

Lecture (2)

Feedback Amplifiers

Definition of Feedback

- ❑ A small part of the output signal is fed-back to the input signal to be Added to it (positive Feedback) or Subtracted from it (negative feedback).

Types of Feedback

❑ Positive Feedback:

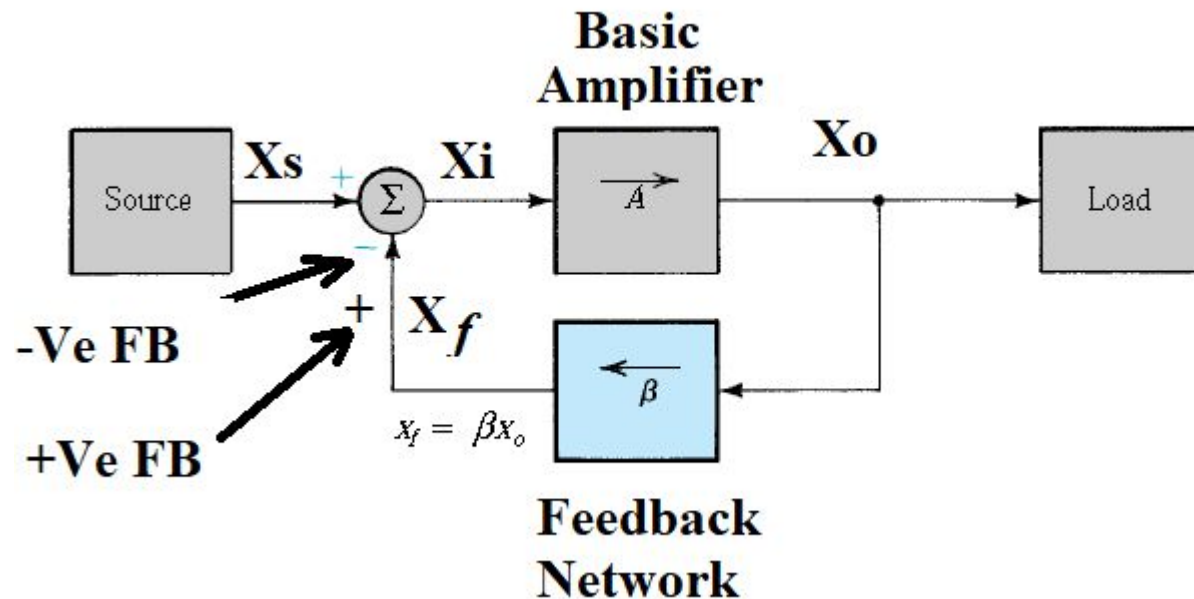
Used to build Oscillators.

❑ Negative feedback:

Used to build Stable amplifier gain.

$$A_f = \frac{A}{1+A\beta} \quad \text{-ve FB}$$

$$A_f = \frac{A}{1-A\beta} \quad \text{+ve FB}$$



A : The Amplifier Gain

β : The feedback factor

A_f : The Feedback gain

(gain after FB, closed loop gain)

Advantages of Negative Feedback

1. Desensitize the gain:

Make the value of the gain less sensitive to variations in the values of circuit components, such as might be caused by changes in temperature.

2. Reduce nonlinear distortion:

Make the output proportional to the input (in other words, make the gain constant, independent of signal level).

3. Reduce the effect of noise:

Minimize the contribution to the output of unwanted electric signals generated, either by the circuit components themselves, or by extraneous interference.

4. Control the input and output resistances:

Raise or lower the input and output resistances by the selection of an appropriate feedback topology.

5. Extend the bandwidth of the amplifier.

Disadvantages of Negative Feedback

- ❑ Reduce the amplifier Gain.



Negative Feedback Amplifiers

$$A_f = \frac{A}{1+A\beta}$$

- ❑ X_s : The signal source (input of the circuit)
- ❑ X_i : The amplifier input
- ❑ X_o : The amplifier output
- ❑ X_f : The feedback signal

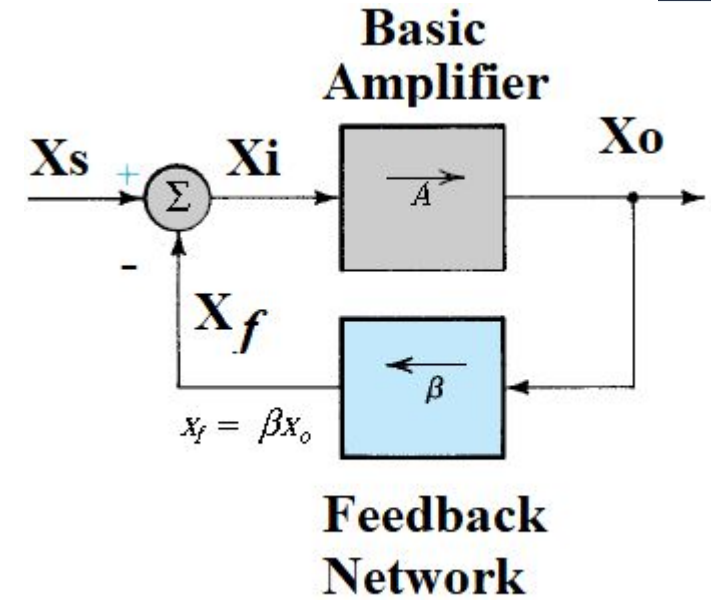
$$A = \frac{X_o}{X_i}$$

$$\beta = \frac{X_f}{X_o}$$

$$A = \frac{X_o}{X_i}$$

$$A_f = \frac{X_o}{X_s} = \frac{A}{1+A\beta} \approx \frac{1}{\beta}$$

The feedback gain A_f is nearly independent of the amplifier gain A



Amplifier gain $A \rightarrow$ Very high
Feedback factor $\beta < 1$
Then,

$$A\beta \gg 1$$

Some Properties of Negative Feedback

Gain Desensitivity:

$$A_f = \frac{A}{1+A\beta} \rightarrow \frac{dA_f}{dA} = \frac{1}{(1+A\beta)^2} \rightarrow dA_f = \frac{dA}{(1+A\beta)^2}$$

$$\text{Then, } \frac{dA_f}{A_f} = \frac{dA}{(1+A\beta)^2} \times \frac{1+A\beta}{A} = \frac{1}{1+A\beta} \frac{dA}{A} \rightarrow \frac{dA_f}{A_f} = \frac{1}{1+A\beta} \frac{dA}{A}$$

Gain Sensitivity (S)

$$S = \frac{dA_f/A_f}{dA/A} = \frac{1}{1+A\beta}$$

Gain Desensitivity (D)

$$D = 1/S = 1 + A\beta$$

dA_f : Amount of change of the feedback gain.

