



PENGOLAHAN SINYAL DIGITAL

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THE DISCRETE-TIME FOURIER ANALYSIS

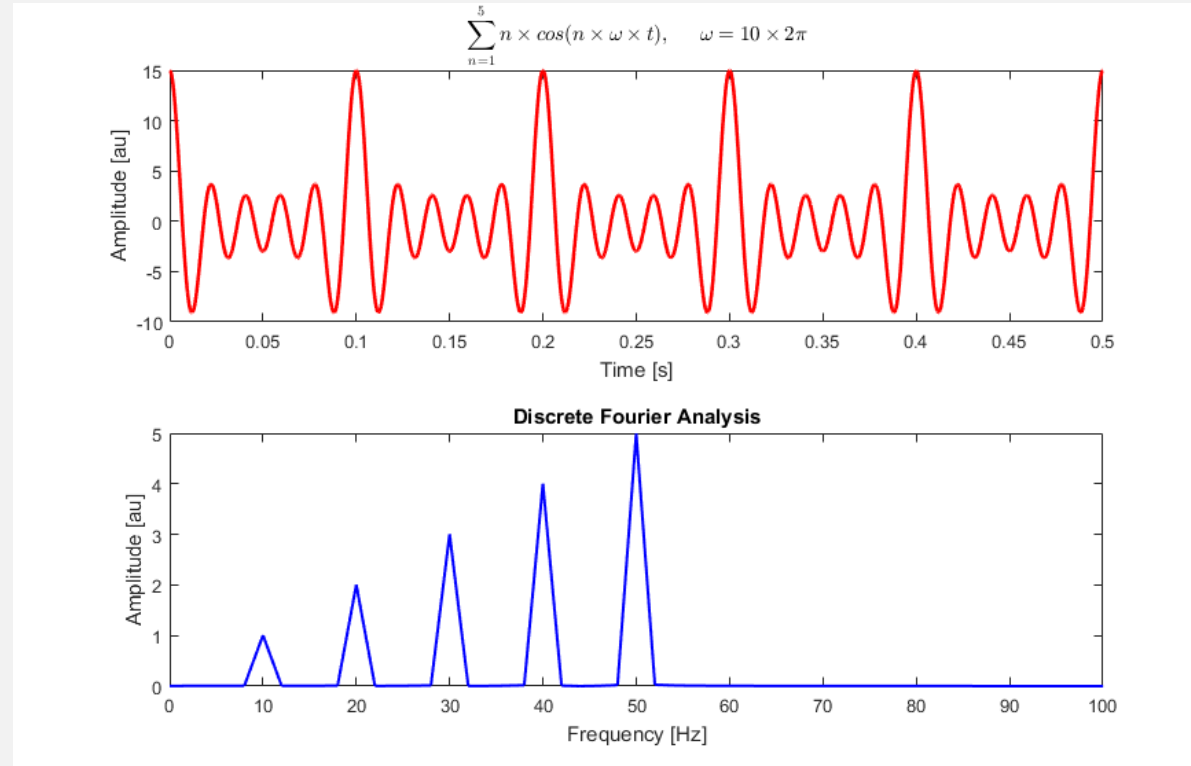
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FOURIER TRANSFORM

- Fourier transform decompose signal into its frequency component



DISCRETE-TIME FOURIER TRANSFORM

- A linear and time-invariant system can be represented using its response to the unit sample sequence that

$$x(n) \longrightarrow \boxed{h(n)} \longrightarrow y(n) = h(n) * x(n)$$

- A linear and time-invariant system can be represented using the complex exponential signal set

$$\{e^{j\omega n}\}$$

DISCRETE-TIME FOURIER TRANSFORM

- General form of DTFT

$$X(e^{j\omega}) \triangleq \mathcal{F}[x(n)] = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$

- The inverse discrete-time Fourier transform (IDTFT) of $X(e^{j\omega})$ is given by

$$x(n) \triangleq \mathcal{F}^{-1}[X(e^{j\omega})] = \frac{1}{2\pi} \sum_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

DISCRETE-TIME FOURIER TRANSFORM

- The operator $\mathcal{F}[\cdot]$ transforms a discrete signal $x(n)$ into a complex-valued continuous function $X(e^{j\omega})$ of real variable ω , called a digital frequency, which is measured in radians/sample

EXAMPLE

- Determine the discrete-time Fourier transform of $x(n) = (0.5)^n u(n)$.
- The sequence $x(n)$ is absolutely summable; therefore its discrete-time Fourier transform exists.

$$\begin{aligned} X(e^{j\omega}) &= \sum_{-\infty}^{\infty} x(n) e^{-j\omega n} = \sum_0^{\infty} (0,5)^n e^{-j\omega n} \\ &= \sum_0^{\infty} (0,5e^{-j\omega})^n = \frac{1}{1 - 0,5e^{-j\omega}} = \frac{e^{j\omega}}{e^{j\omega} - 0,5} \end{aligned}$$

THE GEOMETRIC SERIES

- A one-sided exponential sequence of the form

$$\{\alpha^n, \quad n \geq 0\}$$

- The series converges for $|\alpha| < 1$, while the sum of its components converges to

$$\sum_{n=0}^{\infty} \alpha^n \rightarrow \frac{1}{1-\alpha}, \quad \text{for } |\alpha| < 1$$

DTFT PAIRS

<i>Signal Type</i>	<i>Sequence $x(n)$</i>	<i>DTFT $X(e^{j\omega})$, $-\pi \leq \omega \leq \pi$</i>
Unit impulse	$\delta(n)$	1
Constant	1	$2\pi\delta(\omega)$
Unit step	$u(n)$	$\frac{1}{1 - e^{-j\omega}} + \pi\delta(\omega)$
Causal exponential	$\alpha^n u(n)$	$\frac{1}{1 - \alpha e^{-j\omega}}$
Complex exponential	$e^{j\omega_0 n}$	$2\pi\delta(\omega - \omega_0)$
Cosine	$\cos(\omega_0 n)$	$\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$
Sine	$\sin(\omega_0 n)$	$j\pi[\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$
Double exponential	$\alpha^{ n } u(n)$	$\frac{1 - \alpha^2}{1 - 2\alpha \cos(\omega) + \alpha^2}$

