# Energy-Efficient Distributed Learning With Coarsely Quantized Signals

Alireza Danaee D, Rodrigo C. de Lamare D, and Vítor H. Nascimento

Abstract—In this work, we present an energy-efficient distributed learning framework using low-resolution ADCs and coarsely quantized signals for Internet of Things (IoT) networks. In particular, we develop a distributed quantization-aware leastmean square (DQA-LMS) algorithm that can learn parameters in an energy-efficient fashion using signals quantized with few bits while requiring a low computational cost. We also carry out a statistical analysis of the proposed DQA-LMS algorithm that includes a stability condition. Simulations assess the DQA-LMS algorithm against existing techniques for a distributed parameter estimation task where IoT devices operate in a peer-to-peer mode and demonstrate the effectiveness of the DQA-LMS algorithm.

Index Terms—Distributed learning, energy-efficient signal processing, adaptive algorithms, coarse quantization.

### I. INTRODUCTION

ISTRIBUTED signal processing algorithms are of great relevance for statistical inference in wireless networks and applications such as wireless sensor networks (WSNs) [1] and the Internet of Things (IoT) [2]. These techniques deal with the extraction of information from data collected at nodes that are distributed over a geographic area. Prior work on distributed approaches has studied protocols for exchanging information [3]–[5], adaptive learning algorithms [6], [7], the exploitation of sparse measurements [8], [9], topology adaptation [10], compensation methods for highly correlated input signals [11], and robust techniques against interference and noise [12]. Although there are many studies on the need for data exchange and signaling among nodes as well as their complexity, prior work on energy-efficient techniques is rather limited.

In this context, energy-efficient signal processing techniques have gained a great deal of interest in the last decade or so due to their ability to save energy and promote sustainable development of electronic systems and devices. Electronic devices often exhibit a power consumption that is dependent on the communication module [13], [14] and from a circuit perspective on analog-to-digital converters (ADCs) and decoders [15]. Reducing the number of bits used to represent digital samples can greatly decrease the energy consumption by ADCs [16]. This is key to devices that are battery operated and wireless networks that must keep the power consumption to a low level for sustainability reasons. In particular,

Manuscript received November 2, 2020; revised December 23, 2020; accepted January 10, 2021. Date of publication January 13, 2021; date of current version February 11, 2021. This work was supported in part by CNPq, CAPES, FAPERJ and by the ELIOT Project (FAPESP 2018/12579-7, ANR-18-CE40-0030). The associate editor coordinating the review of this manuscript and approving it for publication was Dr. Arash Mohammadi. (Corresponding author: Alireza Danaee.)

Alireza Danaee and Rodrigo C. de Lamare are with the CETUC, Pontifical Catholic University of Rio de Janeiro, Rio de Janeiro 22451-900, Brazil (e-mail: alireza@cetuc.puc-rio.br; rodrigo.delamare@york.ac.uk).

Vítor H. Nascimento is with the Escola Politécnica, University of São, Sao Paulo 05508-970, Brazil (e-mail: vitor@ieee.org).

Digital Object Identifier 10.1109/LSP.2021.3051522

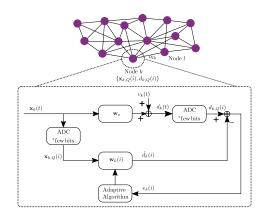


Fig. 1. A distributed adaptive IoT network.

prior work on energy efficiency has reported many contributions in signal processing for communications and electronic systems that operate with coarsely quantized signals [17]–[22].

In this work, we propose an energy-efficient distributed learning framework using low-resolution ADCs and coarsely quantized signals for IoT networks. In particular, we devise a distributed quantization-aware least-mean square (DQA-LMS) algorithm that can learn parameters in an energy-efficient way using signals quantized using few bits with a low computational cost. We also develop a statistical analysis of the DQA-LMS algorithm that includes a stability condition. Simulations assess the DQA-LMS algorithm against existing techniques for a distributed parameter estimation task with IoT devices.

This letter is structured as follows: Section II introduces the signal model and states the problem. Section III details the proposed DQA-LMS algorithm, whereas Section IV analyzes DQA-LMS. Section V shows and discusses the simulation results and Section VI draws the conclusions of this work.

# II. SIGNAL MODEL AND PROBLEM STATEMENT

We consider an IoT network consisting of N nodes or agents, which run distributed signal processing techniques to perform the desired tasks, as depicted in Fig. 1. The model adopted considers a desired signal  $d_k(i)$ , at each time i, described by

$$d_k(i) = \mathbf{w}_o^H \mathbf{x}_k(i) + v_k(i), \quad k = 1, 2, \dots, N,$$
 (1)

where  $\mathbf{w}_o \in \mathbb{C}^{M \times 1}$  is the parameter vector that the agents must estimate,  $\mathbf{x}_k(i) \in \mathbb{C}^{M \times 1}$  is the regressor and  $v_k(i)$  represents Gaussian noise with zero mean and variance  $\sigma^2_{v,k}$  at node k. We adopt the Adapt-then-Combine (ATC) diffusion rule as it outperforms the incremental and consensus protocols [3], [4]. At each node k and time i, based on the local data  $\{d_k(i), \mathbf{x}_k(i)\}$  and the estimated parameter vectors  $\mathbf{h}_l(i)$  from its neighborhood, the parameter vector with local estimates  $\mathbf{w}_k(i)$  is updated. The ATC distributed LMS

1070-9908 © 2021 IEEE. Personal use is permitted, but republication/redistribution requires IEEE permission. See https://www.ieee.org/publications/rights/index.html for more information.

