

Robust Control Of Bevel-Tipped Needle

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Primary Reference:

S. Hans and F. O. M. Joseph, "Robust control of a bevel-tip needle for medical interventional procedures," in IEEE/CAA Journal of Automatica Sinica, vol. 7, no. 1, pp. 244-256, January 2020, doi: 10.1109/JAS.2019.1911660.

References

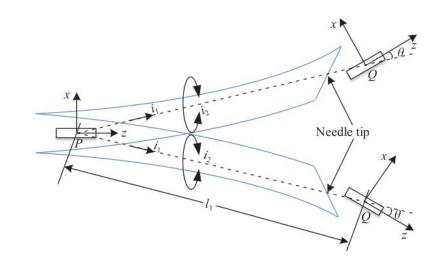
- [1] S. Hans and F. O. M. Joseph, "Robust control of a bevel-tip needle for medical interventional procedures," in IEEE/CAA Journal of Automatica Sinica, vol. 7, no. 1, pp. 244-256, January 2020, doi: 10.1109/JAS.2019.1911660.
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Introduction and Literature Review

- The merits of sliding mode based robust control strategies involve disturbance rejection, parameter uncertainty, and unmodeled dynamics, which are essential for addressing any practical applications
- Many papers have used Sliding-Mode control techniques to drive steerable needle.
- Aims:
 - To implement robust SMC that rejects disturbance
 - Avoid Chattering using ISMC + STA and Continuous Control



Kinematic model and State-Space Equations

$$g_{PQ} = \begin{bmatrix} R_{PQ} & p_{PQ} \\ 0 & 1 \end{bmatrix} \in SE(3)$$

 $Velocity, \ v = \dot{p_1}i_1 + \dot{p_2}i_2$

$$\dot{p_1} = egin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \\ k \\ 0 \end{bmatrix}, \ \dot{p_2} = egin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 1 \end{bmatrix}$$

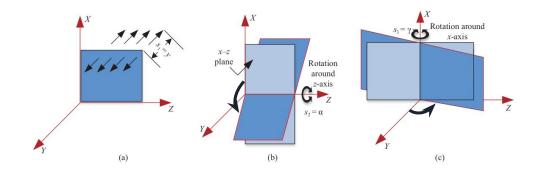
$$v = J\dot{q}, \ q = \begin{bmatrix} x \\ y \\ z \\ \alpha \\ \beta \\ \gamma \end{bmatrix}$$

$$s = [s_1 \ s_2 \ s_3]^T = [y \ \alpha \ \gamma]^T$$

$$\dot{s_1} = -\sin s_2$$

$$\dot{s_2} = k\cos s_3 \sec s_2$$

$$\dot{s_3} = -k\cos s_3 \tan s_2 + i$$



Input-Output Feedback Linearization

$$\dot{\psi} = \frac{\delta h(s)}{\delta s} [f_0(s) + g_0(s)i]$$

$$\Rightarrow L_{f_0}h(s) = \frac{\delta h(s)}{\delta s} f_0(s) = \dot{\psi} = -\sin s_2$$

$$L_{g_0}h(s) = \frac{\delta h(s)}{\delta s} g_0(s) = 0$$

$$L_{f_0}^2h(s) = \frac{\delta L_{f_0}h(s)}{\delta s} f_0(s) = \begin{bmatrix} 0 & -\cos s_2 & 0 \end{bmatrix} \begin{bmatrix} -\sin s_2 \\ k\cos s_3 \sec s_2 \\ -k\cos s_3 \tan s_2 \end{bmatrix} = -k\cos s_3$$

$$L_{g_0}L_{f_0}h(s) = \frac{\delta L_{f_0}h(s)}{\delta s} g_0(s) = \begin{bmatrix} 0 & -\cos s_2 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} = 0$$

$$L_{g_0}L_{f_0}h(s) = \frac{\delta L_{f_0}h(s)}{\delta s} g_0(s) = \begin{bmatrix} 0 & 0 & k\sin s_3 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} = k\sin s_3$$

$$Relative Degree, \ \rho = 3 = n$$

$$L_{f_0}^3h(s) = \frac{\delta L_{f_0}^2h(s)}{\delta s} f_0(s) = \begin{bmatrix} 0 & 0 & k\sin s_3 \end{bmatrix} \begin{bmatrix} -\sin s_2 \\ k\cos s_3 \sec s_2 \\ -k\cos s_3 \tan s_2 \end{bmatrix}$$

