





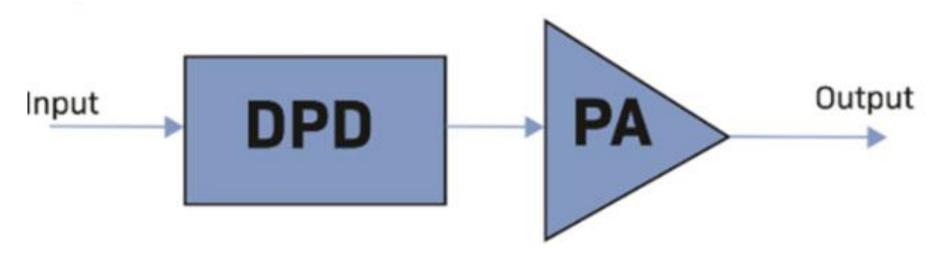


Digital Pre-Distortion

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Introduction

- Power amplifiers (PAs), are inherently nonlinear systems.
- Nonlinearity causes in-band distortion and a spectral regrowth.
- The role of the Digital Pre Distortion (DPD) is to distort the signal in a way that will be, in turn, compensated by the PA.



DPD and PA system

Goals

- The total response of the Pre-distorter and PA will be linear (reduce am/am, am/pm distortions and out of band emission) up to some saturation voltage.
- Using both classical and deep learning methods

AM/AM, AM/PM, OOB emission

AM/AM is the relation between the amplitude of the input signal and the amplitude of the output signal. Ideally, this relation should be linear, but due to nonlinear components in the PA, is usually nonlinear

between the

difference between

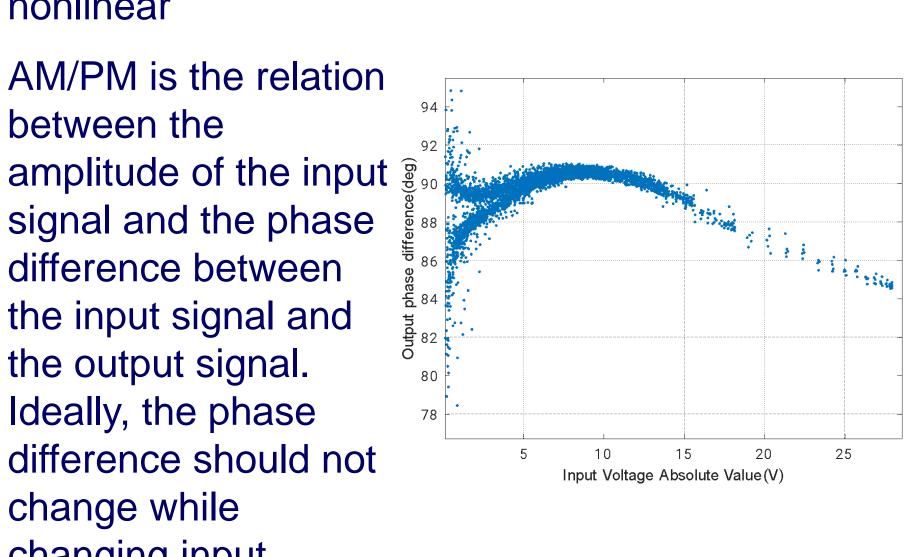
the output signal.

Ideally, the phase

change while

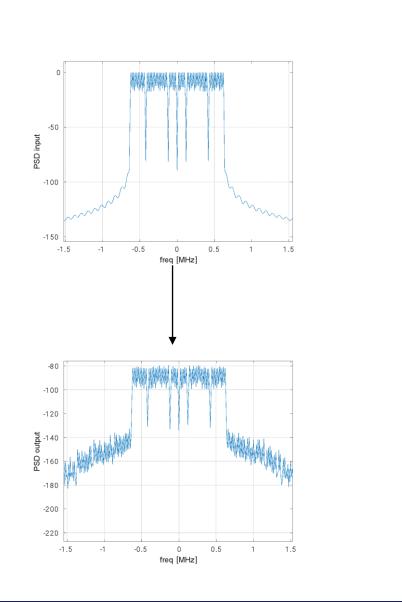
amplitude.

changing input



Input Voltage Absolute Value(V)

Out Of Band (OOB) emission is defined as the difference between the bandwidth of the signal after the PA and the bandwidth of the original signal.