PROPERTIES OF THE DTFT

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PROPERTIES OF THE DTFT

- **Linearity:** The discrete-time Fourier transform is a linear transformation

$$\mathcal{F}[\alpha x_1(n) + \beta x_2(n)] = \alpha \mathcal{F}[x_1(n)] + \beta \mathcal{F}[x_2(n)]$$

- for every α , β , $x_1(n)$, and $x_2(n)$.
- **Time shifting:** A shift in the time domain corresponds to the phase shifting.

$$\mathcal{F}[x(n-k)] = X(e^{j\omega})e^{-j\omega k}$$

PROPERTIES OF THE DTFT

- **Frequency shifting:** Multiplication by a complex exponential corresponds to a shift in the frequency domain

$$\mathcal{F}\left[x(n)e^{j\omega_0 n}\right] = X\left(e^{j(\omega-\omega_0)}\right)$$

- **Conjugation:** Conjugation in the time domain corresponds to the folding and conjugation in the frequency domain

$$\mathcal{F}\left[x^*(n)\right] = X^*\left(e^{-j\omega}\right)$$

PROPERTIES OF THE DTFT

- **Folding:** Folding in the time domain corresponds to the folding in the frequency domain

$$\mathcal{F}[x(-n)] = X(e^{-j\omega})$$

- **Symmetries in real sequences:** We have already studied the conjugate symmetry of real sequences. These real sequences can be decomposed into their even and odd parts

$$x(n) = x_e(n) + x_o(n)$$

$$\mathcal{F}[x_e(n)] = \text{Re}[X(e^{j\omega})]$$

$$\mathcal{F}[x_o(n)] = j \text{Im}[X(e^{j\omega})]$$

PROPERTIES OF THE DTFT

- **Convolution:** This is one of the most useful properties that makes system analysis convenient in the frequency domain

$$\mathcal{F}[x_1(n) * x_2(n)] = \mathcal{F}[x_1(n)] \mathcal{F}[x_2(n)] = X_1(e^{j\omega}) X_2(e^{j\omega})$$

- **Multiplication:** This is a dual of the convolution property

$$\begin{aligned} \mathcal{F}[x_1(n) \cdot x_2(n)] &= \mathcal{F}[x_1(n)] \circledast \mathcal{F}[x_2(n)] \\ &\triangleq \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(e^{j\theta}) X_2(e^{j\omega-\theta}) d\theta \end{aligned}$$

- This convolution-like operation is called a *periodic convolution* and hence denoted by \circledast

PROPERTIES OF THE DTFT

- **Energy:** The energy of the sequence $x(n)$ can be written as

$$\begin{aligned}\mathcal{E}_x &= \sum_{-\infty}^{\infty} |x(n)|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega \\ &= \int_0^{\pi} \frac{|X(e^{j\omega})|^2}{\pi} d\omega\end{aligned}$$

EXAMPLE 1

- Verify the linearity property of DTFT using real-valued finite duration sequences