 $= -k^2 \sin s_3 \cos s_3 \tan s_2$

$$r = T(s) = \xi = \begin{bmatrix} h(s) \\ L_{f_0}h(s) \\ L_{f_0}h(s) \end{bmatrix}$$

$$r = \begin{bmatrix} r_1 \\ r_2 \\ r_3 \end{bmatrix} = \begin{bmatrix} s_1 \\ -\sin s_2 \\ -k\cos s_3 \end{bmatrix}$$

$$\dot{r_1} = r_2$$

$$\dot{r_2} = r_3$$

$$\dot{r_3} = \nu_1 = f(r) + g(r)i$$

$$, where, \ \nu_1 = \gamma(s)(i - \alpha(s))$$

$$\gamma(s) = L_{g_0}L_{f_0}^2h(s) = k\sin s_3$$

$$\alpha(s) = \frac{-L_{f_0}^3h(s)}{L_{g_0}L_{f_0}^2h(s)}$$

$$= \frac{k^2\sin s_3\cos s_3\tan s_2}{k\sin s_3} = k\cos s_3\tan s_2$$

$$\Rightarrow \nu_1 = -k^2\sin s_3\cos s_3\tan s_2 + k\sin s_3i$$

Conventional SMC

$$\dot{r_1} = r_2$$

$$\dot{r_2} = r_3$$

$$\dot{r_3} = \nu_1 = f(r) + g(r)i + d$$

$$f(r) = k \sin \left[\cos^{-1} \left(\frac{-r_3}{k} \right) \right] r_3 \tan \left[\sin^{-1} (-r_2) \right]$$

$$g(r) = k \sin \left[\cos^{-1} \left(\frac{-r_3}{k} \right) \right]$$

$$d = 2 \sin t + 3$$

The sliding surface has been considered as follows:

$$\zeta = c_1 r_1 + c_2 r_2 + r_3
\dot{\zeta} = c_1 r_2 + c_2 r_3 + f(r) + g(r)i + d
i = -g^{-1}(r)[c_1 r_2 + c_2 r_3 + f(r) + K sign(\zeta) + \eta \zeta]
\Rightarrow \dot{\zeta} = -K sign(\zeta) - \eta \zeta + 2 \sin t + 3$$
(8)

The parameters $c_1 = 10$, $c_2 = 5$, K = 6, $\eta = 4$ give a desired rate of convergence of states to zero.

Continuous Finite-time Feedback Control

Theorem 1 [6]. Let $a_1, ... a_n > 0$ be such that the polynomial $\Gamma^n + a_n \Gamma^{n-1} + ... + a_2 \Gamma + a_1$ is **Hurwitz**, and there exists $\epsilon \in (0,1)$ such that, for every $b \in (1-\epsilon,1)$, the **origin** is a **globally finite-time stable equilibrium** for the undisturbed system (6) under the feedback control $i = g^{-1}(r)[u - f(r)]$ by assuming:

$$u = -a_1 |r_1|^{b_1} sign(r_1) - \dots - a_n |r_n|^{b_n} sign(r_n)$$
(9)

, where, $b_1, ..., b_n$ satisfy

$$b_{w-1} = \frac{b_w b_{w+1}}{2b_{w+1} - b_w}, \quad w = 2, ..., n \tag{10}$$

, with, $b_{n+1} = 1$ and $b_n = b$.

$$i = g^{-1}(r)[u - f(r)]$$

$$u_{nom} = -a_1|r_1|^{b_1}sign(r_1) - a_2|r_2|^{b_2}sign(r_2) - a_3|r_3|^{b_3}sign(r_3)$$
(11)

, where, controller gains were chosen as $a_1 = 15$, $a_2 = 23$, $a_3 = 9$, $b_1 = \frac{1}{4}$, $b_2 = \frac{1}{3}$, $b_3 = \frac{1}{2}$.

