## **Automatic Control**

# Loop shaping design of feedback control systems

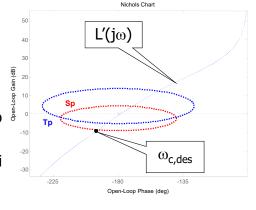
Part I: Lead network

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#### C<sub>T</sub>(s) design procedure: preliminary considerations

After the first step of the design procedure, the steady state controller is obtained  $\rightarrow C_{ss}(s)$ . Before going on with the design of the transient controller  $C_T(s)$ , the following preliminary steps need to be performed.

- 1. consider the loop function obtained after the steady state design  $L'(s) = C_{ss}(s)G(s)$ .
- 2. plot the frequency response of  $L'(j\omega)$  on the Nichols plane and mark the point corresponding to  $\omega_{c,des}$ .
- 3. plot the constant magnitude loci  $T_p$  and  $S_p$  obtained by transient requirements analysis.



#### Time domain requirements translation: resume

The properties of the 2<sup>nd</sup> order prototype model

$$T(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

allowed the translation of the time domain requirements  $\hat{s}$ ,  $t_r$ ,  $t_{s,\alpha\%}$  into the relevant indices of the frequency response of the functions T(s), S(s) and L(s).

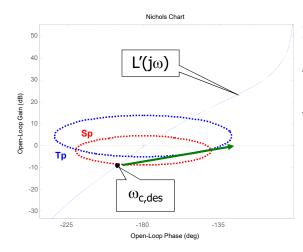
$$\hat{S} \rightarrow \begin{cases} T_p \rightarrow \text{resonant peak of } |T(j\omega)| \\ S_p \rightarrow \text{resonant peak of } |S(j\omega)| \end{cases}$$

$$\left. \begin{array}{c} t_r \\ t_{s,\alpha}\% \end{array} \right\} \rightarrow \omega_{c,des} \rightarrow \text{crossover frequency of } |L(j\omega)|$$

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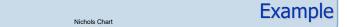
#### Example

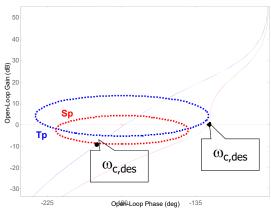


In this case, in order to

- make the course of L'(j $\omega$ ), tangent to the loci  $T_p$  and  $S_p$ ,
- $\bullet \ \ \, \text{make} \,\, \omega_{\text{c,des}} \,\, \text{cross-over} \\ \text{frequency} \\$

phase lead and magnitude increase actions are required in the middle frequency range (i.e. in a suitable neighborhood of  $\omega_{c,des}$ )



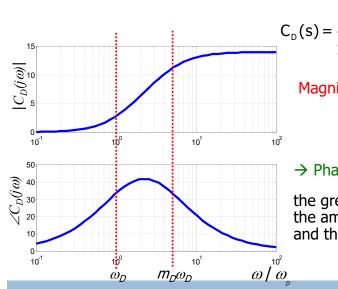


In order to achieve the desired performance, the frequency response of the loop function L'(s) needs to be suitably "shaped"  $\rightarrow$ loop shaping design procedure

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### The lead network: frequency response



 $C_{D}(s) = \frac{1 + \frac{s}{\omega_{D}}}{1 + \frac{s}{m_{D}\omega_{D}}}, \omega_{D} > 0, m_{D} > 1$ 

Magnitude increase

→ Phase lead ←

the greater is m<sub>D</sub>, the larger is the amount of the phase lead and the magnitude increase

The lead network

The just introduced example motivates the use of the

lead network 
$$\rightarrow C_D(s) = \frac{1 + \frac{s}{\omega_D}}{1 + \frac{s}{m_D \omega_D}}, \omega_D > 0, m_D > 1$$

A lead network is described by a proper tf with

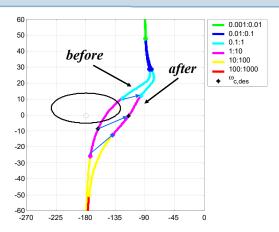
- a real negative zero at − ω<sub>D</sub>
- a real negative pole at m<sub>D</sub> ω<sub>D</sub>

 $\lim_{s\to 0} C_D(s) = \lim_{s\to 0} \frac{1 + \frac{s}{\omega_D}}{1 + \frac{s}{\omega_D}} = 1$ Note also that

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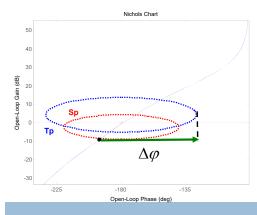
#### The lead network: effects



In the Nichols plane, phase lead and magnitude increase introduced by a lead network produce an oblique shift of the loop function frequency response in the frequency range of interest.

