

Multijoint Torque Control

EE366/CS380: Introduction to Robotics

Dr. Basit Memon

Electrical and Computer Engineering
Habib University

April 24, 2024



Table of Contents

- 1 General model of system
- 2 Independent Joint Control
 - 2.1 Approximate model of system
 - 2.2 PID Controller
 - 2.3 Feedforward
- 3 Centralized Control



Multi-joint control

- We need to obtain a model for the entire robot manipulator - all joints.
- Motion of one link affects the other links. The model is complex.
- Can we just control each joint independently as we did with velocity commands?



Table of Contents

1 General model of system

2 Independent Joint Control

2.1 Approximate model of system

2.2 PID Controller

2.3 Feedforward

3 Centralized Control

- Entire manipulator can be expressed by following rigid body dynamics:

$$J(q)\ddot{q} + C(q, \dot{q})\dot{q} + B_v\dot{q} + g(q) = \tau$$

- q represents joint variable - θ for revolute joint and d for prismatic
- τ is any torque acting on manipulator – it could be desired, i.e. actuator, or undesired disturbance.

Example: Model of two-link manipulator

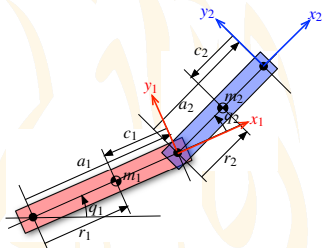


Figure: Two links planar arm

$$\tau_1 = M_{11}(q_2)\ddot{q}_1 + \underbrace{M_{12}(q_2)\ddot{q}_2 + C_1(q_2)\dot{q}_1\dot{q}_2 + C_2(q_2)\dot{q}_2^2}_{\text{disturbance}} + g(q_1, q_2)$$

$$M_{11} = m_1(a_1^2 + 2a_1c_1 + c_1^2) + m_2(a_1^2 + (a_2 + c_2)^2 + (2a_1a_2 + 2a_1c_2)\cos q_2)$$

$$M_{12} = m_2(a_2 + c_2)(a_2 + c_2 + a_1\cos q_2)$$

$$C_1 = -2a_1m_2(a_2 + c_2)\sin q_2$$

$$C_2 = -a_1m_2(a_2 + c_2)\sin q_2$$

$$g = (a_1m_1 + a_1m_2 + c_1m_1)\cos q_1 + (a_2m_2 + c_2m_2)\cos(q_1 + q_2)$$

Figure: Dynamics for net torque at joint 1

Example: Model of two-link manipulator

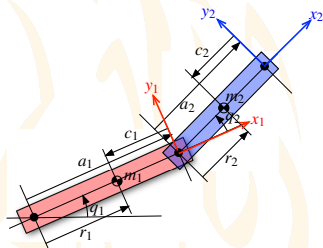


Figure: Two links planar arm

$$\tau_1 = \overset{J_l}{\boxed{M_{11}(q_2)}}\ddot{q}_1 + \underbrace{M_{12}(q_2)\ddot{q}_2 + \boxed{C_1(q_2)\dot{q}_1\dot{q}_2 + C_2(q_2)\dot{q}_2^2}}_{\text{disturbance}} + \boxed{g(q_1, q_2)} \overset{\text{Gravity}}{}$$

$$M_{11} = m_1(a_1^2 + 2a_1c_1 + c_1^2) + m_2(a_1^2 + (a_2 + c_2)^2 + (2a_1a_2 + 2a_1c_2)\cos q_2)$$

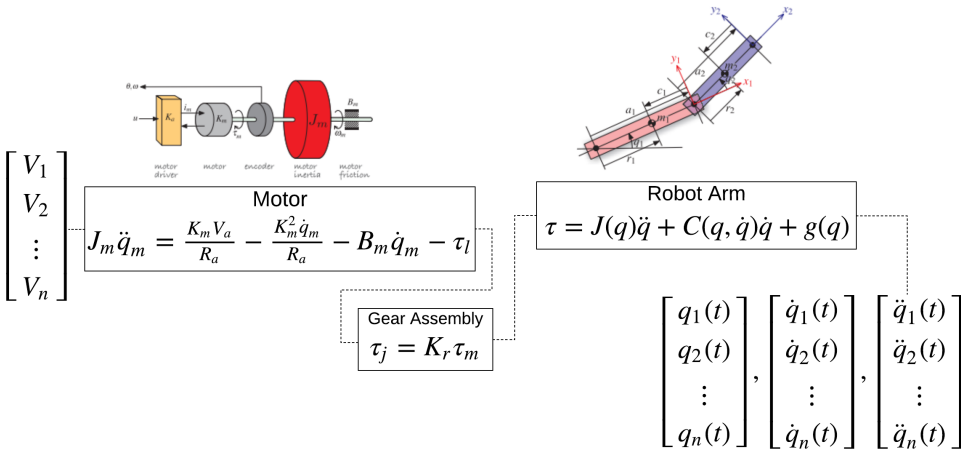
$$M_{12} = m_2(a_2 + c_2)(a_2 + c_2 + a_1\cos q_2)$$

