



# PENGOLAHAN SINYAL DIGITAL

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# SYSTEM REPRESENTATION IN THE Z-DOMAIN

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# THE SYSTEM FUNCTION

- The system function  $H(z)$  is given by

$$H(z) \triangleq \mathcal{Z}[h(n)] = \sum_{n=-\infty}^{\infty} h(n) z^{-n}; \quad R_{h-} < |z| < R_{h+}$$

- Using the convolution property of the z-transform, the output transform  $Y(z)$  is given by

$$Y(z) = H(z)X(z) \quad : \quad \text{ROC}_y = \text{ROC}_h \cap \text{ROC}_x$$

# THE SYSTEM FUNCTION

- Therefore a linear and time invariant system can be represented in the z-domain by

$$X(z) \rightarrow \boxed{H(z)} \rightarrow Y(z) = H(z)X(z)$$

# SYSTEM FUNCTION FROM THE DIFFERENCE EQUATION REPRESENTATION

- LTI systems are described by a difference equation

$$y(n) + \sum_{k=1}^N a_k y(n-k) = \sum_{\ell=1}^M b_{\ell} x(n-\ell)$$

- System Function Representation

$$Y(z) + \sum_{k=1}^N a_k z^{-k} Y(z) = \sum_{\ell=0}^M b_{\ell} z^{-\ell} X(z)$$

$$H(z) \triangleq \frac{Y(z)}{X(z)} = \frac{\sum_{\ell=0}^M b_{\ell} z^{-\ell}}{1 + \sum_{k=1}^N a_k z^{-k}} = \frac{B(z)}{A(z)} = \frac{b_o z^{-M} \left( z^M + \dots + \frac{b_M}{b_o} \right)}{z^{-N} \left( z^N + \dots + a_N \right)}$$

# SYSTEM FUNCTION FROM THE DIFFERENCE EQUATION REPRESENTATION

- After factorization

$$H(z) = b_o z^{N-M} \frac{\prod_{\ell=1}^N (z - z_{\ell})}{\prod_{k=1}^N (z - p_k)}$$

- $z_{\ell}$  are the system zeros
  - $p_k$  are the system poles
- 
- $H(z)$  (an LTI system) can also be represented in the z-domain using a pole-zero plot

# TRANSFER FUNCTION REPRESENTATION

- A frequency response function or transfer function  $H(z)$  on the unit circle  $z = e^{j\omega}$

$$H(e^{j\omega}) = b_o z e^{j(N-M)\omega} \frac{\prod_{\ell=1}^N (e^{j\omega} - z_{\ell})}{\prod_{k=1}^N (e^{j\omega} - p_k)}$$

- The factor  $(e^{j\omega} - z_{\ell})$  can be interpreted as a *vector* in the complex  $z$ -plane from a zero  $z$  to the unit circle at  $z = e^{j\omega}$
- The factor  $(e^{j\omega} - p_k)$  can be interpreted as a vector from a pole  $p_k$  to the unit circle at  $z = e^{j\omega}$ .

# TRANSFER FUNCTION REPRESENTATION

- The magnitude response function

$$\left| H(e^{j\omega}) \right| = |b_o| \frac{|e^{j\omega} - z_1| \dots |e^{j\omega} - z_M|}{|e^{j\omega} - p_1| \dots |e^{j\omega} - p_N|}$$

- a product of the lengths of vectors from zeros to the unit circle *divided* by the lengths of vectors from poles to the unit circle and *scaled* by  $|b_o|$ .



# TRANSFER FUNCTION REPRESENTATION

- The phase response function

$$\angle H(e^{j\omega}) = [0 \text{ or } \pi] + [(N - M)\omega] + \sum_1^M \angle(e^{j\omega} - z_k) - \sum_1^N \angle(e^{j\omega} - p_k)$$

- a sum of a constant factor, a linear-phase factor, and a nonlinear-phase factor (angles from the “zero vectors” *minus* the sum of angles from the “pole vectors”).

