

# Electric Circuits II: (ELCT 401)



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**Lecture 3: AC Analysis**



# Sinusoidal Steady-State Analysis



## Objectives

**To perform circuit analysis using the following techniques:**

1. Basic Approach (KCL and KVL – Circuit reduction)
2. Nodal Analysis
3. Mesh Analysis
4. Superposition Theorem
5. Source Transformation
6. Thevenin and Norton Equivalent Circuits
7. Bridge Networks

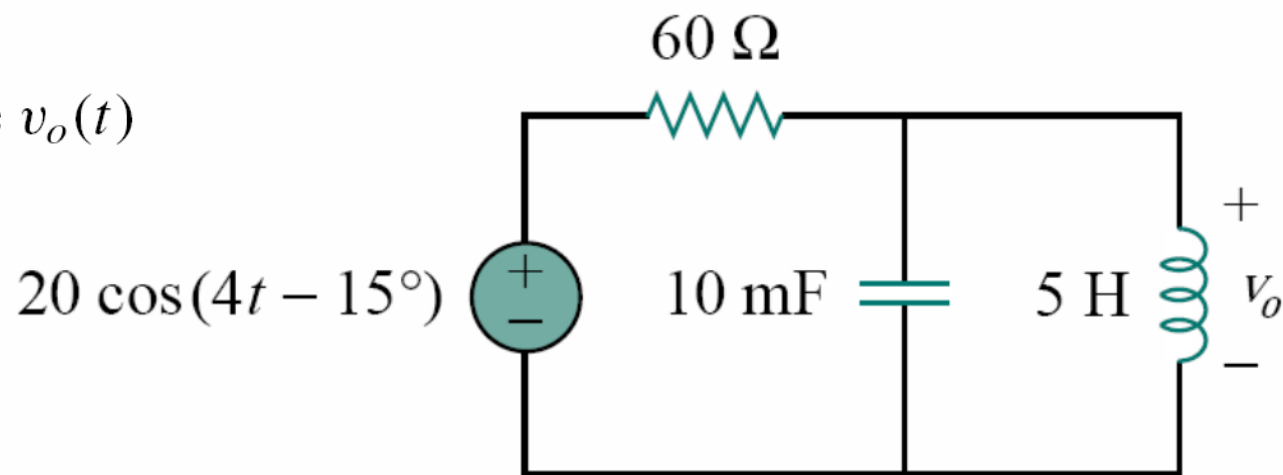


## Steps to Analyze AC Circuits:

1. Transform the circuit to the phasor (frequency) domain
2. Solve using an appropriate circuit technique such as nodal analysis, mesh current analysis, superposition, etc...
3. Transform the resulting phasors back into the time domain

## Example

Determine  $v_o(t)$

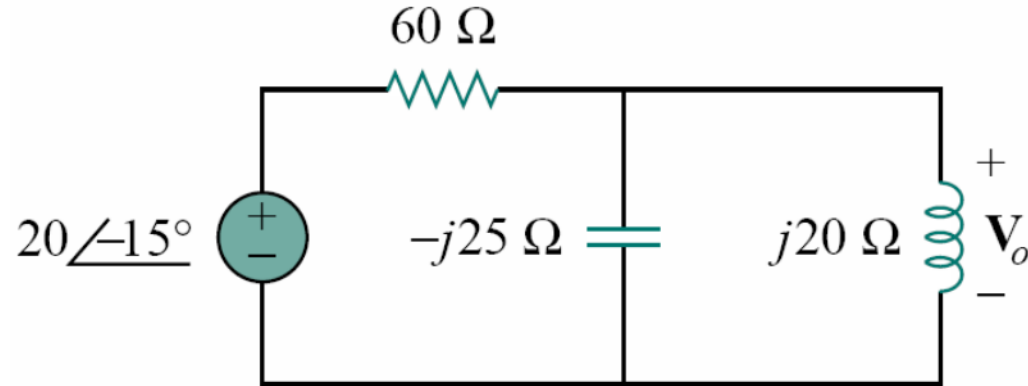


## Solution

$$v_s = 20 \cos(4t - 15^\circ) \quad \Rightarrow \quad \mathbf{V}_s = 20 \angle -15^\circ \, \text{V}, \quad \omega = 4$$

$$10 \, \text{mF} \quad \Rightarrow \quad \frac{1}{j\omega C} = \frac{1}{j4 \times 10 \times 10^{-3}} \\ = -j25 \, \Omega$$

$$5 \, \text{H} \quad \Rightarrow \quad j\omega L = j4 \times 5 = j20 \, \Omega$$



$$\mathbf{Z}_1 = 60\ \Omega$$

$$\mathbf{Z}_2 = -j25 \parallel j20 = \frac{-j25 \times j20}{-j25 + j20} = j100\ \Omega$$

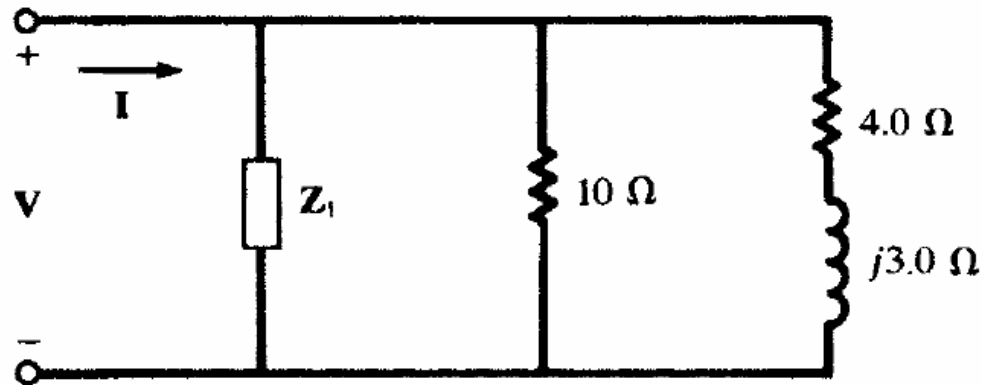
$$\begin{aligned} \mathbf{V}_o &= \frac{\mathbf{Z}_2}{\mathbf{Z}_1 + \mathbf{Z}_2} \mathbf{V}_s = \frac{j100}{60 + j100} (20\angle -15^\circ) \\ &= (0.8575\angle 30.96^\circ) (20\angle -15^\circ) = 17.15\angle 15.96^\circ\ \text{V}. \end{aligned}$$

$$v_o(t) = 17.15 \cos(4t + 15.96^\circ)\text{V}$$



## Example

Find  $\mathbf{Z}_1$  in the three-branch network if  $\mathbf{I} = 31.5 \angle 24.0^\circ \text{ A}$   
for an applied voltage  $\mathbf{V} = 50.0 \angle 60.0^\circ \text{ V}$ .



## Solution

$$\mathbf{Y} = \frac{\mathbf{I}}{\mathbf{V}} = 0.630 \angle -36.0^\circ = 0.510 - j0.370 \text{ S}$$

$$0.510 - j0.370 = \mathbf{Y}_1 + \frac{1}{10} + \frac{1}{4.0 + j3.0}$$

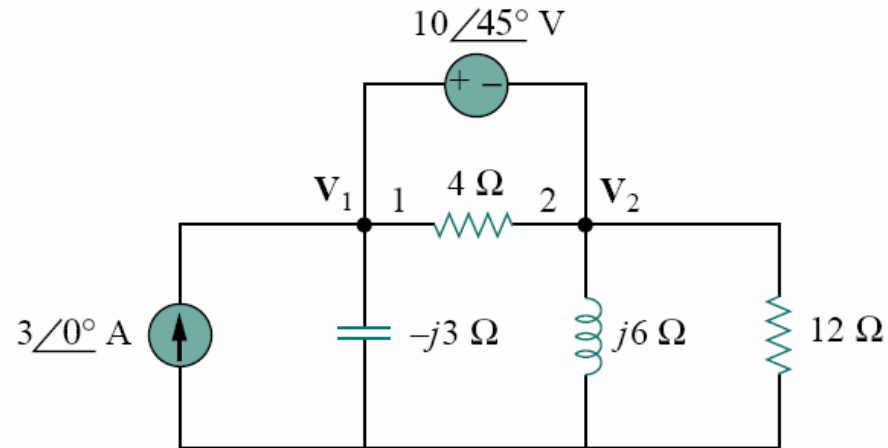
$$\mathbf{Y}_1 = 0.354 \angle -45^\circ \text{ S and } \mathbf{Z}_1 = 2.0 + j2.0 \Omega.$$



## Nodal Analysis

### Example:

Using nodal analysis, find  $V_1$  and  $V_2$ .



### Solution:

$$-3 + \frac{V_1 - V_2}{4} + \frac{V_1}{-j3} + \frac{V_2 - V_1}{4} + \frac{V_2}{j6} + \frac{V_2}{12} = 0 \quad (1)$$

$$V_1 - V_2 = 10\angle 45^\circ \quad (2)$$

Solve 1 & 2:

$$V_2 = 31.41 \angle -87.18^\circ \text{ V}$$

$$V_1 = V_2 + 10\angle 45^\circ = 25.78 \angle -70.48^\circ \text{ V}$$

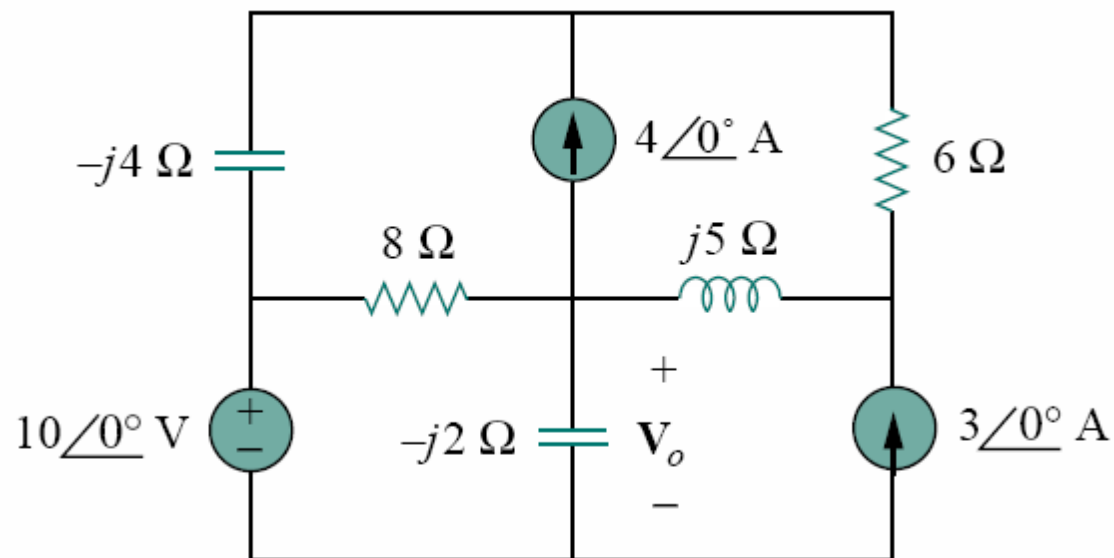




# Mesh Analysis

## Example

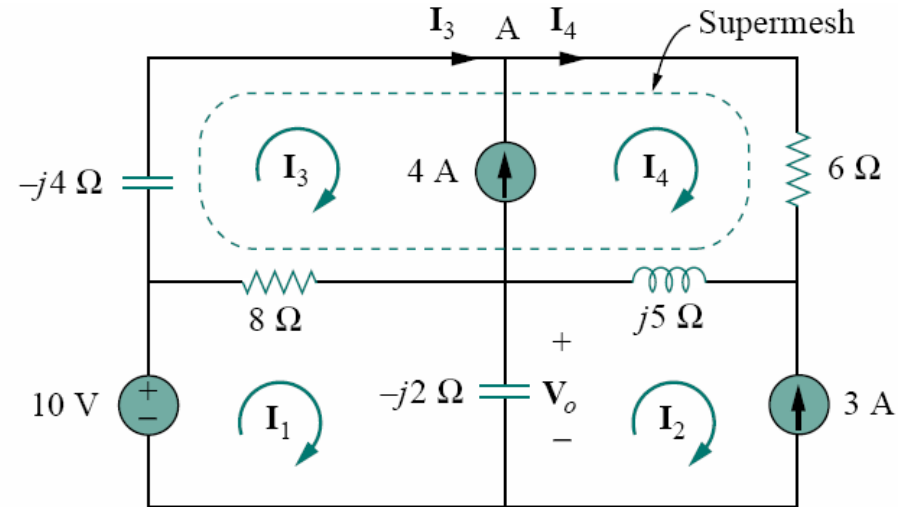
Find  $V_o$  in the following figure using mesh analysis.





## Solution

- Mesh 1:  $-10 + 8(I_1 - I_3) - j2(I_1 - I_2) = 0$
- Mesh 2:  $I_2 = -3$
- Super mesh:  
 $-j4I_3 + 6I_4 + j5(I_4 - I_2) + 8(I_3 - I_1) = 0$
- Constraint:  $I_4 - I_3 = 4$



Solving the 4 equations:

$$I_1 = \frac{41}{145} - j\frac{523}{145} \text{ A}, \quad I_2 = -3 \text{ A}, \quad I_3 = -\frac{271}{145} - j\frac{642}{145} \text{ A}, \quad I_4 = \frac{309}{145} - j\frac{642}{145} \text{ A}$$

$$V_0 = -2j(I_1 - I_2) = -7.2134 - j6.568 = 9.756 \angle 222.32^\circ \text{ V}$$



## Superposition Theorem

If a circuit has sources operating at different frequencies, then:

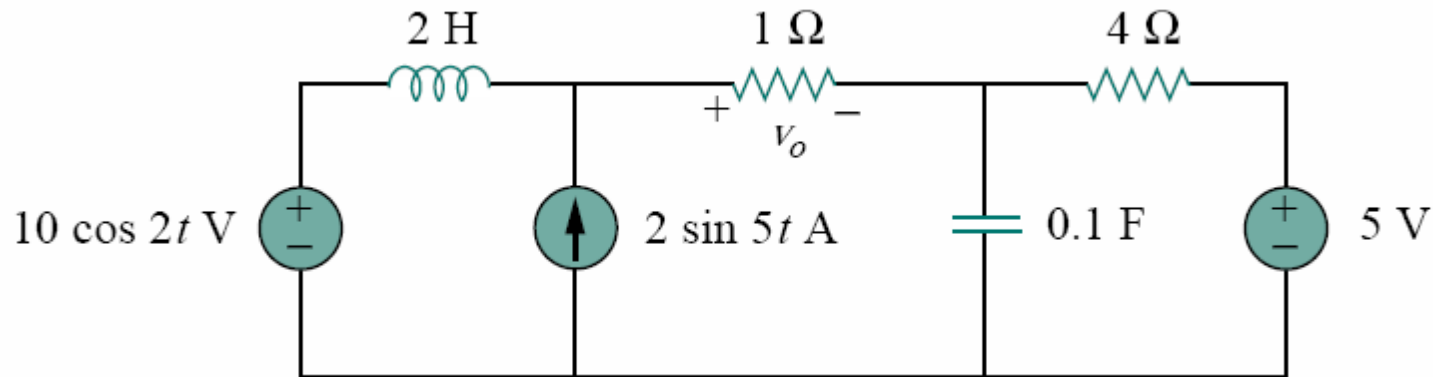
- The separate phasor circuit for each frequency must be solved independently, and
- The total response is the sum of the time-domain responses of all the individual phasor circuits