$$C_1 = -2a_1m_2(a_2 + c_2)\sin q_2$$

$$C_2 = -a_1m_2(a_2 + c_2)\sin q_2$$

$$g = (a_1m_1 + a_1m_2 + c_1m_1)\cos q_1 + (a_2m_2 + c_2m_2)\cos(q_1 + q_2)$$

Figure: Dynamics for net torque at joint 1



Torque balancing equation could be written on arm side

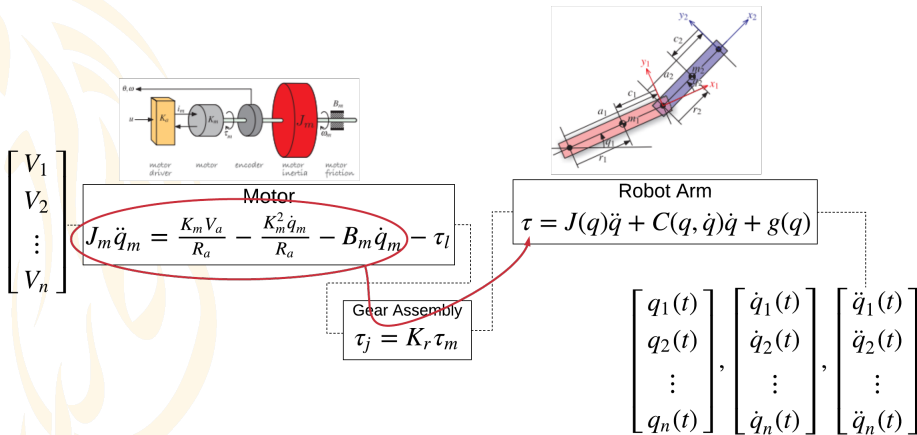


Figure: Motor Reflected on the arm side

Or, torque balancing equation could be written on motor side

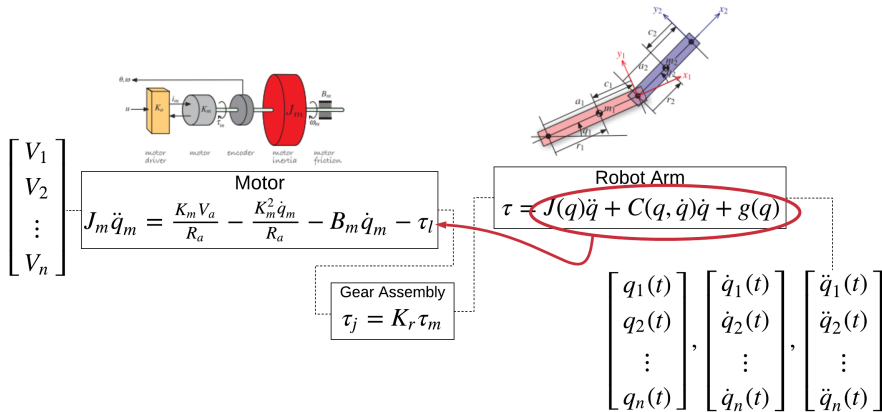


Figure: Arm Reflected on the motor side

$$\begin{aligned}
 J_m \ddot{q}_m &= \frac{K_m V_a}{R_a} - \left(\frac{K_m^2}{R_a} + B_m \right) \dot{q}_m - \tau_l \\
 &= \frac{K_m V_a}{R_a} - \left(\frac{K_m^2}{R_a} + B_m \right) \dot{q}_m - \frac{J(q)\ddot{q} + C(q, \dot{q})\dot{q} + g(q)}{K_r}
 \end{aligned}$$

