

Simulation Results

1.1 Block Diagram of the System

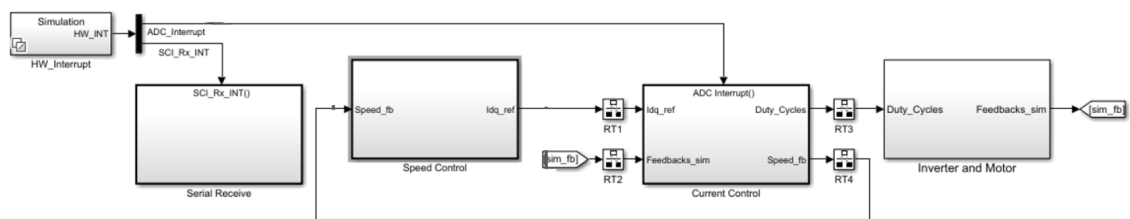


Figure 1.1: Block Diagram of the System

1.2 Speed Control Subsystem

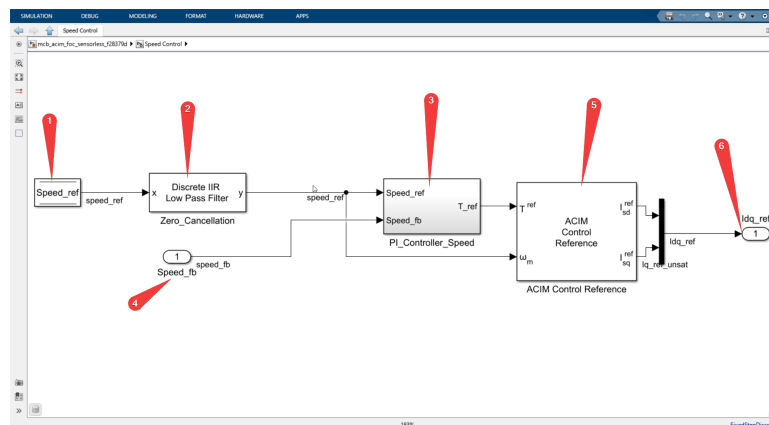


Figure 1.2: Speed Control System

The speed control system shown in Figure 1.2 consists of a DataStoreRead block that holds the speed reference value received from the host

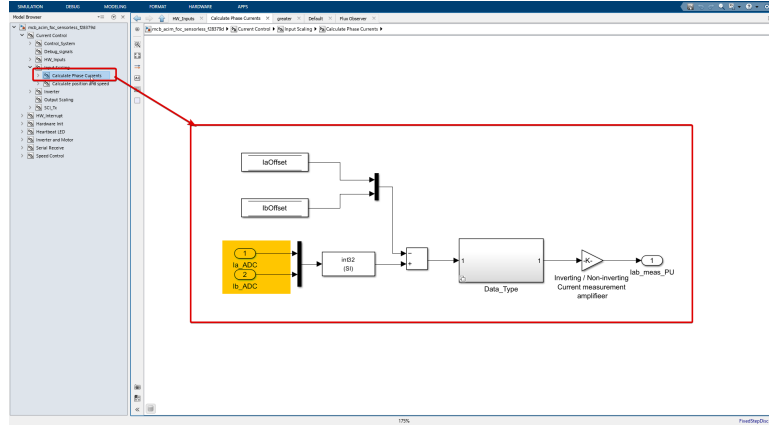


Figure 1.3: Current Measurement

computer, a Discrete IIR lowpass filter block to cancel the zeros in the system, and a Discrete PID Controller with anti-windup block that takes the speed reference and feedback values as inputs and generates the torque reference as output. The ACIM Control reference block then takes the torque reference and speed reference as inputs and generates the I_{sdref} and I_{sqref} values, which are the reference values for the current control loop. The DQ limiter block is used to limit the magnitude of the vector represented in the d-q reference frame, with the option to prioritize either the d-axis or q-axis component.

