

1) The continuous plant:

$$G(s) = \frac{k_a k_t k_e}{(J_e s + B_e) s}$$

$$G(s) = \frac{k_a k_t k_e}{J_e} \left(\frac{1}{(s + \frac{B_e}{J_e}) s} \right)$$

Then the discrete version is:

$$G(z) = \frac{k_a k_t k_e}{J_e} \mathcal{Z} \left\{ \frac{1}{s^2 (s + \frac{B_e}{J_e})} \right\} (1 - z^{-1}) k_d$$

$$= \frac{k_a k_t k_e}{J_e} (1 - z^{-1}) \mathcal{Z} \left(\frac{J_e}{B_e} \frac{1}{s^2} + \frac{-J_e^2}{B_e^2} \frac{1}{s} + \frac{J_e^2}{B_e^2} \frac{1}{s + \frac{B_e}{J_e}} \right) k_d$$

$$= \frac{k_a k_t k_e}{J_e} (1 - z^{-1}) \left(\frac{J_e}{B_e} \frac{T z^{-1}}{(1 - z^{-1})^2} - \left(\frac{J_e}{B_e} \right)^2 \frac{1}{(1 - z^{-1})} + \left(\frac{J_e}{B_e} \right)^2 \frac{1}{1 - z^{-1} e^{-\frac{B_e T}{J_e}}} \right) k_d$$

plug in number:

$$= \frac{5.8017E-5 - 5.805E-5 z^{-1}}{1 - 1.998253 z^{-1} + 0.998253 z^{-2}} k_d$$

MATLAB has similar result

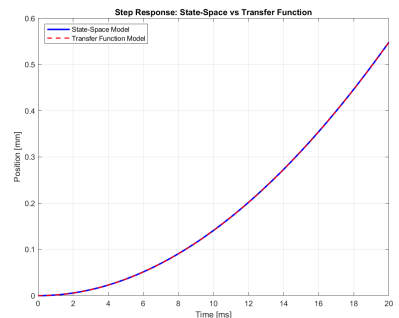
ans =

$$\frac{5.805e-05 z + 5.801e-05}{z^2 - 1.998 z + 0.9983}$$

2) 2nd order system, let's choose w and x_a

$$\begin{bmatrix} \dot{w} \\ \dot{x}_a \end{bmatrix} = A \begin{bmatrix} w \\ x_a \end{bmatrix} + B \begin{bmatrix} v_{in} \\ \frac{T_d}{u} \end{bmatrix}$$

$$x_a = C x + D u$$



$$x_a = \begin{bmatrix} x \\ \dot{x} \end{bmatrix}$$

$$J_e \dot{\omega}(t) + B_e \omega(t) = K_a k_t v_{in} - T_d$$

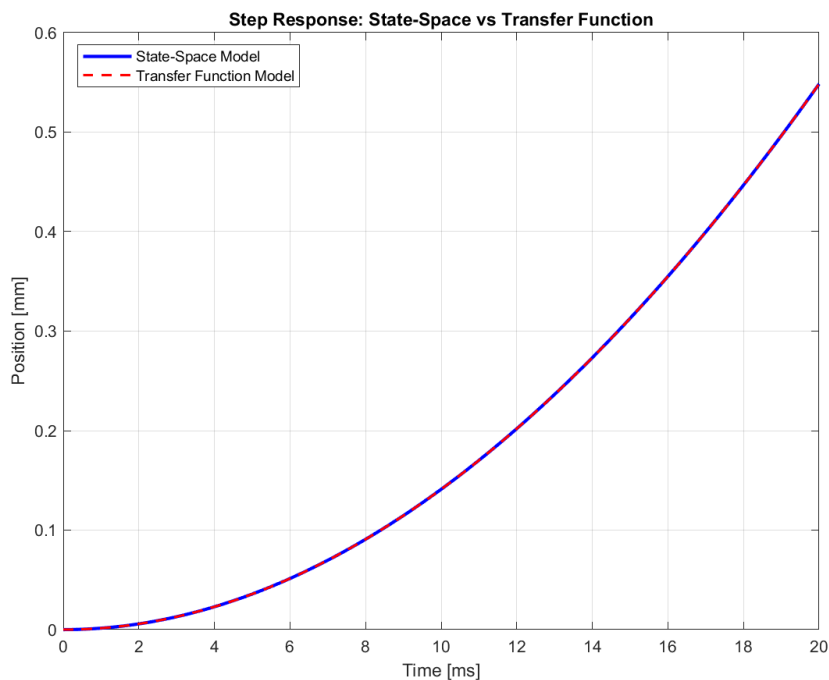
$$\dot{\omega}(t) = \frac{K_a k_t}{J_e} v_{in} - \frac{1}{J_e} T_d - \frac{B_e}{J_e} \omega$$