ISMC With Discontinuous control

 $U_{nom} \rightarrow Performance$

 $U_{discon} \rightarrow Disturbance Rejection$

$$u = u_{nom} + u_{discon}$$

$$u_{discon} = -a_4 sign(\zeta)$$

Sliding variable,
$$\zeta = r_3 - r_{30} - \int_0^t u_{nom} dt$$
, r_{30} is initial condition of r_3

$$\Rightarrow \dot{\zeta} = u + d - u_{nom} = 0$$

$$\Rightarrow u_{nom} + u_{discon} + d - u_{nom} = 0$$

$$\Rightarrow u_{discon} = -d$$

$$\Rightarrow [-a_4 sign(\zeta)]_{eq} = -d$$

$$a_4 = 10$$

Higher Order Sliding Mode Control

- Higher Order SMC means that the equation for the higher order derivative of sliding variable must contain the discontinuity such that all lower derivatives converge to zero in finite-time.
- The **sliding manifold** in this type of control is not just $\zeta=0$, but, a set :
- $\Phi = \{\zeta = 0, d^e \zeta / dt^e = 0\}$
- e goes from 1 to m-1, where, m is the order of sliding mode

The assumptions in STA are as follows [7]:

Sliding variable,
$$\dot{\zeta} = h_1(x) + g_1(x)u + d(t)$$

$$u = -\lambda |\zeta|^{1/2} sign(\zeta) + \nu$$

$$\dot{\nu} = \begin{cases} -u, & |u| > U_M \\ -\alpha sign(\zeta), & |u| \leq U_M \end{cases}$$

$$|\dot{h}_1| + |\dot{g}_1|U_M + |\dot{d}(t)| \leq C$$

$$0 < K_m \leq g_1(x) \leq K_M$$

$$|h_1| < q|g_1|U_M, \ q \in (0, 1)$$

$$|\dot{d}(t)| < C_1$$

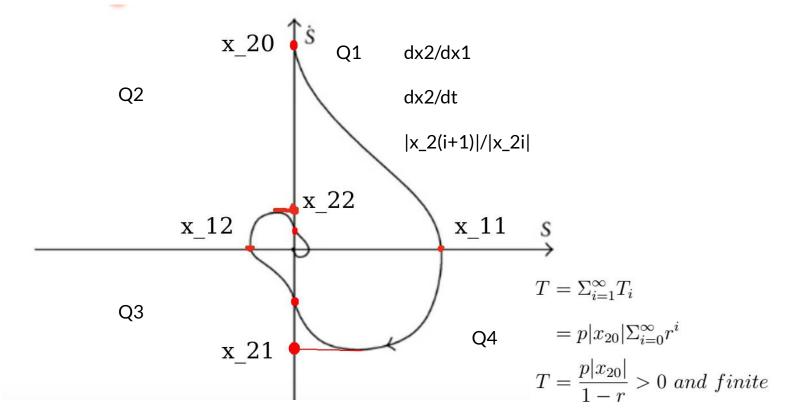
We know that d(t) satisfies the assumptions $|d(t)| \leq C_2$ and $|\dot{d}(t)| \leq C_1$, with, $C_1 = \tau$ and $C_2 = d_m$ as given in the paper [1]. The **conditions on controller gains** are given below:

$$K_m \alpha > C$$

$$\lambda > \sqrt{\frac{2}{K_m \alpha - C}} \times \frac{K_m \alpha + C}{K_m} \times \frac{1 + q}{1 - q}$$
(16)

 $|d(t)| \leq C_2$

Finite-time Stability of STA (Majorant Curve Analysis)