# EXAMPLE

- Given a causal system

$$y(n) = 0.9y(n - 1) + x(n)$$

- Determine  $H(z)$  and sketch its pole-zero plot.
- Plot  $|H(e^{j\omega})|$  and  $\angle H(e^{j\omega})$ .
- Determine the impulse response  $h(n)$ .

# EXAMPLE

- The difference equation can be put in the form

$$y(n) - 0.9y(n - 1) = x(n)$$

$$H(z) = b_o z^{N-M} \frac{\prod_{\ell=1}^N (z - z_{\ell})}{\prod_{k=1}^N (z - p_k)}$$

$$H(z) = \frac{1}{1 - 0.9z^{-1}}; \quad |z| > 0.9$$

- since the system is causal.
- There is one pole at 0.9 and one zero at the origin

# EXAMPLE

- illustrate using the `zplane` function

```
>> b = [1, 0]; a = [1, -0.9]; zplane(b,a)
```

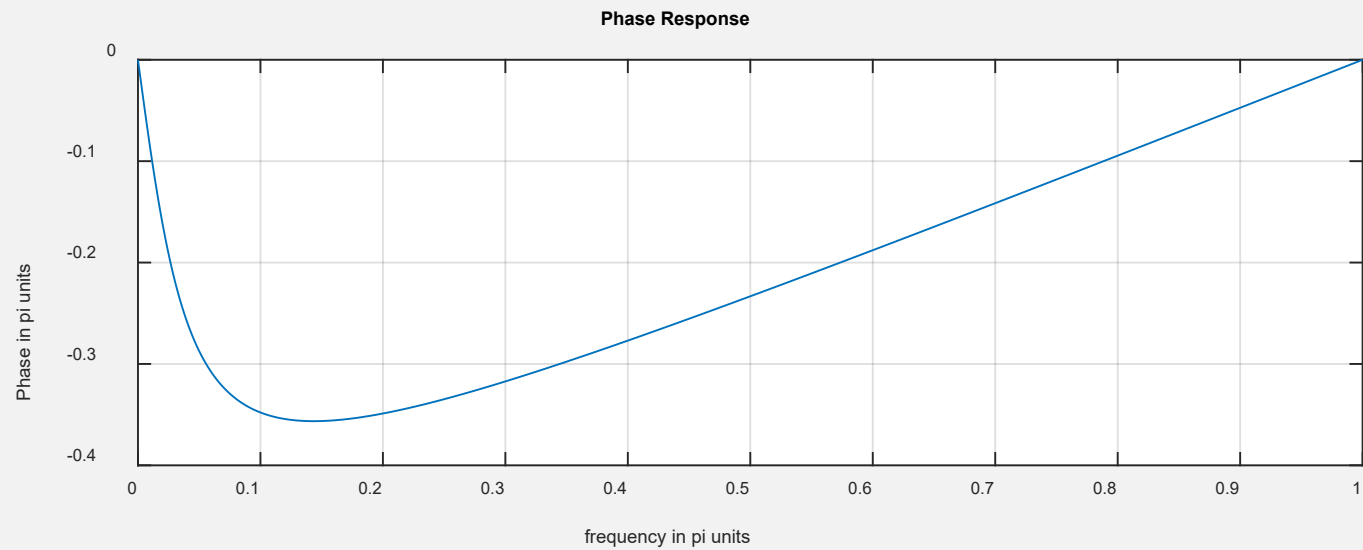
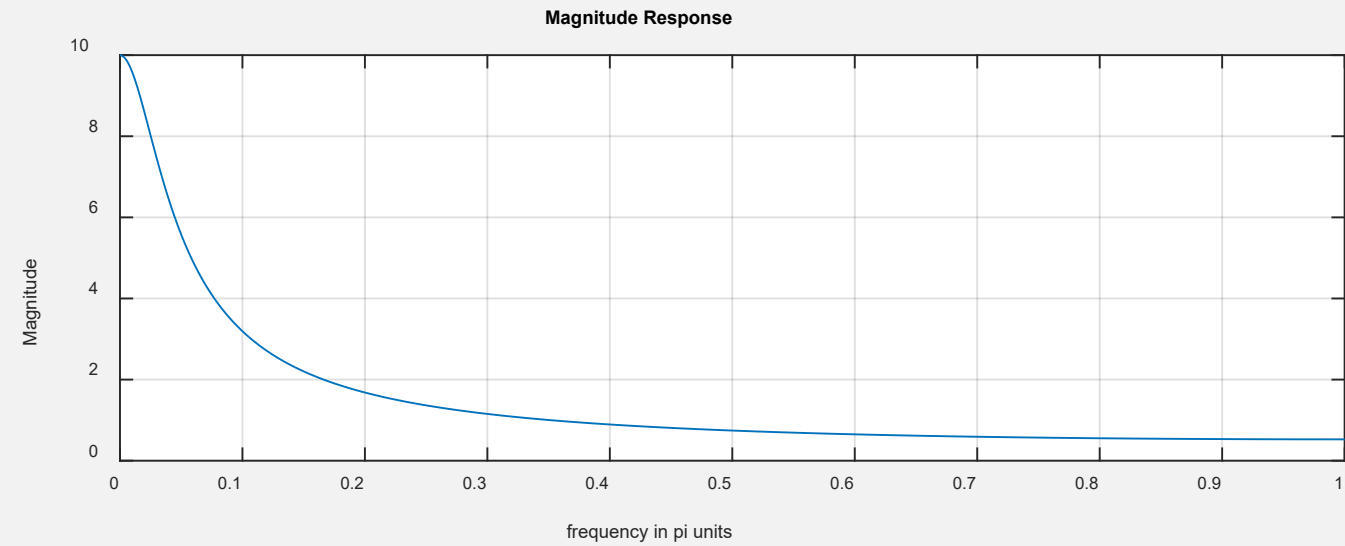