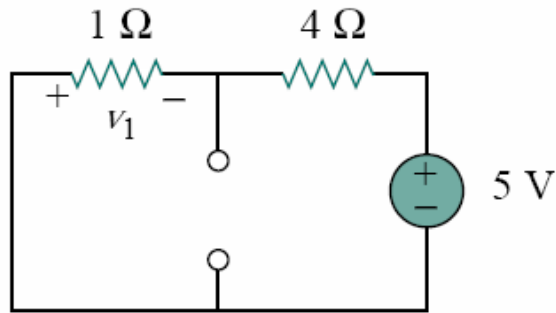
# Superposition Theorem

## Example 5

Calculate  $v_o(t)$  in the circuit shown below using the superposition theorem

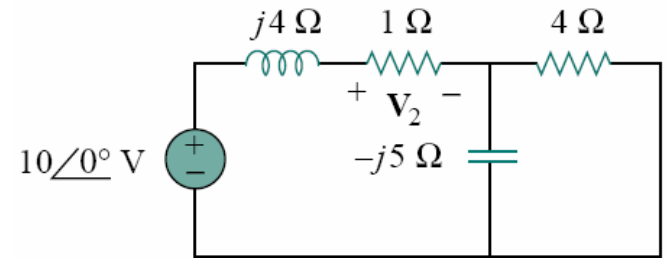


$$v_o = v_1 + v_2 + v_3$$



(a)

$$-v_1 = \frac{1}{1+4}(5) = 1 \text{ V}$$

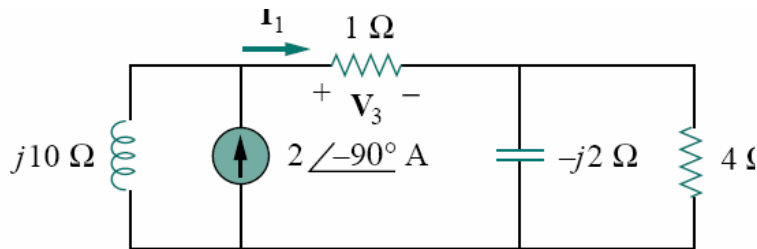


(b)  $\omega = 2 \text{ rad/s}$

Voltage divider

$$V_2 = \frac{1}{1 + j4 + \mathbf{Z}} (10 \angle 0^\circ) = \frac{10}{3.439 + j2.049} = 2.498 \angle -30.79^\circ$$

$$v_2 = 2.498 \cos(2t - 30.79^\circ)$$



$\omega = 5 \text{ rad/s}$  (c)

Current divider

$$V_3 = I_1 \times 1 = \frac{j10}{1.8 + j8.4} (-j2) = 2.328 \angle -77.91^\circ \text{ V}$$

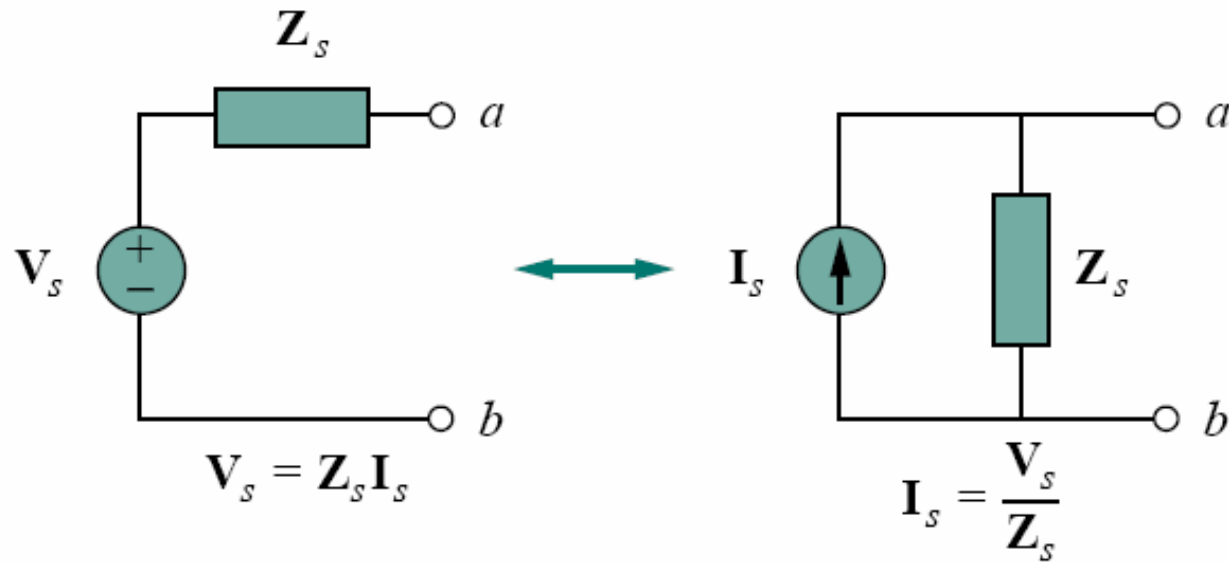
$$v_3 = 2.33 \cos(5t - 80^\circ) = 2.33 \sin(5t + 10^\circ) \text{ V}$$

