

### 3. Electromechanical Energy Conversion

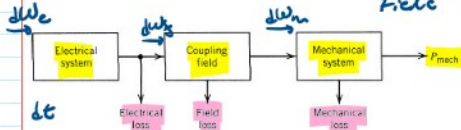
Tuesday, October 27, 2020 3:46 AM

#### 3.1 Energy Conversion Process

- Electromechanical Converter System:

1. Electric System
2. Mechanical System
3. Coupling Field

Elect Energy = Mech Energy + Increase in stored Energy in Coupling Field + Energy loss



$dW_e = E_{\text{elec}} \times \text{flowing to system}$

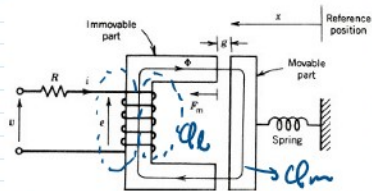
$dW_f = E \text{ supplied to field} \rightarrow \text{if core losses } \downarrow, dW_f = \Delta E \text{ stored field}$

$dW_m = E \text{ converted to mech form} \rightarrow \text{if friction \& windage } \downarrow, dW_m = E \text{ mech output}$

$$dW_e = dW_m + dW_f$$

#### 3.2 Field Energy

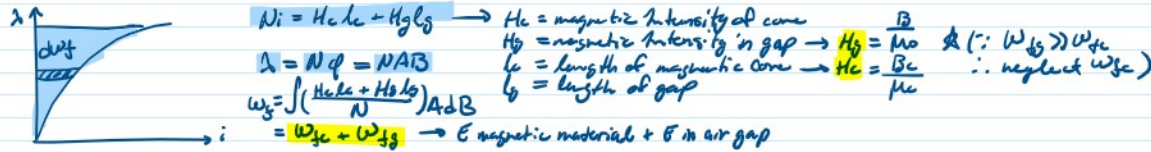
- Faraday & KVL:  $v = ri + \frac{d\lambda}{dt} = ri + e$
- Flux linkage:  $\lambda = N\phi = N(Lm + \phi_l) \rightarrow \text{Magnetizing Flux: } \phi_m = \frac{Ni}{R_m}$
- Inductances:  $\lambda = \left(\frac{N^2}{R_l} + \frac{N^2}{R_m}\right)i = (L_l + L_m)i$  Flux leakage:  $\phi_l = \frac{Ni}{R_l}$   
 $\rightarrow \text{Magnetizing } L = \frac{N^2}{R_m}$   
 $\rightarrow \text{Leakage } L$
- Reluctance:  $R_m = R_c + \lambda R_m$



When Static:  $dW_m = 0 \therefore dW_e = dW_f$

$$e = \frac{d\lambda}{dt} \quad dW_e = e i dt = d\lambda i \quad dW_f = i d\lambda$$

$$\text{flux linkage increased } 0 \rightarrow \lambda: W_f = \int_0^\lambda e i dt = \int_0^\lambda i d\lambda$$

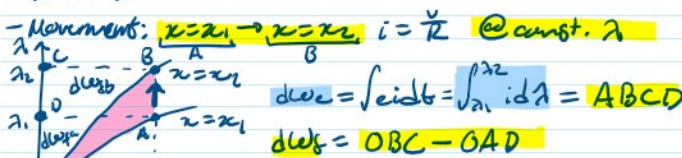
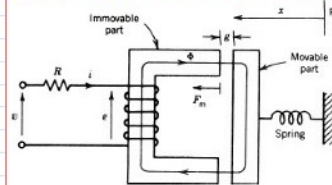


#### 3.2.1 Energy, Coenergy

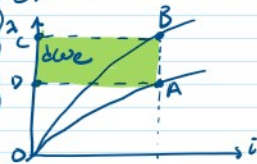
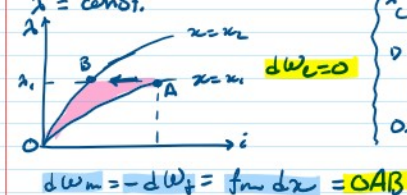
- Air gap:  $\lambda - i \sim \text{linear}$
- Coenergy:  $W'_f = \int_0^i \lambda di$   
 $\hookrightarrow W'_f > W_f$  if  $\lambda - i$  non-linear  
 $\hookrightarrow W'_f = W_f$  if  $\lambda - i$  linear



#### 3.3 Mechanical Force in the Electromagnetic System



- If quick movement, flux linkage  $\lambda = \text{const.}$



$$dW_m = dW_e - dW_f = ABCD - (OBC - OAD) = ABCD - OBC + OAD$$

$$\therefore dW_m = OAB \quad W = \int \lambda di$$

Change in Mech Work:  $dW_m = dW_e = f_m dx$   
 Under const. current  
 $f_m = \frac{\partial W'_f(i, x)}{\partial x} \bigg|_{i = \text{const.}}$   
 $\hookrightarrow$  mech force causing displacement  $dx$

- For rapid motion: Elec input  $= i d\lambda = 0 \therefore \lambda = \text{const.}$
- Mech output  $E$  supplied by Field Energy



$$f_m = \frac{1}{2\lambda} \quad \lambda = \text{const.}$$

- For rapid motion: Elec input =  $i d\lambda = 0 \therefore \lambda = \text{const.}$
- Mech output  $\dot{E}$  supplied by field Energy