# EXAMPLE

- Determine the magnitude and phase of  $H(e^{j\omega})$  using `freqz` function.
- Take 100 points along the upper half of the unit circle

```
[H,w] = freqz(b,a,100); magH = abs(H); phaH = angle(H);  
subplot(2,1,1);plot(w/pi,magH);grid  
xlabel('frequency in pi units'); ylabel('Magnitude');  
title('Magnitude Response')  
subplot(2,1,2);plot(w/pi,phaH/pi);grid  
xlabel('frequency in pi units'); ylabel('Phase in pi units');  
title('Phase Response')
```

# EXAMPLE

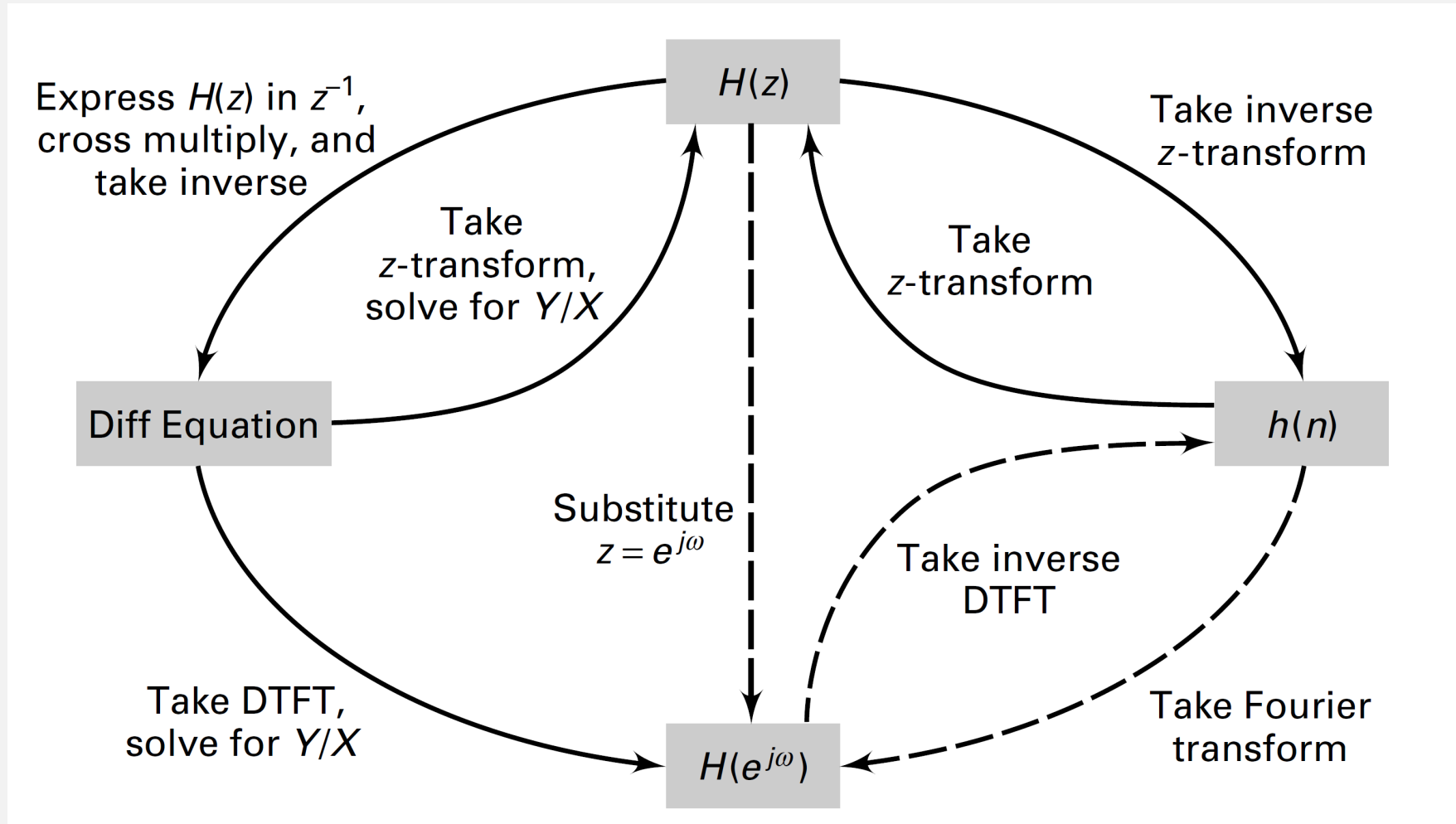


# EXAMPLE

- From the z-transform in Table

$$\begin{aligned}h(n) &= Z^{-1} \left[ \frac{1}{1 - 0,9z^{-1}}; \quad |z| > 0,9 \right] \\&= (0,9)^n u(n)\end{aligned}$$

# RELATIONSHIPS BETWEEN SYSTEM REPRESENTATIONS





# Z-DOMAIN LTI STABILITY

- An LTI system is stable if and only if the unit circle is in the ROC of  $H(z)$ .

# Z-DOMAIN CAUSAL LTI STABILITY

- A causal LTI system is stable if and only if the system function  $H(z)$  has all its poles inside the unit circle.

# EXAMPLE

- A causal LTI system is described by the following difference equation:

$$y(n) = 0.81y(n - 2) + x(n) - x(n - 2)$$

- Determine
  - the system function  $H(z)$ ,
  - the unit impulse response  $h(n)$ ,
  - the unit step response  $v(n)$ , that is, the response to the unit step  $u(n)$ , and
  - the frequency response function  $H(e^{j\omega})$ , and plot its magnitude and phase over  $0 \leq \omega \leq \pi$ .

# EXAMPLE

- Since the system is causal, the ROC will be outside a circle with radius equal to the largest pole magnitude.
- Taking the z-transform of both sides of the difference equation

$$H(z) = \frac{1 - z^{-2}}{1 - 0,81z^{-2}} = \frac{1 - z^{-2}}{(1 + 0,9z^{-1})(1 - 0,9z^{-1})}, \quad |z| > 0,9$$