ISMC + STA

$$u = u_{nom} + u_{STC}$$

$$\dot{\zeta} = u + d - u_{nom}$$

$$\dot{\zeta} = u_{STC} + d$$

$$u_{STC} = -a_5 |\zeta|^{1/2} sign(\zeta) + \Theta$$

$$\dot{\zeta} = -a_5 |\zeta|^{1/2} sign(\zeta) + \Theta + d$$

$$\dot{\Theta} = -a_6 sign(\zeta)$$

$$O = \Theta + d$$

$$\dot{\zeta} = -a_5 |\zeta|^{1/2} sign(\zeta) + O$$

$$\dot{O} = -a_6 sign(\zeta) + d$$

$$\dot{\zeta} = u + d - u_{nom} = 0$$

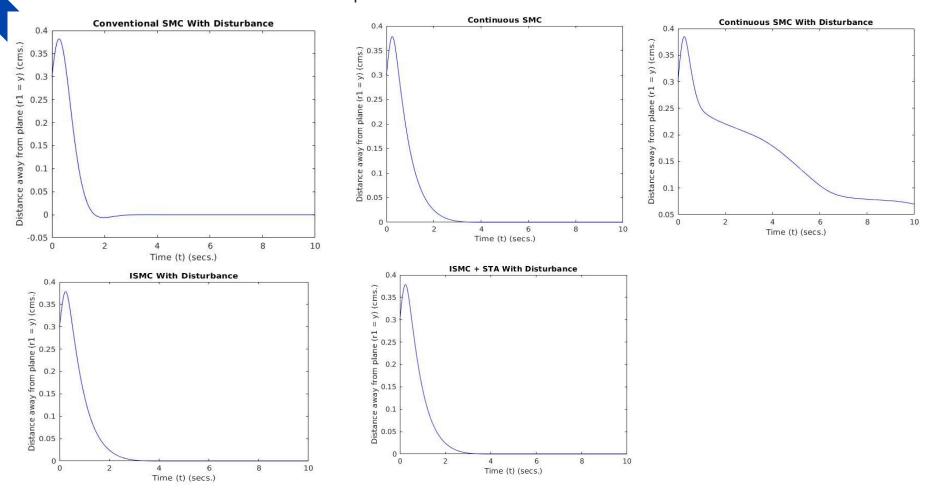
$$\Rightarrow u + d = u_{nom}$$

$$u_{STC} = -d$$

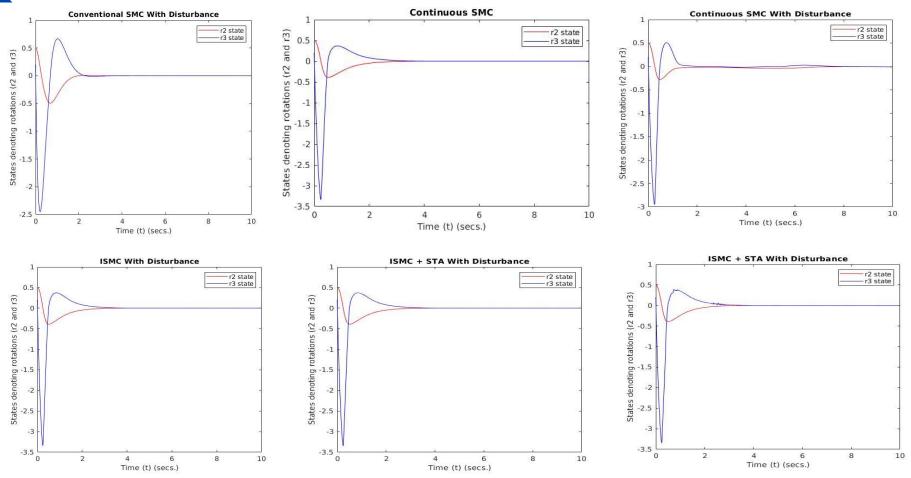
 $a_5 = 0.13$

 $a_{\lambda} = 300$

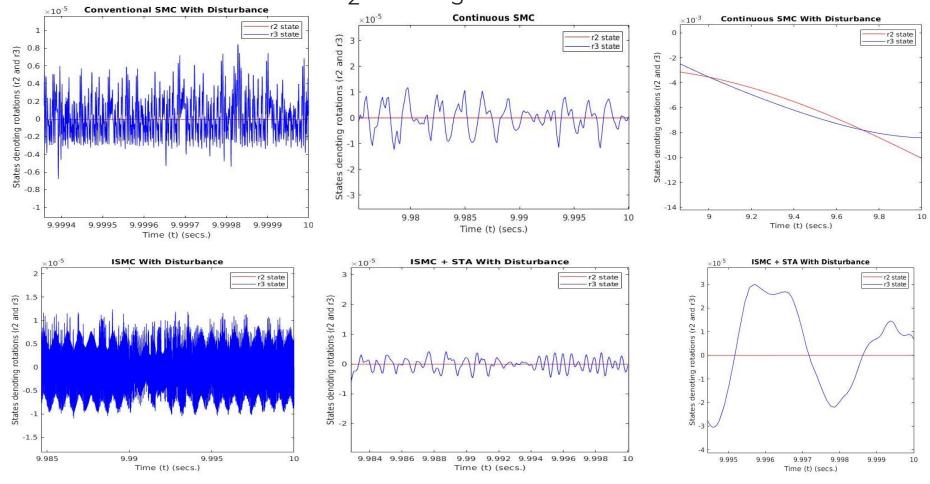
Simulation: r₁ State (Distance from x-z plane)