(DLMS) algorithm consists of the recursions:

$$\mathbf{h}_k(i) = \mathbf{w}_k(i-1) + \mu_k \mathbf{x}_k(i) e_k^*(i), \quad \mathbf{w}_k(i) = \sum_{l \in \mathcal{N}_k} a_{lk} \mathbf{h}_l(i),$$

where  $\mathbf{h}_k(i)$  and  $\mathbf{w}_k(i)$  contain the intermediate and the local estimates of  $\mathbf{w}_o$  at node k and time i, respectively,  $e_k(i) = d_k(i) - \hat{d}_k(i) = d_k(i) - \mathbf{w}_k^H(i-1)\mathbf{x}_k(i)$  is the error between the output of the adaptive filter,  $\hat{d}_k(i)$ , and the desired signal,  $d_k(i)$ , at time i,  $\mu_k$  is the step-size for node k,  $\mathcal{N}_k$  is the set of neighbor nodes connected to node k, and  $a_{lk}$  are the combination coefficients of neighbor nodes at node k such that

$$a_{lk}=0 \text{ if } l \notin \mathcal{N}_k, \ a_{lk}>0 \text{ if } l \in \mathcal{N}_k, \ \text{and} \sum_{l \in \mathcal{N}_k} a_{lk}=1.$$
 (2)

As shown in Fig. 1, as the measurement data at each node and the unknown system are analog and each agent processes local data  $\{d_k(i), \mathbf{x}_k(i)\}$  digitally, we need two ADCs in each agent. One concern is that as the number of agents increases, the power consumption will grow considerably when using high-resolution ADCs for each agent. This motivates us to quantize signals using few bits. Therefore, the problem we are interested in solving is how to design energy-efficient distributed learning algorithms that can cost-effectively operate with coarsely quantized signals.

# III. PROPOSED DQA-LMS ALGORITHM

Let  $\mathbf{x}_{k,Q} = Q_b(\mathbf{x}_k)$  denote the b-bit quantized output of an ADC at node k, described by a set of  $2^b + 1$  thresholds  $\mathcal{T}_b = \{\tau_0, \tau_1, \ldots, \tau_{2^b}\}$ , such that  $-\infty = \tau_0 < \tau_1 < \cdots < \tau_{2^b} = \infty$ , and the set of  $2^b$  labels  $\mathcal{L}_b = \{l_0, l_1, \ldots, l_{2^b-1}\}$  where  $l_p \in (\tau_p, \tau_{p+1}]$ , for  $p \in [0, 2^b - 1]$  [18]. Let us assume that  $\mathbf{x}_k \sim \mathcal{CN}(\mathbf{0}, \mathbf{R}_{x_k})$ , where  $\mathbf{R}_{x_k} \in \mathbb{C}^{M \times M}$  is the covariance matrix of  $\mathbf{x}_k$ . We now use Bussgang's theorem [23] to derive a model for the quantized vector  $\mathbf{x}_{k,Q}$ , which we will use later to derive our DQA-LMS algorithm. Employing Bussgang's theorem,  $\mathbf{x}_{k,Q}$  can be decomposed as

$$\mathbf{x}_{k,Q} = \mathbf{G}_{k,b}\mathbf{x}_k + \mathbf{q}_k,\tag{3}$$

where the quantization distortion  $\mathbf{q}_k$  is uncorrelated with  $\mathbf{x}_k$ , and  $\mathbf{G}_{k,b} \in \mathbb{R}^{M \times M}$  is a diagonal matrix described by

$$\mathbf{G}_{k,b} = \operatorname{diag}(\mathbf{R}_{x_k})^{-\frac{1}{2}} \sum_{j=0}^{2^b - 1} \frac{l_j}{\sqrt{\pi}} \left[ \exp(-\tau_j^2 \operatorname{diag}(\mathbf{R}_{x_k})^{-1}) - \exp(-\tau_{j+1}^2 \operatorname{diag}(\mathbf{R}_{x_k})^{-1}) \right].$$
(4)

Note that this signal decomposition is also applied to the desired signal,  $d_{k,Q}$ , which is the output of the second ADC in the system, and for the particular case that  $\mathbf{R}_{x_k} = \mathbb{E}[\mathbf{x}_k \mathbf{x}_k^H] = \sigma_{x,k}^2 \mathbf{I}_M$ ,  $\mathbf{G}_{k,b}$  becomes  $g_{k,b}\mathbf{I}_M$ . However, to minimize the mean square error (MSE) between  $\mathbf{x}_k$  and  $\mathbf{x}_{k,Q}$ , we need to characterize the probability density function (PDF) of  $\mathbf{x}_k$  to find the optimal quantization labels. Since the choice of labels based on the PDF is not practical, we assume the regressor  $\mathbf{x}_k(i)$  is Gaussian, adapt the approach in [18] and approximate the thresholds and labels as follows:

- 1) We generate an auxiliary Gaussian random variable with unit variance and then use the Lloyd-Max algorithm [24], [25] to find a set of thresholds  $\widetilde{\mathcal{T}}_b = \{\tau_1, \dots, \tau_{2^b-1}\}$  and labels  $\widetilde{\mathcal{L}}_b = \{\widetilde{l}_0, \dots, \widetilde{l}_{2^b-1}\}$  that minimize the MSE between the unquantized and the quantized signals.
- 2) We complete the set of thresholds  $\mathcal{T}_b$  by adding  $\tau_0 = -\infty$  and  $\tau_{2^b} = \infty$  to the set  $\widetilde{\mathcal{T}}_b$ .
- 3) We rescale the labels such that the variance of the auxiliary random variable is 1. To do this, we multiply each label in

the set  $\widetilde{\mathcal{L}}_b$  by

$$\alpha = \left(2\sum_{j=0}^{2^{b}-1} \tilde{l}_{j}^{2} (\Phi(\sqrt{2\tau_{j+1}^{2}}) - \Phi(\sqrt{2\tau_{j}^{2}})\right)^{-1/2}$$
 (5)

to produce a set of suboptimal labels  $\mathcal{L}_b = \alpha \mathcal{L}_b$ , where  $\Phi(.)$  is the cumulative distribution function (CDF) of a standard Gaussian random variable.