# EXAMPLE

- Using the MATLAB script for the partial fraction expansion

$$H(z) = \frac{1 - z^{-2}}{1 - 0,81z^{-2}} = \frac{1 - z^{-2}}{(1 + 0,9z^{-1})(1 - 0,9z^{-1})}, \quad |z| > 0,9$$

```
>> b = [1,0,-1]; a = [1,0,-0.81]; [R,p,C] = residuez(b,a);  
R = -0.1173 -0.1173  
p = -0.9000 0.9000  
C = 1.2346
```

$$H(z) = 1,2346 - 0,1173 \frac{1}{1 + 0,9z^{-1}} - 0,1173 \frac{1}{1 - 0,9z^{-1}}, \quad |z| > 0,9$$

**TABLE 4.1** *Some common  $z$ -transform pairs*

<i>Sequence</i>	<i>Transform</i>	<i>ROC</i>
$\delta(n)$	1	$\forall z$
$u(n)$	$\frac{1}{1 - z^{-1}}$	$ z  > 1$
$-u(-n - 1)$	$\frac{1}{1 - z^{-1}}$	$ z  < 1$
$a^n u(n)$	$\frac{1}{1 - az^{-1}}$	$ z  >  a $
$-b^n u(-n - 1)$	$\frac{1}{1 - bz^{-1}}$	$ z  <  b $
$[a^n \sin \omega_0 n] u(n)$	$\frac{(a \sin \omega_0) z^{-1}}{1 - (2a \cos \omega_0) z^{-1} + a^2 z^{-2}}$	$ z  >  a $
$[a^n \cos \omega_0 n] u(n)$	$\frac{1 - (a \cos \omega_0) z^{-1}}{1 - (2a \cos \omega_0) z^{-1} + a^2 z^{-2}}$	$ z  >  a $
$na^n u(n)$	$\frac{az^{-1}}{(1 - az^{-1})^2}$	$ z  >  a $
$-nb^n u(-n - 1)$	$\frac{bz^{-1}}{(1 - bz^{-1})^2}$	$ z  <  b $

# EXAMPLE

$$H(z) = 1,2346 - 0,1173 \frac{1}{1 + 0,9z^{-1}} - 0,1173 \frac{1}{1 - 0,9z^{-1}}, \quad |z| > 0,9$$

- from Table

$$h(n) = 1,2346 \delta(n) - 0,1173 \left\{ 1 + (-1)^n \right\} (0,9)^n u(n)$$

# EXAMPLE

- From Table 4.1

$$Z[u(n)] = U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1$$

$$V(z) = H(z)U(z)$$

$$= \left[ \frac{(1 + z^{-1})(1 - z^{-1})}{(1 + 0,9z^{-1})(1 - 0,9z^{-1})} \right] \left[ \frac{1}{1 - z^{-1}} \right], \quad |z| > 0,9 \cap |z| > 1$$

$$= \frac{1 + z^{-1}}{(1 + 0,9z^{-1})(1 - 0,9z^{-1})}, \quad |z| > 0,9$$

$$= 1,0556 \frac{1}{1 - 0,9z^{-1}} - 0,0556 \frac{1}{1 + 0,9z^{-1}}, \quad |z| > 0,9$$

# EXAMPLE

- Finally,

$$v(n) = \left[ 1,0556(0,9)^n - 0,556(-0,9)^n \right] u(n)$$

- There is a pole-zero cancellation at  $z = 1$ .
- This has two implications.
  - First, the ROC of  $V(z)$  is still  $\{|z| > 0.9\}$  and not  $\{|z| > 0.9 \cap |z| > 1 = |z| > 1\}$
  - The step response  $v(n)$  contains no steady-state term  $u(n)$ .

# EXAMPLE

- The frequency response function  $H(e^{j\omega})$
- Substituting  $z = e^{j\omega}$  in  $H(z)$ ,

