameters ($g_m r_o$)

3 Draw small signal equivalent circuit → DC Voltage → SC
→ DC Current → OC

4 Determine the amp parameters → R_{in} and R_{out} → Coupling and Bypass → SC
→ A_v and A_i

④ R_{in}/R_{out} shortcuts.

- We need to find equivalent impedance looking from Gate, Source, and Drain

21
Page

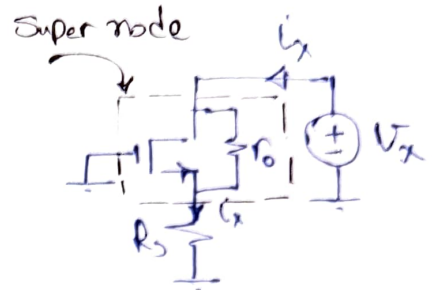
1 looking from Gate.

$$R_{LFG} = \infty \text{ at low freq}$$

2 looking from Drain

- If G and B are ac connected

$$\therefore V_{bs} = V_{gs} \rightarrow g_m \text{ and } g_{mb} \text{ added}$$



Assume Transistor is a Super node

$$\therefore V_{gs} = -V_s = -i_x R_s$$

Apply KCL @ Drain

$$\therefore i_x = (g_m + g_{mb}) V_{gs} + \frac{V_x - i_x R_s}{r_o}$$

$$\therefore i_x = -(g_m + g_{mb}) i_x R_s + \frac{V_x}{r_o} - \frac{i_x R_s}{r_o}$$

$$\therefore i_x [1 + (g_m + g_{mb}) R_s] = \frac{V_x}{r_o}$$

$$\therefore R_{LFD} = r_o [1 + (g_m + g_{mb}) R_s]$$

Active Load
 - If $R_s = 0$ (G and S are ac connected)
 \therefore Mod equivalent to r_o

3 looking from Source

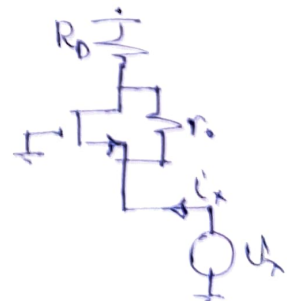
Apply KCL @ Source

$$\therefore i_x = \frac{V_x - V_D}{r_o} - g_m V_{gs} - g_{mb} V_{bs}$$

$$\therefore i_x = \frac{V_x}{r_o} - \frac{i_x R_D}{r_o} - (g_m + g_{mb}) V_x$$

$$\therefore i_x \left(1 + \frac{R_D}{r_o}\right) = V_x \left[\frac{1}{r_o} + (g_m + g_{mb})\right]$$

$$\therefore \frac{V_x}{i_x} = \frac{1}{g_m + g_{mb}} \left(1 + \frac{R_D}{r_o}\right)$$

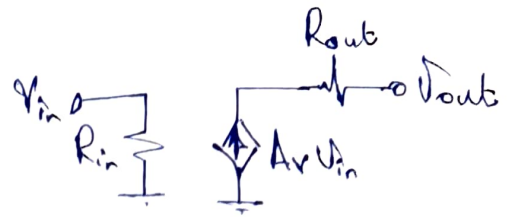


⑤ G_m Rout method

1. Thevenin model of Amp

$$- R_{in} = \frac{V_{in}}{I_{in}}$$

$$- R_{out} = \frac{V_x}{I_x} \text{ at } V_{in} = 0$$



Page 21

$$\therefore \text{DC Voltage gain} \rightarrow V_{Thevenin} = V_{out\ DC} = A_v V_{in}$$

$$\therefore A_v = \frac{V_{out\ DC}}{V_{in}}$$

2. Norton model of Amp

$$- I_{Norton} = I_{out\ SC} = G_m V_{in}$$

$$\rightarrow G_m = \frac{I_{out\ SC}}{V_{in}}$$

$$\therefore A_v = G_m R_{out}$$



$$- A_v = \frac{I_{out\ SC}}{I_{in}} = G_m R_{in}$$

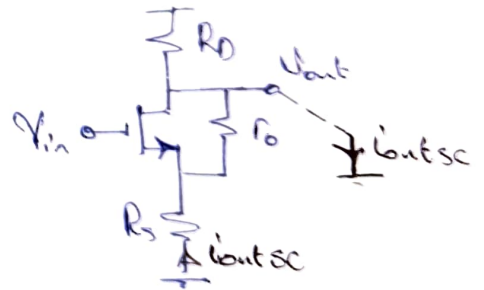
⑥ Common Source (CS)

$$- \text{Apply KCL at } S \text{ where } V_{gs} = V_{in} + I_{out\ SC} R_S$$

$$\therefore I_{out\ SC} + g_m (V_{in} + I_{out\ SC} R_S)$$

$$+ g_{mb} (I_{out\ SC} R_S) + \frac{I_{out\ SC} R_S}{r_o} = 0$$

$$\therefore G_m = \frac{I_{out\ SC}}{V_{in}} = \frac{-g_m}{1 + (g_m + g_{mb}) R_S}$$



$$- R_{out} = R_D \parallel r_o [1 + (g_m + g_{mb}) R_S]$$

$$\therefore A_v = G_m R_{out} \begin{cases} \text{if } R_D \gg r_o \text{ at DC, } A_v = -g_m r_o \\ \text{if } R_D \ll R_{D\ DC}, A_v = \frac{-g_m R_D}{1 + (g_m + g_{mb}) R_S} \\ \text{if } R_S = 0, A_v = -g_m (R_D \parallel r_o) \end{cases}$$