We generate these thresholds and labels offline to build  $G_{k,b}$  for the proposed DQA-LMS algorithm in what follows.

# A. Derivation of DQA-LMS

We consider  $\mathbf{x}_k(t)$  and  $d_k(t)$  as the analog input and output of the unknown system  $\mathbf{w}_o$  at node k. Let  $\mathbf{x}_k(i)$  and  $d_k(i)$  denote the high-precision sampled versions of  $\mathbf{x}_k(t)$  and  $d_k(t)$ , and  $\mathbf{x}_{k,Q}(i)$  and  $d_k(i)$ , respectively. We assume that the input signal at each node is Gaussian with zero mean and covariance matrix  $\mathbf{R}_{x_k} = E[\mathbf{x}_k \mathbf{x}_k^H] = \sigma_{x_k}^2 \mathbf{I}_M$  for  $k = 1, 2, \ldots, N$ . Using (3), we can decompose  $\mathbf{x}_{k,Q}(i)$  and  $d_{k,Q}(i)$  as

$$\mathbf{x}_{k,Q}(i) = g_{k,b}(i)\mathbf{x}_k(i) + \mathbf{q}_{x,k}(i),$$

$$d_{k,Q}(i) = Q(d_k(i)) \approx g_{k,b}(i)d_k(i) + q_k(i)$$

$$= g_{k,b}(i)\mathbf{w}_o^H \mathbf{x}_k(i) + \hat{q}_k(i),$$

$$(7)$$

where  $\hat{q}_k(i) = g_{k,b}(i)v_k(i) + q_k(i)$  and  $g_{k,b}(i)$  are built from an estimate of  $\mathbf{R}_{x_k}$  given by  $\widehat{\mathbf{R}}_{x_k} = \mathbf{x}_k \mathbf{x}_k^H$  [26] that depends on the choice of  $\mathbf{x}_k$  due to (1). Because the adaptive algorithm receives a quantized signal,  $\mathbf{x}_{k,Q}$ , and the signal is assumed to be wide-sense stationary, at each time instant, we estimate  $\sigma_{x,k}^2$  using the variance of the received input,  $\sigma_{x_{k,Q}}^2$  and the distortion factor of the *b*-bit quantization,  $\rho_{k,b}$ , such that  $\sigma_{x,k}^2 \approx \sigma_{x_{k,Q}}^2 + \rho_{k,b}$ , where  $\rho_{k,b} \approx \frac{\pi\sqrt{3}}{2}2^{-2b}$  [19] for a Gaussian signal using non-uniform quantization to obtain the scalar  $g_{k,b}(i)$ .

We show next that a learning algorithm based directly on (7) is biased for estimating  $\mathbf{w}_o$ , and show how to correct for this bias. For this, let  $\beta_k(i)$  be a coefficient to be chosen shortly, and define  $\hat{d}_k(i) = \beta_k(i) \mathbf{w}_k^H(i-1) \mathbf{x}_{k,Q}(i)$  and construct an MSE cost function as described by

$$J_{k}(\mathbf{w}_{k}(i)) = \mathbb{E}[|e_{k,Q}(i)|^{2}] = \mathbb{E}[|d_{k,Q}(i) - \hat{d}_{k}(i)|^{2}]$$
$$= \mathbb{E}[|d_{k,Q}(i) - \beta_{k}(i)\mathbf{w}_{k}^{H}(i-1)\mathbf{x}_{k,Q}(i)|^{2}], \quad (8)$$

which depends only on the observed quantized quantities  $d_{k,Q}(i)$  and  $\mathbf{x}_{k,Q}(i)$ . For  $\beta_k(i)=1$  as in DLMS, the quantization of  $d_k(i)$  would result in biased estimates of  $\mathbf{w}_o$ . In the following we show how to optimally choose  $\beta_k(i)$  to reduce the bias. The proposed gradient-descent recursion to perform distributed learning based on (8) is described by

$$\mathbf{h}_k(i) = \mathbf{w}_k(i-1) - \mu_k \nabla J_k(\mathbf{w}_k(i-1)). \tag{9}$$

To compute the gradient of (9), we write the error in (8) as

$$e_{k,Q}(i) = d_{k,Q}(i) - \beta_k(i)\mathbf{w}_k^H(i-1)\mathbf{x}_{k,Q}(i)$$

$$= g_{k,b}(i)\mathbf{w}_o^H\mathbf{x}_k(i) + \hat{q}_k(i) - \beta_k(i)\mathbf{w}_k^H(i-1)$$

$$(g_{k,b}(i)\mathbf{x}_k(i) + \mathbf{q}_{x,k}(i))$$

$$= g_{k,b}(i)(\mathbf{w}_o^H - \beta_k(i)\mathbf{w}_k^H(i-1))\mathbf{x}_k(i)$$

$$- \beta_k(i)\mathbf{w}_k^H(i-1)\mathbf{q}_{x,k}(i) + \hat{q}_k(i).$$
(10)