$$v_o = v_1 + v_2 + v_3$$

$$v_o(t) = -1 + 2.498 \cos(2t - 30.79^\circ) + 2.33 \sin(5t + 10^\circ) \text{ V}$$

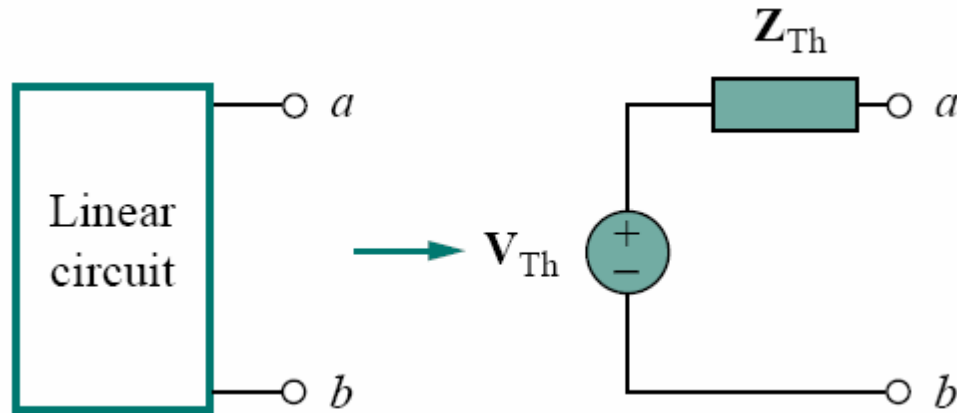


## Source Transformation

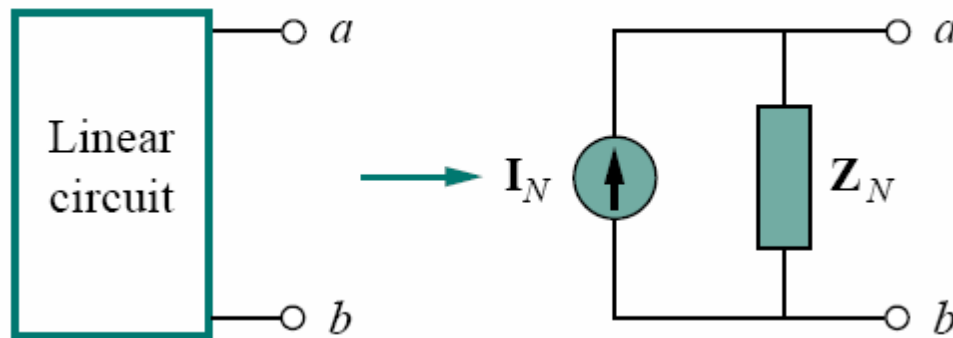




## Thevenin and Norton Equivalent Circuits



Thevenin transform



Norton transform



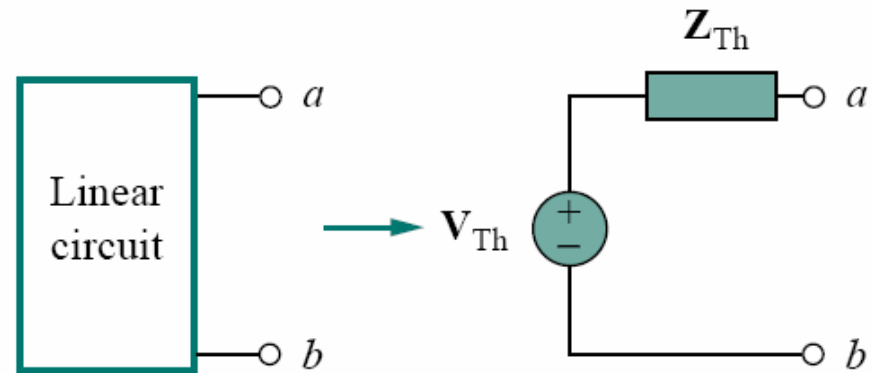
## Thevenin and Norton Equivalent Circuits

