

# IRAMS10UP60B \*\*MOTION\*\*\* Series 10A, 600V with Internal Shunt Resistor

Integrated Power Hybrid IC for Appliance Motor Drive Applications.

#### Description

International Rectifier's IRAMS10UP60B is an Integrated Power Module developed and optimized for electronic motor control in appliance applications such as washing machines and refrigerators. Plug N Drive technology offers an extremely compact, high performance AC motor-driver in a single isolated package for a very simple design. An internal shunt is also included and offers easy current feedback and overcurrent monitor for precise and safe operation. A built-in temperature monitor and over-current protection, along with the short-circuit rated IGBTs and integrated under-voltage lockout function, deliver high level of protection and fail-safe operation. The integration of the bootstrap diodes for the high-side driver section, and the single polarity power supply required to drive the internal circuitry, simplify the utilization of the module and deliver further cost reduction advantages.

#### **Features**

- Internal Shunt Resistor
- Integrated Gate Drivers and Bootstrap Diodes
- Temperature Monitor
- Fully Isolated Package
- Low V<sub>CE(on)</sub> Non Punch Through IGBT Technology
- Undervoltage lockout for all channels
- Matched propagation delay for all channels
- Schmitt-triggered input logic
- Cross-conduction prevention logic
- Lower di/dt gate driver for better noise immunity
- Motor Power range 0.4~0.75kW / 85~253 Vac
- Isolation 2000V<sub>RMS</sub>/1min and CTI > 600V
- Recognized by UL (E252584), RoHS Compliant

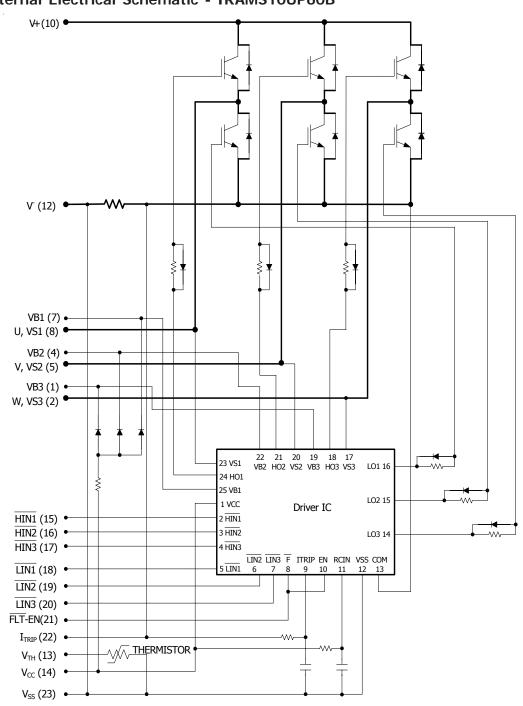


**Absolute Maximum Ratings** 

Parameter	Description	Value	Units
V <sub>CES</sub> / V <sub>RRM</sub>	IGBT/Diode Blocking Voltage	600	
V <sup>+</sup>	Positive Bus Input Voltage	450	V
I <sub>O</sub> @ T <sub>C</sub> =25°C	RMS Phase Current (Note 1)	10	
I <sub>O</sub> @ T <sub>C</sub> =100°C	RMS Phase Current (Note 1)	5	Α
$I_{O}$	Pulsed RMS Phase Current (Note 2)	15	
F <sub>PWM</sub>	PWM Carrier Frequency	20	kHz
$P_D$	Power dissipation per IGBT @ T <sub>C</sub> =25°C	27	W
$V_{\rm ISO}$	Isolation Voltage (1min)	2000	$V_{RMS}$
T <sub>J</sub> (IGBT & Diodes)			0.0
T <sub>J</sub> (Driver IC)	Operating Junction temperature Range	-40 to +150	°C
Т	Mounting torque Range (M3 screw)	0.5 to 1.0	Nm

Note 1: Sinusoidal Modulation at V $^+$ =400V, T $_J$ =150°C, F $_{PWM}$ =20kHz, Modulation Depth=0.8, PF=0.6, See Figure 3. Note 2: t $_P$ <100ms; T $_C$ =25°C; F $_{PWM}$ =20kHz. Limited by I $_{BUS-ITRIP}$ , see Table "Inverter Section Electrical Characteristics"