We assume that  $\mathbf{R}_{x_k} = \mathbb{E}[\mathbf{x}_k \mathbf{x}_k^H] = \sigma_{x,k}^2 \mathbf{I}_M$  and  $\mathbf{R}_{q,k} = \mathbb{E}[\mathbf{q}_{x,k} \mathbf{q}_{x,k}^H] = \sigma_{q,k}^2 \mathbf{I}_M$ . Substituting (10) in (8) and taking

Task	Multiplications	Additions	Divisions	Exponentiations
$g_{k,b}(i) = \frac{1}{\sqrt{\sigma_{x_k}^2}} \sum_{j=0}^{2^b - 1} \frac{l_j}{\sqrt{\pi}} \left[ \exp\left(\frac{-\tau_j^2}{\sigma_{x_k}^2}\right) - \exp\left(\frac{-\tau_{j+1}^2}{\sigma_{x_k}^2}\right) \right]$	$2^{b+1} + 1$	$2^{b} - 1$	$2^{b} + 1$	$2^b$
$\beta_k(i) = \frac{g_{k,b}(i)\sigma_{x,k}^2}{g_{k,b}(i)\sigma_{x,k}^2 + \sigma_{q,k}^2}$	2	1	1	0
$\hat{d}_{k,Q}(i) = \beta_k(i) \mathbf{w}_k^H(i) \mathbf{x}_{k,Q}(i)$	M+1	M-1	0	0
$e_{k,Q}(i) = d_{k,Q}(i) - \hat{d}_{k,Q}(i)$	0	1	0	0
$\mathbf{h}_{k}(i+1) = \mathbf{w}_{k}(i) + \mu_{k}\beta_{k}(i)e_{k,Q}^{*}(i)\mathbf{x}_{k,Q}(i)$	M+2	M	0	0
$\mathbf{w}_k(i+1) = \sum_{l \in \mathcal{N}_k} a_{lk} \mathbf{h}_l(i+1)$	$n_k M$	$n_k M$	0	0
Total (at node k)	$(2+n_k)M + 2^{b+1} + 6$	$(2+n_k)M+2^b$	$2^{b} + 2$	$2^b$

TABLE I
COMPUTATIONAL COMPLEXITY PER TIME INSTANT

the expected value of (9), we have

$$\mathbb{E}[\mathbf{h}_{k}(i)] = [\mathbf{I}_{M} - \mu_{k} g_{k,b}^{2}(i)\beta_{k}(i)\mathbf{R}_{x_{k}} - \mu_{k} g_{k,b}(i)\beta_{k}(i)\mathbf{R}_{q,k}]$$

$$\cdot \mathbb{E}[\mathbf{w}_{k}(i-1)] + \mu g_{k,b}^{2}(i)\mathbf{R}_{x_{k}} \mathbf{w}_{o}^{H}. \tag{11}$$

Substituting the values of  $\mathbf{R}_{x_k}$  and  $\mathbf{R}_{q,k}$  and taking the limit on (11), we obtain

$$\lim_{i \to +\infty} \mathbb{E}[\mathbf{h}_k(i)] = \frac{1}{\beta_k(i)} \frac{g_{k,b}(i)\sigma_{x,k}^2}{g_{k,b}(i)\sigma_{x,k}^2 + \sigma_{g,k}^2} \mathbf{w}_o. \tag{12}$$

We conclude that the solution is unbiased if we choose

$$\beta_k(i) = \frac{g_{k,b}(i)\sigma_{x,k}^2}{g_{k,b}(i)\sigma_{x,k}^2 + \sigma_{q,k}^2}.$$
 (13)

The gradient of  $|e_{k,Q}(i)|^2$  with respect to  $\mathbf{w}_k^H$  is  $\nabla J_k(\mathbf{w}_k(i-1)) = -\frac{g_{k,b}(i)\sigma_{x,k}^2}{g_{k,b}(i)\sigma_{x,k}^2+\sigma_{q,k}^2}\mathbf{x}_{k,Q}(i)e_{k,Q}^*(i)$ . After organizing the terms of the gradient, we obtain the DQA-LMS algorithm:

$$\mathbf{h}_{k}(i) = \mathbf{w}_{k}(i-1) + \mu_{k} \frac{g_{k,b}(i)\sigma_{x,k}^{2}}{g_{k,b}(i)\sigma_{x,k}^{2} + \sigma_{a,k}^{2}} \mathbf{x}_{k,Q}(i)e_{k,Q}^{*}(i),$$

$$\mathbf{w}_k(i) = \sum_{l \in \mathcal{N}_k} a_{lk} \mathbf{h}_l(i), \tag{14}$$

$$e_{k,Q}(i) = d_{k,Q}(i) - \frac{g_{k,b}(i)\sigma_{x,k}^2}{g_{k,b}(i)\sigma_{x,k}^2 + \sigma_{a,k}^2} \mathbf{w}_k^H(i-1)\mathbf{x}_{k,Q}(i) \,,$$

$$g_{k,b}(i) = \frac{1}{\sqrt{\sigma_{x_k}^2}} \sum_{j=0}^{2^{b-1}} \frac{l_j}{\sqrt{\pi}} \left[ \exp\left(\frac{-\tau_j^2}{\sigma_{x_k}^2}\right) - \exp\left(\frac{-\tau_{j+1}^2}{\sigma_{x_k}^2}\right) \right],$$
(15)

and  $\sigma_{x,k}^2 \approx \sigma_{x_{k,Q}}^2 + \rho_{k,b}$ . The scalar  $g_{k,b}$  can be computed offline when  $\mathbf{R}_{x,k}$  is known and wide-sense stationary and must be estimated online when  $\mathbf{R}_{x,k}$  is unknown or non-stationary.