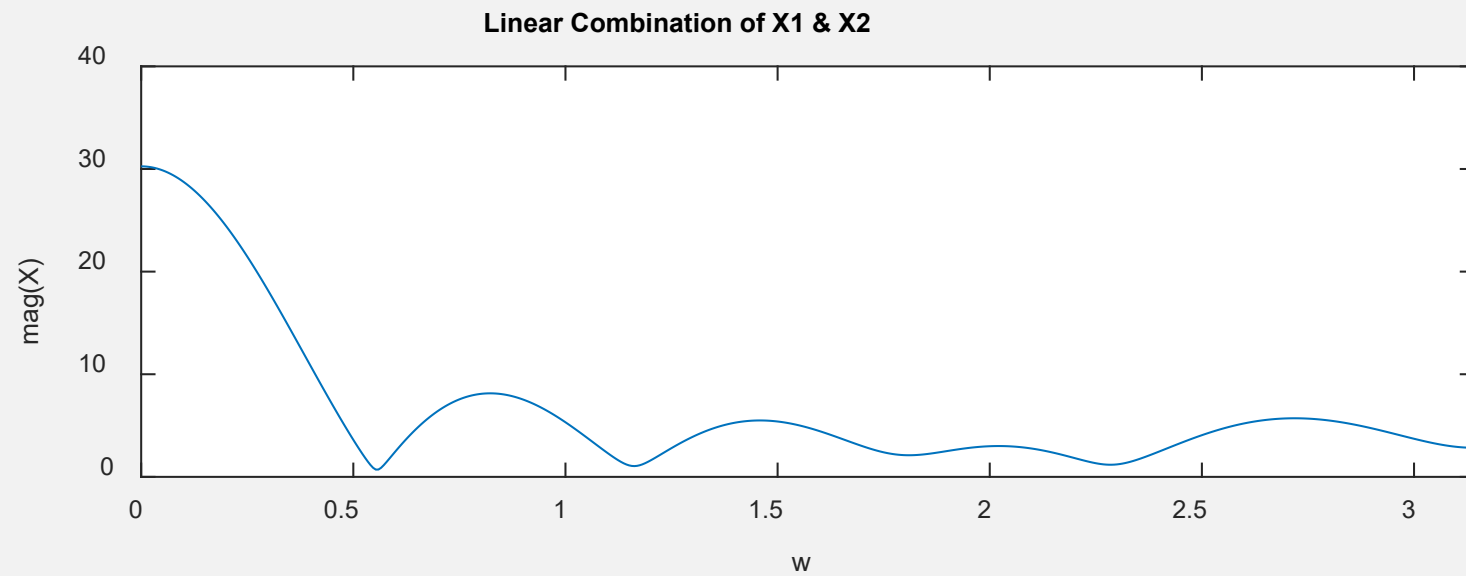
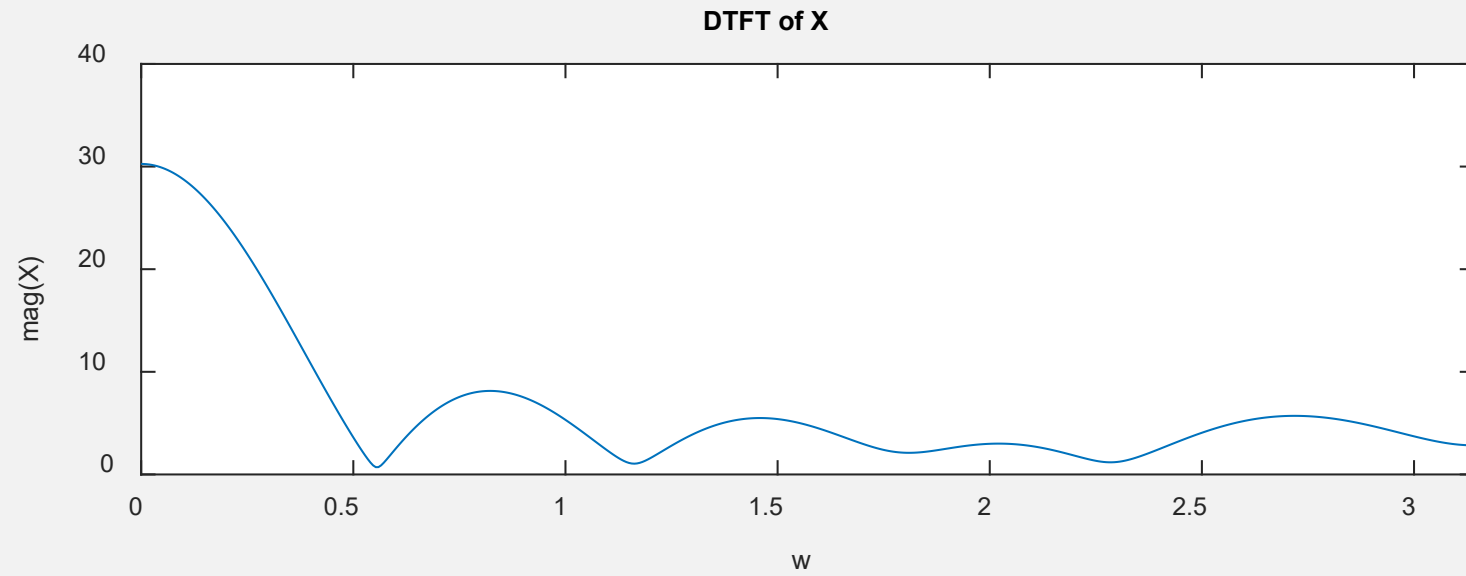
$$\mathcal{F}[\alpha x_1(n) + \beta x_2(n)] = \alpha \mathcal{F}[x_1(n)] + \beta \mathcal{F}[x_2(n)]$$

- Which $x_1(n)$ and $x_2(n)$ are two random sequences uniformly distributed between $[0, 1]$ over $0 \leq n \leq 10$.

EXAMPLE 1

```
x1 = rand(1,11); x2 = rand(1,11); n = 0:10;  
alpha = 2; beta = 3; k = 0:500; w = (pi/500)*k;  
X1 = x1 * (exp(-j*pi/500)).^(n'*k); % DTFT of x1  
X2 = x2 * (exp(-j*pi/500)).^(n'*k); % DTFT of x2  
x = alpha*x1 + beta*x2; % Linear combination of x1 & x2  
X = x * (exp(-j*pi/500)).^(n'*k); % DTFT of x  
  
% verification  
X_check = alpha*X1 + beta*X2; % Linear Combination of X1 & X2  
error = max(abs(X-X_check)) % Difference
```