1.3 Current Control Subsystem

1.3.1 Current measurement

The current measurement part of the Input Scaling block shown in Figure 1.3

converts the I_{aADC} and I_{bADC} inputs from the ADCs to the appropriate data type (int32) and removes the offsets ($I_{aoffset}$ and $I_{boffset}$) that were previously calibrated. The signals then go through a series of gain blocks to convert them to per-unit (PU) values. The first gain block converts the ADC voltage to the actual voltage, the second gain

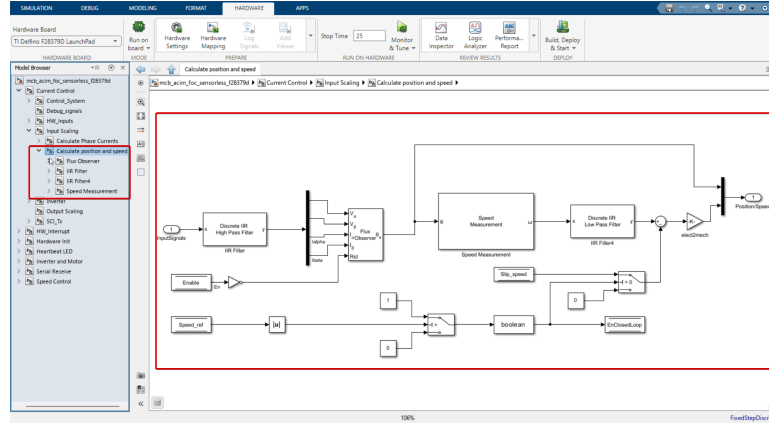


Figure 1.4: Position and Speed Estimation

block converts the voltage to current using the inverter's current sense voltage per amp, and the third gain block converts the current to PU using the motor's base current.

1.3.2 Position and Speed Estimation

The position and speed estimation part of the Input Scaling block shown in Figure 1.4 uses the $V I_{fb}$ signal, which contains the motor's voltage and current in the $\alpha\beta$ reference frame. This signal goes through a high-pass filter to remove low-frequency noise. The filtered signals are then fed into the Flux Observer block, which estimates the stator flux position. The estimated stator flux position is then used in the Speed Estimation block to calculate the motor's speed. The estimated speed is filtered using a low-pass IIR filter to remove high-frequency noise. The final rotor speed is calculated by subtracting the slip speed from the estimated stator flux speed and then dividing by the number of pole pairs to get the mechanical speed.

1.3.3 Control System

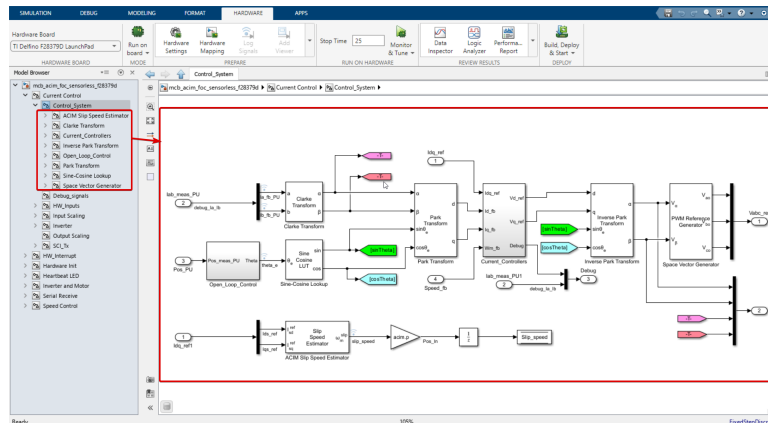


Figure 1.5: Current Control System

The control system shown in Figure 1.5 implemented in the inner current loop consists of two main components: the current controllers and the DQO transformation.

The current controllers use Proportional-Integral (PI) controllers to minimize the error between the reference currents (I_{dq_ref}) and the feedback currents (I_{d_fb} and I_{q_fb}). The output of the PI controllers is then limited, filtered, and adjusted using a feedforward controller and a saturation function to ensure safe and reliable operation of the system. The adjusted d-axis current is then used to generate the reference voltages (V_{d_ref} and V_{q_ref}) in the DQ frame. These reference voltages are then transformed into the alpha-beta frame using the stator flux position (θ) and fed into the PWM reference generator block, which generates the three-phase voltage references for the inverter. This inner current loop ensures that the actual currents closely follow the desired reference currents, which is crucial for the overall performance and stability of the control system.