

北京航空航天大学

2013—2014 学年 第一学期期末

Modern Control Engineering

考试 A 卷

班 级 _____ 学 号 _____

姓 名 _____ 成 绩 _____

2014 年 1 月 15 日

班号_____ 学号_____ 姓名_____ 成绩_____

Examination Questions

1. (15 points) A network is shown in Figure 1:

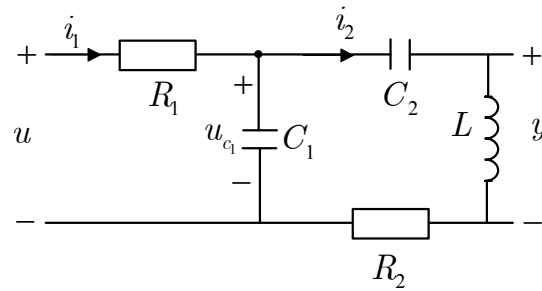


Figure 1

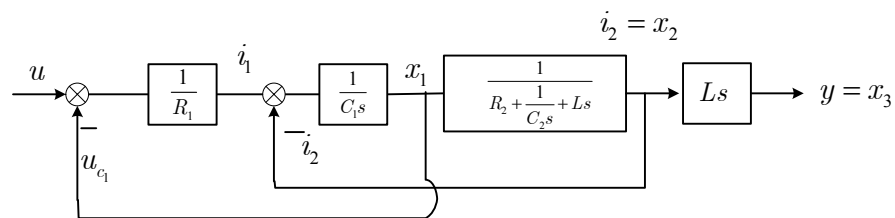
where u is the input signal, y , the voltage across the inductor L , is the output signal, and i_1 and i_2 are the currents passing through the resistor R_1 and the capacitor C_2 , respectively.

(1). Draw the block diagram according to Figure 1;

(2). If $R_1 = R_2 = 1\Omega$, $C_1 = C_2 = 1F$ and $L = 1H$, find the transfer function $Y(s)/U(s)$.

Solution:

(1) By applying the concept of complex impedance, it is easy to obtain the diagram of the network:



(2) Letting $R_1 = R_2 = 1\Omega$, $C_1 = C_2 = 1F$ and $L = 1H$, the transfer function from the above diagram is

$$G(s) = \frac{s^2}{s^3 + 2s^2 + 3s + 1}$$

2. (15 points, 5 points each)

(1). Let the closed loop transfer function of a system be

$$\frac{C(s)}{R(s)} = \frac{1}{(s+1)(s+2)}$$

Determine its unit step response $c(t)$.

(2). Let the transfer function of a system be

$$\frac{C(s)}{R(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}, \quad (\zeta \geq 0, \omega_n > 0)$$

where $0 < \zeta < 1$. Write the response $c(t)$ when the input signal is a unit-step function.

(3). The block diagram of a system is shown in Figure 2. Design a feedforward compensator $G_c(s)$, such that the steady-state error e_{ss} of the system is zero when the input signal is $r(t) = t, t \geq 0$ (The error is defined as $e = r - c$).

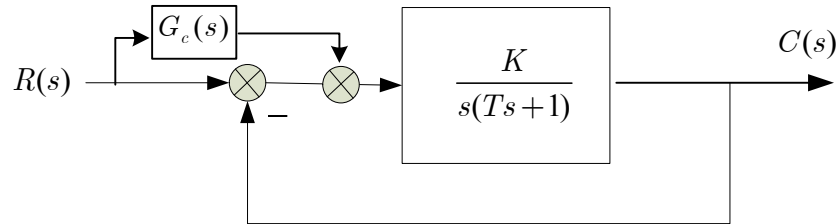


Figure 2

4. (15 points, 5 points each).

(1). The system transfer function is

$$\frac{Y(s)}{X(s)} = G(s) = \frac{4(s+1)}{(s+2)(s+3)}$$

If the input signal $x(t) = A \sin \omega t$, determine the system output $y(t)$ when the

system reaches the steady-state.

- (2). The *Nyquist* curves of two open-loop transfer functions are shown in Figure 4, where P denotes the number of unstable open-loop poles. Determine the stability of the two systems by using *Nyquist* stability criterion.

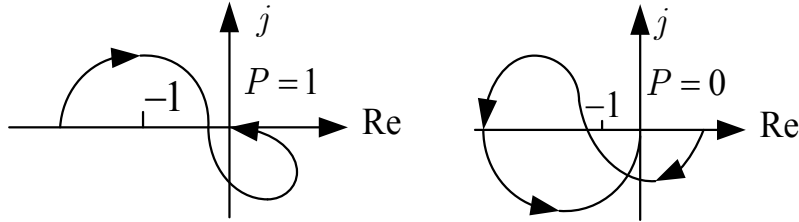


Figure 4

3. **(15 points)** The block diagram of a system is shown in Figure 5:

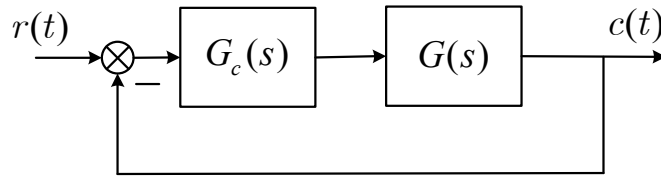


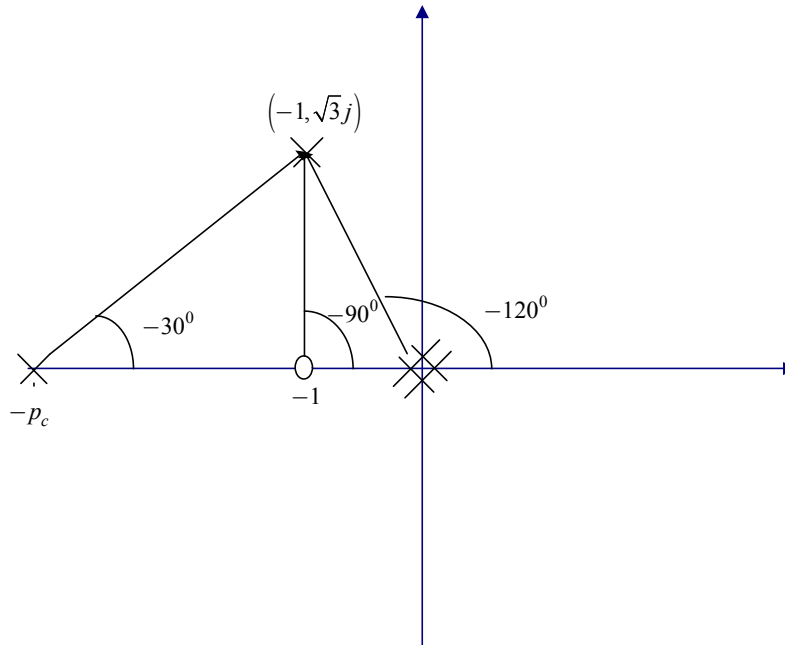
Figure 5

where $G(s) = \frac{1}{s^2}$ and the transfer function of the compensator is

$$G_c(s) = \frac{K_c^*(s+1)}{(s+p_c)}$$

- (1). By using root locus method to determine the values of K_c^* and p_c , so that a pair of closed-loop poles are located at $-1 \pm \sqrt{3}j$ after the compensation (hint: using angle and magnitude conditions);
- (2). With the determined p_c , draw the root loci as $K_c^* : 0 \rightarrow +\infty$.

Solution: (1) From the phase angle condition, we obtain that $p_c = 4$.



由此图易得 $p_c = 4$ 。

根据模值方程可得 $K_c^* = 8$ 。

(2) 开环传递函数极点 0,0,-4, 零点-1;

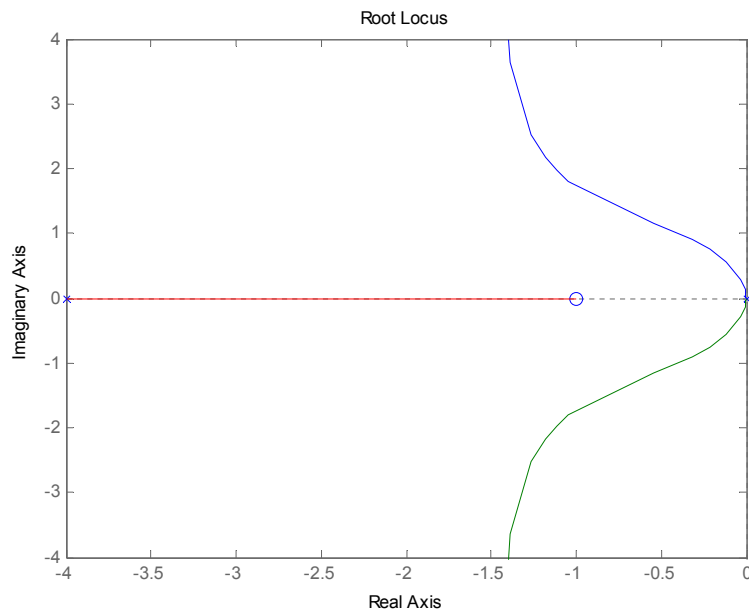
实轴上根轨迹 $(-4, -1)$;

渐近线 $\varphi_a = \frac{\pi}{2}, -\frac{\pi}{2}$, $\sigma_a = -1.5$;

起始角 $\frac{\pi}{2}, -\frac{\pi}{2}, 0$;

分离点坐标 $\frac{1}{d+1} = \frac{2}{d} + \frac{1}{d+4}$, 舍去

由以上法则可绘制出如下根轨迹



5. (10 points) The open-loop transfer function of a unity-feedback system is

$$G(s) = \frac{3}{s(s-1)}$$

Draw its *Nyquist* curve and determine

closed-loop stability by using *Nyquist* stability criterion.

