

Sheet 3
Active Filters

1. Design KHN Filter to realize BPF, $f_0 = 10 \text{ KHz}$, $BW = 200 \text{ Hz}$, choose $C = 10 \text{ nF}$, what is the value of center freq. gain.

Sol:

$$f_0 = 10 \text{ KHz}, \quad BW = 200 \text{ Hz}, \quad C = 10 \text{ nF}$$

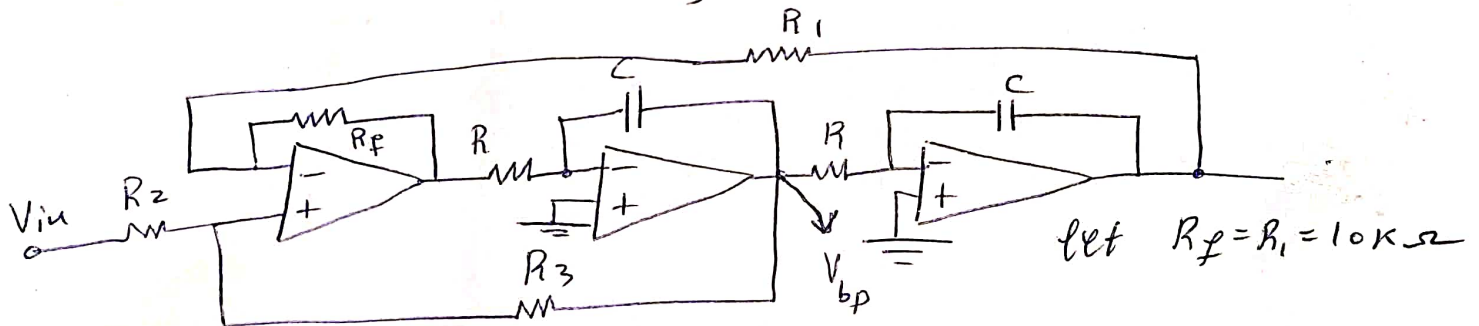
$$\Rightarrow Q = \frac{f_0}{BW} = \frac{10 \times 10^3}{200} = 50$$

$$\because f_0 = \frac{1}{2\pi RC} \Rightarrow R = \frac{1}{2\pi \times 10 \times 10^{-9} \times 10 \times 10^3} = 1.59 \text{ K}\Omega$$

$$\Rightarrow K = 2 - \frac{1}{Q} = 2 - \frac{1}{50} = 1.98$$

$$\Rightarrow \frac{R_3}{R_2} = 2Q - 1 \Rightarrow \text{let } R_2 = 1 \text{ K}\Omega$$

$$\Rightarrow R_3 = 2(50 - 1)R_2 = 98 \text{ K}\Omega$$



* Center Frequency :

$$\Rightarrow V_{BP} = -\frac{\omega_0}{s} V_{HP} = \left(-\frac{\omega_0}{s}\right) \frac{K s^2}{s^2 + \left(\frac{\omega_0}{Q}\right)s + \omega_0^2} (V_{in})$$

$$\Rightarrow \frac{V_{BP}}{V_{in}} = \frac{-K \omega_0 s}{s^2 + \left(\frac{\omega_0}{Q}\right)s + \omega_0^2} = \frac{-jK \omega \omega_0}{-\omega^2 + j\frac{\omega_0}{Q}\omega + \omega_0^2}$$

$$\Rightarrow A_m = \left| \frac{V_{BP}}{V_{in}} \right|_{\omega=\omega_0} = \frac{K \omega_0 \omega_0}{\sqrt{\left(\omega_0^2 - \omega_0^2\right)^2 + \left(\frac{\omega_0^2}{Q}\right)^2}} = \frac{K \omega_0^2}{\frac{\omega_0^2}{Q}} = KQ$$

$$\Rightarrow A_m = KQ = (1.98)(50) = 99$$

2. Design KHN Filter to realize HPF with $f_0 = 10 \text{ KHz}$
 $Q = 2$, choose $C = 1 \text{ nF}$. what is the value of high
 frequency gain obtained? what is the center frequency
 gain of the BPF that is simultaneously available at
 the o/p of the 1st integrator.