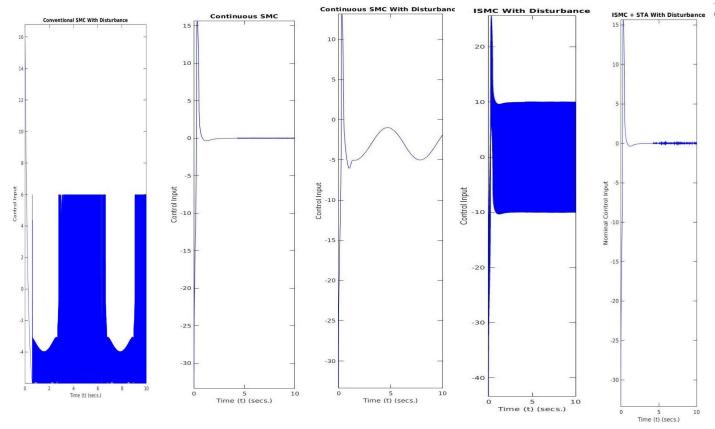
Simulation: r_2 and r_3 States (Representing Rotations)

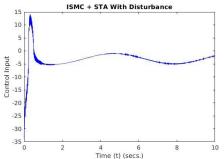


Simulation: r_2 and r_3 States Zoomed

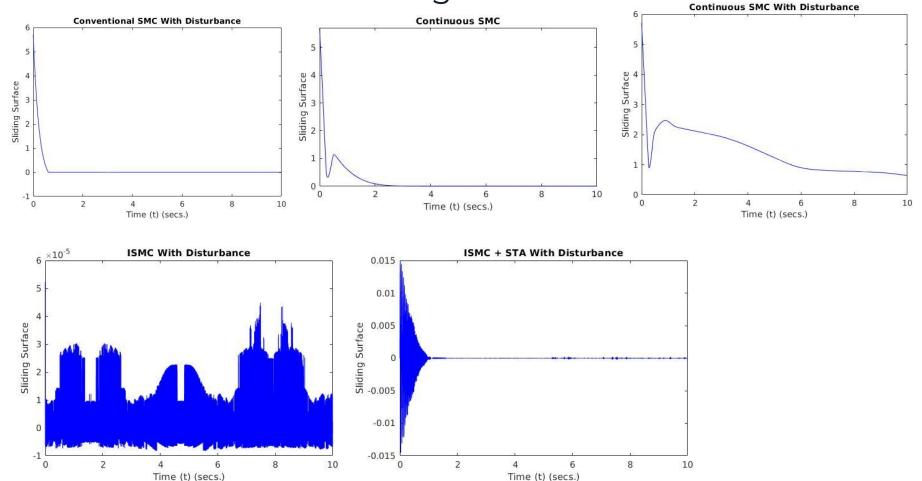


Simulation: Control Input "u"

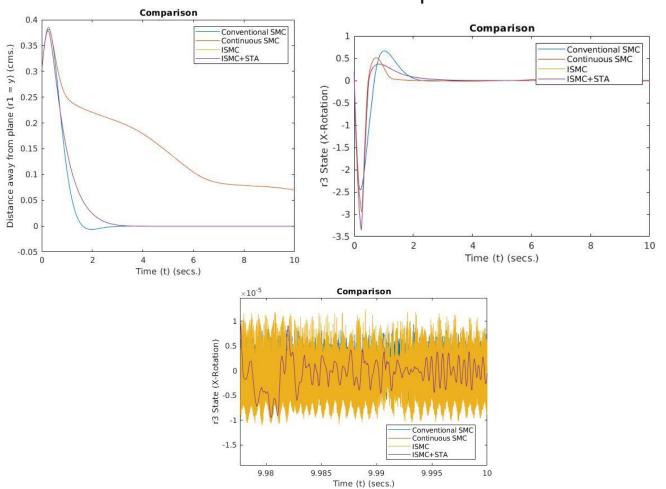


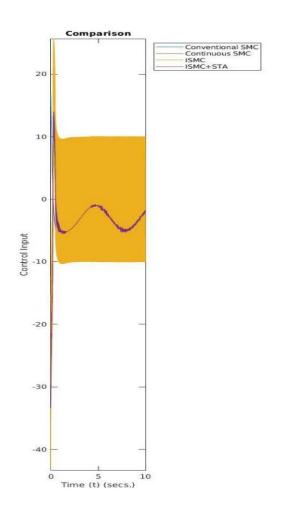


Simulation: Sliding Variable



Comparison





THANKS Q&A