

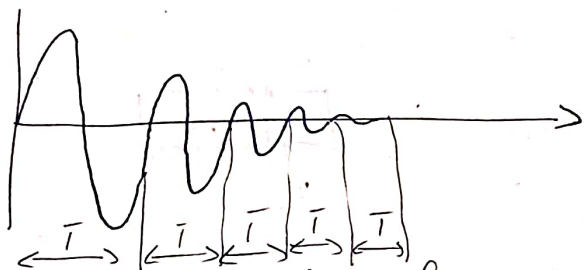
Q-factor :

- Quality Factor is a dimensionless parameter that describes how "underdamped" an oscillator is.

$$Q = \frac{\overset{\text{Initial}}{\uparrow} \text{Energy stored in an resonator}}{\text{Energy lost in one radian}}$$

$$= 2\pi \frac{\text{Energy stored}}{\text{Energy dissipated per cycle}} = \omega \times \frac{\text{Max. Energy stored}}{\text{power loss}}$$

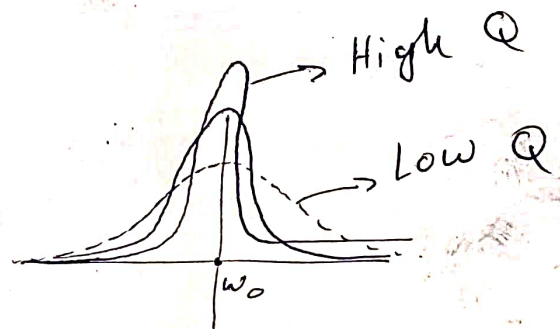
Ex:



\Rightarrow 5 cycles then the energy is completely dissipated
 $\hookrightarrow Q \approx 5$

Another definition for Q:

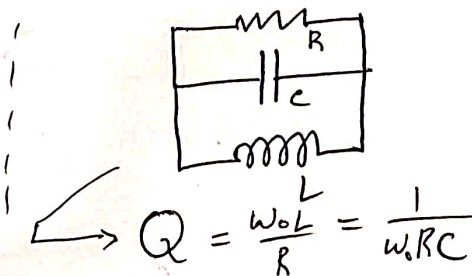
$$Q = \frac{\omega_0}{BW}$$



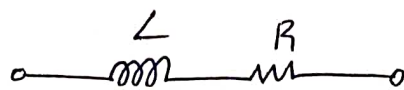
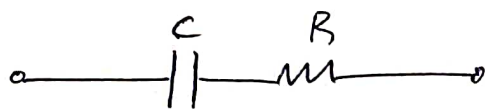
For an RLC-network :



$$\hookrightarrow Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 RC}$$



The capacitors & Inductors used are not ideal
 & can be modeled as :



→ So, the Resistance in series (L or C) used to Model the ohmic loss in the inductor turns/capacitor plates : → these losses limits the Quality Factor when used in an RLC ct.

→ so, we define for an inductor/cap. : (Quality Factor)

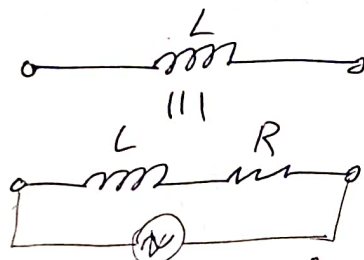
$\because X_C \gg R$
 $V_m \approx \frac{I_m}{\omega C}$

$$Q = \frac{\frac{1}{2} V_m^2 C (2\pi)}{\left(\frac{I_m}{\sqrt{2}}\right)^2 R \cdot T} \approx \frac{\frac{I_m^2}{\omega^2 C} \cdot \frac{1}{2} (2\pi)}{\frac{I_m^2}{2} R \cdot \left(\frac{1}{f}\right)}$$

$$\Rightarrow Q = \frac{\omega}{\omega^2 C R} = \frac{1}{\omega C R}$$

$$\Rightarrow \boxed{Q_C = \frac{X_C}{R}}$$

↓
 capacitor
 Quality
 Factor



$$Q_L = \frac{\text{Max. stored energy}}{\text{energy dissipated in one cycle}} \times 2\pi$$