#### Internal Electrical Schematic - IRAMS10UP60B





**Absolute Maximum Ratings (Continued)** 

Symbol	Parameter	Min	Max	Units	Conditions
$I_{BDF}$	Bootstrap Diode Peak Forward Current		4.5	Α	t <sub>P</sub> = 10ms, T <sub>J</sub> = 150°C, T <sub>C</sub> =100°C
P <sub>BR Peak</sub>	Bootstrap Resistor Peak Power (Single Pulse)		80	W	$t_P$ =100 $\mu$ s, $T_C$ =100 $^{\circ}$ C ESR / ERJ series
V <sub>S1,2,3</sub>	High side floating supply offset voltage	V <sub>B1,2,3</sub> - 25	V <sub>B1,2,3</sub> +0.3	٧	
V <sub>B1,2,3</sub>	High side floating supply voltage	-0.3	600	٧	
V <sub>CC</sub>	Low Side and logic fixed supply voltage	-0.3	20	٧	
$V_{IN}$ , $V_{EN}$ , $V_{ITRIP}$	Input voltage LIN, HIN, EN, $I_{Trip}$	-0.3	Lower of (V <sub>SS</sub> +15V) or V <sub>CC</sub> +0.3V	٧	

# Inverter Section Electrical Characteristics @T<sub>J</sub>= 25°C

Symbol	Parameter	Min	Тур	Max	Units	Conditions
V <sub>(BR)CES</sub>	Collector-to-Emitter Breakdown Voltage	600			V	V <sub>IN</sub> =5V, I <sub>C</sub> =250μA
$\Delta V_{(BR)CES}$ / $\Delta T$	Temperature Coefficient of Breakdown Voltage		0.57		V/°C	V <sub>IN</sub> =5V, I <sub>C</sub> =1.0mA (25°C - 150°C)
V <sub>CE(ON)</sub>	Collector-to-Emitter Saturation		1.70	2.00	V	$I_C=5A$ , $V_{CC}=15V$
VCE(ON)	Voltage		2.00	2.40	\ \ \	$I_{C}=5A$ , $V_{CC}=15V$ , $T_{J}=150$ °C
T	Zero Gate Voltage Collector		5	80	μA	V <sub>IN</sub> =5V, V <sup>+</sup> =600V
$I_{CES}$	Current		10		μΑ	V <sub>IN</sub> =5V, V <sup>+</sup> =600V, T <sub>J</sub> =150°C
V <sub>FM</sub>	Diada Farward Voltago Drop		1.80	2.35	V	I <sub>C</sub> =5A
V FM	Diode Forward Voltage Drop		1.30	1.70	\ \ \	I <sub>C</sub> =5A, T <sub>J</sub> =150°C
V	Bootstrap Diode Forward Voltage			1.25	V	I <sub>F</sub> =1A
$V_{BDFM}$	Drop			1.10	\ \ \	I <sub>F</sub> =1A, T <sub>J</sub> =150°C
R <sub>BR</sub>	Bootstrap Resistor Value		2		Ω	T <sub>J</sub> =25°C
$\Delta R_{BR}/R_{BR}$	Bootstrap Resistor Tolerance			±5	%	T <sub>J</sub> =25°C
I <sub>BUS_TRIP</sub>	Current Protection Threshold (positive going)	13.1		16.4	Α	$T_J$ =-40°C to 125°C See fig. 2