Solⁿ

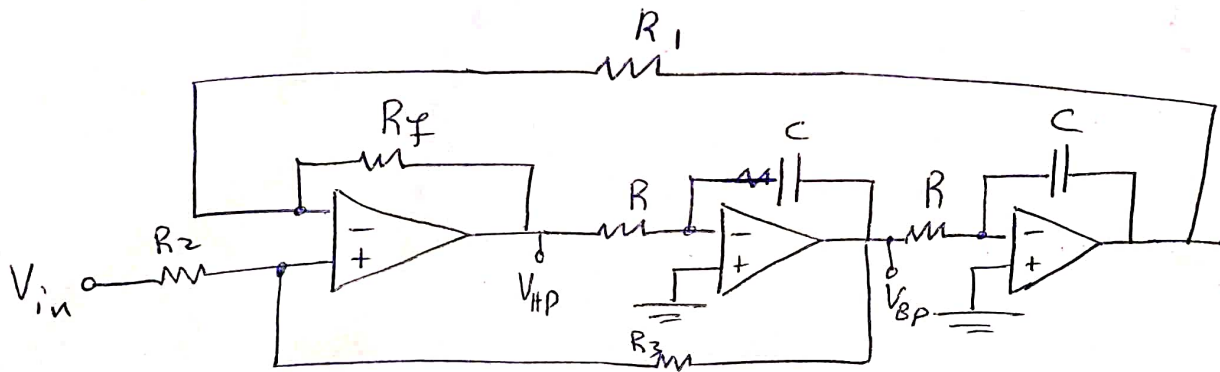
$$f_0 = 10 \text{ KHz}, Q = 2, C = 1 \text{ nF}$$

$$\Rightarrow f_0 = \frac{1}{2\pi RC} \Rightarrow R = \frac{1}{2\pi(10^4)(10^{-9})} = 15.9 \text{ K}\Omega$$

$$\therefore K = 2 - \frac{1}{Q} = 2 - \frac{1}{2} = 1.5$$

$$\therefore \frac{R_3}{R_2} = 2Q - 1 = 0 \Rightarrow R_3 = 0$$

$$\text{Let } R_2 = 10 \text{ K}$$



* at High frequency $\therefore \frac{V_{HP}}{V_{in}} = \frac{Ks^2}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2}$

$$\text{at } s \rightarrow \infty: \left| \frac{V_{HP}}{V_{in}} \right|_{s \rightarrow \infty} = \left| \frac{Ks^2}{s^2 + \frac{\omega_0}{Q}s + \omega_0^2} \right|_{s \rightarrow \infty} = K$$

* center freq gain of BPF \therefore

$$\Rightarrow \left| \frac{V_{BP}}{V_{in}} \right| = \left| \frac{-jK\omega\omega_0}{-\omega^2 + j\frac{\omega_0}{Q}\omega + \omega_0^2} \right|$$

$$\Rightarrow \left| \frac{V_{BP}}{V_{in}} \right|_{\omega=\omega_0} = \frac{K\omega_0^2}{\sqrt{(\omega_0^2 - \omega_0^2)^2 + \left(\frac{\omega_0^2}{Q}\right)^2}} = KQ = (1.5)(2) = 3$$

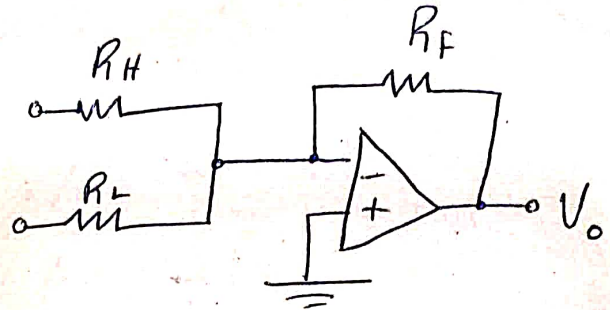
3. Use KHN ct. with an o/p of a summing amplifier to design a band-stop notch filter with $f_0 = 5 \text{ KHz}$, $f_u = 8 \text{ KHz}$, $Q = 5$ & dc gain of 3, select $C = 1 \text{ nF}$.