Using the gear ratio, $q = \frac{q_m}{K_r}$:

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = \frac{K_m V_a}{R_a} - \left(\frac{K_m^2}{R_a} + B_m + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}$$

Manipulator is modeled by coupled non-linear ODE :(

$$J(q) = \begin{bmatrix} m_1 a_1^2 + m_2(a_1^2 + 2a_1 a_2 \cos q_2 + a_2^2) & m_2(a_1 a_2 \cos q_2 + a_2^2) \\ m_2(a_1 a_2 \cos q_2 + a_2^2) & m_2 a_2^2 \end{bmatrix}$$
$$C(q, \dot{q}) = \begin{bmatrix} -2m_2 a_1 a_2 \sin q_2 \dot{q}_2 & \dot{q}_2 \\ m_2 a_1 a_2 \dot{q}_1 \sin q_2 & 0 \end{bmatrix}$$
$$g(q) = \begin{bmatrix} (m_1 + m_2) a_1 g \cos q_1 + m_2 g a_2 \cos(q_1 + q_2) \\ m_2 g a_2 \cos(q_1 + q_2) \end{bmatrix}$$

- Previously J was constant matrix. Now all of these are dependent on q . A nonlinear model completely!
- Can we control each motor/joint separately?



Table of Contents

- 1 General model of system
- 2 Independent Joint Control
 - 2.1 Approximate model of system
 - 2.2 PID Controller
 - 2.3 Feedforward
- 3 Centralized Control



Is independent joint control possible?

- Doesn't seem so since the J matrix is coupled.
- Unless, we can decouple it.
- Note that this will be an approximate model!
- Might as well try for linear approximate model.

Large gear ratio results in diagonal decoupled J

- If the gear ratio is large, $\frac{J(q)}{K_r^2}$ can be approximated by a constant diagonal matrix.

$$\frac{J(q)}{K_r^2} = \frac{1}{K_r^2} \begin{bmatrix} m_1 a_1^2 + m_2(a_1^2 + 2a_1 a_2 \cos q_2 + a_2^2) & m_2(a_1 a_2 \cos q_2 + a_2^2) \\ m_2(a_1 a_2 \cos q_2 + a_2^2) & m_2 a_2^2 \end{bmatrix}$$

Terms are dependent on q , but choose maximum inertia possible over all q .

- Now we can treat each joint independently – Decentralized control.

$$\frac{J(q)}{K_r^2} \approx \begin{bmatrix} I_1 & 0 \\ 0 & I_2 \end{bmatrix}$$

Nonlinear terms treated as disturbances, given large gear ratio.

- If the gear ratio is large and \dot{q} is small, $\frac{C(q, \dot{q})\dot{q}_m}{K_r^2}$ is also very small.

$$\frac{C(q, \dot{q})\dot{q}_m}{K_r^2} = \frac{1}{K_r} \begin{bmatrix} -2m_2 a_1 a_2 \sin q_2 \dot{q}_2 & \dot{q}_2 \\ m_2 a_1 a_2 \dot{q}_1 \sin q_2 & 0 \end{bmatrix} \dot{q}_m$$

- We can model this and gravity term as disturbance for control problem:

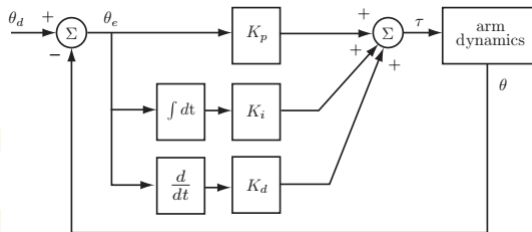
$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = \frac{K_m V_a}{R_a} - \left(\frac{K_m^2}{R_a} + B_m + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}$$

Approximately, and only approximately

$$J_s \ddot{q}_m = u - B_s \dot{q}_m - D$$

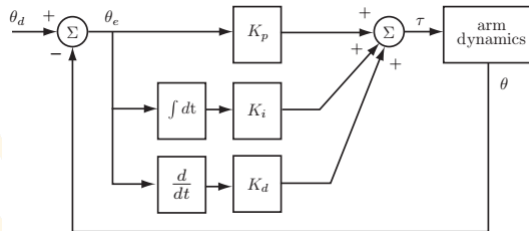
- In order to highlight the effect of K_r , the inverse is displayed as a reciprocal but don't be misled. K_r is a matrix in general and its inverse would be taken.