A_f : The feedback gain.

$\frac{dA_f}{A_f}$: Percentage change in the feedback gain.

A : Amplifier gain (open-loop gain without FB).

dA : Amount of change of the amplifier gain.

$\frac{dA}{A}$: Percentage change in the amplifier (open-loop) gain.

Example

A designer is required to achieve a closed loop gain of $25 \pm 1\%$ v/v using a basic amplifier whose gain variations is $\pm 10\%$. What is the nominal values of the amplifier gain A and the feedback factor β .

Solution:

$$* A_f = 25 \pm 1\%$$

$$\therefore \frac{\Delta A_f}{A_f} = \pm 1\% = \pm 0.01$$

$$* \frac{\Delta A}{A} = \pm \frac{10}{100} = \pm 0.1$$

$$S = \frac{\Delta A_f}{A_f} / \frac{\Delta A}{A} = \frac{1}{1 + A\beta}$$

$$\therefore \frac{0.01}{0.1} = \frac{1}{1 + A\beta}$$

$$\therefore \boxed{1 + A\beta = 10}$$

$$* A_f = 25 = \frac{A}{1 + A\beta} \rightarrow 25 = \frac{A}{10}$$

$$\boxed{A = 250}$$

$$* 1 + A\beta = 10$$

$$\beta = \frac{9}{A} = \frac{9}{250} = 0.036$$

Example

An amplifier with a nominal gain $A = 1000$ V/V exhibits a gain change of 10% as the temperature changes from 25°C to 75°C . It is required to constrain the change to 0.1 % by applying negative feedback. Calculate the largest closed loop gain possible (A_f).

Solution:

$$* A = 1000, \quad \frac{\Delta A}{A} = \frac{10}{100} = 0.1$$

$$* \text{Required } \frac{\Delta A_f}{A_f} = \frac{0.1}{100} = 0.001$$

$$\text{Sensitivity } S = \frac{\Delta A_f / A_f}{\Delta A / A} = \frac{1}{1 + A\beta}$$

$$\therefore \frac{0.001}{0.1} = \frac{1}{1 + A\beta}$$

$$\therefore \boxed{1 + A\beta = 100}$$

$$\therefore A_f = \text{closed loop gain} = \frac{A}{1 + A\beta}$$

$$\therefore A_f = \frac{1000}{100}$$

$$\boxed{A_f = 10} \#$$



Types of Amplifiers

1. Voltage Amplifier (Series-Shunt FB)

$$X_i = V_i, X_o = V_o, X_f = V_f, X_s = V_s$$

$$A = \frac{X_o}{X_i} = \frac{V_o}{V_i}, \quad \beta = \frac{X_f}{X_o} = \frac{V_f}{V_o}$$

$$A_f = \frac{X_o}{X_s} = \frac{V_o}{V_s} = \frac{A}{1 + A\beta}$$

□ Feedback input resistance (R_{if})

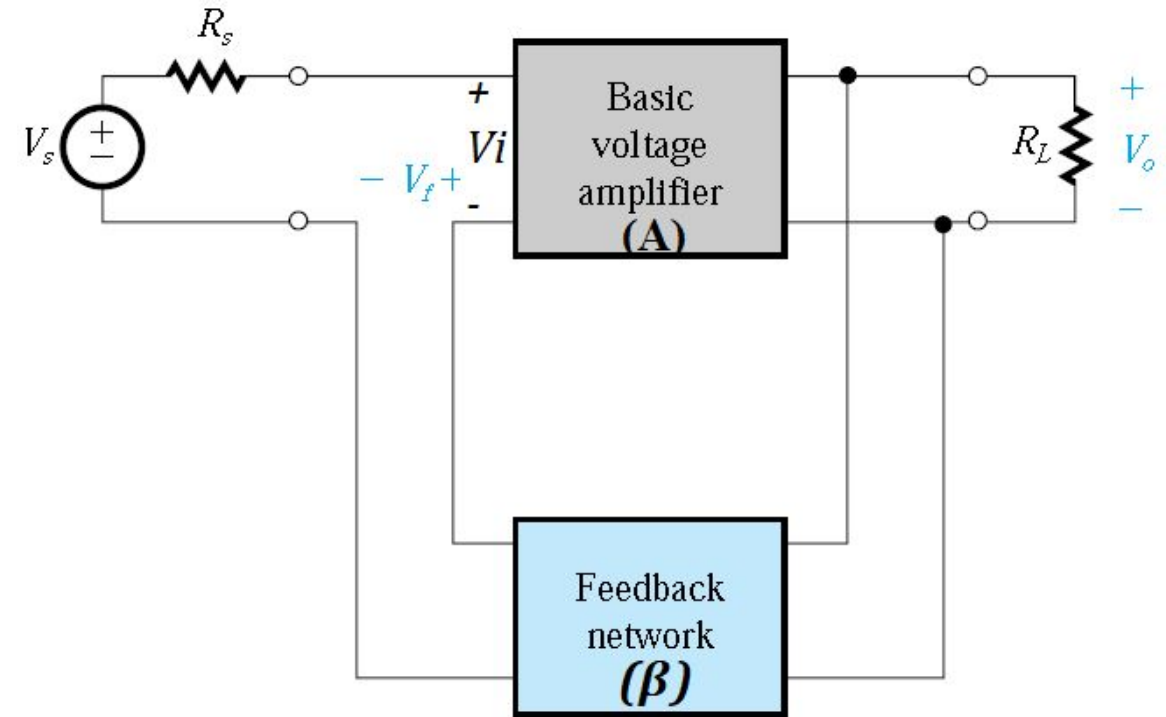
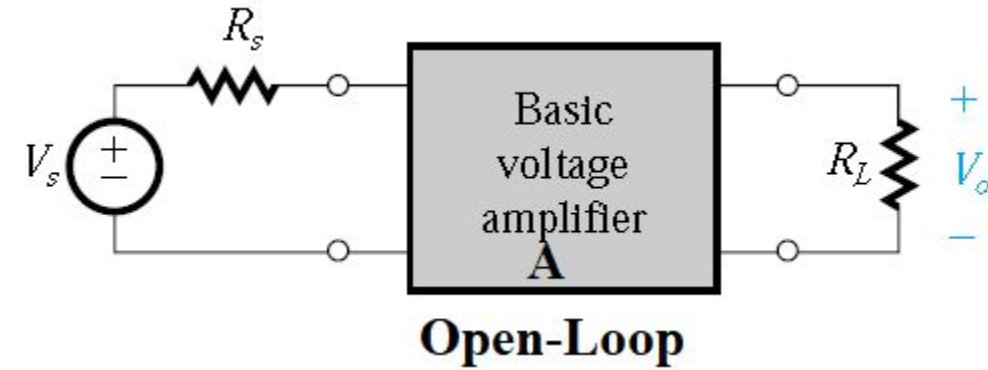
$$R_{if} = R_i (1 + A\beta)$$