6. (15 points) The block diagram of a system is shown in Figure 6 with $r(t) = t$ ($t \geq 0$) ($e(t) = r(t) - c(t)$).

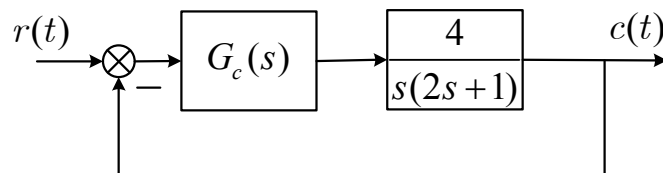


Figure 6

Design the compensator $G_c(s)$ such that there is no steady-state error.

7. (15 points) The open-loop transfer function of a unity-feedback system is

$$G_p(s) = \frac{2000}{s(s+2)(s+20)}$$

The asymptotic Bode magnitude curve of $20\log|G_c(j\omega)G_p(j\omega)|$ is shown in Figure 7, where $G_c(s)$ is a series compensator.

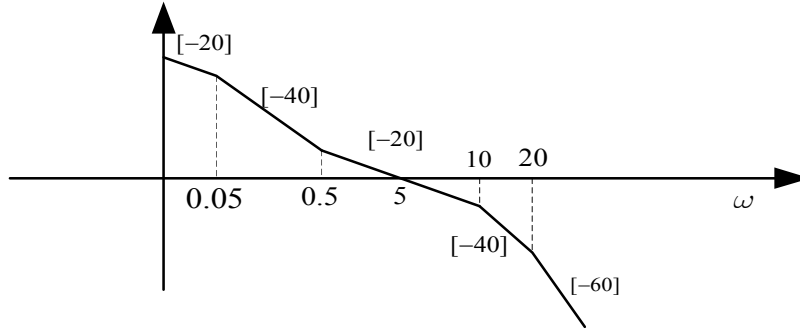


Figure 7: The asymptotic Bode magnitude curve of $20\log|G_c(j\omega)G_p(j\omega)|$

- (1). Determine the transfer function of $G_c(s)$. What kind of a compensator is it?
- (2). Draw the asymptotic Bode magnitude curve of $G_c(s)$ for which the corner frequencies and slopes are required;
- (3). Calculate the phase margin of the compensated system.

解: $G_p(s) = \frac{50}{s\left(\frac{1}{2}s+1\right)\left(\frac{1}{20}s+1\right)}$

$$(1) \quad G_e(s) = \frac{K\left(\frac{1}{0.5}s+1\right)}{s\left(\frac{1}{0.05}s+1\right)\left(\frac{1}{10}s+1\right)\left(\frac{1}{20}s+1\right)}$$

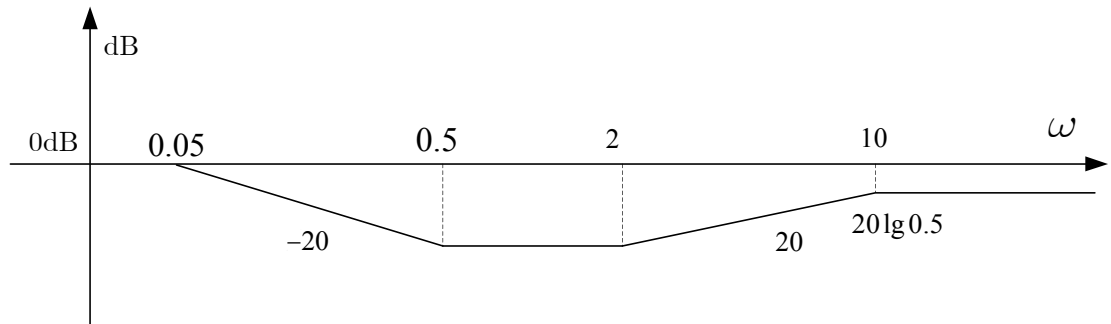
$$\text{由 } \frac{K \frac{1}{0.5}}{5 \times \frac{1}{0.05}} = 1 \Rightarrow K = 50$$

$$\text{故 } G_c(s) = \frac{G_e(s)}{G_p(s)} = \frac{\left(\frac{1}{0.5}s+1\right)\left(\frac{1}{2}s+1\right)}{\left(\frac{1}{0.05}s+1\right)\left(\frac{1}{10}s+1\right)}$$

滞后超前校正。

(1) 幅频特性如下图所示。

$$\text{当 } \omega > 10 \text{ 时, 由 } |G_c| = \frac{\frac{1}{0.5} \times \frac{1}{2}}{\frac{1}{0.05} \times \frac{1}{10}} = 0.5$$



$$(3) \quad \gamma = 180^\circ - 90^\circ + \arctan \frac{5}{0.5} - \arctan \frac{5}{0.05} - \arctan \frac{5}{10} - \arctan \frac{5}{20} = 44^\circ$$