$$\Rightarrow Q_L = \frac{\frac{1}{2} L I_m^2}{\left(\frac{1}{\sqrt{2}} I_m\right)^2 R \cdot \underbrace{T}_{\text{periodic time} = \frac{1}{f}}} \times 2\pi$$

$$= \frac{\frac{1}{2} L I_m^2 (2\pi)}{\frac{1}{2} I_m^2 R / f}$$

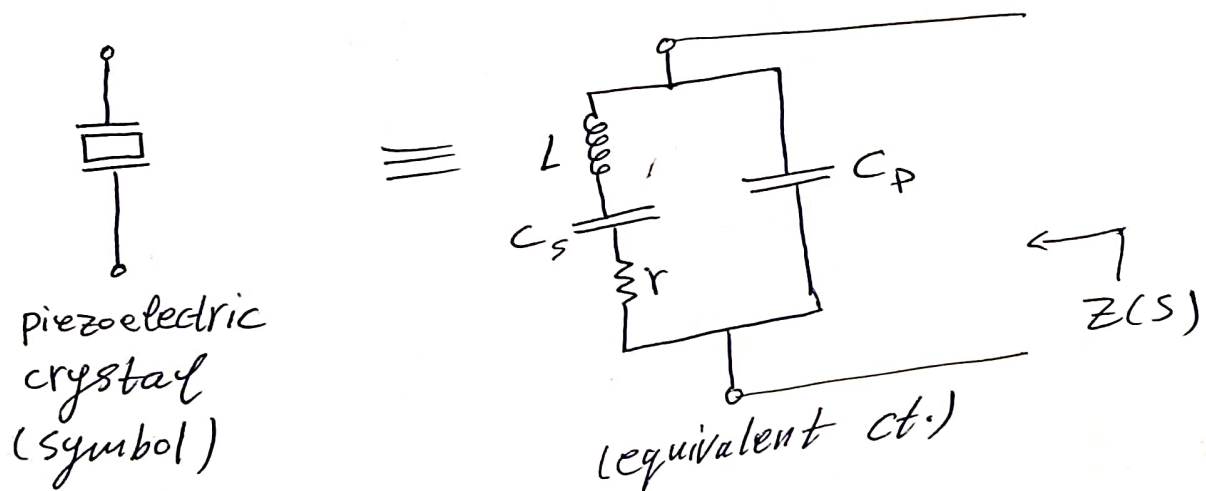
$$\Rightarrow Q_L = \frac{2\pi f L}{R} = \frac{\omega L}{R} = \frac{X_L}{R}$$

$$\Rightarrow \boxed{Q_L = \frac{X_L}{R}}$$

↓
 Inductor
 Quality
 Factor

crystal oscillator :

piezoelectric crystals have electromechanical - resonance characteristics that are very stable (with time & temperature) & highly selective (have high Q).



$L \rightarrow$ typically hundreds of Henrys

$C_s \rightarrow$ very small (≈ 0.0005 PF) cap.

$r \rightarrow$ representing Q of inductor ($\frac{\omega_0 L}{r}$) (few hundred/thousands)

$C_p \rightarrow$ typically few pico Farads.

$\therefore Q$ is very high $\rightarrow r$ is negligible
($X_L \gg r$)

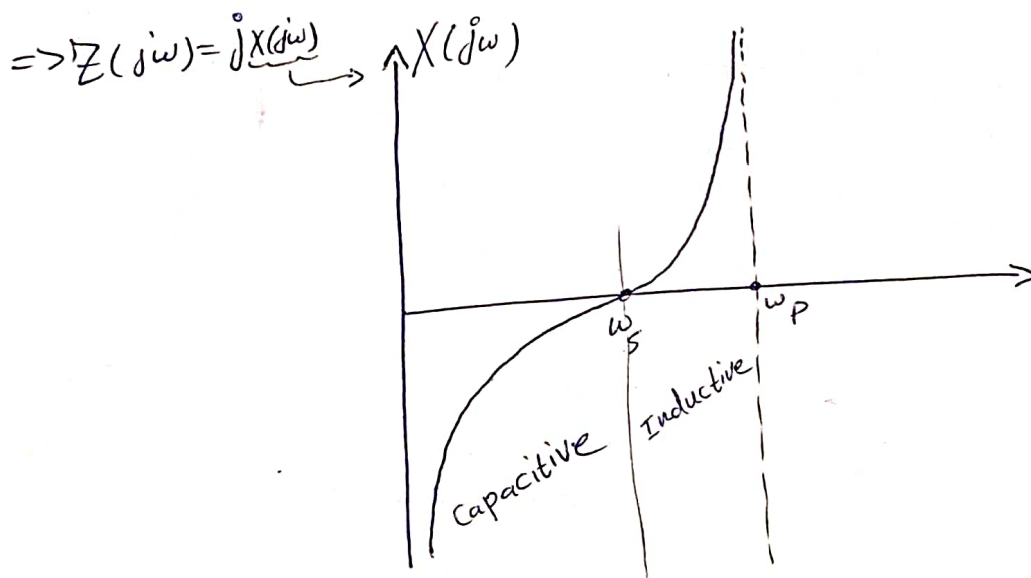
$$\begin{aligned} \Rightarrow Z(s) &= \frac{1}{sC_p} \parallel \left(sL + \frac{1}{sC_s} \right) \\ &= \frac{\frac{1}{sC_p} \left(sL + \frac{1}{sC_s} \right)}{\frac{1}{sC_p} + sL + \frac{1}{sC_s}} \\ &= \frac{1}{sC_p} \frac{\left(sL + \frac{1}{sC_s} \right)}{\frac{1}{sC_p} + sL + \frac{1}{sC_s}} \end{aligned}$$

$$\begin{aligned}
 \Rightarrow Z(s) &= \frac{1}{sC_p} \cdot \frac{s^2 LC_s + 1}{\frac{C_s}{C_p} + s^2 LC_s + 1} \\
 &= \frac{1}{sC_p} \cdot \frac{\left(s^2 + \frac{1}{LC_s}\right)}{\frac{1}{LC_p} + s^2 + \frac{1}{LC_s}} \\
 &= \left(\frac{1}{sC_p}\right) \frac{\left[s^2 + \frac{1}{LC_s}\right]}{\left[s^2 + \left(\frac{C_s + C_p}{LC_s C_p}\right)\right]}
 \end{aligned}$$

$$\Rightarrow Z(j\omega) = \left(\frac{1}{j\omega C_p}\right) \frac{-\omega^2 + \omega_s^2}{-\omega^2 + \omega_p^2}$$

where ; $\omega_s = \frac{1}{\sqrt{LC_s}}$, $\omega_p = \frac{1}{\sqrt{L\left(\frac{C_s C_p}{C_s + C_p}\right)}}$

$$\Rightarrow Z(j\omega) = -j\left(\frac{1}{\omega C_p}\right) \frac{\omega^2 - \omega_s^2}{\omega^2 - \omega_p^2}$$



∴ $C_p \gg C_s \Rightarrow \omega_p \approx \frac{1}{\sqrt{LC_s}}$

So, at ~~the~~ $\omega_s < \omega < \omega_p \rightarrow$ the crystal can be used as an inductor.

∴ $\omega_p \approx \omega_s = \frac{1}{\sqrt{LC_s}}$

$\Rightarrow \omega_0 \approx \frac{1}{\sqrt{LC_s}}$

Q1. For the shown pierce oscillator. Let C_1 be variable in range 1 pF to 10 pF & Let C_2 be fixed at 10 pF. Find the range over which the oscillation frequency can be tuned. [For the crystal: $L = 0.52 \text{ H}$, $C_s = 0.012 \text{ pF}$ & $C_p = 4 \text{ pF}$].

Sol:

∴ we know from colpitts oscillator:

$$\omega_0 = \frac{1}{\sqrt{L \frac{C_1 C_2}{C_1 + C_2}}}$$

→ where $\frac{C_1 C_2}{C_1 + C_2} \Rightarrow$ is the total cap. seen by the inductor.