If a load impedance  $Z_L$  is connected between the terminals  $a$  and  $b$ , the maximum power is transferred to the load when:

$$Z_L = Z_{th}^*$$

In this case, the maximum power transferred to the load is given by:

$$P_{max} = \frac{|V_{th-max}|^2}{8R_{th}}$$

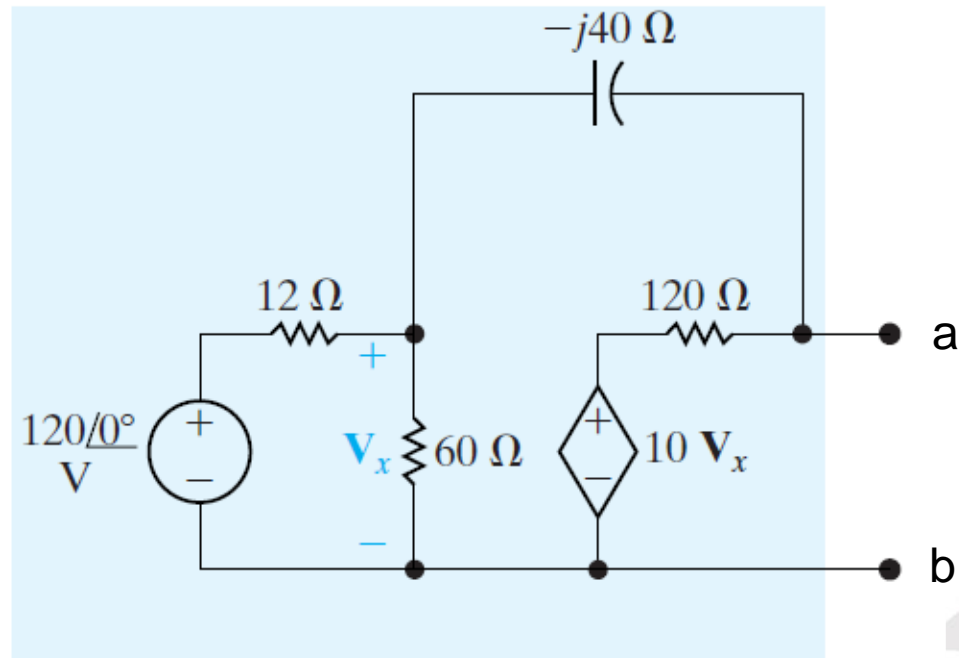






## Example

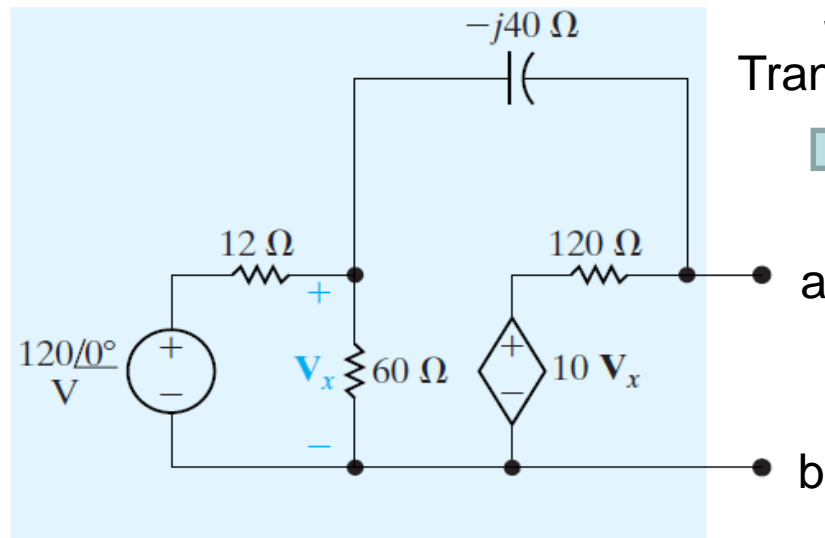
Find the Thevenin equivalent circuit between terminals a & b



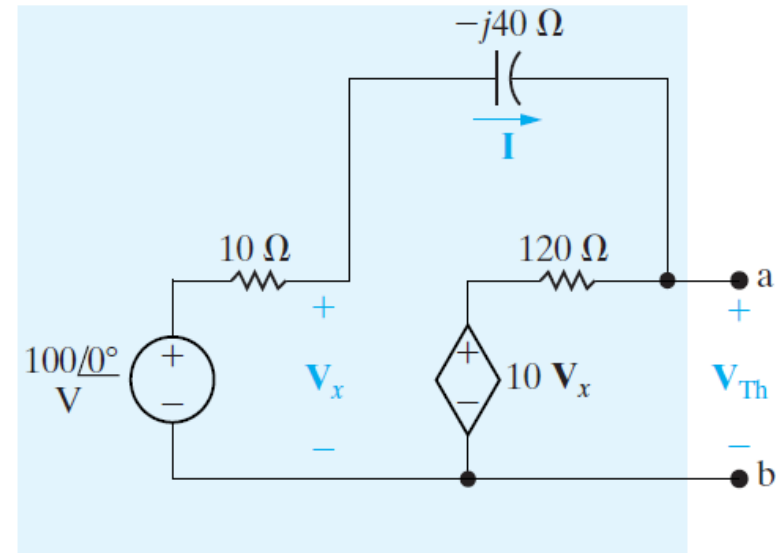
This means that we need to find  $V_{Th}$  and  $Z_{Th}$