EXAMPLE 1



EXAMPLE 2

- Verify the sample shift property

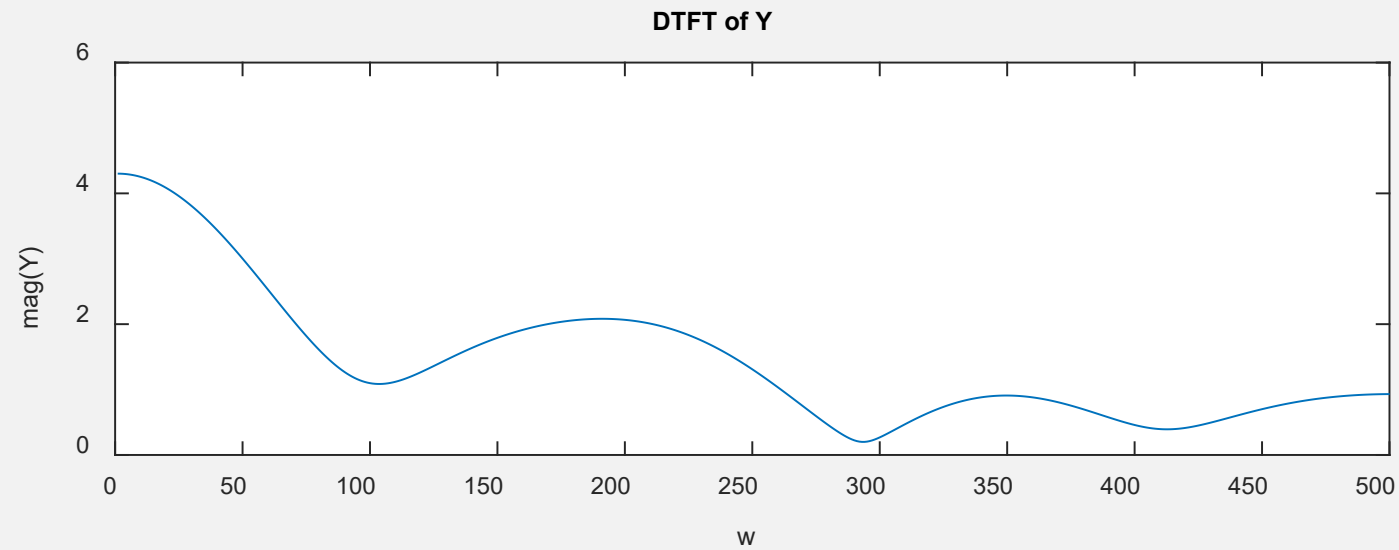
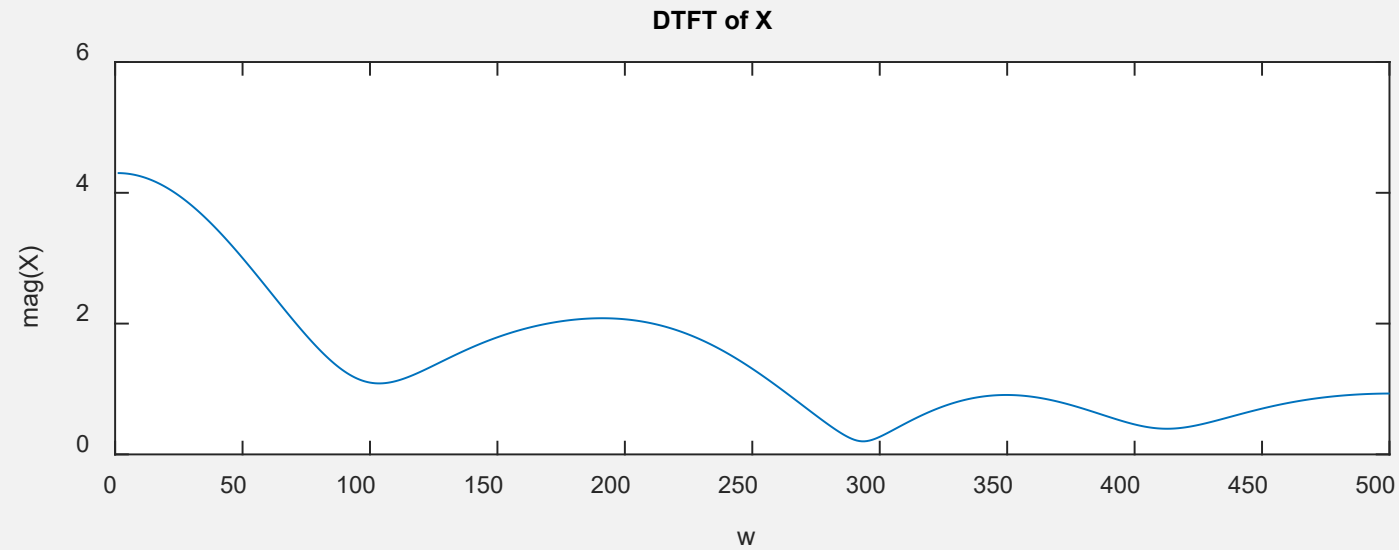
$$\mathcal{F}[x(n-k)] = X(e^{j\omega})e^{-j\omega k}$$

- which $x(n)$ is a random sequence uniformly distributed between $[0, 1]$ over $0 \leq n \leq 10$ and let $y(n) = x(n-2)$.

EXAMPLE 2

```
x = rand(1,11); n = 0:10;  
k = 0:500; w = (pi/500)*k;  
X = x * (exp(-j*pi/500)).^(n'*k); % DTFT of x  
  
% signal shifted by two samples  
y = x; m = n+2;  
Y = y * (exp(-j*pi/500)).^(m'*k); % DTFT of y  
  
% verification  
Y_check = (exp(-j*2).^w).*X; % multiplication by exp(-j2w)  
error = max(abs(Y-Y_check)) % Difference
```


EXAMPLE 2



EXAMPLE 3

- Verify the frequency shift property

$$\mathcal{F}\left[x(n)e^{j\omega_0 n}\right] = X\left(e^{j(\omega-\omega_0)}\right)$$