Volterra series as a PA & **DPD** model

Volterra series are used in order to model systems that are both nonlinear and have memory. Memory Polynomial model is a simplification of the full Volterra model, with less coefficients

$$y_{MP}(n) = \sum_{k=0}^{K-1} \sum_{m=0}^{M-1} h_{km} x(n-m) |x(n-m)|^k$$

This model can be represented efficiently in matrix form as:

$$y = X\theta_{MP}$$

$$\begin{bmatrix} y(n) \\ \vdots \\ y(n+N) \end{bmatrix} = \begin{bmatrix} x(n) & \cdots & x(n-m)|x(n-m)|^k \\ \vdots & \ddots & \vdots \\ x(n+N) & \cdots & x(n+N-m)|x(n+N-m)|^k \end{bmatrix} \begin{bmatrix} h_{00} \\ \vdots \\ h_{km} \end{bmatrix}$$

The coefficients vector θMP can be found using the well-known method of minimizing the least squares error between the calculated output and measured output.

$$\hat{\theta}_{MP} = (X^H X)^{-1} X y$$

At the same way, we can model our DPD using Volterra series:

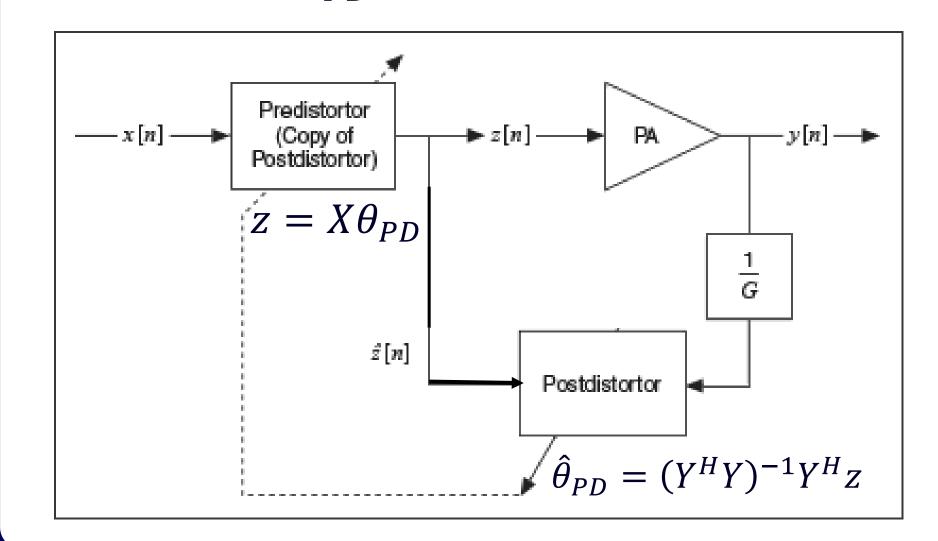
$$x = Y\theta_{PD}$$

In order to calculate the coefficient array Θ *PD*, the same set of measurements is used, but it is scaled and used in reverse order:

$$y_{scaled} = y \cdot \frac{\max(|X_{in}|)}{\max(|y|)}$$

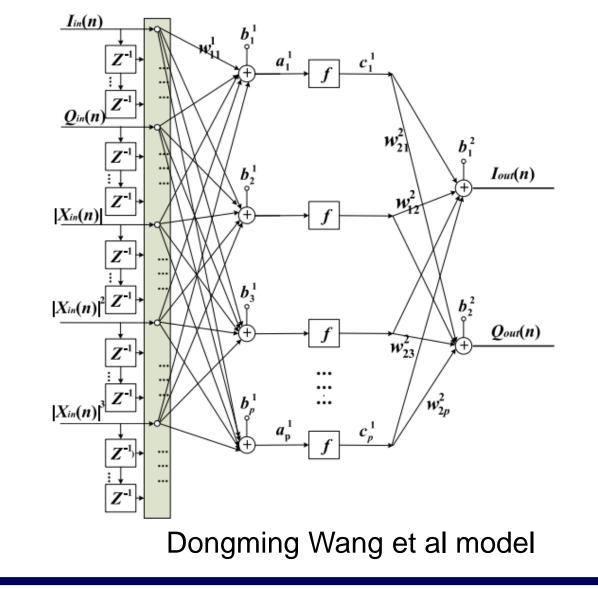
The coefficients vector $\theta_{-}MP$ can be found using the same algorithm as used at the pa modelling:

$$\hat{\theta}_{PD} = (Y^H Y)^{-1} Y^H x$$



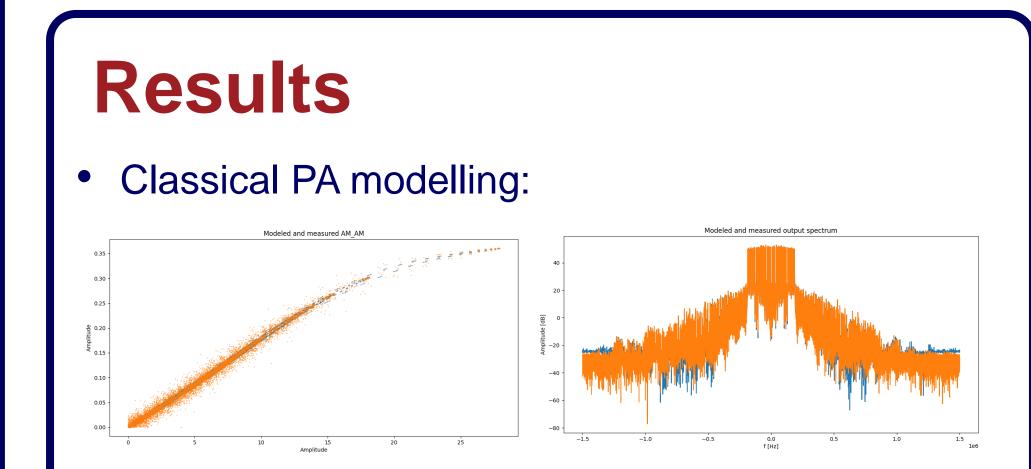
NNs as a PA & DPD model

- Due to their strong adaptive nature and approximation capability, NNs are very attractive for the behavioral modeling of PAs.
- The proposed model, which takes into account both nonlinear and memory effects:

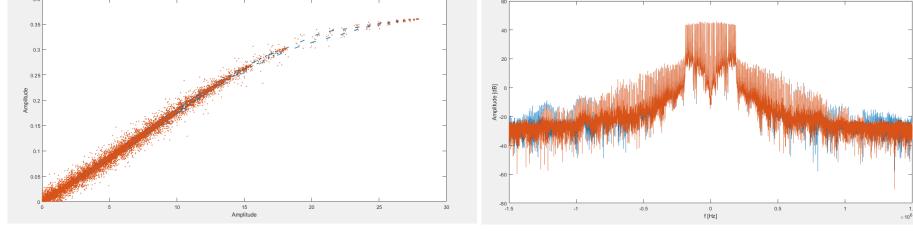


PA NN proposed architecture feedforward (22-40-2) NN modelled PA **DPD NN proposed architecture**

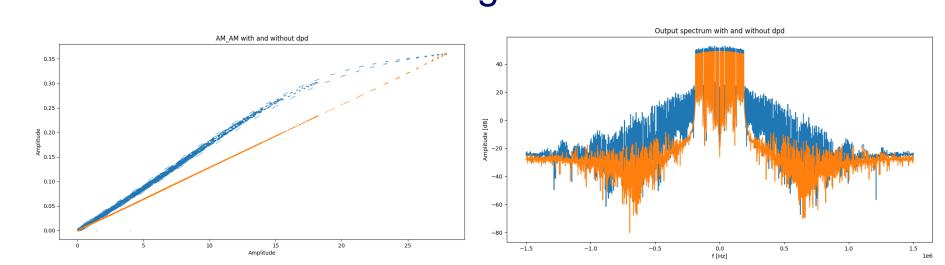
feedforward (22-30-40-30-20-10-2) NN modelled DPD



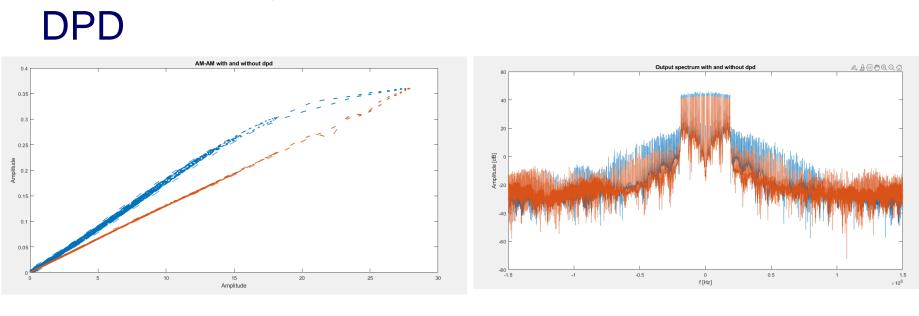
• feedforward (22-40-2) NN modelled PA:



Classical DPD modelling:



feedforward (22-30-40-30-20-10-2) NN modelled DPD



 In order to test AM/AM performance, we calculated the normalized MMSE between desired output (Gain*input) and modeled output with DPD, and R^2 of linear regression of the AM/AM curvature.

DPD Model	MMSE	R^2	OOB emission
22-30-40-30-20-10-2	-17.4 dB	0.9996	-8 dB
Classical method	-22 dB	0.9998	-19.5 dB

Conclusions

- The methods described in this report can model a real power amplifier and a digital pre distorter that their total response together will be, approximately, linear
- classical methods should be considered and tested seriously before using NN solutions.
- We used data that contains relatively low content of high power, and was relatively narrowband. In the future this methods should be examined in the communication lab with more suitable measurements.