## Solution:



Source  
Transformation



Finding  $V_{TH}$

$$V_x = 100 - 10I.$$

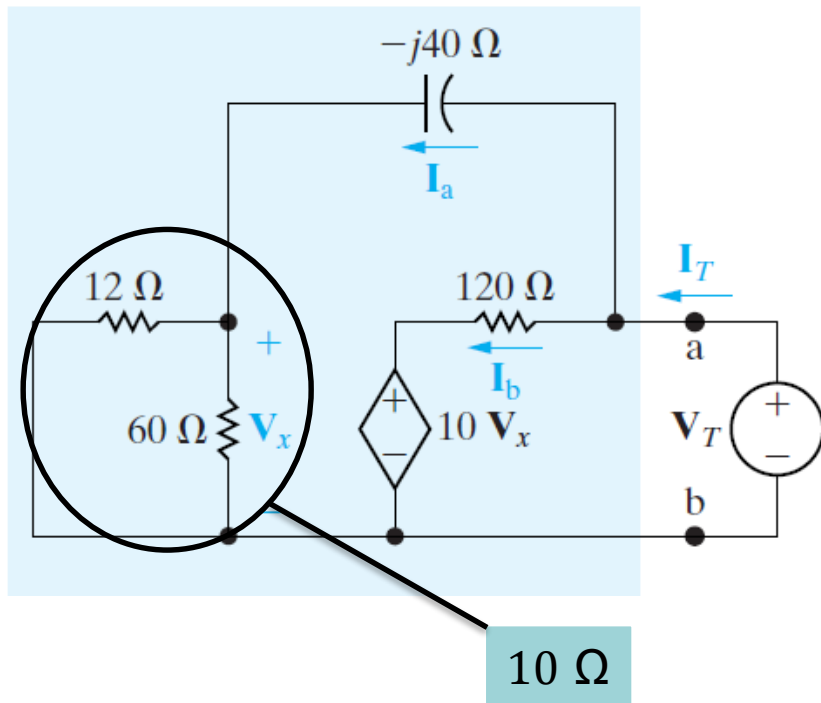
$$I = \frac{-900}{30 - j40} = 18 \angle -126.87^\circ \text{ A}.$$

$$V_x = 100 - 180 \angle -126.87^\circ = 208 + j144 \text{ V}.$$

$$\begin{aligned} V_{Th} &= 10V_x + 120I = 2080 + j1440 + 120(18) \angle -126.87^\circ \\ &= 784 - j288 = 835.22 \angle -20.17^\circ \text{ V}. \end{aligned}$$



Finding  $Z_{TH}$



$$\mathbf{I}_a = \frac{\mathbf{V}_T}{10 - j40}, \quad \mathbf{V}_x = 10\mathbf{I}_a,$$

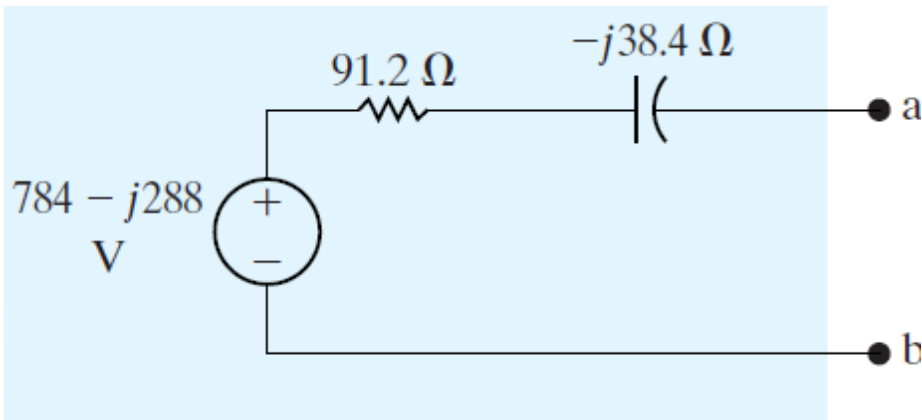
$$\mathbf{I}_b = \frac{\mathbf{V}_T - 10\mathbf{V}_x}{120}$$

$$\mathbf{I}_T = \mathbf{I}_a + \mathbf{I}_b$$

$$= \frac{\mathbf{V}_T}{10 - j40} \left( 1 - \frac{9 + j4}{12} \right)$$

$$= \frac{\mathbf{V}_T(3 - j4)}{12(10 - j40)},$$

$$Z_{Th} = \frac{\mathbf{V}_T}{\mathbf{I}_T} = 91.2 - j38.4 \, \Omega.$$





## Bridge Networks

- A bridge is a 4-arm circuit, connected as shown below
- It is used to measure unknown impedances
- We place the unknown impedance as  $Z_4$ , a known impedance as  $Z_1$ , then we keep varying  $Z_2$  and  $Z_3$  until  $V$  and  $I$  are equal to zero
- At balance, the unknown quantities can be calculated

For balance,  $\mathbf{I}$  and  $\mathbf{V} = 0$ .