- If S and B are ac Connected (for PMOS)

$$A_v = \frac{-R_D \parallel R_{LFD}}{\frac{1}{g_m} + R_S} = - \frac{\text{Drain res}}{\frac{1}{g_m} + \text{Source res}}$$

Page 21

- If $R_S \gg \frac{1}{g_m}$ and $R_D \ll R_{LFD}$: $A_v \approx -\frac{R_D}{R_S} \rightarrow$ linear gain

- R_S reduces G_m (Source Degeneration)

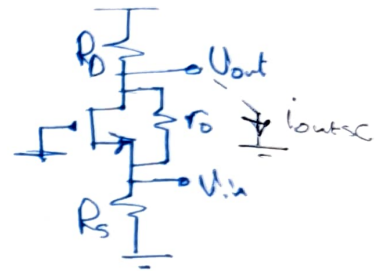
But improves Linearity

④ Common Gate (CG)

- Apply KCL at D where $U_{gs} = U_{bs} = -U_{in}$

$$\therefore I_{outsc} + (g_m + g_{mb})(-U_{in}) - \frac{U_{in}}{r_o} = 0$$

$$\therefore G_m = \frac{I_{outsc}}{U_{in}} \approx g_m + g_{mb}$$



- $R_{out} \approx R_D \parallel r_o$ (Why?)

$$\begin{aligned} - A_v = G_m R_{out} & \rightarrow A_v = (g_m + g_{mb}) r_o \rightarrow \text{if } R_D \rightarrow \infty \text{ ac OC} \\ & \rightarrow A_v \approx (g_m + g_{mb}) R_D \rightarrow \text{if } R_D \ll r_o \\ & \rightarrow A_i = G_m R_{in} \approx 1 \text{ (Current Buffer)} \end{aligned}$$

⑤ Common Drain (CD)

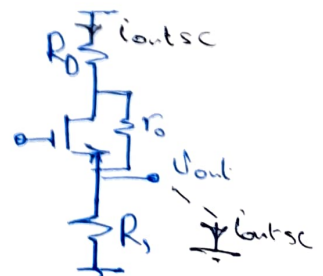
- Apply KCL at D

$$\therefore I_{outsc} - g_m U_{in} + \frac{I_{outsc} R_D}{r_o} = 0$$

$$\therefore G_m = \frac{g_m}{1 + R_D/r_o}$$

$$- R_{out} \approx R_S \parallel \frac{1}{g_m + g_{mb}} \left(1 + \frac{R_D}{r_o}\right)$$

$$- \text{If } R_S \gg R_{LFS} : A_v \approx \frac{g_m}{g_m + g_{mb}} < 1$$



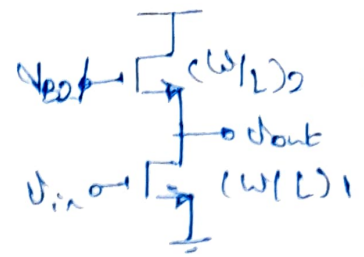
⑨ Ques

1 Find A_v in terms of W/L

$$A_v = G_m R_{out}$$

$$= g_{m1} \cdot \left(\frac{1}{g_{m2}} \parallel r_{o1} \right)$$

$$= \frac{g_{m1}}{g_{m2}} = \sqrt{\frac{(W/L)_1}{(W/L)_2}}$$

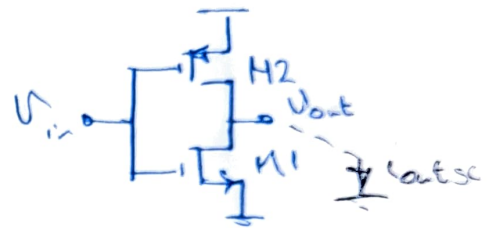


Page 51

2 Find A_v using $G_m R_{out}$

apply KCL @ output node

$$\begin{aligned} \therefore V_{outSC} &= -g_{m2} V_{in} - g_{m1} V_{in} \\ &= -V_{in} (g_{m1} + g_{m2}) \end{aligned}$$



$$\therefore G_m = - (g_{m1} + g_{m2})$$

$$\therefore R_{out} = r_{o2} \parallel r_{o1}$$

$$\therefore A_v = G_m R_{out}$$