## Inverter Section Switching Characteristics @ T<sub>J</sub>= 25°C

Symbol	Parameter	Min	Тур	Max	Units	Conditions	
E <sub>ON</sub>	Turn-On Switching Loss		200	235		I <sub>C</sub> =5A, V <sup>+</sup> =400V	
E <sub>OFF</sub>	Turn-Off Switching Loss		75	100	1	V <sub>CC</sub> =15V, L=2mH	
E <sub>TOT</sub>	Total Switching Loss		275	335	μJ	Energy losses include "tail" and diode reverse recovery	
E <sub>REC</sub>	Diode Reverse Recovery energy		15	25		diode reverse recovery	
t <sub>RR</sub>	Diode Reverse Recovery time		70	100	ns	See CT1	
E <sub>ON</sub>	Turn-On Switching Loss		300	360		I <sub>C</sub> =5A, V <sup>+</sup> =400V	
E <sub>OFF</sub>	Turn-off Switching Loss		135	165		V <sub>CC</sub> =15V, L=2mH, T <sub>J</sub> =150°C	
Етот	Total Switching Loss		435	525	μЈ	Energy losses include "tail" and diode reverse recovery	
E <sub>REC</sub>	Diode Reverse Recovery energy		30	40		diode reverse recovery	
t <sub>RR</sub>	Diode Reverse Recovery time		100	145	ns	See CT1	
$Q_G$	Turn-On IGBT Gate Charge		29	44	nC	I <sub>C</sub> =15A, V <sup>+</sup> =400V, V <sub>GE</sub> =15V	
RBSOA	Reverse Bias Safe Operating Area	FULL SQUARE			$T_J$ =150°C, $I_C$ =5A, $V_P$ =600V $V^+$ = 450V $V_{CC}$ =+15V to 0V See CT3		
SCSOA	Short Circuit Safe Operating Area	10			μs	$T_J=150$ °C, $V_P=600$ V, $V^+=360$ V, $V_{CC}=+15$ V to 0V See CT2	
I <sub>CSC</sub>	Short Circuit Collector Current		47		А	$T_J$ =150°C, $V_P$ =600V, $t_{SC}$ <10 $\mu$ s $V^+$ = 360V, $V_{GE}$ =15V $V_{CC}$ =+15V to 0V See CT2	

## **Recommended Operating Conditions Driver Function**

The Input/Output logic timing diagram is shown in Figure 1. For proper operation the device should be used within the recommende conditions. All voltages are absolute referenced to COM/ $I_{TRIP}$ . The  $V_S$  offset is tested with all supplies biased at 15V differential (Note 3)

Symbol	Definition	Min	Max	Units
V <sub>B1,2,3</sub>	High side floating supply voltage	V <sub>S</sub> +12	V <sub>S</sub> +20	V
V <sub>S1,2,3</sub>	High side floating supply offset voltage	Note 4	450	v
V <sub>CC</sub>	Low side and logic fixed supply voltage	12	20	V
V <sub>ITRIP</sub>	I <sub>TRIP</sub> input voltage	V <sub>SS</sub>	V <sub>SS</sub> +5	v
V <sub>IN</sub>	Logic input voltage LIN, HIN	V <sub>SS</sub>	V <sub>SS</sub> +5	V
V <sub>EN</sub>	Logic input voltage EN	V <sub>SS</sub>	V <sub>SS</sub> +5	V

Note 3: For more details, see IR21363 data sheet

Note 4: Logic operational for  $V_s$  from COM-5V to COM+600V. Logic state held for  $V_s$  from COM-5V to COM- $V_{BS}$ . (please refer to DT97-3 for more details)



#### **Static Electrical Characteristics Driver Function**

 $V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS1,2,3}$ )=15V, unless otherwise specified. The  $V_{IN}$  and  $I_{IN}$  parameters are referenced to COM/ $I_{TRIP}$  and are applicable to all six channels. (Note 3)