# B. Computational Complexity and Energy Consumption

Table I shows the computational complexity of the DQA-LMS algorithm in terms of the number of multiplications and additions at node k per time instant, where  $n_k$  is the number of neighbor nodes connected to node k. At each time instant, DQA-LMS performs a few more operations ( $\approx O(2^b)$ ) than DLMS. Note that we compute  $g_{k,b}(i)$  online since this is more appropriate for non-stationary input data. However, one can compute  $\mathbf{G}_{k,b}$  offline if an estimate of  $\mathbf{R}_{x_k}$  in (4) is available.

However, the extra complexity of DQA-LMS allows the system to work in a more energy-efficient way. In order to assess the power savings by low resolution quantization, we consider a network with N nodes in which each node uses two ADCs. The power consumption of each ADC is  $P_{ADC}(b) = cB2^b$  [27], where B

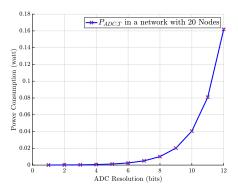


Fig. 2. Power consumption of the ADCs in an adaptive IoT network.

is the bandwidth (related to the sampling rate), b is the number of quantization bits of the ADC, and c is the power consumption per conversion step. Therefore, the total power consumption of the ADCs in the network is

$$P_{ADC,T}(b) = 2NcB2^b$$
 (watts). (16)

Fig. 2 shows an example of the total power consumption of ADCs in a narrowband IoT (NB-IoT) network running diffusion adaptation consisting of 20 nodes with bandwidth  $B=200\,\mathrm{kHz}$  [28] and considering the power consumption per conversion step of each ADC,  $c=494\,\mathrm{fJ}$ , as in [29].

# IV. ANALYSIS OF DQA-LMS

In this section, we find sufficient conditions for all local estimates to converge in the mean to the unknown parameter vector  $\mathbf{w}_o$  by using the evolution of the weight error vectors [4]. Let us consider the global quantities of the network:  $\mathbf{W}_o \triangleq [\mathbf{w}_o, \dots, \mathbf{w}_o]_{(NM \times 1)}$ ,  $\mathbf{d}_Q(i) \triangleq [d_{1,Q}(i), \dots, d_{N,Q}(i)]^T$ ,  $\mathbf{v}_i \triangleq [v_1(i), \dots, v_N(i)]^T$ ,  $\mathbf{X}_Q(i) \triangleq \mathrm{diag}[\mathbf{x}_{1,Q}^T(i), \dots, \mathbf{x}_{N,Q}^T(i)]$ .

Using these quantities, the global form of (1) is given by  $\mathbf{d}_Q(i) = \mathbf{W}_o^H \mathbf{X}_Q(i) + \mathbf{v}(i)$ . Defining  $\mathbf{B}(i)$ ,  $\mathbf{W}(i)$  and  $\mathbf{H}(i)$  as the global quantities for, respectively,  $\beta_k(i)$ ,  $\mathbf{h}_k(i)$  and  $\mathbf{w}_k(i)$ , we can express (14) as

$$\mathbf{H}(i) = \mathbf{W}(i-1) + \mathbf{D}\mathbf{B}(i)\mathbf{X}_{Q}(i)(\mathbf{d}_{Q}(i)$$
$$-\mathbf{B}(i)\mathbf{W}^{H}(i-1)\mathbf{X}_{Q}(i))^{*}, \quad \mathbf{W}(i) = \mathbf{C}\mathbf{H}(i), \quad (17)$$

which can be written in a compact form as

$$\mathbf{W}(i) = \mathbf{C}\mathbf{W}(i-1) + \mathbf{C}\mathbf{D}\mathbf{B}(i)\mathbf{X}_{Q}(i)(\mathbf{d}_{Q}(i)$$
$$-\mathbf{B}(i)\mathbf{W}^{H}(i-1)\mathbf{X}_{Q}(i))^{*}, \tag{18}$$

where  $\mathbf{D} \triangleq \operatorname{diag}\{\mu_1 I_M, \dots, \mu_N I_M\}$  and  $\mathbf{C}$  is an  $MN \times MN$  matrix based on the combination coefficients,  $a_{lk}$ , defined as

$$\mathbf{A} \triangleq [a_{ij}], \qquad \mathbf{C} \triangleq \mathbf{A} \otimes \mathbf{I}_{M}. \tag{19}$$

