DSP

IDFT:

$$IDFT[X(K)] = rac{1}{N} \sum_{k=0}^{N-1} X(K) W_N^{-kn}$$

DFT:

$$DFT[x(n)] = \sum_{n=0}^{N-1} x(n)W_N^{kn}$$

IDFT using DFT: DFT of conjugate of the input, and the output is the conjugate divided by N.

$$IDFT = rac{1}{N}(DFT[X^*(K)])^*$$

Properties

Circular Time Shift:

Use IDFT equation to prove.

$$DFT[x(n-m)_N] = w_N^{mk}X(K)$$

Conjugate Symmetry:

• Use DFT equation to prove

$$X(K) = X^*(N - K)$$

Circular Frequency Shift:

• Use DFT equation to prove.

$$IDFT[X(K-l)_N] = w_N^{-ln} x(n)$$

Complex Conjugate:

Use DFT equation to prove

$$DFT[x^*(n)] = X^*((-K))_N = X^*(N-K)$$

Parseval Theorem:

$$\sum_{n=0}^{N-1} |x(n)|^2 = rac{1}{N} \sum_{k=0}^{N-1} |X(K)|^2$$

Periodicity:

$$x(n+N)$$

FFT

Number of multiplications = $rac{N}{2}log_2N$

Number of additions = Mog_2N

Composite Radix:

• 3X2 : then 3 summations going from 0 to 1

$$\sum_{n=0}^{1} x(3n)w_6^{3nk} + \sum_{n=0}^{1} x(3n+1)w_6^{(3n+1)k} + \sum_{n=0}^{1} x(3n+2)w_6^{(3n+2)k}$$

2X3: then 2 summations going from 0 to 2

$$\sum_{n=0}^2 x(2n)w_6^{2nk} + \sum_{n=0}^2 x(2n+1)w_6^{(2n+1)k}$$

Butterworth Filter

1. LP o LP

$$s
ightarrow rac{\Omega_p}{\Omega_{LP}} s$$

2. LP o HP

$$s
ightarrow rac{\Omega_P \Omega_{HP}}{s}$$

3. LP o BP

$$s o \Omega_p rac{s^2 + \Omega_l \Omega_u}{s(\Omega_u - \Omega_l)}$$

4. LP o BS

$$s o \Omega_p rac{(\Omega_u - \Omega_l)}{s^2 + \Omega_l \Omega_u}$$

NOTE: Centre frequency $\Omega_o = \sqrt{\Omega_l \Omega_u}$

$$|H(\Omega)|^2=rac{1}{1+(rac{\Omega}{\Omega_c})^{2N}}$$

$$P_k=\pm~e^{j(2k+1+N)\pi/2N}$$

$$N \geq rac{log[rac{rac{1}{A_s^2}-1}{rac{1}{A_p^2}-1}]}{2log(rac{\Omega_s}{\Omega_p})} or rac{log[rac{10^{0.1A_sdb}-1}{10^{0.1A_pdb}-1}]}{2log(rac{\Omega_s}{\Omega_p})}$$

IIT:

$$egin{aligned} \Omega &= rac{\omega}{T} \ &rac{1}{S-P_k}
ightarrow rac{1}{1-e^{P_kT}Z^{-1}} \ &rac{b}{(s+a)^2+b^2}
ightarrow rac{e^{-aT}sinbT}{1-e^{-aT}cosbT}rac{Z^{-1}}{Z^{-1}} \end{aligned}$$

BLT:

$$\Omega = rac{2}{T}tan(rac{\omega}{2})$$

$$s o rac{2}{T}(rac{1-Z^{-1}}{1+Z^{-1}})$$

Chebysev Filters

$$egin{aligned} \Omega_p' &= rac{\Omega_p}{\Omega_p} \ \Omega_s' &= rac{\Omega_s}{\Omega_p} \ &\epsilon &= \sqrt{rac{1}{A_p^2} - 1} = \sqrt{10^{0.1 A_p db} - 1} \ |H(\Omega)|_{in \ db} &= -20 log(\epsilon) - 6(N-1) - 20 \ N \ log \ \Omega_s' \ &S_k &= \sigma_k + j \Omega_k \ &\sigma_k &= -sinh[rac{1}{N} sinh^{-1}(rac{1}{\epsilon})] sin(rac{2K-1}{2N})\pi \ &\Omega_k &= cosh[rac{1}{N} sinh^{-1}(rac{1}{\epsilon})] cos(rac{2K-1}{2N})\pi \ &k &= b_o \ if \ N \ odd \ &rac{b_o}{\sqrt{1+\epsilon^2}} \ if \ N \ even \end{aligned}$$

Windowing techniques

$$h_d(n) = rac{1}{2\pi} \int_{-3\pi/4}^{3\pi/4} e^{jw(n- au)}$$

Hanning: -44 db

$$0.5 - 0.5 \; cos(rac{2\pi n}{N-1})$$

Hamming; -53 db

$$0.54 - 0.46~cos(rac{2\pi n}{N-1})$$

• Rectangular: -21 db

1

Frequency Sampling:

$$H_d(\omega) = e^{-j\omega(rac{N-1}{2})}; \; |\omega| \leq \; \omega_c \ To \; sample, \; \; \omega = rac{2\pi k}{N}$$

$$h(n) = rac{1}{N}[H(0) + 2\sum_{k=1}^{rac{N-1}{2}} Re(H(k)e^{jrac{2\pi kn}{N}})]$$

Block Diagrams

 $Direct\ form\ I o Lattice:$

$$K_m=a_m(m)$$
 $a_{m-1}(i)=rac{a_m(i)-a_m(m)a_m(m-i)}{1-K_m^2}$

 $Lattice \rightarrow Direct\ form\ I:$

$$egin{aligned} a_m(0) &= 1 \ a_m(m) &= K_m \ for \ m &= 1, 2, ..., M \end{aligned}$$