Sol

$$\therefore f_0 = 5 \text{ KHz}$$

$$f_u = 8 \text{ KHz}$$

$$\Rightarrow \left(\frac{f_u}{f_0} \right)^2 = \frac{R_H}{R_L}$$



$$\left(\frac{R_H}{R_L} \right) = \left(\frac{\omega_u}{\omega_0} \right)^2$$

$$\text{Let } R_H = 10 \text{ K}\Omega \Rightarrow \left(\frac{8}{5} \right)^2 = \frac{10 \text{ K}}{R_L}$$

$$\Rightarrow \boxed{R_L = 3.906 \text{ K}\Omega}$$

* For the KHN filter :

$$f_0 = \frac{1}{2\pi RC}, \quad C = 1 \text{ nF}$$

$$\Rightarrow R = \frac{1}{2\pi (5 \times 10^3) (1 \times 10^{-9})} = 31.83 \text{ K}\Omega$$

$$Q = 5 \Rightarrow K = 2 - \frac{1}{Q} = 2 - \frac{1}{5} = 1.8$$

$$\frac{R_3}{R_2} = 2Q - 1 \Rightarrow \text{Let } R_2 = 10 \text{ K}\Omega$$

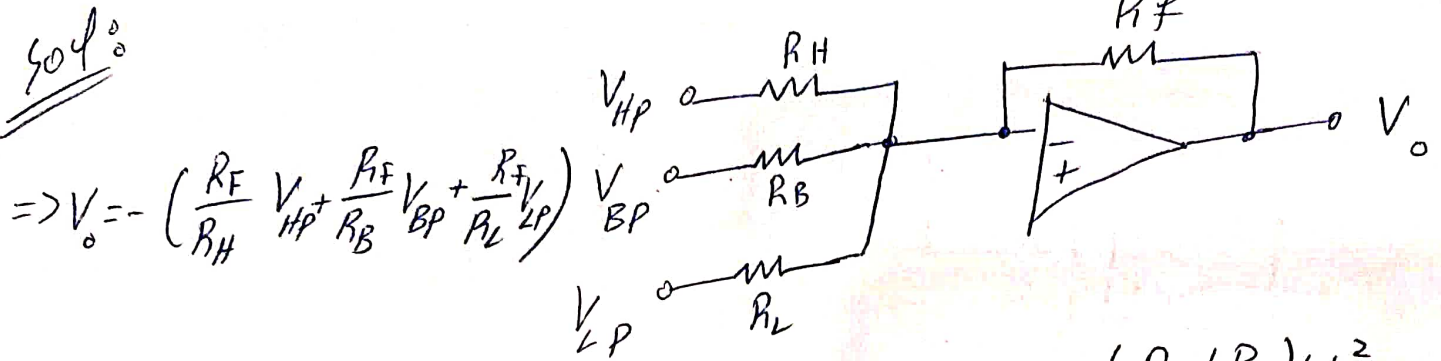
$$R_3 = (10 - 1)(10 \text{ K}\Omega) = 90 \text{ K}\Omega$$

$$\Rightarrow \frac{V_o}{V_i} = \frac{-K (R_F/R_H) s^2 + (R_F/R_L) \omega_0^2}{s^2 + s(\omega_0/Q) + \omega_0^2}$$

$$\text{at } s \rightarrow 0 \Rightarrow \left. \frac{V_o}{V_i} \right|_{s \rightarrow 0} = -K \left(\frac{R_F}{R_L} \right) \Rightarrow K \frac{R_F}{R_L} = 3 \Rightarrow R_F = 21 \text{ K}\Omega$$

dc gain

4. Use the KHN ct. with summing amplifier to get flat gain of all pass filter.



$$\Rightarrow V_O = - \left(\frac{R_F}{R_H} V_{HP} + \frac{R_F}{R_B} V_{BP} + \frac{R_F}{R_L} V_{LP} \right)$$

$$\Rightarrow \frac{V_O}{V_i} = \frac{-K \left(\frac{R_F}{R_H} s^2 - s \left(\frac{R_F}{R_B} \right) \omega_0 + \left(\frac{R_F}{R_L} \right) \omega_0^2 \right)}{s^2 + s \left(\frac{\omega_0}{Q} \right) + \omega_0^2}$$

$$= -K \left(\frac{R_F}{R_H} \right) \cdot \frac{s^2 - s \left(\frac{R_H}{R_B} \right) \omega_0 + \left(\frac{R_H}{R_L} \right) \omega_0^2}{s^2 + s \left(\frac{\omega_0}{Q} \right) + \omega_0^2}$$