Using the independence assumption [4] that states that  $\mathbf{x}_{k,Q}(i)$  and  $v_k(i)$  are i.i.d. in time and space with  $\sigma^2_{v,k} = \mathbb{E}[|v_k(i)|^2]$ , and  $v_k(i)$  is independent of  $\mathbf{x}_{k,Q}(i)$ , we define the weight error vector,  $\widetilde{\mathbf{w}}_k(i)$  and its global vector  $\widetilde{\mathbf{w}}(i)$  as

$$\widetilde{\mathbf{W}}(i) \triangleq \mathbf{W}_o - \mathbf{W}(i). \tag{20}$$

Note that using diffusion combination policies for  $a_{lk}$ , we have  $\mathbf{CW}_o = \mathbf{W}_o$  [4]. Subtracting  $\mathbf{W}_o$  from the left-hand side and  $\mathbf{CW}_o$  from the right-hand side of (18), we have

$$\widetilde{\mathbf{W}}(i) = \mathbf{C}\mathbf{W}_o - \mathbf{C}\mathbf{W}(i-1) - \mathbf{C}\mathbf{D}\mathbf{B}(i)\mathbf{X}_Q(i)$$

$$= \mathbf{C}\widetilde{\mathbf{W}}(i-1) - \mathbf{C}\mathbf{D}\mathbf{B}(i)\mathbf{X}_Q(i)$$

$$(\mathbf{X}_Q^*(i)\mathbf{B}(i)^*\widetilde{\mathbf{W}}(i-1) + \mathbf{v}^*(i))$$

$$= \mathbf{C}(\mathbf{I}_{MN} - \mathbf{D}\mathbf{B}(i)\mathbf{X}_Q(i)\mathbf{X}_Q^*(i)\mathbf{B}(i)^*)\widetilde{\mathbf{W}}(i-1)$$

$$- \mathbf{C}\mathbf{D}\mathbf{B}(i)\mathbf{X}_Q(i)\mathbf{v}^*(i). \tag{21}$$

Taking the expectation of both sides of (21), we have

$$\mathbb{E}[\widetilde{\mathbf{W}}(i)] = C(\mathbf{I}_{MN} - \mathbf{D}\mathbf{R}_Q)\mathbb{E}[\widetilde{\mathbf{W}}(i-1)], \tag{22}$$

where  $C \triangleq \mathbb{E}[\mathbf{C}]$ ,  $\mathbf{R}_Q \triangleq \mathrm{diag}\{\mathbf{R}_{1,Q},\dots,\mathbf{R}_{N,Q}\}$  and  $\mathbf{R}_{k,Q} = \mathbb{E}[\mathbf{B}_k(i)\mathbf{x}_{k,Q}^*(i)\mathbf{x}_{k,Q}^*(i)\mathbf{B}_k(i)^*]$ .

To ensure stability of the recursion in (22) with the independence assumption and using combinations that satisfy (2), there exist sufficiently small step-sizes  $\mu_k < \mu_{max}$  such that

$$\|\mathbb{E}[\widetilde{\mathbf{W}}(i)]\|_{bl_{\infty}} \leq \|C\|_{bl_{\infty}}.\|\mathbf{E}_i\|_{bl_{\infty}}.\|\mathbb{E}[\widetilde{\mathbf{w}}(i-1)]\|_{bl_{\infty}}, \qquad (23)$$
 where  $\|.\|_{bl_{\infty}}$  denotes the block maximum norm [30] and  $\mathbf{E}_i = \mathbf{I}_{MN} - \mathbf{D}\mathbf{R}_Q$ . In order for DQA-LMS to converge, we hold (23) such that  $\|\mathbf{E}\|_{bl_{\infty}} < 1$  and  $\|C_i\|_{bl_{\infty}} \leq 1$  for all  $i \geq 0$ . It is proven in [30] that possibly random, time-varying convex combinations generated by ATC or CTA diffusion algorithms ensure  $\|C_i\|_{bl_{\infty}} \leq 1$ . Therefore, to find sufficient conditions on step-sizes, we must have  $\mathbf{I}_{MN} - \mathbf{D}\mathbf{R}_Q < 1$ .

We now employ the eigenvalue decomposition  $\mathbf{R}_{k,Q} = \Phi_{k,Q} \mathbf{\Lambda}_{k,Q} \Phi_{k,Q}^H$ , where  $\mathbf{\Lambda}_{k,Q}$  is an  $M \times M$  diagonal matrix consisting of the eigenvalues  $\{\lambda_{(k,Q)_1}, \dots, \lambda_{(k,Q)_M}\}$  of  $\mathbf{R}_{k,Q}$ , and the matrix  $\Phi_{k,Q}$  is an  $M \times M$  square matrix whose columns are the eigenvectors  $\{\phi_{(k,Q)_1}, \dots, \phi_{(k,Q)_M}\}$  of  $\mathbf{R}_{k,Q}$  associated with these eigenvalues. We define  $\Phi_Q \triangleq \mathrm{diag}\{\Phi_{1,Q}, \dots, \Phi_{N,Q}\}$  and  $\mathbf{\Lambda}_Q \triangleq \mathrm{diag}\{\mathbf{\Lambda}_{1,Q}, \dots, \mathbf{\Lambda}_{N,Q}\}$ . Since  $\mathbf{\Phi}_Q^H \mathbf{D} \mathbf{\Phi}_Q = \mathbf{D}$ , the condition on the step size can be written as  $\|\mathbf{I}_{MN} - \mathbf{D} \mathbf{\Lambda}_Q\|_{\infty} < 1$ , which yields

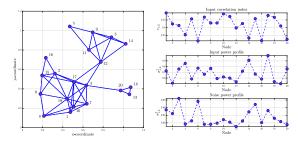
$$\begin{split} \|\mathbf{I}_{MN} - \mathbf{D}\mathbf{\Lambda}_Q\|_{\infty} &= \max_{1 \leq k \leq N} \|\mathbf{I}_M - \mu_k \mathbf{\Lambda}_{k,Q}\| \\ &= \max_{1 \leq k \leq N} \max_{1 \leq m \leq M} |1 - \mu_k \lambda_{(k,Q)_m}| < 1, \end{split}$$

where  $\lambda_{(k,Q)_m}$  is the mth diagonal eigenvalue of  $\mathbf{R}_{k,Q}$ . Therefore, the stability condition for DQA-LMS is given by

$$0 < \mu_k < \frac{2}{\lambda_{\max}(\mathbf{R}_{k,O})}$$
 for all  $k = 1, 2, \dots, N$ . (24)

