

Extracting single dispersion Cole–Cole impedance model parameters using an integrator setup

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Abstract In this letter, we report an integrator-based setup to extract the four parameters that characterize a single-dispersion Cole–Cole impedance model without direct measurement of the complex impedance. Experimental results in the range 100 Hz–1 MHz using apple and plumb fruits are given and results are compared with numerical simulations of the acquired model.

Keywords Cole–Cole model · Bioelectrical impedance · Biomedical instrumentation · Circuit theory · Bioimpedance

1 Introduction

The Cole–Cole impedance model [1, 2] is widely used for characterizing the electrochemical properties of biological tissues [3]. The model comprises three hypothetical circuit elements: a low frequency resistor R_0 , a high frequency resistor R_∞ and a Constant Phase Element (CPE), arranged as shown in the feedback path of the op amp in Fig. 1. The CPE is also known as the fractional capacitor [4] and its impedance is $Z_{CPE} = 1/(j\omega C)^\alpha$ where C is the capacitance

and α is its order ($0 < \alpha \leq 1$). The Cole–Cole impedance is given by

$$Z = R_\infty + \frac{R_0 - R_\infty}{1 + (j\omega\tau)^\alpha} = Z' + jZ'' \quad (1)$$

and $(j\omega)^\alpha = \omega^\alpha [\cos(\alpha\pi/2) + j\sin(\alpha\pi/2)]$.

To characterize a particular tissue [5], it is required to find the values of the four parameters (R_0 , R_∞ , τ , α). Note that $1/\tau$ is known as the characteristic frequency of the tissue where $\tau = [(R_0 - R_\infty)C]^{1/\alpha}$. To extract these values, an impedance analyzer is required to measure the impedance of the tissue under consideration and a plot is then constructed relating the imaginary part of the impedance Z'' and the real part Z' . Usually, least squares regression is applied to the acquired data points and a circular arc is obtained, as shown in Fig. 2, from which R_0 and R_∞ can be directly found. The angle ϕ_{CPE} (see Fig. 2) is equal to $\alpha\pi/2$ and hence measuring ϕ_{CPE} enables calculating α . Finally, the frequency at which Z'' has its maximum value is measured. This frequency is the characteristic tissue frequency and is therefore equal to $1/\tau$.

The impedance measuring technique thus far has been employed in all tissue characterization applications using the Cole–Cole model [6–8] although it is clear that acquiring Z' and Z'' requires the use of an expensive impedance analyzer and post processing of the data to draw the circular arc. Cheap alternatives include using data acquisition cards and custom software modules which implement the necessary signal processing [9].

In this work we introduce a simple integrator-based measurement technique which enables extracting the values of the parameter set (R_0 , R_∞ , τ , α) without having to measure the impedance. We apply the developed theory to extract the tissue properties of two apples and a plumb fruit. The simplicity of the proposed method is evident

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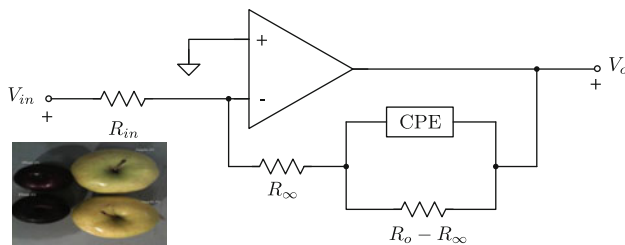


Fig. 1 Integrator setup for extracting the Cole–Cole model parameters for the fruits shown in the inset

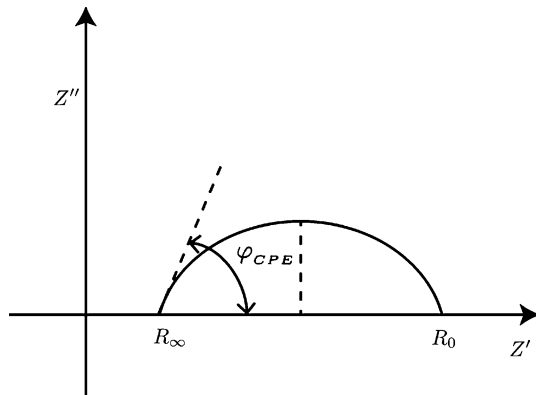


Fig. 2 Impedance loci used to extract the parameter set (\$R_0, R_{\infty}, \tau, \alpha\$) in classical impedance-based analysis

when compared to the impedance measuring technique used in [10] to model several fruits and vegetables.