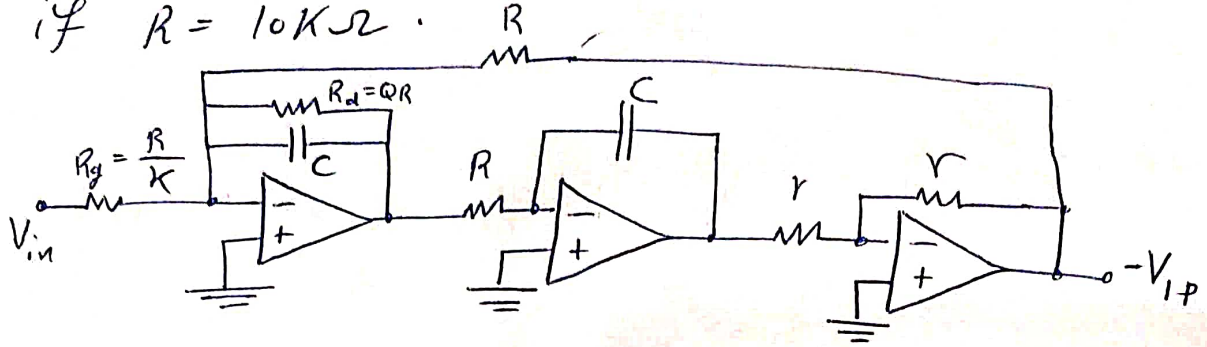
* For an all pass filter :

$$T(s) = \text{gain} \times \frac{s^2 - s \left(\frac{\omega_0}{Q} \right) + \omega_0^2}{s^2 + s \frac{\omega_0}{Q} + \omega_0^2}$$

$$\Rightarrow \boxed{\frac{R_H}{R_B} = \frac{1}{Q}} \quad , \quad \frac{R_H}{R_L} = 1 \Rightarrow \boxed{R_H = R_L}$$

$$\text{Flat gain} = -K \left(\frac{R_F}{R_H} \right)$$

5. Use the Tow-Thomas filter shown to design a 2nd order BPF with $f_0 = 10\text{KHz}$, $Q = 20$ & unity center gain if $R = 10\text{K}\Omega$.



Sol:

$$f_0 = \frac{1}{2\pi C \sqrt{R_2 R_3}}, \quad R_2 = R_3 = R$$

$$\Rightarrow f_0 = \frac{1}{2\pi C R} = \frac{1}{2\pi C (10\text{K})} = 10 \times 10^3$$

$$\Rightarrow C = \frac{1}{2\pi \times 10 \times 10^3 \times (10 \times 10^3)} = 1.59 \text{ nF}$$

$$Q = \frac{R_1}{\sqrt{R R}} = \frac{R_d}{R} \Rightarrow R_d = QR = (20)(10\text{K})$$

$$\Rightarrow R_d = 200 \text{ K}\Omega$$

$$A_m = \frac{R_d}{R_g} = \frac{200 \text{ K}}{R_g} = 1$$

$$\Rightarrow R_g = 200 \text{ K}\Omega$$

6. use the Tow-Thomas biquad Filter shown in the previous problem to analyze the TF of the 2nd order low pass filter & design the filter $f_0 = 10 \text{ KHz}$ & center frequency gain of 50: if $R = 10 \text{ K}\Omega$, give values of C , R_d & R_g .

Sol:

$$\therefore V_{LP} = V_{BP} \left(-\frac{1}{sRC} \right)$$

$$\Rightarrow T_{LP} = \left(-\frac{1}{sRC} \right) \frac{-\left(\frac{1}{R_g C} \right) s}{s^2 + \frac{1}{R_d C} s + \frac{1}{R^2 C^2}}$$

$$= \frac{\frac{1}{RR_g C^2}}{s^2 + \frac{1}{R_d C} s + \frac{1}{R^2 C^2}} = \frac{a_0}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

$$\Rightarrow \boxed{\omega_0 = \frac{1}{RC}} \quad , \quad \frac{\omega_0}{Q} = \frac{1}{R_d C} \Rightarrow Q = \frac{\omega_0}{R_d C}$$

$$\boxed{Q = \frac{1}{RR_d C^2}}$$

$$\Rightarrow |T_{LP}|_{\omega=\omega_0} = \frac{a_0}{\sqrt{(\omega_0^2/\omega_0^2)^2 + \left(\frac{\omega_0^2}{Q}\right)^2}} = \frac{a_0}{\omega_0^2/Q} = \frac{a_0 Q}{\omega_0^2}$$

$$= \frac{\frac{1}{RR_g C^2} \left(\frac{Q}{\omega_0} \right)}{\omega_0}$$

$$= \frac{R_d C^2}{RR_g C^2} \cdot \frac{1}{RC}$$

$$= \frac{R_d}{R_g}$$