#### The lead network: basic guidelines for design

$$C_{D}(s) = \frac{1 + \frac{s}{\omega_{D}}}{1 + \frac{s}{m_{D}\omega_{D}}}, \omega_{D} > 0, m_{D} > 1$$



- Quantify the amount of the phase lead  $\Delta \varphi$  needed at  $\omega_{\text{c,des}}$ in order to shift the value of  $\angle \text{L}'(j\omega_{\text{c,des}})$  outside the "influence" of the constant magnitude loci  $T_n$  and  $S_n$ .
- $m_D$  is choosen on the basis of the required value of  $\Delta \varphi$ .
- $\omega_D$  is fixed to obtain that the phase lead  $\Delta \varphi$  occurs exactly at  $\omega_{c,des}$ .
- magnitude adjustments (if needed) are obtained in a successive step.
- a systematic procedure for the choice of  $m_D$  and  $\omega_D$  can be established using the universal lead network diagrams →

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ω/ω

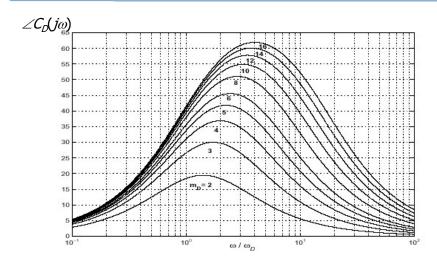
The lead network: magnitude design diagram

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 $|C_D(j\omega)|_{dB}$ 

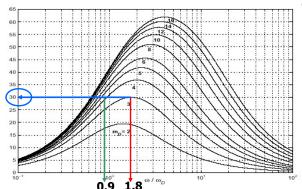
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### The lead network: phase design diagram



# The lead network: design

Example: suppose that a phase lead of  $\Delta \varphi = 30^{\circ}$  is required at  $\omega_{c,des} = 3 \ rad/s$ 



different choices can be made

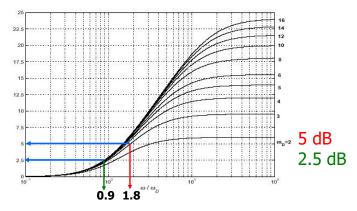
$$m_D = 3$$
;  $\omega_{norm} = \omega/\omega_D = 1.8$   
 $(\omega/\omega_D)|_{\omega=\omega_C,des} = 1.8$   
 $\omega_D = \omega_{c,des}/1.8 = 1.67$   
rad/s

$$m_D$$
= 4;  $\omega_{norm} = \omega/\omega_D$  = 0.9  
 $(\omega/\omega_D)|_{\omega=\omega_C,des}$  = 0.9  
 $\omega_D$ =  $\omega_{c,des}/0.9$ =3.33  
rad/s

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#### The lead network: design

The corresponding magnitude increases are



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#### Lead network: design example 2

A plant to be controlled is described by the following transfer function

 $G(s) = \frac{s+1}{s^2(s-1)}$ 

design a cascade controller C(s) in order to satisfy the following requirements.

$$|e_r^{\infty}| = 0$$
,  $r(t) = 2\varepsilon(t)$ ,  $|y_{d_a}^{\infty}| \le 0.1$ ,  $|q_a(t)| = \varepsilon(t) \rightarrow C_{SS}(s) = 10$ 

• 
$$\hat{s} \le 25\% \rightarrow T_p = 2.67dB, S_p = 4.35dB$$

• 
$$t_r \le 0.1 \text{ s}, t_{s,1\%} \le 0.7 \text{ s} \rightarrow \omega_{c,des} = 18 \text{ rad / s}$$

#### Lead network: design example 1

A plant to be controlled is described by the following transfer function

$$G(s) = \frac{2}{(1+0.2s)(1+0.1s)}$$

design a cascade controller C(s) in order to satisfy the following requirements.

• 
$$\left| \mathbf{e}_{r}^{\infty} \right| \leq 0.1$$
,  $r(t) = t\varepsilon(t) \rightarrow C_{SS}(s) = \frac{5}{s}$ 

• 
$$\hat{s} \le 30\% \rightarrow T_p = 3.67 dB, S_p = 5.1 dB$$

• 
$$t_r \le 0.3 s \rightarrow \omega_{c,des} = 7 \text{ rad/s}$$

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#### Multiple lead network

• In principle, the maximum phase lead that can be introduced by a lead network is 90° corresponding to the ideal case  $m_D \rightarrow \infty$ ; in this case, the lead network tf degenerates in the non-proper form:

$$C_{D}(s) = \frac{1 + \frac{s}{\omega_{D}}}{1 + \frac{s}{m_{D}\omega_{D}}} \xrightarrow{m_{D} \to \infty} 1 + \frac{s}{\omega_{D}}$$

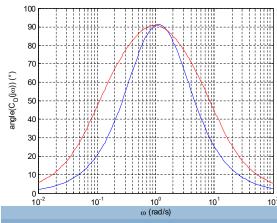
• In practice, it is suggested to use a multiple lead network (e.g. double) when the required phase lead is greater than 60°:

$$C_{D}(s) = \frac{1 + \frac{s}{\omega_{D1}}}{1 + \frac{s}{m_{D1}\omega_{D1}}} \frac{1 + \frac{s}{\omega_{D2}}}{1 + \frac{s}{m_{D2}\omega_{D2}}}$$

### Multiple lead network

$$C_{D}(s) = \frac{1 + \frac{s}{\omega_{D1}}}{1 + \frac{s}{m_{D1}\omega_{D1}}} \cdot \frac{1 + \frac{s}{\omega_{D2}}}{1 + \frac{s}{m_{D2}\omega_{D2}}}$$

Parameters  $m_{D1}$  and  $m_{D2}$  of a double lead network are chosen in order to introduce the required phase increase at the desired cross-over frequency



#### Example:

$$\Delta \varphi = 90^{\circ} @ 1 \text{ rad/s}$$

$$m_{_{D1}}=m_{_{D2}}=6\,$$

$$\omega_{\mathsf{norm},1} = \omega_{\mathsf{norm},2} = 2.2$$

$$m_{_{D1}} = m_{_{D2}} = 12$$

$$\omega_{\mathsf{norm},1} = 10, \omega_{\mathsf{norm},2} = 1.3$$

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