Since  $\mathbf{I} = 0$ ,

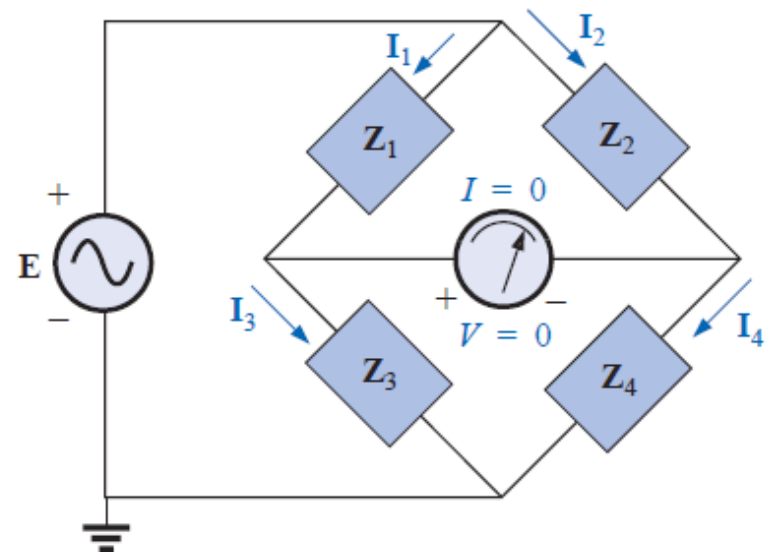
$$\mathbf{I}_1 = \mathbf{I}_3$$

$$\mathbf{I}_2 = \mathbf{I}_4$$

for  $\mathbf{V} = 0$ ,

$$\mathbf{I}_1 \mathbf{Z}_1 = \mathbf{I}_2 \mathbf{Z}_2 \quad (1)$$

$$\mathbf{I}_3 \mathbf{Z}_3 = \mathbf{I}_4 \mathbf{Z}_4 \quad (2)$$



(1)/(2)

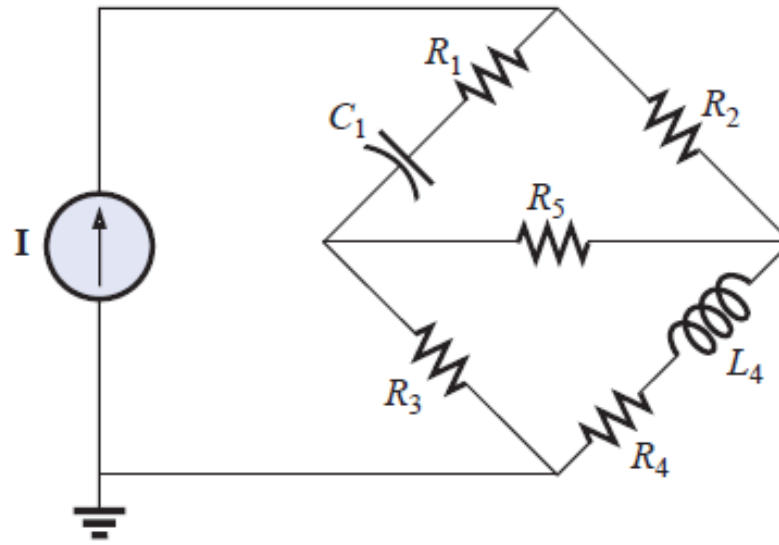
$$\frac{\mathbf{Z}_1}{\mathbf{Z}_3} = \frac{\mathbf{Z}_2}{\mathbf{Z}_4}$$



## Example

Hay's Bridge

$L_4$  &  $R_4$  are the unknowns



At Balance:

$$\mathbf{Z_2 Z_3 = Z_4 Z_1}$$

$$R_2 R_3 = (R_4 + j X_L)(R_1 - j X_C)$$

$$R_2 R_3 = R_1 R_4 + j (R_1 X_L - R_4 X_C) + X_C X_L$$

Therefore

$$R_2 R_3 = R_1 R_4 + X_C X_L$$

$$0 = R_1 X_L - R_4 X_C$$

Since,  $X_L = \omega L$  and  $X_C = \frac{1}{\omega C}$

$$\mathbf{Z_1 = R_1 - j X_C}$$

$$\mathbf{Z_2 = R_2}$$

$$\mathbf{Z_3 = R_3}$$

$$\mathbf{Z_4 = R_4 + j X_L}$$

$$R_2 R_3 = R_1 R_4 + \frac{L}{C}$$

$$R_1 \omega L = \frac{R_4}{\omega C}$$

$$L = \frac{C R_2 R_3}{1 + \omega^2 C^2 R_1^2}$$

$$R_4 = \frac{\omega^2 C^2 R_1 R_2 R_3}{1 + \omega^2 C^2 R_1^2}$$