PID reduces error to zero for step trajectory and disturbances.



- It was shown that PID control reduces error to zero, if desired trajectory is ramp or step function.
- It was shown that PID control reduces effects of disturbance to zero, if disturbance is step function.

What is the equation for arm dynamics here?



$$J_s \ddot{q}_m = u - B_s \dot{q}_m - D$$

OR

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = u - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}?$$



PID designed using approx. model, but reality doesn't change!

- When I'm designing:

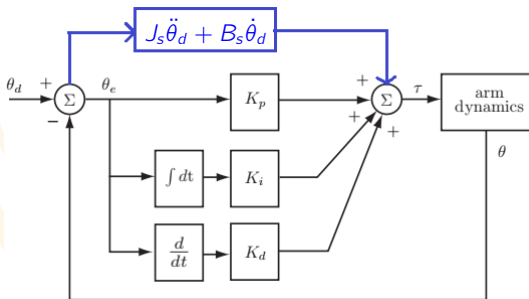
$$J_s \ddot{q}_m = u - B_s \dot{q}_m - D$$

- $u = K_p \theta_e + K_i \int_0^t \theta_e dt + K_d \dot{\theta}_e$
- Find K_p , K_i , and K_d to minimize error and converge fast.

- Install sensors to measure θ and $\dot{\theta}$
- Program a controller to output $K_p \theta_e + K_i \int_0^t \theta_e dt + K_d \dot{\theta}_e$ to the motor and arm.
- This is equivalent to:

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = K_p \theta_e + K_i \int_0^t \theta_e dt + K_d \dot{\theta}_e - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}$$

Feedforward Control allows tracking of any arbitrary trajectory.



- The feedforward function is determined using approximate system model.
- The feedback controller can be used to reject disturbance (PD or PID).



Table of Contents

- 1 General model of system
- 2 Independent Joint Control
 - 2.1 Approximate model of system
 - 2.2 PID Controller
 - 2.3 Feedforward
- 3 Centralized Control



Strategies – Centralized Control

- When $J(q)$ cannot be made a diagonal matrix (e.g. motors are directly connected to links without gears), we'll have to design controller for all joints together.
- Coupling of dynamics cannot be avoided:

$$J(q)\ddot{q} = \tau$$

Still ignoring the other terms.

- We have a multi-input, multi-output problem.
- While PID controller is still a good controller for this problem, the approach discussed for designing PID is no longer suitable, and one typically uses state-space methods.

- Dynamics of an arm in all glory:

$$J(q)\ddot{q} + C(q, \dot{q})\dot{q} + B\dot{q} + g(q) = u.$$

- By using gear mechanisms, we can decouple inertia matrix and make it independent of q . By assuming that \dot{q} and \ddot{q} are small, we can ignore $C(q, \dot{q})$ term. But, we still have to treat $g(q)$ as disturbance.

- Find K_p and K_d for $u = K_p q_e + K_d \dot{q}_e$, assuming that system is modeled by $J_s \ddot{q}_m = u - B_s \dot{q}_m - D$

- **Implementation:**

- Measure q and \dot{q} .
- We know form of $g(q)$. Substitute q to find $g(q)$.
- Let $u = K_p q_e + K_d \dot{q}_e + \frac{g(q)}{K_r}$

$$u = K_P q_e + K_D \dot{q}_e + \frac{g(q)}{K_r}$$

- Complete system dynamics are:

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = K_P(\theta_d - \theta) + K_D(\dot{\theta}_d - \dot{\theta}) + \frac{g(q)}{K_r} - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}$$

- The dynamics used for design are now closer to actual dynamics

- Why stop at gravity term? Let's get rid of all annoying elements.

- $u =$ Controller for achieving desired trajectory +
Inverses of annoying terms in dynamics

- System dynamics are:

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = u - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r} + d$$

- Let $u =$ Controller for achieving desired trajectory + $\left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m + \frac{g(q)}{K_r}$



Inverse Dynamics change dynamics exactly to linear system.