□ Feedback output resistance (R_{of})

$$R_{of} = \frac{R_o}{1 + A\beta}$$

R_i input resistance without feedback

R_o output resistance without feedback



2. Current Amplifier (Shunt-Series FB)

$$X_i = I_i, X_o = I_o, X_f = I_f, X_s = I_s$$

$$A = \frac{X_o}{X_i} = \frac{I_o}{I_i}, \quad \beta = \frac{X_f}{X_o} = \frac{I_f}{I_o}$$

$$A_f = \frac{X_o}{X_s} = \frac{I_o}{I_s} = \frac{A}{1 + A\beta}$$

□ Feedback input resistance (R_{if})

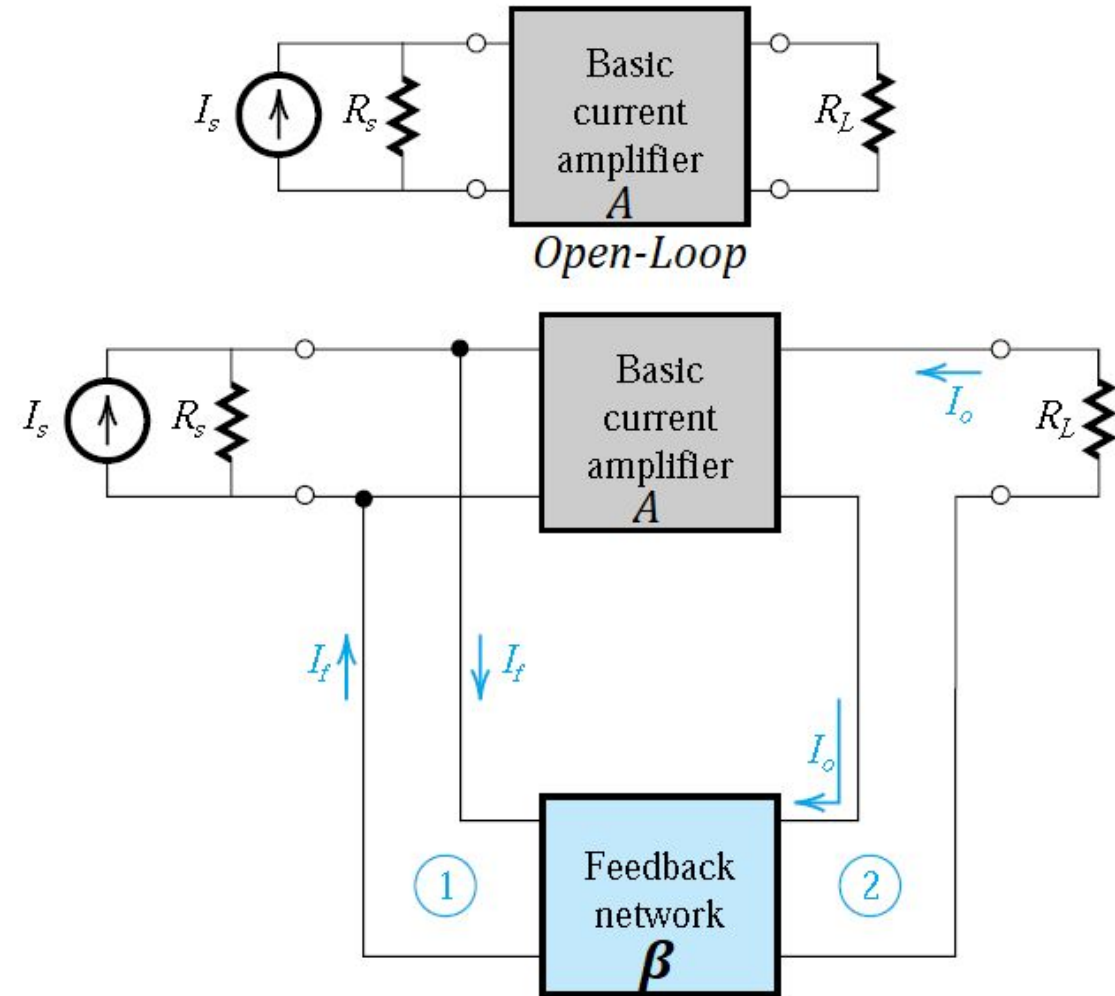
$$R_{if} = \frac{R_i}{1 + A\beta}$$

□ Feedback output resistance (R_{of})

$$R_{of} = R_o (1 + A\beta)$$

R_i input resistance without feedback

R_o output resistance without feedback



3. Trans-conductance Amplifier (Series-Series FB)

$$X_i = V_i, X_o = I_o, X_f = V_f, X_s = V_s$$

$$A = \frac{X_o}{X_i} = \frac{I_o}{V_i}, \quad \beta = \frac{X_f}{X_o} = \frac{V_f}{I_o}$$

$$A_f = \frac{X_o}{X_s} = \frac{I_o}{V_s} = \frac{A}{1 + A\beta}$$

□ Feedback input resistance (R_{if})