- Which

$$x(n) = \cos(\pi n / 2), \quad 0 \leq n \leq 100$$

$$y(n) = e^{j\pi n/4} x(n)$$

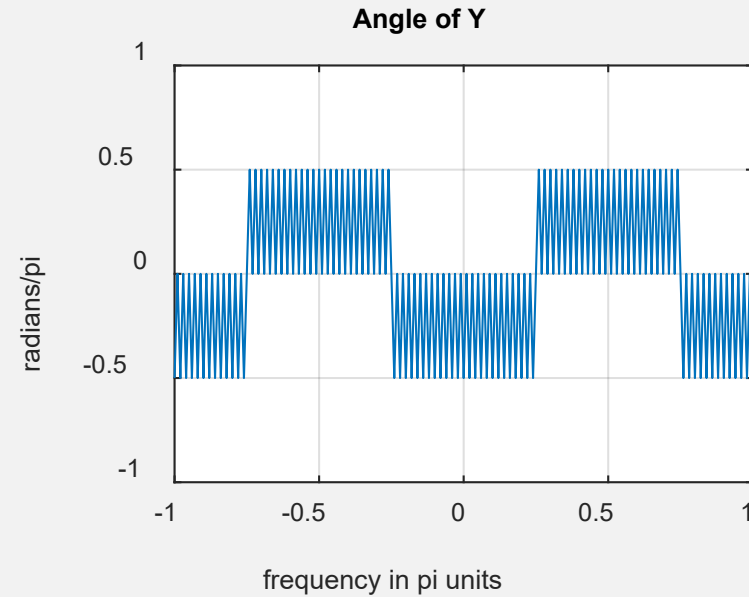
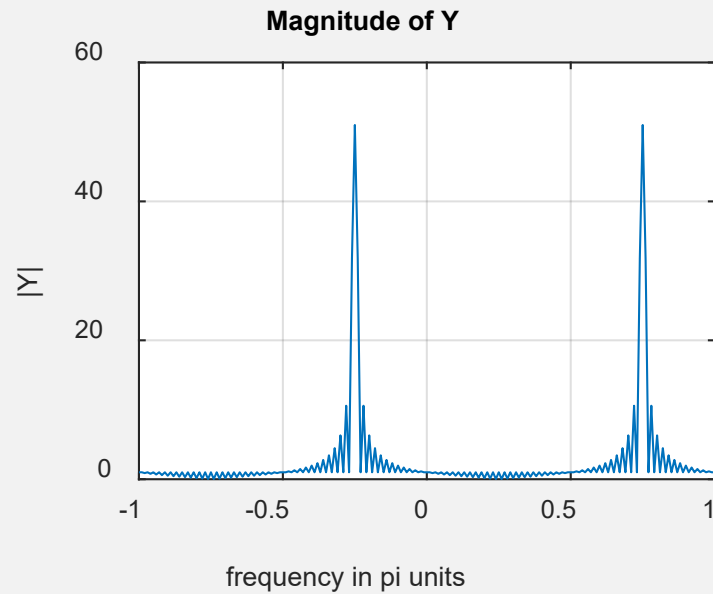
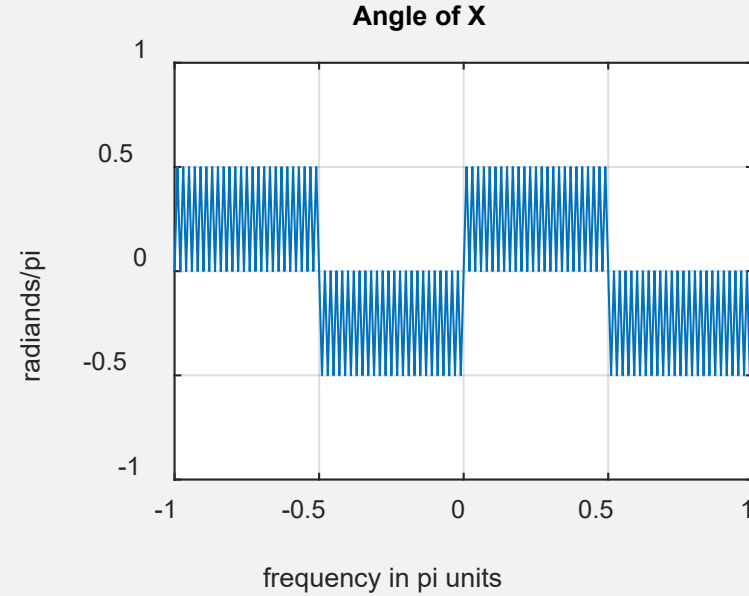
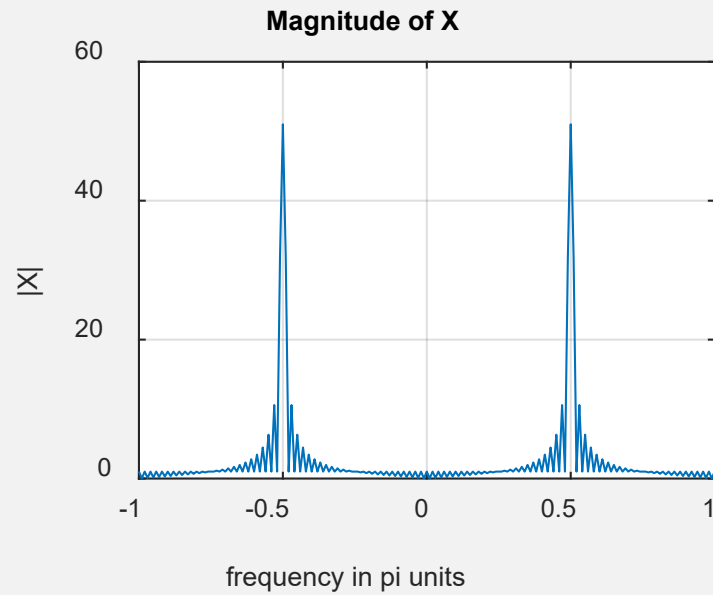
EXAMPLE 3

```
n = 0:100; x = cos(pi*n/2);  
k = -100:100; w = (pi/100)*k; % frequency between -pi and +pi  
X = x * (exp(-j*pi/100)).^(n'*k); % DTFT of x  
y = exp(j*pi*n/4).*x; % signal multiplied by exp(j*pi*n/4)  
Y = y * (exp(-j*pi/100)).^(n'*k); % DTFT of y
```

EXAMPLE 3

```
% Graphical verification
subplot(2,2,1); plot(w/pi,abs(X)); grid; axis([-1,1,0,60])
xlabel('frequency in pi units'); ylabel('|X|')
title('Magnitude of X')
subplot(2,2,2); plot(w/pi,angle(X)/pi); grid; axis([-1,1,-1,1])
xlabel('frequency in pi units'); ylabel('radiands/pi')
title('Angle of X')
subplot(2,2,3); plot(w/pi,abs(Y)); grid; axis([-1,1,0,60])
xlabel('frequency in pi units'); ylabel('|Y|')
title('Magnitude of Y')
subplot(2,2,4); plot(w/pi,angle(Y)/pi); grid; axis([-1,1,-1,1])
xlabel('frequency in pi units'); ylabel('radians/pi')
title('Angle of Y')
```