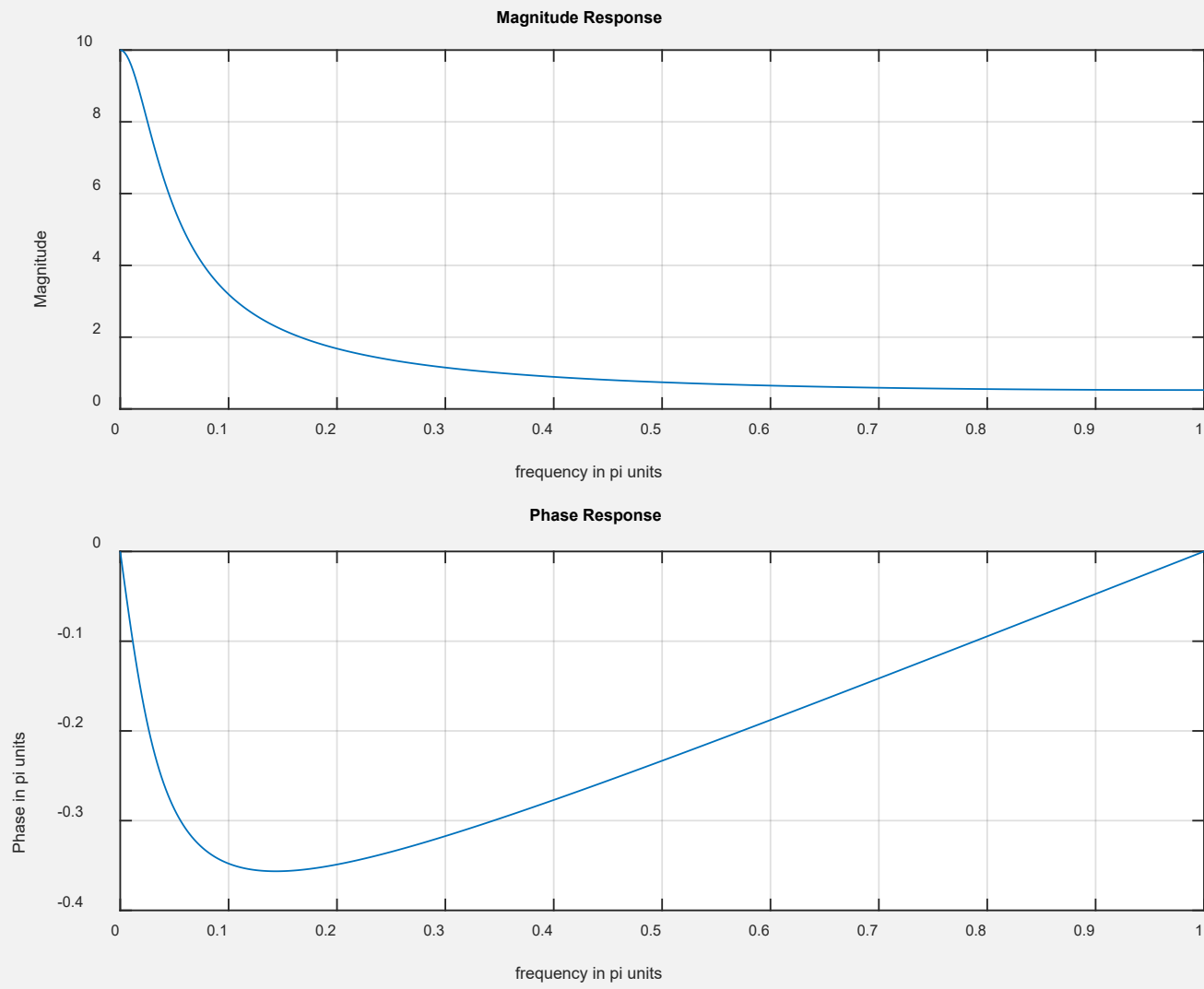
$$H(e^{j\omega}) = \frac{1 - e^{-j2\omega}}{1 - 0,81e^{-j2\omega}}$$



# EXAMPLE

```
% MATLAB script to compute and plot responses
w = [0:1:500]*pi/500; H = freqz(b,a,w);
magH = abs(H); phaH = angle(H);
subplot(2,1,1); plot(w/pi,magH); grid
xlabel('frequency in pi units'); ylabel('Magnitude')
title('Magnitude Response')
subplot(2,1,2); plot(w/pi,phaH/pi); grid
xlabel('frequency in pi units'); ylabel('Phase in pi units')
title('Phase Response')
```

# EXAMPLE





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