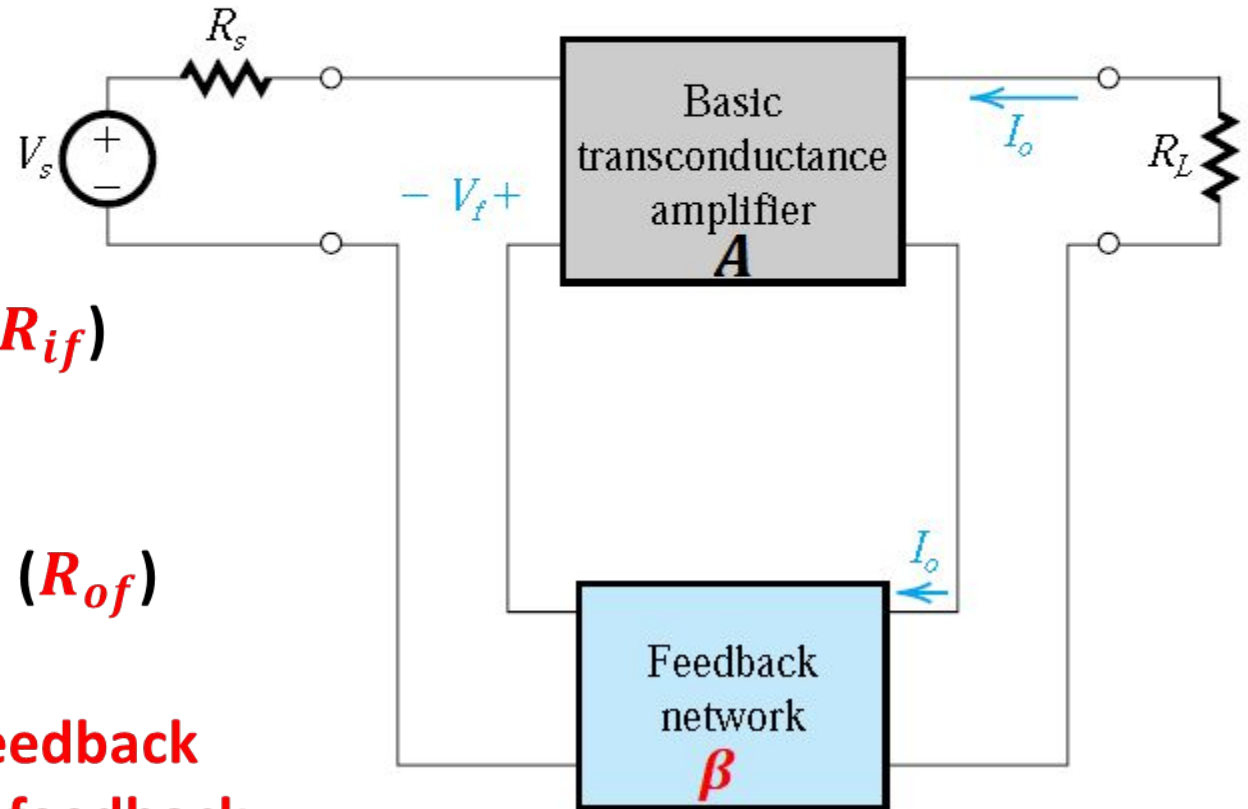
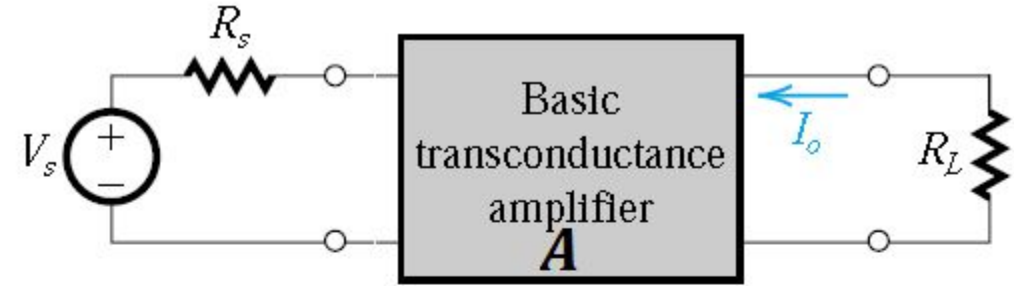
$$R_{if} = R_i (1 + A\beta)$$

□ Feedback output resistance (R_{of})

$$R_{of} = R_o (1 + A\beta)$$

R_i input resistance without feedback

R_o output resistance without feedback



4. Trans-Resistance Amplifier (Shunt-Shunt FB)

$$X_i = I_i, X_o = V_o, X_f = I_f, X_s = I_s$$

$$A = \frac{X_o}{X_i} = \frac{V_o}{I_i}, \quad \beta = \frac{X_f}{X_o} = \frac{I_f}{V_o}$$

$$A_f = \frac{X_o}{X_s} = \frac{V_o}{I_s} = \frac{A}{1 + A\beta}$$

□ Feedback input resistance (R_{if})

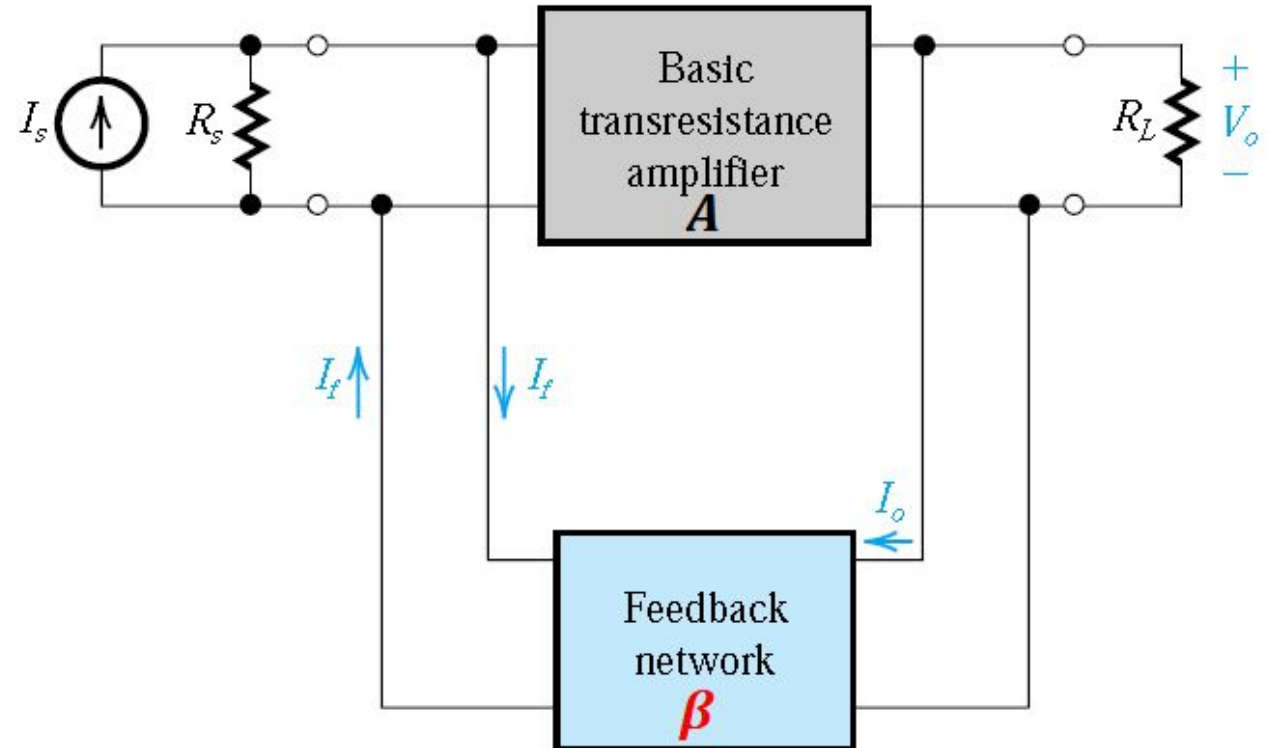
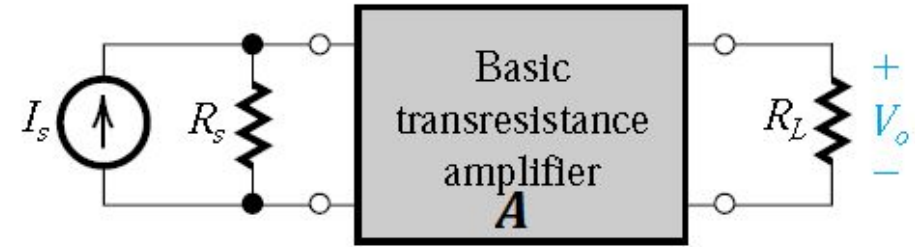
$$R_{if} = \frac{R_i}{1 + A\beta}$$