EXAMPLE 3



EXAMPLE 4

- Verify the conjugation property

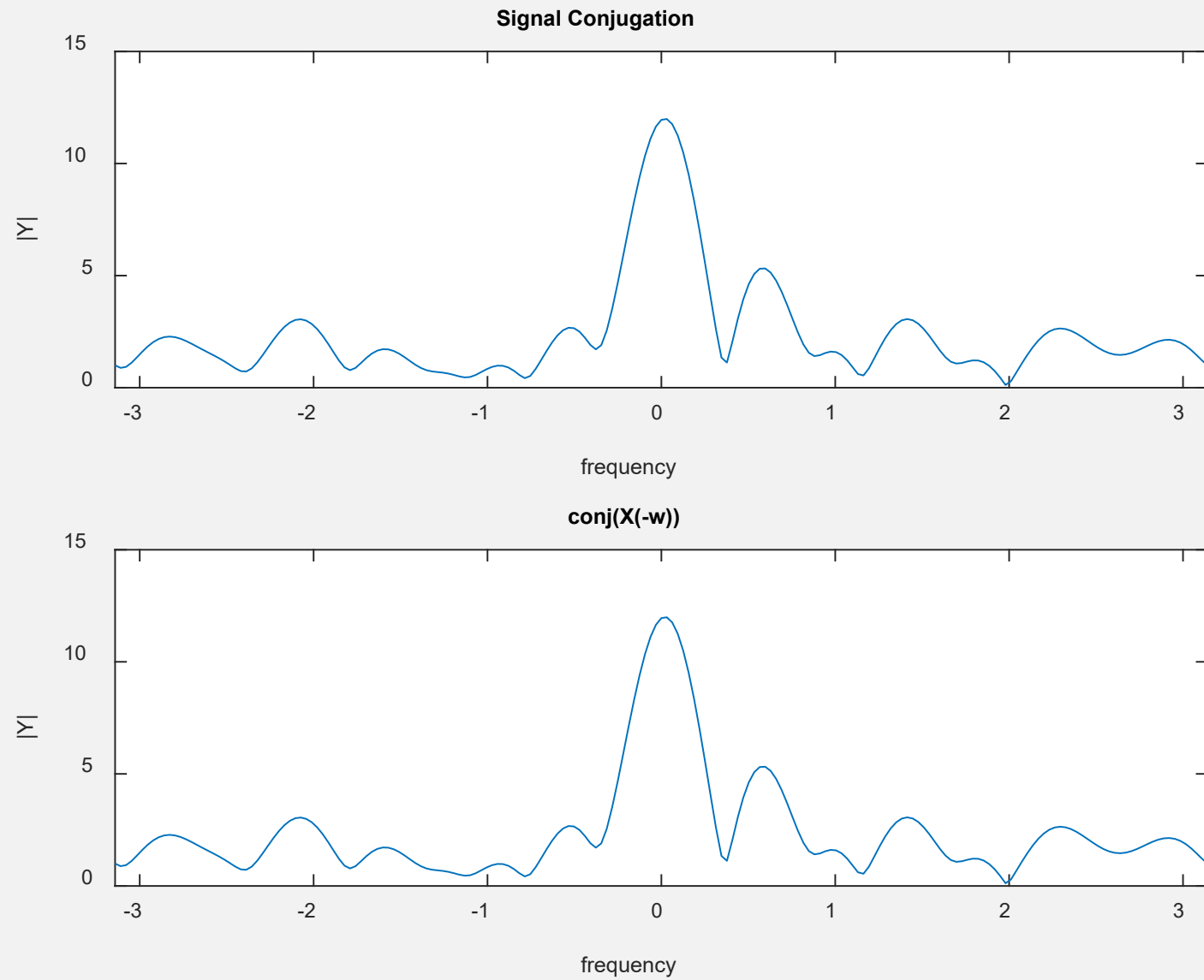
$$\mathcal{F}[x^*(n)] = X^*(e^{-j\omega})$$

- Which $x(n)$ is a complex-valued random sequence over $-5 \leq n \leq 10$ with real and imaginary parts uniformly distributed between $[0, 1]$

EXAMPLE 4

```
n = -5:10; x = rand(1,length(n)) + j*rand(1,length(n));  
k = -100:100; w = (pi/100)*k; % frequency between -pi and +pi  
X = x * (exp(-j*pi/100)).^(n'*k); % DTFT of x  
  
% conjugation property  
y = conj(x); % signal conjugation  
Y = y * (exp(-j*pi/100)).^(n'*k); % DTFT of y  
  
% verification  
Y_check = conj(fliplr(X)); % conj(X(-w))  
error = max(abs(Y-Y_check)) % Difference
```