$$\dot{x}_a = k_e \omega$$

$$\begin{bmatrix} \dot{\omega} \\ \dot{x}_a \end{bmatrix} = \begin{bmatrix} -\frac{B_e}{J_e} & 0 \\ k_e & 0 \end{bmatrix} \begin{bmatrix} \omega \\ x_a \end{bmatrix} + \begin{bmatrix} \frac{K_a k_t}{J_e} & -\frac{1}{J_e} \\ 0 & 0 \end{bmatrix} \begin{bmatrix} v_{in} \\ T_d \end{bmatrix}$$

$$x_a = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} \omega \\ x_a \end{bmatrix}$$

The following is the MATLAB result



Both the state space and TF Model are close

3)

a) The equation is:

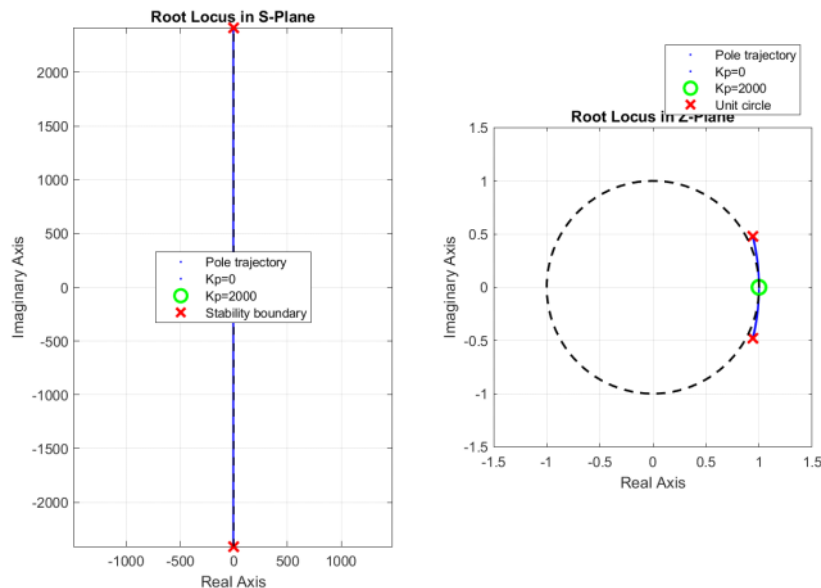
$$G_c = \frac{k_p G_o}{1 + k_p G_o}$$

... for the root

$$1 + K_p G = 0$$

So the root locus solves for the root $1 + K_p G = 0$

Below are the figure from Matlab:

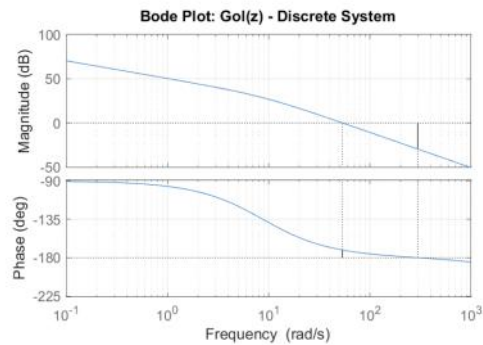
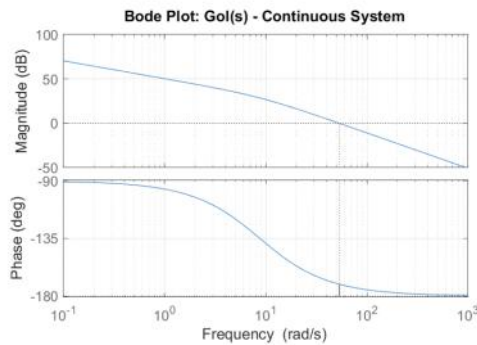


In continuous domain, the poles begin in the negative real axis with $K_p=0$, and as K_p increases our poles begin to go to the centroid of the poles. Then after they meet, they will begin to diverge to infinite in the imaginary axis. If you look at this graph, the system is stable (goes to 0 as times goes to infinity). One thing to note is that if you keep increasing K_p , imaginary number increases, this means that there will be more oscillations, which is what we expect.

For the z plane, the root goes around the circle, and if K_p is high enough, we will see a pole that goes outside the unit circle. That means the system is unstable for high enough K_p , a very different behavior than the continuous system.

h) The bode plots are:

b) The bode plots are:

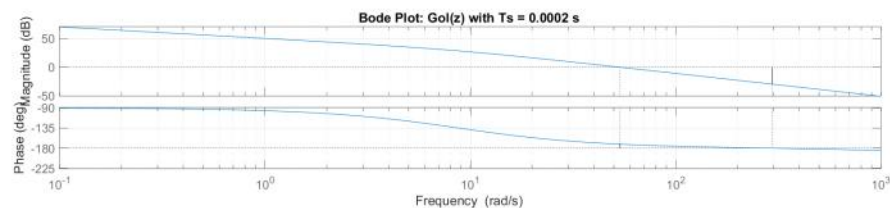
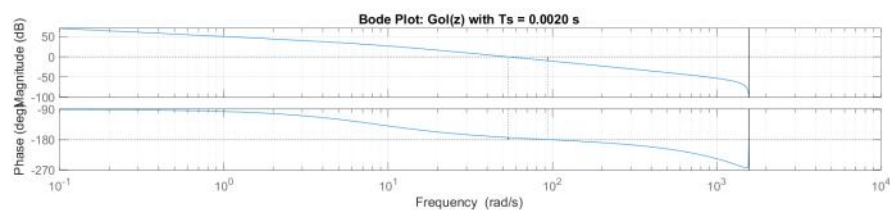
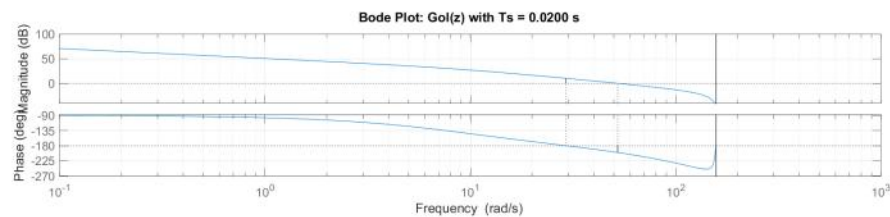


From the matlab command:

	gain margin	phase margin
continuous	∞	9.28°
discrete	29.58 db	8.97°

despite using a $2E-4$ s, there is still a gain margin in discrete system, while it's infinite in continuous system. The phase margin is also different.

c)



As you can see from the bode plot, they even look like a different system, with different phase and gain margin. Note that for low sampling time, they get much more unstable at high frequency (makes sense, as the delay added by our discrete system will be much more noticeable if we give a high frequency input).