□ Feedback output resistance (R_{of})

$$R_{of} = \frac{R_o}{1 + A\beta}$$

R_i input resistance without feedback

R_o output resistance without feedback



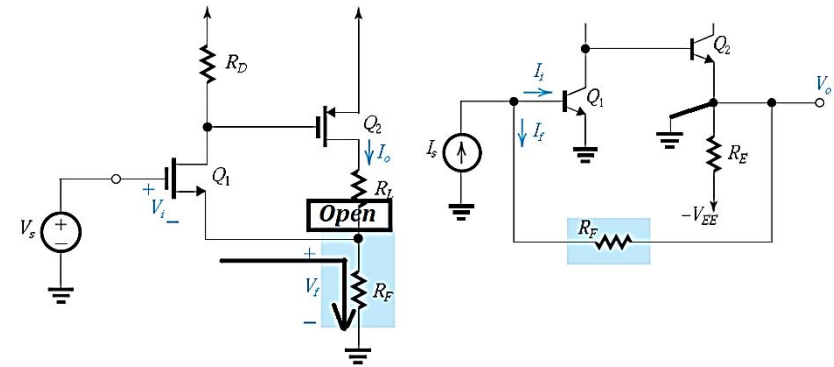
Analysis of the Feedback Amplifier Circuits

Steps:

1. Draw the input Circuit **Without Feedback (Cancel the output)**

Output topology **Shunt** → Out = 0 (short $V_o = 0$)

Output topology **Series** → Out = Open ($I_o = 0$)



2. Draw the Output Circuit **Without Feedback (Cancel the Input)**

Input topology **Shunt** → Input of amplifier = 0 (short input)

Input topology **Series** → Input of amplifier = Open

And calculate the feedback factor β

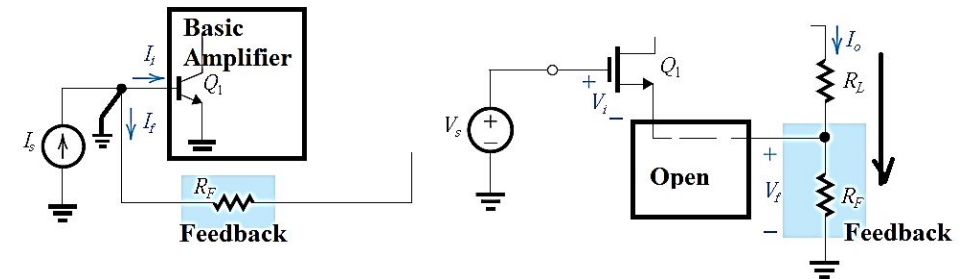
3. Draw the Total circuit without feedback (Input and output)

4. Draw the Small signal model and analyze the circuit and find:

- Open Loop gain $A = \frac{X_o}{X_s}$ (and you had calculated β in step 2)
- The open-loop input resistance R_i
- The open-loop output resistance R_o

5. The Feedback gain is then $A_f = \frac{A}{1+A\beta}$

6. Calculate R_{if} and R_{of} according to the topology.



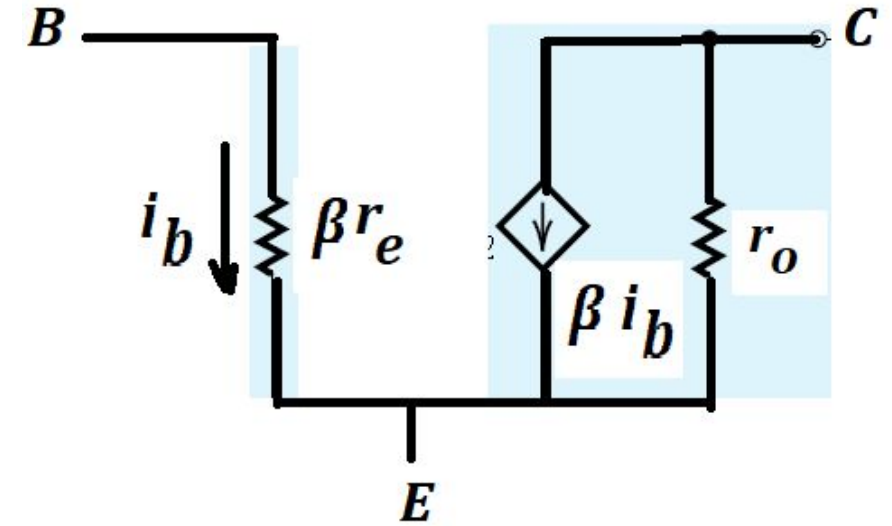
Small Signal AC Model

1. BJT Model

$$\square \quad r_e = \frac{r_\pi}{\beta} = \frac{V_T}{I_C}$$

$V_T = 0.025 \text{ V}$ at room temp.

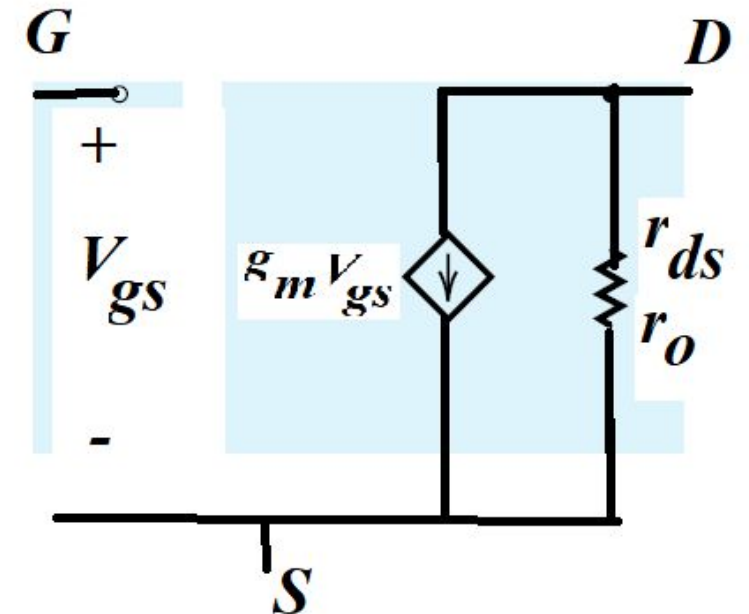
$$\square \quad r_o = \frac{V_A}{I_C}$$



2. MOSFET Model

$$\square \quad g_m = \sqrt{2 K I_D}$$

$$\square \quad r_o = r_{ds} = \frac{V_A}{I_D}$$



Feedback Topologies

1. Series-Series Feedback:

Input= Voltage (V_s) □ Voltage Source

Output = Current (I_o) □ Amp.out is a Current Source

2. Series-Shunt Feedback:

Input= Voltage (V_s) □ Voltage Source

Output = Voltage (V_o) □ Amp.out is a Voltage Source

3. Shunt-Series Feedback:

Input= Current (I_s) □ Current Source

Output = Current (I_o) □ Amp.out is a Current Source

4. Shunt-Shunt Feedback:

Input= Current (I_s) □ Current Source

Output = Voltage (V_o) □ Amp.out is a Voltage Source

Series Input □ Voltage
Shunt Input □ Current

Series Out □ Current
Shunt Out □ Voltage



Input and Output Resistances (R_{in} and R_{out})

□ R_{in} The input resistance excluding R_s

- Series Input topology $\rightarrow R_{in} = R_{if} - R_s$

- Shunt Input Topology $\rightarrow R_{in} = \frac{1}{\frac{1}{R_{if}} - \frac{1}{R_s}}$

□ R_{out} The Output resistance excluding R_L

- Series Output topology $\rightarrow R_{out} = R_{of} - R_L$

- Shunt Output Topology $\rightarrow R_{out} = \frac{1}{\frac{1}{R_{of}} - \frac{1}{R_L}}$



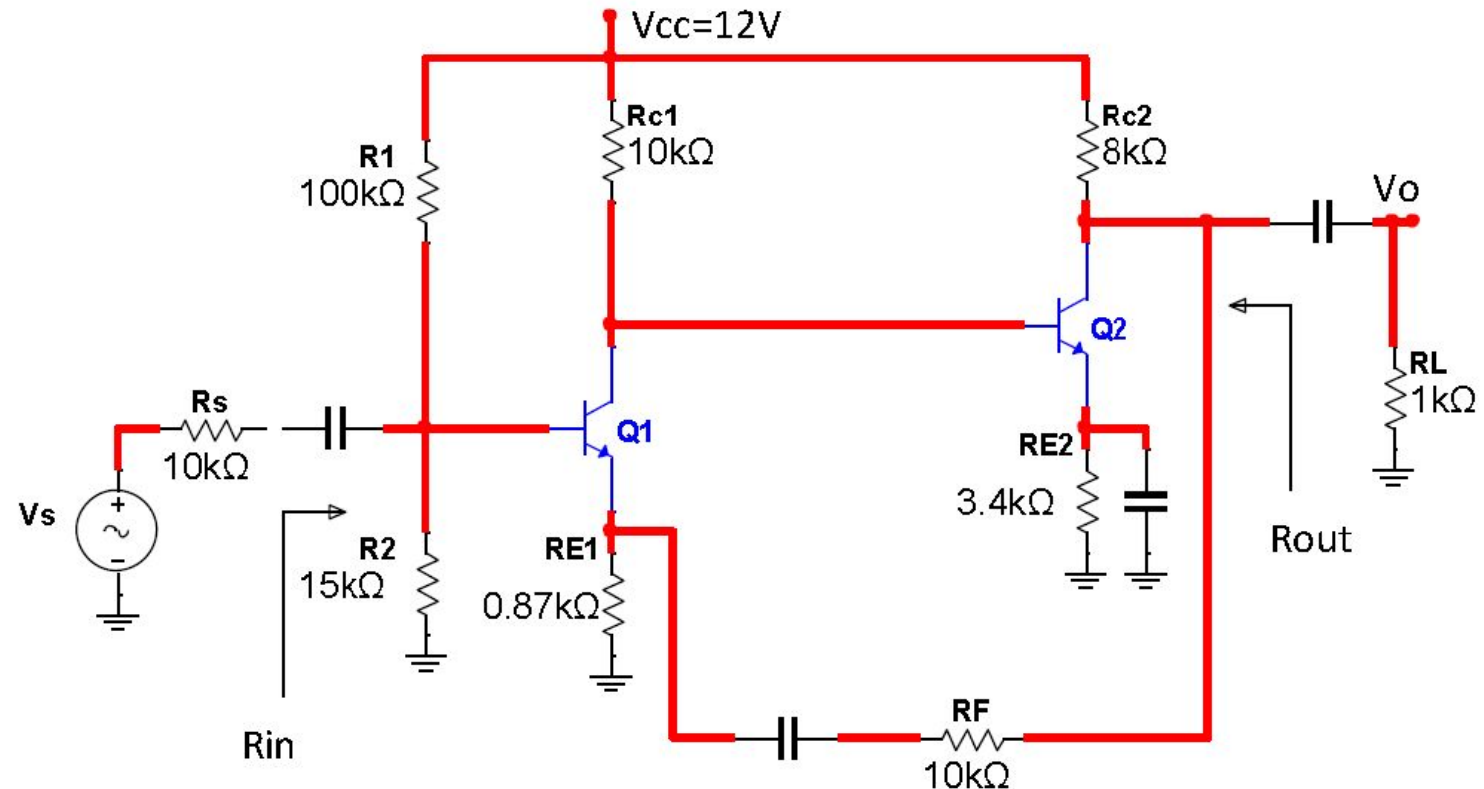
Example

Series-Shunt

The circuit shown in Figure represents a Series-Shunt Feedback amplifier circuit. Analyze the circuit and Calculate:

- The feedback factor β and the open loop gain A .
- The feedback gain A_f .
- The feedback input and output resistances (R_{if} and R_{of}).
- R_{in} and R_{out} .

Given: $\beta_1 = \beta_2 = 100$, $r_{e1} = 26 \Omega$, $r_{e2} = 68 \Omega$, $r_{o1} = 100k \Omega$ and $r_{o2} = 261.54k \Omega$.



Solution:

(a) The feedback factor β and the open loop gain A .