# V. SIMULATION RESULTS

In this section, we assess the performance of the DQA-LMS algorithm for a parameter estimation problem in an IoT network with N=20 nodes. The impulse response of the unknown system has M=8 taps, is generated randomly and normalized to one. The input signals  $\mathbf{x}_k(i)$  at each node are generated by passing a white Gaussian noise process with variance  $\sigma_{x,k}^2$  through a first order autoregressive model with transfer function  $\frac{1}{1-r_{x,k}z^{-1}}$  where  $r_{x,k} \in (0.3,0.5)$  are the correlation coefficients and quantized using



(a) Distributed network structure (b) Variances and correlation coefficients

Fig. 3. A wireless network with N = 20 nodes.

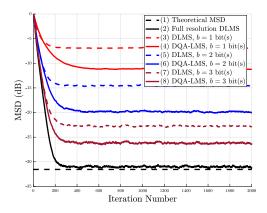


Fig. 4. The MSD curves for the DLMS and DQA-LMS algorithms.

Lloyd-Max quantization scheme to generate  $\mathbf{x}_{k,Q}(i)$ . The noise samples of each node are drawn from a zero mean white Gaussian process with variance  $\sigma_{v,k}^2$ . Fig. 3 plots the network details.

The simulated mean-square deviation (MSD) learning curves are obtained by ensemble averaging over 100 independent trials. We choose the same step sizes for all agents, i.e.,  $\mu_k = 0.05$ . The combining coefficients  $a_{lk}$  are computed by the Metropolis rule. The evolution of the ensemble-average learning curves,  $\frac{1}{N}\mathbb{E}[\|\widetilde{\mathbf{w}}_i\|^2]$ , for the ATC diffusion strategy using different numbers of bits is assessed. The theoretical MSD of the DLMS with the same step size  $\mu$  and the Metropolis rule applied to  $a_{lk}$  is approximated by  $\frac{\mu M}{N^2} \sum_{k=1}^{N} \sigma_{v,k}^2$  [5] and shown by curve 1. Curve 2 shows the standard DLMS performance assuming full resolution ADCs to perform system identification. Curves 3, 5 and 7 show the MSD evolution of the standard DLMS with low resolution signals coarsely quantized with 1, 2 and 3 bits, respectively. Curves 4, 6 and 8 show the MSD performance of the proposed DQA-LMS algorithm that improves the error measurement confronted with coarsely quantized signals. The performance of the proposed DQA-LMS algorithm is closer to the DLMS while it reduces about 90% of the power consumption by ADCs in the network (see Fig. 2).

# VI. CONCLUSION

In this letter, we have proposed an energy-efficient framework for distributed learning and developed the DQA-LMS algorithm using low-resolution ADCs for adaptive IoT networks. DQA-LMS has comparable computational cost to the full-resolution DLMS algorithm while it enormously reduces the power consumption of the ADCs in the network. Simulations have shown the close performance of DQA-LMS to the DLMS algorithm despite dealing with coarsely quantized signals.

# REFERENCES

- J. B. Predd, S. B. Kulkarni, and H. V. Poor, "Distributed learning in wireless sensor networks," *IEEE Signal Process. Mag.*, vol. 23, no. 4, pp. 56–69, Jul. 2006.
- [2] M. M. Rana, W. Xiang, and E. Wang, "IOT-based state estimation for microgrids," *IEEE Internet Things J.*, vol. 5, no. 2, pp. 1345–1346, Apr. 2018.
- [3] R. Olfati-Saber, J. A. Fax, and R. M. Murray, "Consensus and cooperation in networked multi-agent systems," *Proc. IEEE*, vol. 95, no. 1, pp. 215–233, Jan. 2007.
- [4] C. G. Lopes and A. H. Sayed, "Diffusion least-mean squares over adaptive networks: Formulation and performance analysis," *IEEE Trans. Signal Process.*, vol. 56, no. 7, pp. 3122–3136, Jul. 2008.
- [5] A. H. Sayed, S.-Y. Tu, J. Chen, X. Zhao, and Z. J. Towfic, "Diffusion strategies for adaptation and learning over networks: An examination of distributed strategies and network behavior," *IEEE Signal Process. Mag.*, vol. 30, no. 3, pp. 155–171, May 2013.
- [6] S. Xu, R. C. de Lamare, and H. V. Poor, "Distributed estimation over sensor networks based on distributed conjugate gradient strategies," *IET Signal Process.*, vol. 10, no. 3, pp. 291–301, 2016.
- [7] T. G. Miller, S. Xu, R. C. de Lamare, and H. V. Poor, "Distributed spectrum estimation based on alternating mixed discrete-continuous adaptation," *IEEE Signal Process. Lett.*, vol. 23, no. 4, pp. 551–555, Apr. 2016.
- [8] S. Xu, R. C. de Lamare, and H. V. Poor, "Distributed compressed estimation based on compressive sensing," *IEEE Signal Process. Lett.*, vol. 22, no. 9, pp. 1311–1315, Sep. 2015.
- [9] T. G. Miller, S. Xu, R. C. de Lamare, V. H. Nascimento, and Y. Zakharov, "Sparsity-aware distributed conjugate gradient algorithms for parameter estimation over sensor networks," in *Proc. 49th Asilomar Conf. Signals*, Syst. Comput., 2015, pp. 1556–1560.
- [10] S. Xu, R. C. de Lamare, and H. V. Poor, "Adaptive link selection algorithms for distributed estimation," *EURASIP J. Adv. Signal Process.*, vol. 2015, no. 1, pp. 1–22, 2015.
- [11] S. Zhang and W. X. Zheng, "Distributed separated-decorrelation LMS algorithms over sensor networks with noisy inputs," *IEEE Trans. Signal Process.*, vol. 68, pp. 4163–4177, 2020.
- [12] Y. Yu, H. Zhao, R. C. de Lamare, Y. Zakharov, and L. Lu, "Robust distributed diffusion recursive least squares algorithms with side information for adaptive networks," *IEEE Trans. Signal Process.*, vol. 67, no. 6, pp. 1566–1581, Mar. 2019.
- [13] C. Han, J. M. Jornet, E. Fadel, and I. F. Akyildiz, "A cross-layer communication module for the Internet of Things," *Comput. Netw.*, vol. 57, no. 3, pp. 622–633, 2013.
- [14] I. Utlu, O. F. Kilic, and S. S. Kozat, "Resource-aware event triggered distributed estimation over adaptive networks," *Digit. Signal Process.*, vol. 68, pp. 127–137, 2017.