* Similarly the total cap. seen by L^0

$$\Rightarrow C_T = \frac{C_s \left[C_p + \frac{C_1 C_2}{C_1 + C_2} \right]}{C_s + C_p + \frac{C_1 C_2}{C_1 + C_2}}$$

C_s series with $\left[C_p \parallel \left(\frac{C_1 C_2}{C_1 + C_2} \right) \right]$
series comb. of C_1 & C_2 .

- for $C_1 = 1 \text{ pF}$

$$\Rightarrow C_{T1} = \frac{0.012 \left(4 + \frac{10 \times 1}{10 + 1} \right)}{0.012 + 4 + \frac{10 \times 1}{10 + 1}} = 0.01197 \text{ pF}$$

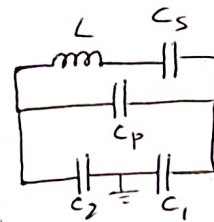
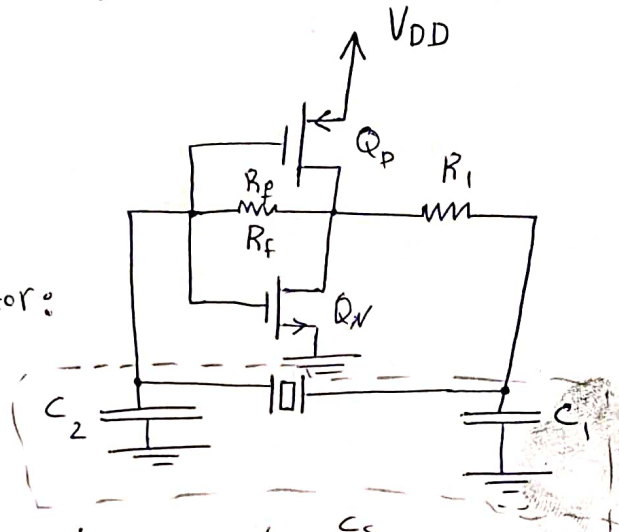
- for $C_1 = 10 \text{ pF}$:

$$\Rightarrow C_{T2} = \frac{0.012 \left(4 + \frac{10 \times 10}{10 + 10} \right)}{0.012 + 4 + \frac{10 \times 10}{10 + 10}} = 0.01198 \text{ pF}$$

$$\Rightarrow f_0 = \frac{1}{2\pi \sqrt{L C_T}} \Rightarrow f_{01} = \frac{1}{2\pi \sqrt{0.52 \times 0.01197 \times 10^{-12}}} = 2.0172 \text{ MHz}$$

$$\Rightarrow f_{02} = \frac{1}{2\pi \sqrt{0.52 \times 0.01198 \times 10^{-12}}} = 2.0162 \text{ MHz}$$

$$\Rightarrow \Delta f = 1 \text{ KHz} \rightarrow (\text{very small})$$



EX: A 2 MHz quartz crystal is specified to have
 $L = 0.52 \text{ H}$, $C_s = 0.012 \text{ pF}$, $C_p = 4 \text{ pF}$ &
 $r = 120 \Omega$ Find f_s , f_p & Q :

sol:

$$\Rightarrow f_s = \frac{1}{2\pi\sqrt{LC_s}} = \frac{1}{2\pi\sqrt{0.52 \times 0.012 \times 10^{-12}}}$$
$$= 2.015 \text{ MHz}$$

$$\Rightarrow f_p = \frac{1}{2\pi\sqrt{L \frac{C_s C_p}{C_s + C_p}}} = \frac{1}{2\pi\sqrt{0.52 \frac{0.012 \times 10^{-12} \times 4 \times 10^{-12}}{(0.012 + 4) \times 10^{-12}}}}$$
$$= 2.018 \text{ MHz}$$

$$\Rightarrow Q = \frac{\omega_0 L}{R} = \frac{2 \times 10^6 \times 2\pi \times 0.52}{120}$$
$$= 54,454$$