2 Integrator-based setup

Figure 1 shows a simple op amp inverting integrator with an input resistance \$R_{in}\$. The Cole impedance is assumed in the feedback path and hence it can be shown that the transfer function of this filter is

$$T(s) = \frac{V_o}{V_i} = -\frac{G_1 + G_2(\tau s)^\alpha}{1 + (\tau s)^\alpha} \quad (2)$$

where \$G_1 = R_0/R_{in}\$, \$G_2 = R_{\infty}/R_{in}\$ and \$s = j\omega\$. We then note the following:

- at \$\omega = 0 \rightarrow |T(j\omega)| = G_1\$. Hence, measuring the DC gain of the transfer function and knowing \$R_{in}\$ yields \$R_0\$.
- at \$\omega = \infty \rightarrow |T(j\omega)| = G_2\$. Hence, measuring the high frequency gain and knowing \$R_{in}\$ yields \$R_{\infty}\$.
- the \$-3\$ dB point at which \$|T(j\omega_{3dB})| = G_1/\sqrt{2}\$ is given by

$$\omega_{-3dB} = \frac{1}{\tau} \left[a \left(\sqrt{\frac{b + \cos(\alpha\pi)}{2}} - \cos\left(\frac{\alpha\pi}{2}\right) \right) \right]^{1/\alpha} = \frac{f_1(\alpha)}{\tau} \quad (3a)$$

$$\text{with } a = \frac{G_1(G_1 - 2G_2)}{G_1^2 - 2G_2^2} \quad \text{and} \quad b = \frac{G_1(3G_1 - 4G_2)}{(G_1 - 2G_2)^2} \quad (3b)$$

- the phase angle \$\angle T(j\omega)\$ always exhibits a minimum value at a frequency \$\omega_{\phi_{min}}\$ given by

$$\omega_{\phi_{min}} = \frac{1}{\tau} \left(\sqrt{\frac{G_1}{G_2}} \right)^{1/\alpha} = \frac{f_2(\alpha)}{\tau} \quad (4)$$

and at this frequency, the magnitude \$|T(j\omega_{\phi_{min}})|\$ is uniquely equal to \$\sqrt{G_1 G_2}\$. Therefore, without plotting the phase of the transfer function, it is possible to locate \$\omega_{\phi_{min}}\$ from the magnitude response.

- from (3a, 3b) and (4), it is seen that the ratio \$\omega_{-3dB}/\omega_{\phi_{min}} = f_1(\alpha)/f_2(\alpha)\$ is independent of \$\tau\$ and a function only of \$\alpha\$. Therefore assuming \$\omega_{-3dB}/\omega_{\phi_{min}} = p\$, \$\alpha\$ can be found by numerically solving the equation

$$\left(\frac{\sqrt{G_1 G_2} (G_1 - 2G_2)}{(G_1^2 - 2G_2^2)} \right) \times \left(\sqrt{\frac{b + \cos(\alpha\pi)}{2}} - \cos\left(\frac{\alpha\pi}{2}\right) \right) = p^\alpha \quad (5)$$

Knowing \$\alpha\$, \$\tau\$ could then be found from (4).

3 Experimental validation

Figure 1 was constructed using an OP27 op amp and \$R_{in} = 10\$ k\$\Omega\$ to extract the Cole model parameters for two apples and two plumb fruits in the range 100 Hz–1 MHz. Figure 3(a) shows the measured magnitude and phase responses of the integrator transfer function when an apple is used while Fig. 3(b) shows the same responses for the plumb. We have plotted the phase response in order to show the existence of a minimum phase point, as theoretically predicted. However, the phase response is *not needed* to extract any of the model parameters.

From Fig. 3(a) we measure a DC gain of 8.711 dB which yields \$R_0 = 27.261\$ k\$\Omega\$ and a high frequency gain of \$-14.964\$ dB yielding \$R_{\infty} = 1.786\$ k\$\Omega\$. The \$\omega_{-3dB}\$ frequency was located at 1.674 kHz while \$\omega_{\phi_{min}}\$ was located at 19.189 kHz corresponding to a magnitude of \$-3.127\$ dB. Thus \$\omega_{-3dB}/\omega_{\phi_{min}} = 0.087\$ and numerically solving (5) then yields \$\alpha = 0.71\$ and hence \$\tau = 56.47\$ \$\mu\$s. The Matlab magnitude and phase response simulated using these extracted parameters are also plotted in Fig. 3(a) (dashed lines) showing very good agreement with the experimental measurements within the limits of accuracy of the Cole model itself [2]. Measurements were repeated on another apple, and are given in Table 1 where the extracted capacitance \$C\$ is also given.