- [15] A. Mezghani and J. A. Nossek, "Power efficiency in communication systems from a circuit perspective," in *Proc. IEEE Int. Symp. Circuits* Syst., 2011, pp. 1896–1899.
- [16] R. H. Walden, "Analog-to-digital converter survey and analysis," *IEEE J. Sel. Areas Commun.*, vol. 17, no. 4, pp. 539–550, Apr. 1999.
- [17] L. T. N. Landau and R. C. de Lamare, "Branch-and-bound precoding for multiuser MIMO systems with 1-bit quantization," *IEEE Wireless Commun. Lett.*, vol. 6, no. 6, pp. 770–773, Dec. 2017.
- [18] S. Jacobsson, G. Durisi, M. Coldrey, U. Gustavsson, and C. Studer, "Throughput analysis of massive MIMO uplink with low-resolution ades," *IEEE Trans. Wireless Commun.*, vol. 16, no. 6, pp. 4038–4051, Jun. 2017.
- [19] A. Mezghani, M.-S. Khoufi, and J. A. Nossek, "A modified MMSE receiver for quantized MIMO systems," in *Proc. ITG/IEEE WSA*, Vienna, Austria, 2007, pp. 1–5.
- [20] L. T. N. Landau, M. Dörpinghaus, R. C. de Lamare, and G. P. Fettweis, "Achievable rate with 1-bit quantization and oversampling using continuous phase modulation-based sequences," *IEEE Trans. Wireless Commun.*, vol. 17, no. 10, pp. 7080–7095, Oct. 2018.
- [21] Z. Shao, R. C. de Lamare, and L. T. N. Landau, "Iterative detection and decoding for large-scale multiple-antenna systems with 1-bit ADCS," *IEEE Wireless Commun. Lett.*, vol. 7, no. 3, pp. 476–479, Jun. 2018.
- [22] Z. Shao, L. Landau, and R. C. de Lamare, "Adaptive RLS channel estimation and SIC for large-scale antenna systems with 1-bit ADCs," in *Proc. 22nd Int. ITG Workshop Smart Antennas.*, 2018, pp. 1–4.
- [23] J. J. Bussgang, "Crosscorrelation functions of amplitude-distorted Gaussian signals," Research Lab. Electron MIT, Cambridge, MA, USA, Tech. Rep. 216, 1952.
- [24] S. Lloyd, "Least squares quantization in PCM," *IEEE Trans. Inf. Theory*, vol. IT-28, no. 2, pp. 129–137, Mar. 1982.
- [25] J. Max, "Quantizing for minimum distortion," IRE Trans. Inf. Theory, vol. 6, no. 1, pp. 7–12, 1960.
- [26] Y. Li, C. Tao, G. Seco-Granados, A. Mezghani, A. L. Swindlehurst, and L. Liu, "Channel estimation and performance analysis of one-bit massive MIMO systems," *IEEE Trans. Signal Process.*, vol. 65, no. 15, pp. 4075–4089, Aug. 2017.
- [27] O. Orhan, E. Erkip, and S. Rangan, "Low power analog-to-digital conversion in millimeter wave systems: Impact of resolution and bandwidth on performance," in *Proc. Inf. Theory Appl. Workshop.*, 2015, pp. 191–198.
- [28] R. Ratasuk, B. Vejlgaard, N. Mangalvedhe, and A. Ghosh, "NB-IOT system for M2M communication," in *Proc. IEEE Wireless Commun. Netw.* Conf., 2016, pp. 1–5.
- [29] H. Chung, A. Rylyakov, Z. T. Deniz, J. Bulzacchelli, G. Wei, and D. Friedman, "A 7.5-GS/s 3.8-ENOB 52-mW flash ADC with clock duty cycle control in 65 nm CMOS," in *Proc. Symp. VLSI Circuits*, 2009, pp. 268–269.
- [30] N. Takahashi, I. Yamada, and A. H. Sayed, "Diffusion least-mean squares with adaptive combiners: Formulation and performance analysis," *IEEE Trans. Signal Process.*, vol. 58, no. 9, pp. 4795–4810, Sep. 2010.