Symbol	Definition	Min	Тур	Max	Units
V <sub>INH</sub> , V <sub>ENH</sub>	Logic "0" input voltage	3.0			V
V <sub>INL</sub> , V <sub>ENL</sub>	Logic "1" input voltage			0.8	V
V <sub>CCUV+</sub> , V <sub>BSUV+</sub>	V <sub>CC</sub> and V <sub>BS</sub> supply undervoltage Positive going threshold	10.6	11.1	11.6	V
V <sub>CCUV-</sub> , V <sub>BSUV-</sub>	V <sub>CC</sub> and V <sub>BS</sub> supply undervoltage Negative going threshold	10.4	10.9	11.4	V
V <sub>CCUVH</sub> , V <sub>BSUVH</sub>	V <sub>CC</sub> and V <sub>BS</sub> supply undervoltage lock-out hysteresis		0.2		V
$V_{IN,Clamp}$	Input Clamp Voltage (HIN, LIN, $I_{TRIP}$ ) $I_{IN}$ =10 $\mu$ A	4.9	5.2	5.5	V
$I_{QBS}$	Quiescent V <sub>BS</sub> supply current V <sub>IN</sub> =0V			165	μΑ
$I_{QCC}$	Quiescent V <sub>CC</sub> supply current V <sub>IN</sub> =0V			3.35	mA
I <sub>LK</sub>	Offset Supply Leakage Current			60	μΑ
I <sub>IN+</sub> , I <sub>EN+</sub>	Input bias current V <sub>IN</sub> =5V		200	300	μΑ
I <sub>IN-</sub> , I <sub>EN-</sub>	Input bias current V <sub>IN</sub> =0V		100	220	μΑ
I <sub>TRIP+</sub>	I <sub>TRIP</sub> bias current V <sub>ITRIP</sub> =5V		30	100	μΑ
I <sub>TRIP-</sub>	I <sub>TRIP</sub> bias current V <sub>ITRIP</sub> =0V		0	1	μΑ
V(I <sub>TRIP</sub> )	I <sub>TRIP</sub> threshold Voltage	440	490	540	mV
V(I <sub>TRIP</sub> , HYS)	I <sub>TRIP</sub> Input Hysteresis		70		mV
R <sub>ON</sub> ,FLT	Fault Output ON Resistance		50	100	ohm

## **Dynamic Electrical Characteristics**

Driver only timing unless otherwise specified.)

Symbol	Parameter	Min	Тур	Max	Units	Conditions	
T <sub>ON</sub>	Input to Output propagation turn- on delay time (see fig.11)		590		ns	V <sub>CC</sub> =V <sub>BS</sub> = 15V, I <sub>C</sub> =10A, V <sup>+</sup> =400V	
T <sub>OFF</sub>	Input to Output propagation turn- off delay time (see fig. 11)		700		ns	VCC-VBS- 13V, IC-10A, V -400V	
T <sub>FLIN</sub>	Input Filter time (HIN, LIN)	100	200		ns	V <sub>IN</sub> =0 & V <sub>IN</sub> =5V	
T <sub>BLT-Trip</sub>	I <sub>TRIP</sub> Blancking Time	100	150		ns	V <sub>IN</sub> =0 & V <sub>IN</sub> =5V	
$D_T$	Dead Time (V <sub>BS</sub> =V <sub>DD</sub> =15V)	220	290	360	ns	$V_{BS}=V_{CC}=15V$	
M <sub>T</sub>	Matching Propagation Delay Time (On & Off)		40	75	ns	$V_{CC} = V_{BS} = 15V$ , external dead time > 400ns	
T <sub>ITrip</sub>	I <sub>Trip</sub> to six switch to turn-off propagation delay (see fig. 2)			1.75	μs	V <sub>CC</sub> =V <sub>BS</sub> = 15V, I <sub>C</sub> =10A, V <sup>+</sup> =400V	
T	Post I <sub>Trip</sub> to six switch to turn-off		7.7		mc	T <sub>C</sub> = 25°C	
T <sub>FLT-CLR</sub>	clear time (see fig. 2)		6.7		ms	T <sub>C</sub> = 100°C	



#### **Thermal and Mechanical Characteristics**

Symbol	Parameter	Min	Тур	Max	Units	Conditions
R <sub>th(J-C)</sub>	Thermal resistance, per IGBT		4.2	4.7		Flat, greased surface. Heatsink
R <sub>th(J-C)</sub>	Thermal resistance, per Diode		5.5	6.5		compound thermal conductivity
R <sub>th(C-S)</sub>	Thermal resistance, C-S		0.1			1W/mK
C <sub>D</sub>	Creepage Distance	3.2			mm	See outline Drawings

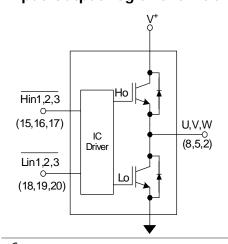
## **Internal Current Sensing Resistor - Shunt Characteristics**