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = \text{Trajectory Controller} + \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m + \frac{g(q)}{K_r} - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r} + d$$

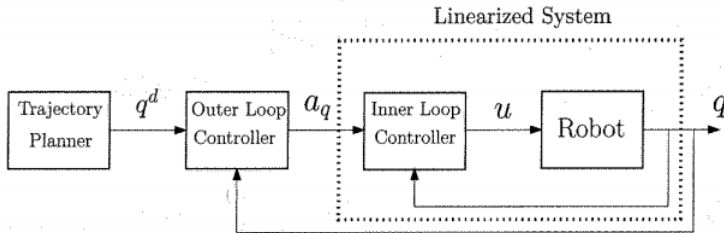
- Can we get rid of inertia term too?
- Let $u = \left(J_m + \frac{J(q)}{K_r^2} \right) a_q + \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m + \frac{g(q)}{K_r}$, where a_q is controller to achieve desired trajectory.
- System dynamics reduce to: $\ddot{q}_m = a_q$. System is now a linear decoupled one. This is exact (not an approximation).
- **Inverse Dynamics or Computed Torque Control.**

Inverse Dynamics implements two controllers.

- Complete system dynamics are:

$$\left(J_m + \frac{J(q)}{K_r^2} \right) \ddot{q}_m = \left(J_m + \frac{J(q)}{K_r^2} \right) a_q + \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m + \frac{g(q)}{K_r} - \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m - \frac{g(q)}{K_r}$$

- New Objective:** Find a_q , so that $\ddot{q}_m = a_q$ gives $q_m = q_d$.





What should be controller a_q ?

- What if $a_q = \ddot{q}_d$?
- Let's be safe, and add feedback in case there are errors in the model or if there is a disturbance.

- Say $a_q = \ddot{q}_d(t) + K_p(q_d - q + K_d(\dot{q}_d - \dot{q}))$.

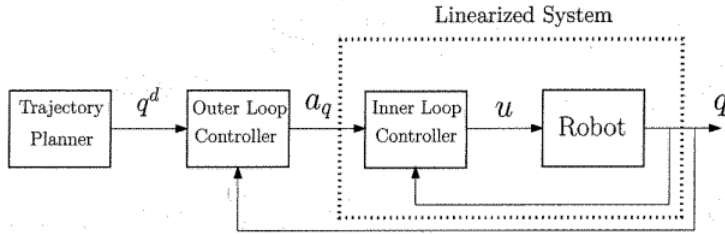
- Since $\ddot{q}_m = a_q$, we have

$$\ddot{q}_e + K_d\dot{q}_e + K_pq_e = 0,$$

where $q_e = q_d - q$.

- We know that, we can choose K_p and K_d to make any error converge to zero at a suitable rate.

Inverse Dynamics: Complete picture



$$u = \left(J_m + \frac{J(q)}{K_r^2} \right) [\ddot{q}_d(t) + K_p(q_d - q) + K_d(\dot{q}_d - \dot{q})] + \left(B_s + \frac{C(q, \dot{q})}{K_r^2} \right) \dot{q}_m + \frac{g(q)}{K_r}$$

- Measure q and \dot{q} .
- Substitute q and \dot{q} to find $g(q)$, $J(q)$, and $C(q, \dot{q})$.

- Substitute in expression for u

- The dynamics of an arm can in general be expressed as:

$$J(q)\ddot{q} + C(q, \dot{q})\dot{q} + B\dot{q} + g(q) = u.$$

- Let $u = \hat{J}(q) a_q + \hat{C}(q, \dot{q})\dot{q} + \hat{B}\dot{q} + \hat{g}(q)$
- Notice the hats on J , C , B , and g . These are our estimates for these quantities. What do we get if we now substitute u ?

$$\ddot{q} = a_q + \eta(q, \dot{q}, a_q),$$

where

$$\eta = J^{-1} (\tilde{J}a_q + \tilde{C}\dot{q} + \tilde{B}\dot{q} + \tilde{g})$$

and

$$(\tilde{\cdot}) = (\cdot) - (\hat{\cdot}).$$

- Will our original $a_q = \ddot{q}_d + K_p(q_d - q) + K_d(\dot{q}_d - \dot{q})$ still work?
- Perhaps. But we would like some guarantees.