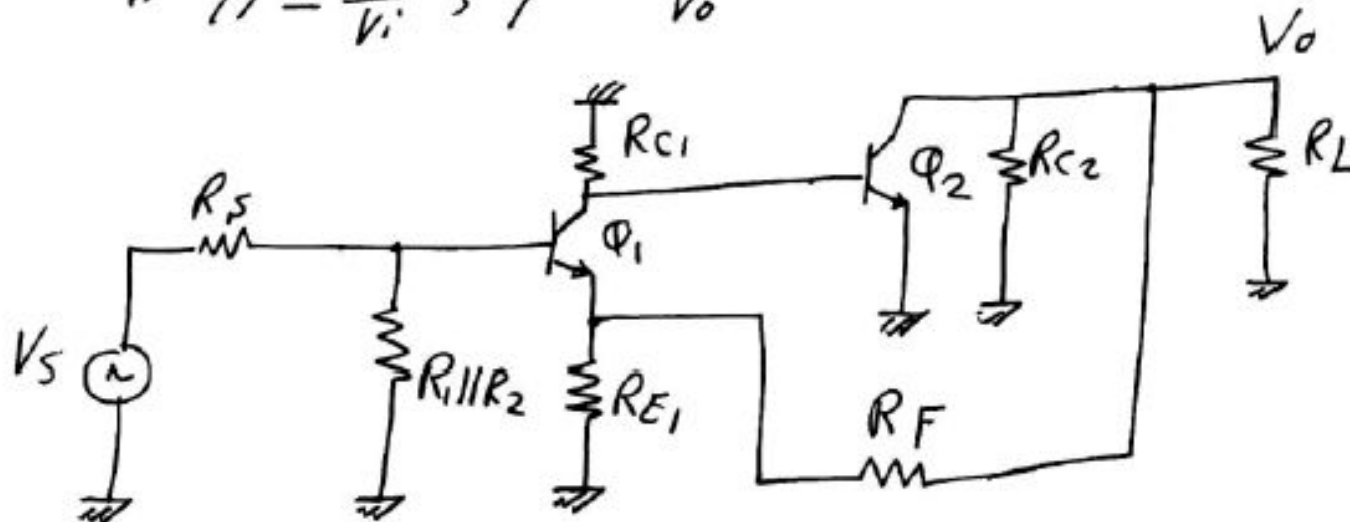
Topology:-

Input \rightarrow Series

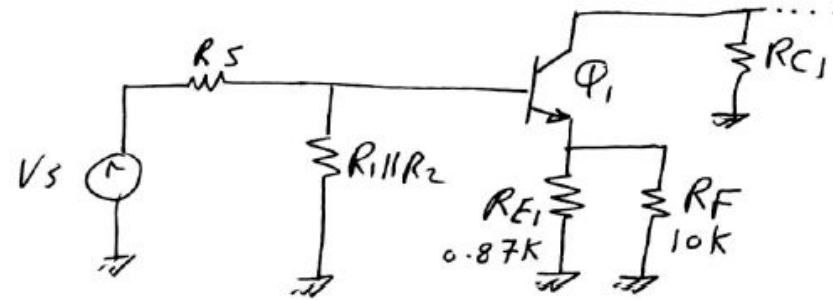
output \rightarrow shunt

\therefore Series — shunt Feedback

$$\therefore A = \frac{V_o}{V_i}, \quad \beta = \frac{V_F}{V_o}$$



* Input Circuit without Feedback:-
 o/p \rightarrow short ($V_o = 0$)



Let $R_x = R_{E1} \parallel R_F = 0.8k$

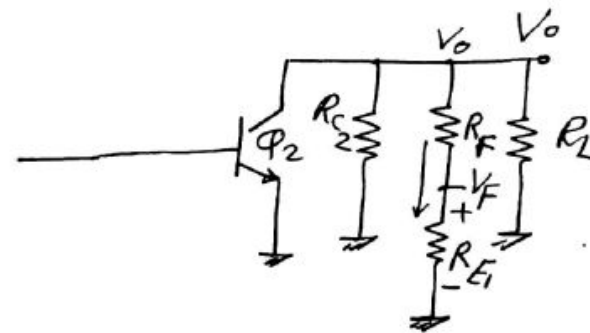
* Output Circuit without Feedback
 I/P \rightarrow open