Fig. 3 Experimental (*solid*) and simulated (*dashed*) transfer function magnitude and phase responses using **a** an apple fruit and **b** a plumb fruit in the integrator of Fig. 1

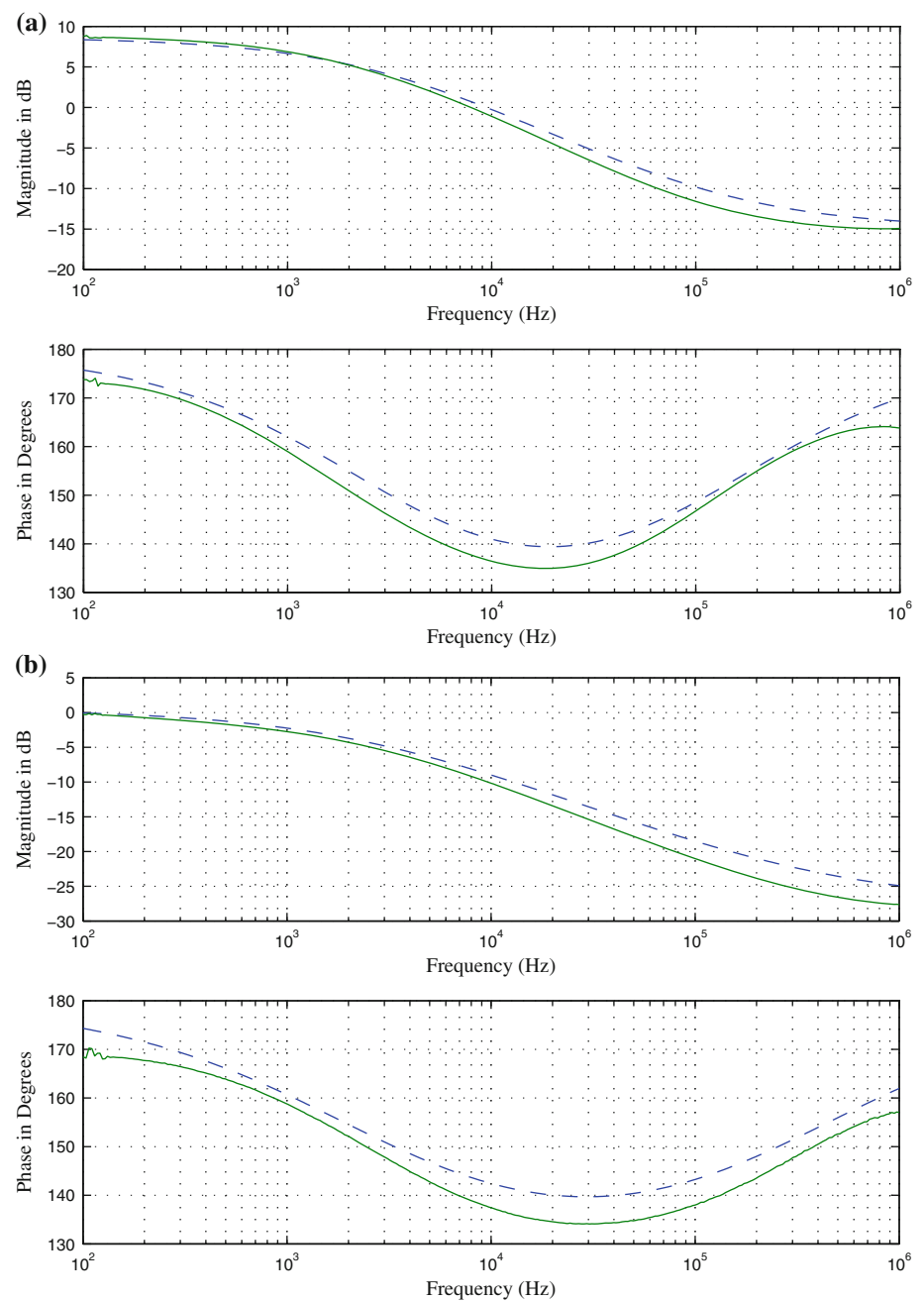


Table 1 Experimentally extracted Cole model parameters (resistors in k Ω)

	R_0	R_∞	α	τ (μ s)	C (nF)
Apple 1	24.141	1.95	0.696	58.56	51.02
Apple 2	27.261	1.786	0.71	56.47	37.63
Plumb 1	10.655	0.424	0.64	67.34	208.5
Plumb 2	15.795	0.408	0.698	69.78	81.12

From Fig. 3(b), obtained using a plumb, the DC gain, high frequency gain, $\omega_{-3\text{dB}}$ and $\omega_{\phi\text{min}}$ were found to be 0.551 dB, -27.445 dB, 1.127 kHz and 29.274 kHz

respectively leading to a parameter set (R_0 , R_∞ , α , τ) = 10.655 k Ω , 424.38 Ω , 0.64, 67.37 μ s. The experimental data is also compared with the numerical simulation using the extracted parameter values.

4 Conclusion

In an integrator setup that incorporates a tissue, and by plotting only the magnitude response, all parameters of the Cole–Cole model which characterize this tissue can be extracted without need for an impedance analyzer. The

only computational overhead is the need to numerically solve (5) for α . To the best of the authors' knowledge, this is the simplest possible method yet reported. It can also be automated using a microcontroller based embedded system.

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