Symbol	Parameter	Min	Тур	Max	Units	Conditions
R <sub>Shunt</sub>	Resistance	33.0	33.3	33.7	mΩ	T <sub>C</sub> = 25°C
T <sub>Coeff</sub>	Temperature Coefficient	0		200	ppm/°C	
P <sub>Shunt</sub>	Power Dissipation			2.2	W	-40°C< T <sub>C</sub> <100°C
T <sub>Range</sub>	Temperature Range	-40		125	°C	

#### **Internal NTC - Thermistor Characteristics**

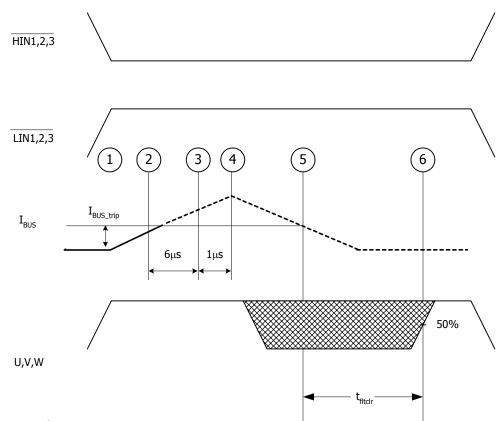
Parameter	Definition	Min	Тур	Max	Units	Conditions	
R <sub>25</sub>	Resistance	97	100	103	kΩ	T <sub>C</sub> = 25°C	
R <sub>125</sub>	Resistance	2.25	2.52	2.80	kΩ	T <sub>C</sub> = 125°C	
В	B-constant (25-50°C)	4165	4250	4335	k	$R_2 = R_1 e^{[B(1/T2 - 1/T1)]}$	
Temperature Range		-40		125	°C		
Typ. Dissipation constant			1		mW/°C	T <sub>C</sub> = 25°C	

# **Input-Output Logic Level Table**



FLT- EN	$I_{TRIP}$	HIN1,2,3	LIN1,2,3	U,V,W
1	0	0	1	V <sup>+</sup>
1	0	1	0	0
1	0	1	1	Off
1	1	Χ	X X	
0	Χ	Χ	Χ	Off





Sequence of events:

- 1-2) Current begins to rise
- 2) Current reaches I<sub>BUS\_Trip</sub> level
   2-3) Current is higher than I<sub>BUS\_Trip</sub> for at least 6μs. This value is the worst-case condition with very low over-current. In case of high current (short circuit), the actual delay will be smaller.
- 3-4) Delay between driver identification of over-current condition and disabling of all outputs
- 4) Current starts decreasing, eventually reaching 0
- 5) Current goes below  $I_{\text{BUS trip'}}$  the driver starts its auto-reset sequence
- 6) Driver is automatically reset and normal operation can resume (over-current condition must be removed
- by the time the drivers automatically resets itself)

Figure 2.  $I_{Trip}$  Timing Waveform

Note 5: The shaded area indicates that both high-side and low-side switches are off and therefore the half-bridge output voltage would be determined by the direction of current flow in the load.