EXAMPLE 4



EXAMPLE 5

- Verify the folding property

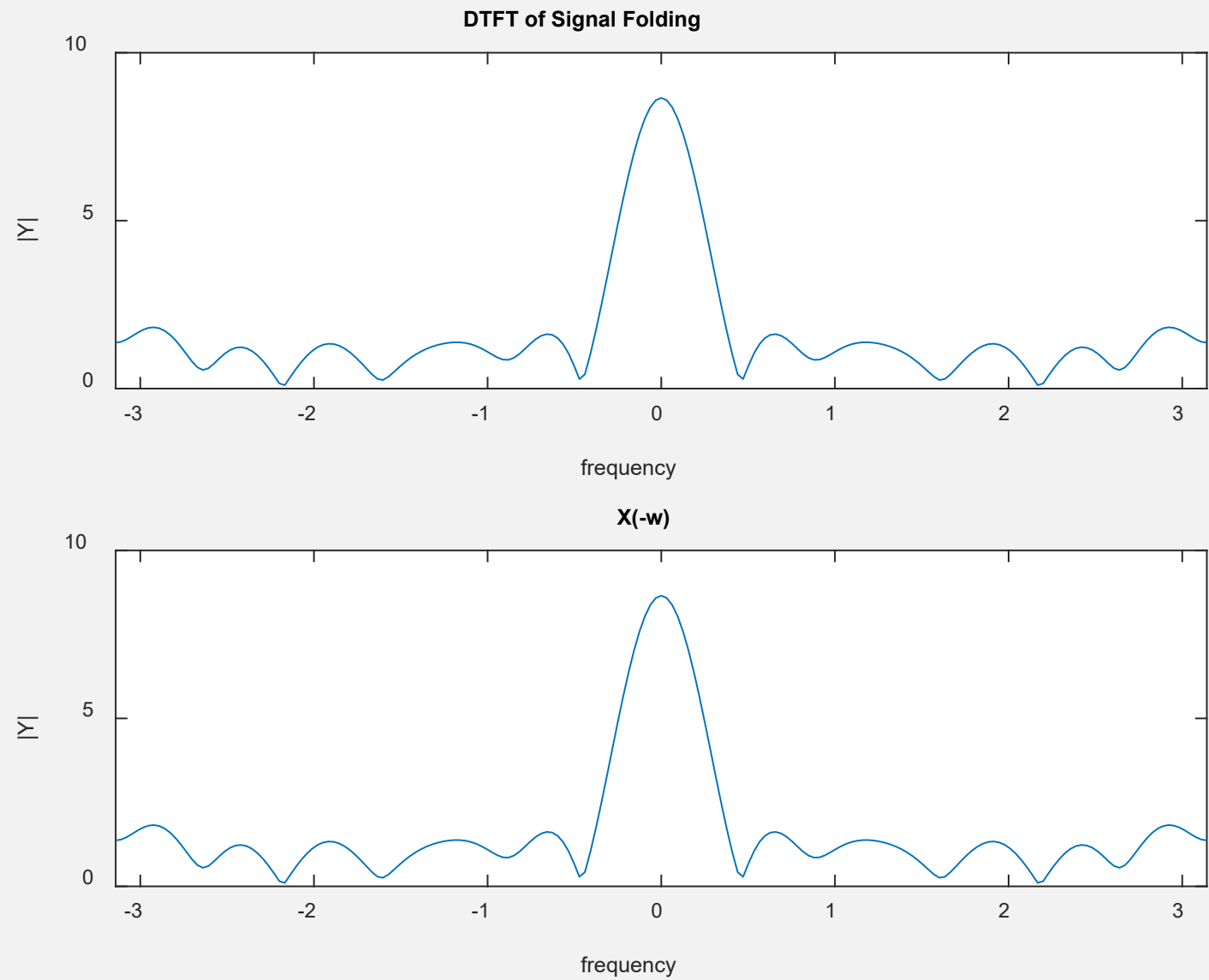
$$\mathcal{F}[x(-n)] = X(e^{-j\omega})$$

- Which $x(n)$ be a random sequence over $-5 \leq n \leq 10$ uniformly distributed between $[0, 1]$.

EXAMPLE 5

```
n = -5:10; x = rand(1,length(n));  
k = -100:100; w = (pi/100)*k; % frequency between -pi and +pi  
X = x * (exp(-j*pi/100)).^(n'*k); % DTFT of x  
  
% folding property  
y = fliplr(x); m = -fliplr(n); % signal folding  
Y = y * (exp(-j*pi/100)).^(m'*k); % DTFT of y  
  
% verification  
Y_check = fliplr(X); % X(-w)  
error = max(abs(Y-Y_check)) % Difference
```

EXAMPLE 5



EXAMPLE 6

- Verify the symmetry property of real signals

$$x(n) = x_e(n) + x_o(n)$$

$$\mathcal{F}[x_e(n)] = \text{Re}[X(e^{j\omega})]$$

$$\mathcal{F}[x_o(n)] = j \text{Im}[X(e^{j\omega})]$$

- Which $x(n) = \sin(\pi n/2)$, $-5 \leq n \leq 10$

EXAMPLE 6

```
n = -5:10; x = sin(pi*n/2);  
k = -100:100; w = (pi/100)*k; % frequency between -pi and +pi  
X = x * (exp(-j*pi/100)).^(n'*k); % DTFT of x  
% signal decomposition  
[xe,xo,m] = evenodd(x,n); % even and odd parts  
XE = xe * (exp(-j*pi/100)).^(m'*k); % DTFT of xe  
XO = xo * (exp(-j*pi/100)).^(m'*k); % DTFT of xo  
% verification  
XR = real(X); % real part of X  
error1 = max(abs(XE-XR)) % Difference  
XI = imag(X); % imag part of X
```

EXAMPLE 6

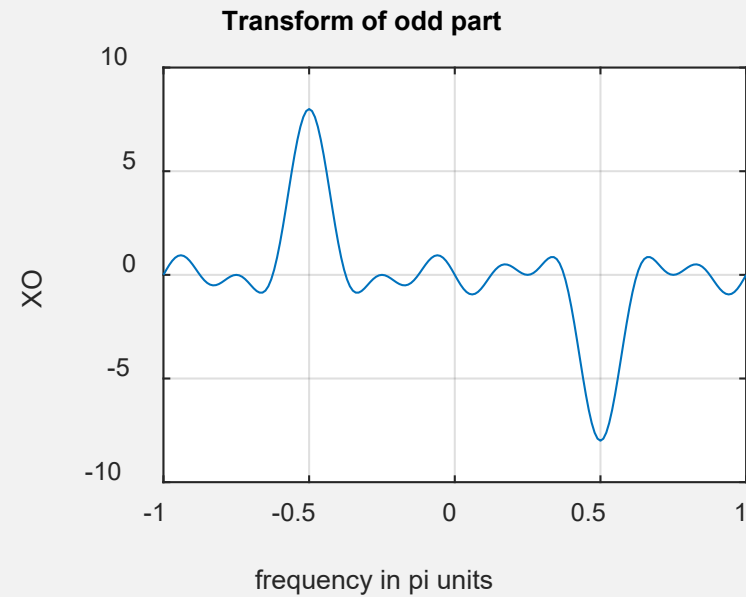
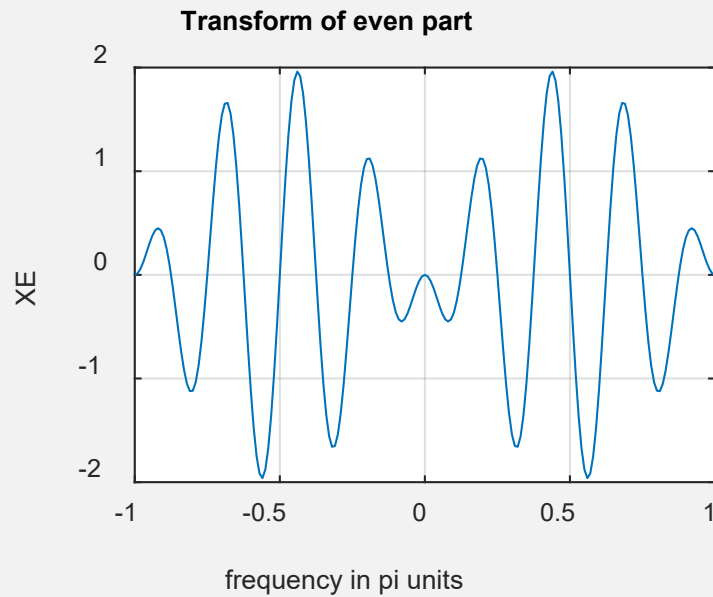
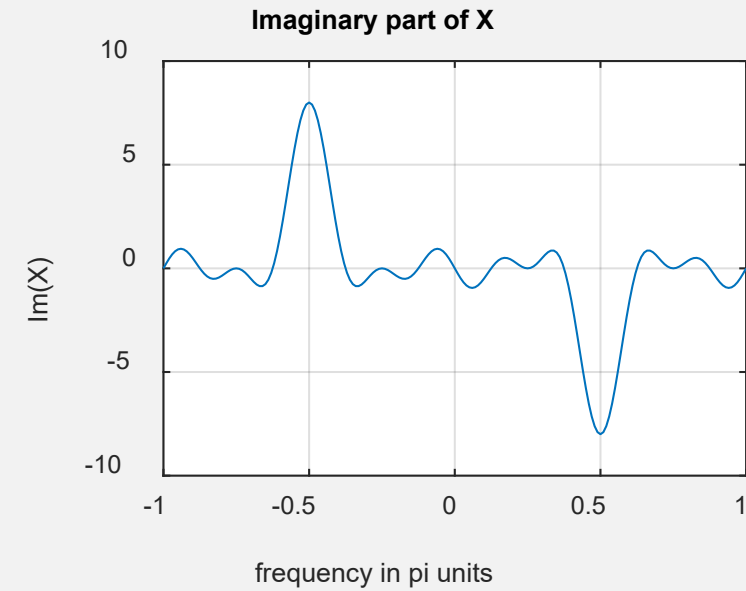
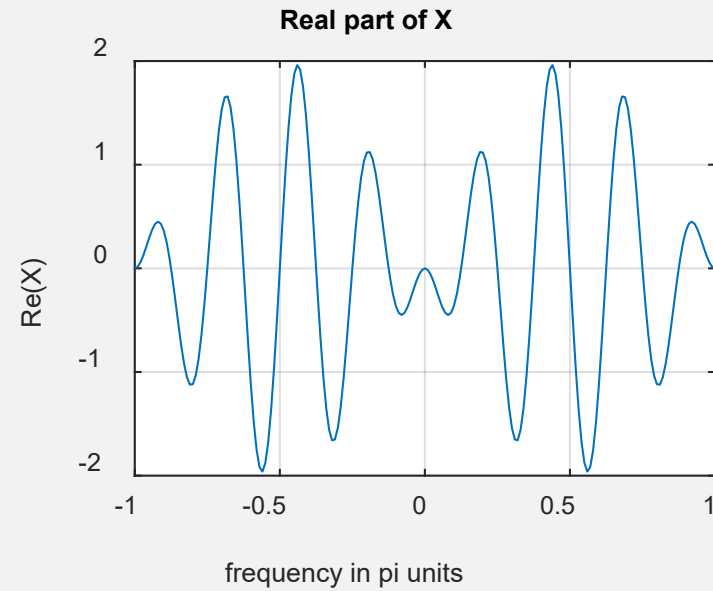
```
error2 = max(abs(XO-j*XI)) % Difference
error2 = 1.8033e-019

% graphical verification
subplot(2,2,1); plot(w/pi,XR); grid; axis([-1,1,-2,2])
xlabel('frequency in pi units'); ylabel('Re(X)');
title('Real part of X')
subplot(2,2,2); plot(w/pi,XI); grid; axis([-1,1,-10,10])
xlabel('frequency in pi units'); ylabel('Im(X)');
title('Imaginary part of X')
```


EXAMPLE 6

```
subplot(2,2,3); plot(w/pi,real(XE)); grid; axis([-1,1,-2,2])  
xlabel('frequency in pi units'); ylabel('XE');  
title('Transform of even part')  
subplot(2,2,4); plot(w/pi,imag(XO)); grid; axis([-1,1,-10,10])  
xlabel('frequency in pi units'); ylabel('XO');  
title('Transform of odd part')
```

MATLAB SCRIPT





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