### 3.3.1 Linear System

- 2P  $R_c \ll R_g$ , ( $\uparrow l_g$ ),  $\lambda - i = \text{linear}$

$$\lambda = L(x)i \rightarrow L(x) = \text{inductance of coil, depends on } x$$

$$W_f = W_f' \int i d\lambda = \int_0^\lambda \frac{\lambda^2}{2L(x)} d\lambda = \frac{\lambda^2}{2L(x)}$$

$$f_m = - \frac{\partial}{\partial x} \left( \frac{\lambda^2}{2L(x)} \right) \Big|_{i=\text{const.}} = \frac{\lambda^2}{2L^2(x)} \frac{dL(x)}{dx} = \frac{1}{2} i^2 \frac{dL(x)}{dx}$$

$$W_i = H_g l_g = \frac{B_g^2}{\mu_0} l_g \rightarrow \text{volume of air gap}$$

$$W_g = \frac{B_g^2}{2\mu_0} \times V_g = \frac{B_g^2}{2\mu_0} A_g l_g$$

$$f_m = \frac{\partial}{\partial x} \left( \frac{B_g^2}{2\mu_0} A_g l_g \right) = \frac{B_g^2}{2\mu_0} (2A_g)$$

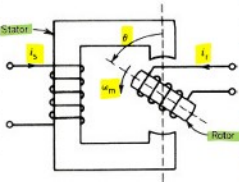
- $\uparrow$ :  $W_g$  supplies  $W_f, W_m$
- $\downarrow$ :  $W_g, W_m$  supplies  $W_e$
- $\rightarrow$ :  $W_m$  supplies  $W_g$
- $\leftarrow$ :  $W_f$  supplies  $W_m$

- Total cross-sectional  $A$  of gap =  $2A_g$
- $\therefore$  Force/unit  $A$  of gap  $\rightarrow$  magnetic pressure  $F_m$

$$F_m = \frac{B_g^2}{2\mu_0} \text{ N/m}^2$$

### 3.4 Rotating Machines

- Rotating electromagnetic system: Stator & Rotor
- Current fed into rotor circuit through brushes & slip rings



$$\text{Find } W_f \text{ using currents } i_s \text{ \& } i_r$$

$$dW_f = e_s i_s dt + e_r i_r dt$$

$$= i_s d\lambda_s + i_r d\lambda_r$$

For linear magnetic system,  $\lambda$  can be expressed w/ inductances  $L$  depending on rotor pos.  $\theta$

$$\lambda_s = L_{ss} i_s + L_{sr} i_r$$

$$\lambda_r = L_{rs} i_s + L_{rr} i_r$$

- $\rightarrow L_{ss}$  = self inductance stator windings
- $L_{rr}$  = self inductance rotor windings
- $L_{sr}, L_{rs}$  = Mutual inductances b/w stator & rotor

- For linear sys:  $L_{sr} = L_{rs}$

$$dW_f = L_{ss} i_s di_s + L_{rr} i_r di_r + L_{sr} d(i_s i_r)$$

$$W_f = \frac{1}{2} L_{ss} i_s^2 + \frac{1}{2} L_{rr} i_r^2 + L_{sr} i_s i_r$$

- Force (Torque) developed in rotational electromagnetic system:

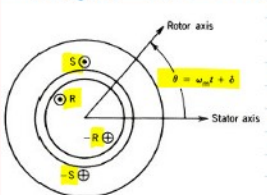
$$T = \frac{\partial W_f}{\partial \theta} \Big|_{i=\text{const.}}$$

$$\text{- For linear sys: } W_f = W_f' \therefore T = \frac{1}{2} i_s^2 \frac{dL_{ss}}{d\theta} + \frac{1}{2} i_r^2 \frac{dL_{rr}}{d\theta} + i_s i_r \frac{dL_{sr}}{d\theta}$$

- Reluctance Torque
- Torques produced in machine (self induct)
- Variation of mutual inductance b/w stator & rotor windings

### 3.5 Cylindrical machines

- 2-pole rotating machine, uniform air gap
- Reluctance magnetic path independent of rotor position
- $L_{ss}, L_{rr}$  are const., no reluctance torque produced,  $L_{sr}$  varies w/ pos.



$$\text{Torque } T = i_s i_r \frac{dL_{sr}}{d\theta}, \quad L_{sr} = M \cos \theta \rightarrow M = \text{peak value of mutual inductance } L_{sr}$$

$$i_s = I_{sm} \cos \omega_s t, \quad i_r = I_{rm} \cos(\omega_r t + \alpha) \rightarrow \omega_s, \omega_r = \text{freq of stator, rotor current}$$

$$\text{Pos. rotor: } \theta = \omega_m t + \delta \rightarrow \omega_m = \text{angular velocity rotor}$$

$$\delta = \text{rotor pos. @ } t=0$$

$$T = -I_{sm} I_{rm} M \cos \omega_s t \cos(\omega_r t + \alpha) \sin(\omega_m t + \delta)$$

$$T_{avg} \rightarrow \text{non zero if: } \omega_m \neq (\omega_s \pm \omega_r)$$

$$T_{avg} \rightarrow \text{if } \omega_m = |\omega_s \pm \omega_r|$$

$$W_m = \int T d\theta$$