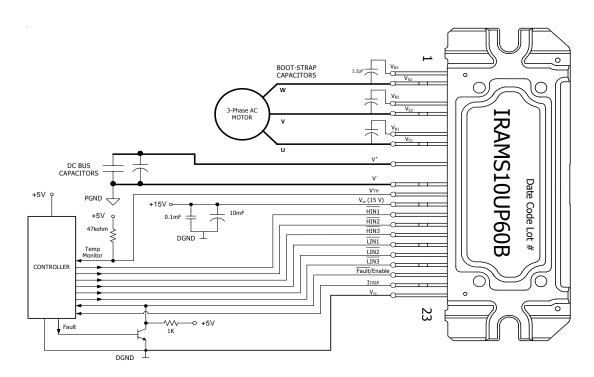
## **Module Pin-Out Description**

Pin	Name	Description
1	V <sub>B3</sub>	High Side Floating Supply Voltage 3
2	W,V <sub>S3</sub>	Output 3 - High Side Floating Supply Offset Voltage
3	NA	none
4	V <sub>B2</sub>	High Side Floating Supply voltage 2
5	V,V <sub>S2</sub>	Output 2 - High Side Floating Supply Offset Voltage
6	NA	none
7	V <sub>B1</sub>	High Side Floating Supply voltage 1
8	U, V <sub>S1</sub>	Output 1 - High Side Floating Supply Offset Voltage
9	NA	none
10	V <sup>+</sup>	Positive Bus Input Voltage
11	NA	none
12	V <sup>-</sup>	Negative Bus Input Voltage
13	V <sub>TH</sub>	Temperature Feedback
14	V <sub>CC</sub>	+15V Main Supply
15	H <sub>IN1</sub>	Logic Input High Side Gate Driver - Phase 1
16	H <sub>IN2</sub>	Logic Input High Side Gate Driver - Phase 2
17	H <sub>IN3</sub>	Logic Input High Side Gate Driver - Phase 3
18	L <sub>IN1</sub>	Logic Input Low Side Gate Driver - Phase 1
19	L <sub>IN2</sub>	Logic Input Low Side Gate Driver - Phase 2
20	L <sub>IN3</sub>	Logic Input Low Side Gate Driver - Phase 3
21	FLT/Enable	Fault Output and Enable Pin
22	I <sub>TRIP</sub>	Current Sense and Itrip Pin
23	V <sub>SS</sub>	Negative Main Supply





#### Typical Application Connection IRAMS10UP60B



- 1. Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible to reduce ringing and EMI problems. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
- 2. In order to provide good decoupling between  $V_{CC}$ - $V_{SS}$  and  $V_{B1,2,3}$ - $V_{S1,2,3}$  terminals, the capacitors shown connected between these terminals should be located very close to the module pins. Additional high frequency capacitors, typically 0.1 $\mu$ F, are strongly recommended.
- 3. Value of the boot-strap capacitors depends upon the switching frequency. Their selection should be made based on IR design tip DN 98-2a, application note AN-1044 or Figure 9. Bootstrap capacitor value must be selected to limit the power dissipation of the internal resistor in series with the  $V_{CC}$ . (see maximum ratings Table on page 3).
- 4. Current sense signal can be obtained from pin 22 and pin 23. Care should be taken to avoid having inverter current flowing through pin 22 to mantain required current measurement accuracy.
- 5. After approx. 8ms the FAULT is reset. (see Dynamic Characteristics Table on page 5).
- 6. PWM generator must be disabled within Fault duration to garantee shutdown of the system, overcurrent condition must be cleared before resuming operation.
- 7. Fault/Enable pin must be pulled-up to +5V.

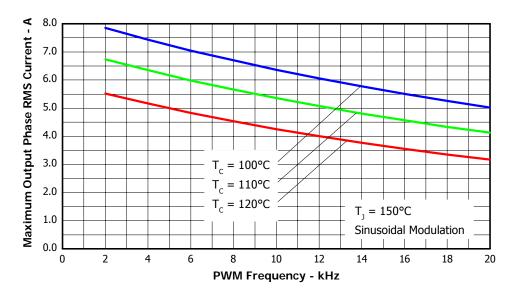


Figure 3. Maximum Sinusoidal Phase Current vs. PWM Switching Frequency  $V^+$ =400V ,  $T_3$ =150°C, Modulation Depth=0.8, PF=0.6

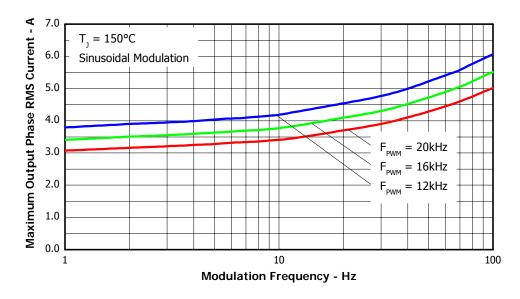


Figure 4. Maximum Sinusoidal Phase Current vs. Modulation Frequency  $V^+=400V$ ,  $T_1=150$ °C,  $T_C=100$ °C, Modulation Depth=0.8, PF=0.6