- $T_s = 0.02$ s: UNSTABLE (negative gain margin)
- $T_s = 0.002$ s: Marginally stable (9.60 dB gain margin)
- $T_s = 0.0002$ s: Most stable (29.58 dB gain margin)

The 0.02s system is even unstable.

$$4) G_{C1} = \frac{k_p G_0}{1 + k_p G_0}$$

$$1 + k_p G_0 = 0$$

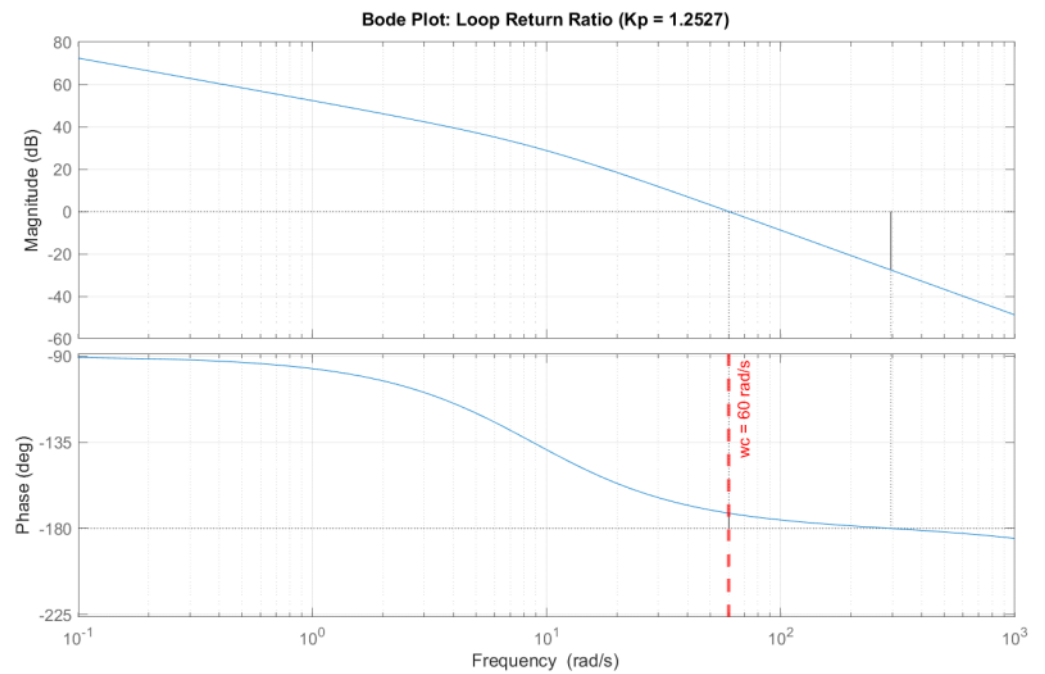
$$k_p = \frac{-1}{G_0}$$

So we just plug in $G_0(60j)$, and get the magnitude

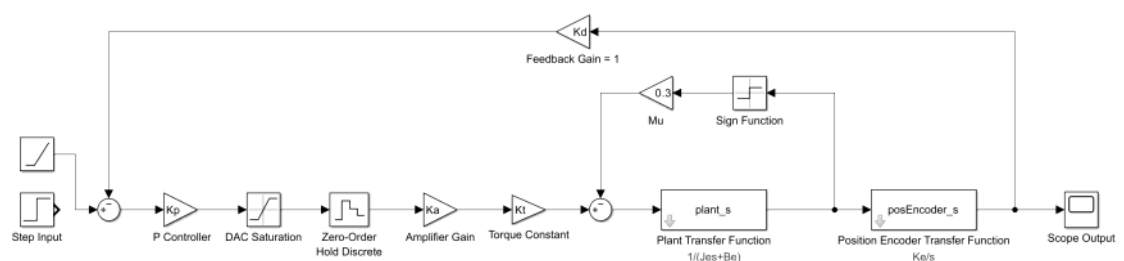
$$k_p = \left| \frac{1}{G_0} \right|$$

$$k_p = 1.2527 \frac{V}{mm}$$

Here is the bode plot:



Below is the Simulink model:



And the resulting graph:

