

# DSP

IDFT:

$$IDFT[X(K)] = \frac{1}{N} \sum_{k=0}^{N-1} X(K) W_N^{-kn}$$

DFT:

$$DFT[x(n)] = \sum_{n=0}^{N-1} x(n) W_N^{kn}$$

**IDFT using DFT:** DFT of conjugate of the input, and the output is the conjugate divided by N.

$$IDFT = \frac{1}{N} (DFT[X^*(K)])^*$$

## Properties

**Circular Time Shift:**

- Use IDFT equation to prove.

$$DFT[x(n - m)_N] = w_N^{mk} X(K)$$

**Conjugate Symmetry:**

- Use DFT equation to prove

$$X(K) = X^*(N - K)$$

**Circular Frequency Shift:**

- Use DFT equation to prove.

$$IDFT[X(K - l)_N] = w_N^{-ln} x(n)$$

**Complex Conjugate:**

- Use DFT equation to prove

$$DFT[x^*(n)] = X^*((-K))_N = X^*(N - K)$$

Parseval Theorem:

$$\sum_{n=0}^{N-1} |x(n)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(K)|^2$$

Periodicity:

$$x(n + N)$$

## FFT

Number of multiplications =  $\frac{N}{2} \log_2 N$

Number of additions =  $N \log_2 N$

Composite Radix:

- 3X2 : then 3 summations going from 0 to 1

$$\sum_{n=0}^1 x(3n)w_6^{3nk} + \sum_{n=0}^1 x(3n+1)w_6^{(3n+1)k} + \sum_{n=0}^1 x(3n+2)w_6^{(3n+2)k}$$

- 2X3: then 2 summations going from 0 to 2

$$\sum_{n=0}^2 x(2n)w_6^{2nk} + \sum_{n=0}^2 x(2n+1)w_6^{(2n+1)k}$$

## Butterworth Filter

1.  $LP \rightarrow LP$

$$s \rightarrow \frac{\Omega_p}{\Omega_{LP}} s$$

2.  $LP \rightarrow HP$

$$s \rightarrow \frac{\Omega_P \Omega_{HP}}{s}$$

3.  $LP \rightarrow BP$

$$s \rightarrow \Omega_p \frac{s^2 + \Omega_l \Omega_u}{s(\Omega_u - \Omega_l)}$$

4.  $LP \rightarrow BS$

$$s \rightarrow \Omega_p \frac{(\Omega_u - \Omega_l)}{s^2 + \Omega_l \Omega_u}$$

**NOTE:** Centre frequency  $\Omega_o = \sqrt{\Omega_l \Omega_u}$

$$|H(\Omega)|^2 = \frac{1}{1 + \left(\frac{\Omega}{\Omega_c}\right)^{2N}}$$

$$P_k = \pm e^{j(2k+1+N)\pi/2N}$$

$$N \geq \frac{\log\left[\frac{\frac{1}{A_s^2}-1}{\frac{1}{A_p^2}-1}\right]}{2\log\left(\frac{\Omega_s}{\Omega_p}\right)} \text{ or } \frac{\log\left[\frac{10^{0.1A_s db}-1}{10^{0.1A_p db}-1}\right]}{2\log\left(\frac{\Omega_s}{\Omega_p}\right)}$$

**IIT:**

$$\Omega = \frac{\omega}{T}$$

$$\frac{1}{S - P_k} \rightarrow \frac{1}{1 - e^{P_k T} Z^{-1}}$$

$$\frac{b}{(s+a)^2 + b^2} \rightarrow \frac{e^{-aT} \sin bT Z^{-1}}{1 - e^{-aT} \cos bT Z^{-1} + e^{-2aT} Z^{-2}}$$

**BLT:**

$$\Omega = \frac{2}{T} \tan\left(\frac{\omega}{2}\right)$$

$$s \rightarrow \frac{2}{T} \left( \frac{1 - Z^{-1}}{1 + Z^{-1}} \right)$$

## Chebyshev Filters

$$\Omega'_p = \frac{\Omega_p}{\Omega_p}$$

$$\Omega'_s = \frac{\Omega_s}{\Omega_p}$$

$$\epsilon = \sqrt{\frac{1}{A_p^2} - 1} = \sqrt{10^{0.1A_p db} - 1}$$

$$|H(\Omega)|_{in\ db} = -20 \log(\epsilon) - 6(N - 1) - 20 N \log \Omega'_s$$

$$S_k = \sigma_k + j\Omega_k$$

$$\sigma_k = -\sinh\left[\frac{1}{N} \sinh^{-1}\left(\frac{1}{\epsilon}\right)\right] \sin\left(\frac{2K-1}{2N}\right)\pi$$

$$\Omega_k = \cosh\left[\frac{1}{N} \sinh^{-1}\left(\frac{1}{\epsilon}\right)\right] \cos\left(\frac{2K-1}{2N}\right)\pi$$

$$k = b_o \text{ if } N \text{ odd}$$

$$\frac{b_o}{\sqrt{1 + \epsilon^2}} \text{ if } N \text{ even}$$

## Windowing techniques

$$h_d(n) = \frac{1}{2\pi} \int_{-3\pi/4}^{3\pi/4} e^{jw(n-\tau)}$$

- **Hanning: -44 db**

$$0.5 - 0.5 \cos\left(\frac{2\pi n}{N-1}\right)$$

- **Hamming; -53 db**

$$0.54 - 0.46 \cos\left(\frac{2\pi n}{N-1}\right)$$

- **Rectangular: -21 db**

$$1$$

Frequency Sampling:

$$H_d(\omega) = e^{-j\omega(\frac{N-1}{2})}; |\omega| \leq \omega_c$$

*To sample,  $\omega = \frac{2\pi k}{N}$*

$$h(n) = \frac{1}{N} \left[ H(0) + 2 \sum_{k=1}^{\frac{N-1}{2}} \operatorname{Re}(H(k) e^{j\frac{2\pi kn}{N}}) \right]$$

## Block Diagrams

*Direct form I  $\rightarrow$  Lattice :*

$$K_m = a_m(m)$$

$$a_{m-1}(i) = \frac{a_m(i) - a_m(m)a_m(m-i)}{1 - K_m^2}$$

*Lattice  $\rightarrow$  Direct form I :*

$$a_m(0) = 1$$

$$a_m(m) = K_m$$

*for  $m = 1, 2, \dots, M$*

$$a_m(i) = a_{m-1}(i) + a_m(m)a_{m-1}(m-i)$$