$$\beta = \frac{V_F}{V_o}$$

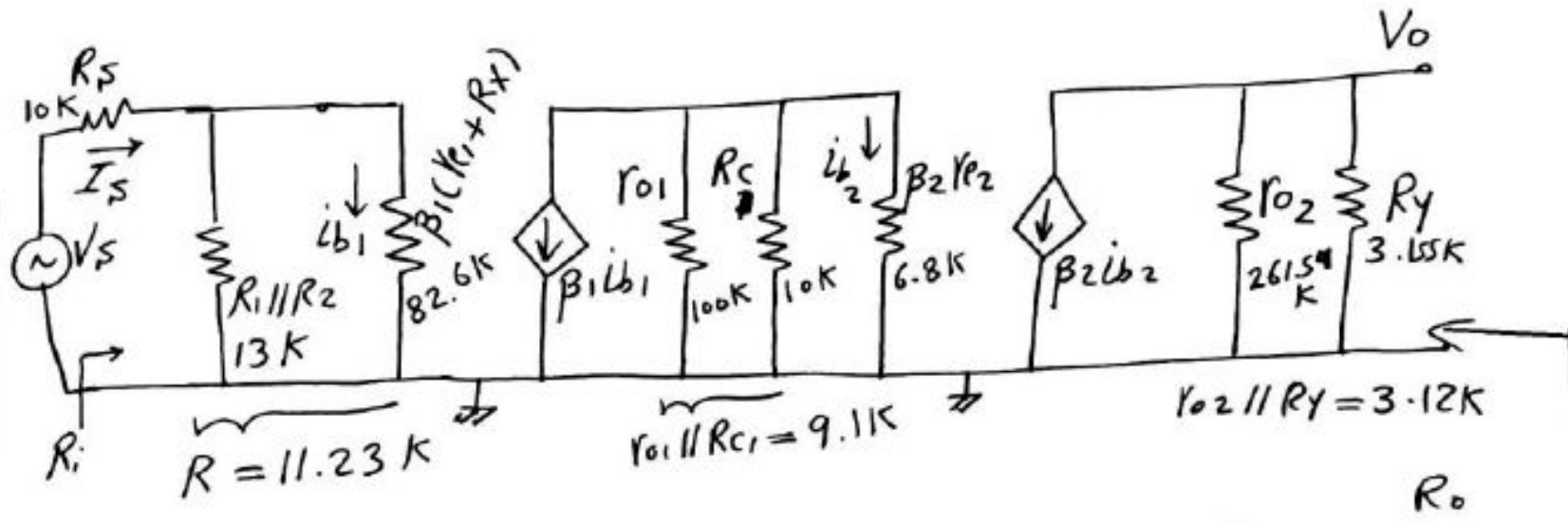
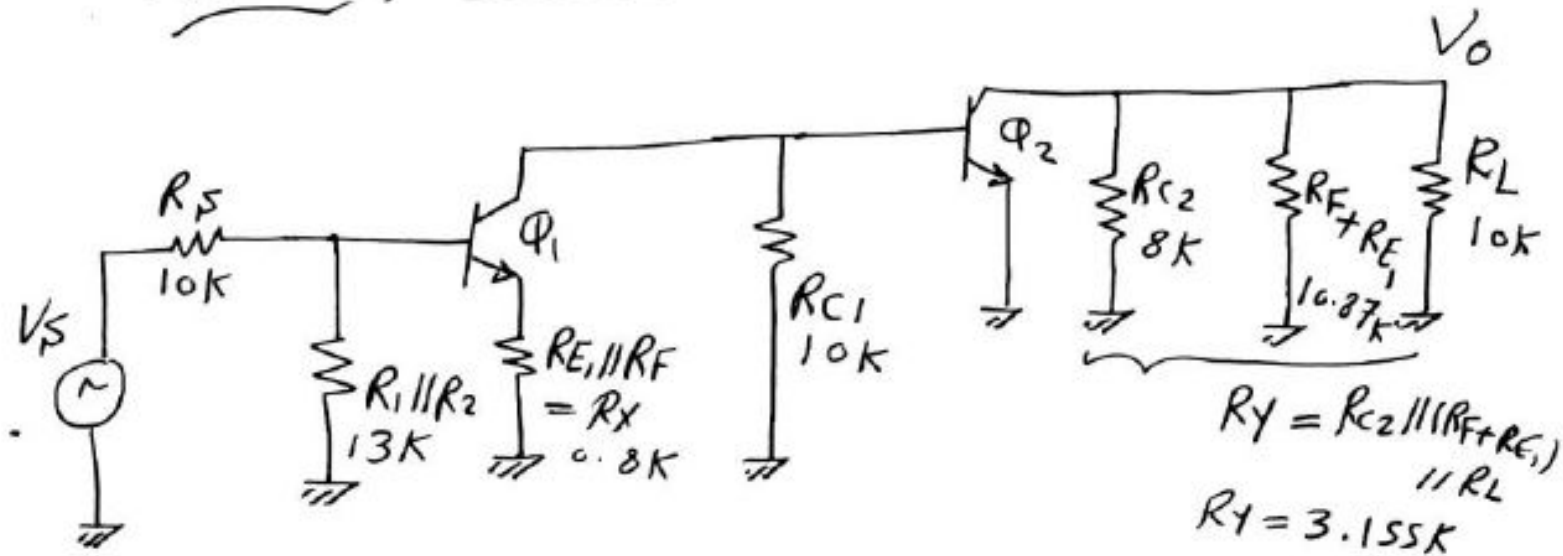
$$* V_F = \frac{V_o}{R_F + R_{E1}} \cdot R_{E1}$$

$$\therefore \beta = \frac{V_F}{V_o} = \frac{R_{E1}}{R_F + R_{E1}} = \frac{0.87}{10 + 0.87} \approx 0.08$$

$$\beta = 0.08$$



Total Circuit without F.B



$$A = \frac{V_O}{V_S} = \frac{V_O}{i_{b2}} \times \frac{i_{b2}}{i_{b1}} \times \frac{i_{b1}}{I_S} \times \frac{I_S}{V_S}$$

$$* V_o = -\beta_2 i_{b2} (r_{o2} \parallel R_Y)$$

$$\therefore \frac{V_o}{i_{b2}} = -\beta_2 (r_{o2} \parallel R_Y) = -311.74$$

$$* i_{b2} = -\beta_1 i_{b1} \frac{(r_{o1} \parallel R_{C1})}{(r_{o1} \parallel R_{C1}) + \beta_2 r_{e2}}$$

$$\therefore \frac{i_{b2}}{i_{b1}} = -\beta_1 \frac{r_{o1} \parallel R_{C1}}{r_{o1} \parallel R_{C1} + \beta_2 r_{e2}} = -57.233$$

$$* i_{b1} = I_S \frac{R_1 \parallel R_2}{(R_1 \parallel R_2) + \beta_1 (r_{e1} + R_X)}$$

$$\frac{i_{b1}}{I_S} = \frac{R_1 \parallel R_2}{(R_1 \parallel R_2) + \beta_1 (r_{e1} + R_X)} \approx 0.136$$

$$* I_S = \frac{V_S}{R_S + (R_1 \parallel R_2 \parallel \beta_1 (r_{e1} + R_x))}$$

$$\frac{I_S}{V_S} = \frac{1}{R_S + [R_1 \parallel R_2 \parallel \beta_1 (r_{e1} + R_x)]} = 0.0471$$

$$\therefore A = \frac{V_o}{V_S} = (-311.74)(-57.233)(0.136)(0.0471,$$

$$A \approx 114.3, \quad \beta = 0.08$$

$$* 1 + A\beta = 10.144$$

b. The feedback gain A_f .

$$A_{vf} = \frac{V_o}{V_s} = \frac{A}{1+A\beta} = \frac{114.3}{10.144} = 11.2677$$

Feedback gain.

c. The feedback input and output resistances (R_{if} and R_{of}).

$$* R_i = R_s + [R_1 \parallel R_2 \parallel \beta_1 (r_{e1} + R_x)]$$

$$\boxed{R_i = 21.23 \text{ k}\Omega}$$

$$* R_{if} = R_i (1 + A\beta) = 21.23 \times 10.144$$

$$R_{if} = 215.36 \text{ k}\Omega$$

$$* R_o = R_y \parallel r_{o2} = 3.12 \text{ k}\Omega$$

$$* R_{of} = \frac{R_o}{1 + A\beta} = \frac{3.12}{10.144}$$

$$R_{of} = 0.3076 \text{ k}\Omega = 307.6 \Omega$$

d. R_{in} and R_{out} .

$$* \{ R_{in} = R_{if} - R_s = 205.36 \text{ k}\Omega$$

$$R_{out} = \frac{1}{\frac{1}{R_{of}} - \frac{1}{R_L}} = \frac{1}{\frac{1}{0.3076} - \frac{1}{10}}$$

$$R_{out} = 0.31736 \text{ k}\Omega = 317.36 \text{ }\Omega$$