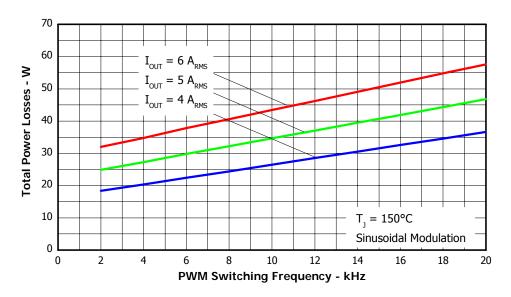


Figure 5. Total Power Losses vs. PWM Switching Frequency, Sinusoidal modulation  $V^{+}{=}400V\ ,\ T_{J}{=}150^{\circ}C,\ Modulation\ Depth{=}0.8,\ PF{=}0.6$ 

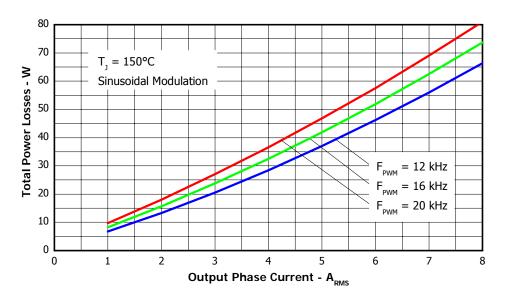


Figure 6. Total Power Losses vs. Output Phase Current, Sinusoidal modulation  $V_{BUS}$ =400V ,  $T_{J}$ =150°C, Modulation Depth=0.8, PF=0.6

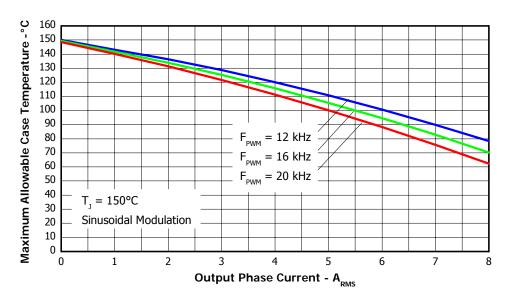


Figure 7. Maximum Allowable Case temperature vs. Output RMS Current per Phase

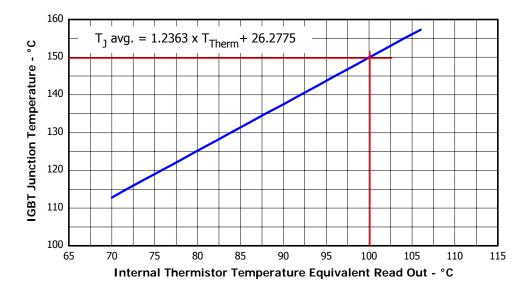


Figure 8. Estimated Maximum IGBT Junction Temperature vs. Thermistor Temperature

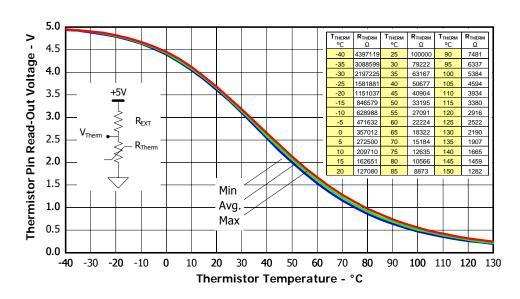


Figure 9. Thermistor Readout vs. Temperature (47kohm pull-up resistor, 5V) and Nominal Thermistor Resistance values vs. Temperature Table.

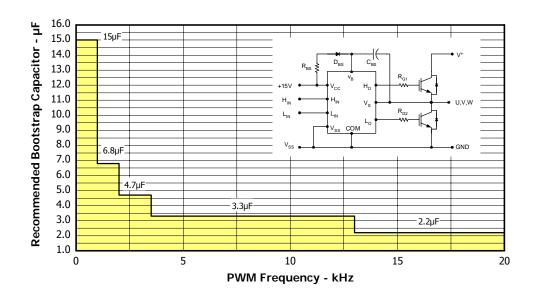


Figure 10. Recommended Bootstrap Capacitor Value vs. Switching Frequency

Figure 11. Switching Parameter Definitions

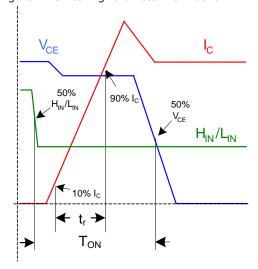


Figure 11a. Input to Output Propagation turn-on Delay Time

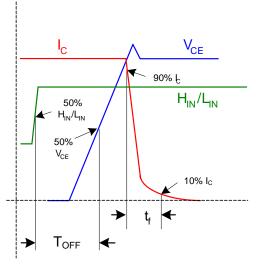


Figure 11b. Input to Output Propagation turn-off Delay Time

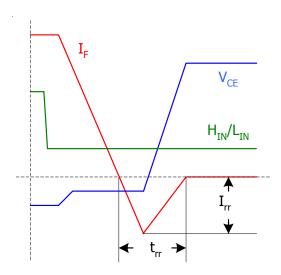
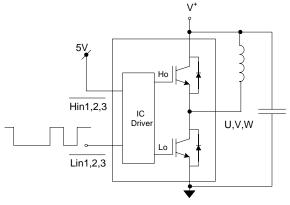


Figure 11c. Diode Reverse Recovery





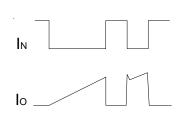
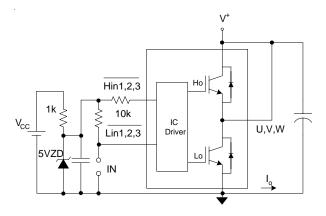


Figure CT1. Switching Loss Circuit



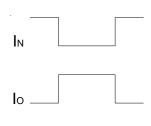
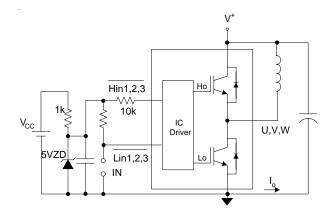


Figure CT2. S.C.SOA Circuit



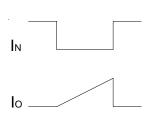
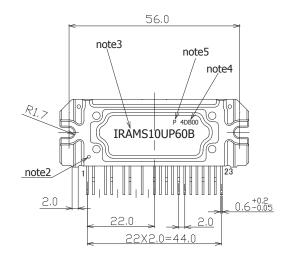
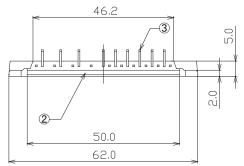


Figure CT3. R.B.SOA Circuit



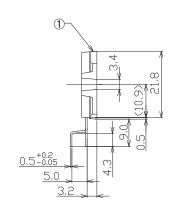
## Package Outline IRAMS10UP60B





Dimensions in mm For mounting instruction see AN-1049

missing pin: 3,6,9,11



note1: Unit Tolerance is  $\pm 0.5$ mm, Unless Otherwise Specified.

note2: Mirror Surface Mark indicates Pin1 Identification.

note3: Part Number Marking.

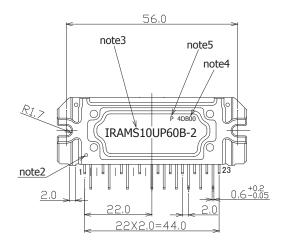
Characters Font in this drawing differs from Font shown on Module.

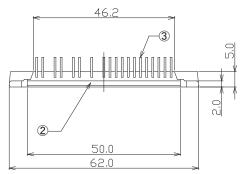
note4: Lot Code Marking.

Characters Font in this drawing differs from Font shown on Module.

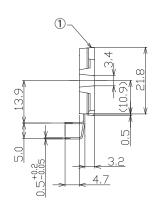
note5: "P" Character denotes Lead Free. Characters Font in this drawing differs from Font shown on Module.

#### Package Outline IRAMS10UP60B-2





Dimensions in mm For mounting instruction see AN-1049 missing pin: 3,6,9,11



note1: Unit Tolerance is <u>+</u>0.5mm, Unless Otherwise Specified.

note2: Mirror Surface Mark indicates Pin1 Identification.

note3: Part Number Marking.

Characters Font in this drawing differs from Font shown on Module.

note4: Lot Code Marking.

Characters Font in this drawing differs from

Font shown on Module.

note5: "P" Character denotes Lead Free.

Characters Font in this drawing differs from

Font shown on Module.



Data and Specifications are subject to change without notice

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