If we're certain that

$$\|\eta\| \leq \rho(\tilde{q}, \dot{\tilde{q}}, t)$$

then we can design a controller

$$a_q = \ddot{q}_d - K_P(q - q_d) - K_D(\dot{q} - \dot{q}_d) + \delta a,$$

which guarantees global convergence of tracking error.



$$\delta a = \begin{cases} -\rho \frac{B^T P e}{\|B^T P e\|} & \text{if } \|B^T P e\| \neq 0 \\ 0 & \text{if } \|B^T P e\| = 0 \end{cases},$$

$$e = (\tilde{q}^T, \tilde{\dot{q}}^T)^T,$$

$$B = (0, I_n)^T,$$

$$Q = PA + A^T P,$$

$$A = \begin{bmatrix} 0 & I_n \\ -K_P & -K_D \end{bmatrix}$$

and $Q > 0$ and symmetric,

$$\rho = \frac{1}{1-\alpha} [\alpha\beta + \|K\| \|e\| + \lambda_H \phi(e, t)],$$

$$1 > \alpha \geq \|J^{-1} \hat{J} - I_n\|$$

$$\beta > \sup_{t \in [0, \infty)} \|\ddot{q}_d(t)\|$$

$$K = (K_P, K_D)$$

$$\lambda_H \geq \|J^{-1}\|$$

$$\phi(e, t) \geq \|\tilde{C}\dot{q} + \tilde{B}\dot{\dot{q}} + \tilde{g}\|$$



Adaptive Inverse Dynamics

- Adaptive controllers estimate controller parameters online and are adapt at dealing with imperfect knowledge of dynamical parameters.
- Recall the dynamics of an arm can be expressed as:

$$J(q)\ddot{q} + C(q, \dot{q})\dot{q} + B\dot{q} + g(q) = u.$$

- Let

$$\begin{aligned} u &= \hat{J}(q) \hat{a}_q + \hat{C}(q, \dot{q})\dot{q} + \hat{B}\dot{q} + \hat{g}(q) \\ &= \hat{J}(q) [\ddot{q}_d - K_P(q - q_d) - K_D(\dot{q} - \dot{q}_d)] + \hat{C}(q, \dot{q})\dot{q} + \hat{B}\dot{q} + \hat{g}(q) \end{aligned}$$

- Adaptive control assumes that \hat{J} , \hat{C} , \hat{B} , \hat{g} are not constants but time varying quantities now, which we'll update at each iteration.

- The entire idea of Adaptive control hinges on linear parameterization property of adaptive dynamics, according to which

$$J(q)\ddot{q} + C(q, \dot{q})\dot{q} + B\dot{q} + g(q) = Y(q, \dot{q}, \ddot{q}) a,$$

where Y , called regressor is a known matrix, and a is a vector of parameters.

- Similarly,

$$\hat{J}(q)\ddot{q} + \hat{C}(q, \dot{q})\dot{q} + \hat{B}\dot{q} + \hat{g}(q) = Y(q, \dot{q}, \ddot{q}) \hat{a}$$

- How does this help us?

$$\begin{aligned} u &= \hat{J}(q) [\ddot{q}_d - K_P(q - q_d) - K_D(\dot{q} - \dot{q}_d)] + \hat{C}(q, \dot{q})\dot{q} + \hat{B}\dot{q} + \hat{g}(q) \\ &= \hat{J}(q) [\ddot{q}_d - K_P(q - q_d) - K_D(\dot{q} - \dot{q}_d)] + Y(q, \dot{q}, \ddot{q})\hat{a} - \hat{J}(q)\ddot{q} \\ &= \hat{J}(q) [(\ddot{q}_d - \ddot{q}) + K_P(q_d - q) + K_D(\dot{q}_d - \dot{q})] + Y(q, \dot{q}, \ddot{q})\hat{a} \end{aligned}$$

- We can find an update equation of the form:

$$\hat{a} = -\Gamma^{-1} Y^T \hat{J}^{-1} B^T P (e_q^T, \dot{e}_q^T)^T$$

- We know that,

$$\dot{X} = J(q)\dot{q}$$

$$\ddot{X} = J(q)\ddot{q} + \dot{J}(q)\dot{q}$$

$$\tau = J^T(q)F$$

where J is analytical Jacobian, X is vector of end-effector positions and orientations, τ is vector of joint torques, and F is vector of end-effector forces and torques.

- Substituting these into $M(q)\ddot{q} + C(q, \dot{q})\dot{q} + B\dot{q} + g(q) = u$, we can show that

$$M(q) [J^{-1}\ddot{X} - J^{-1}\dot{J}J^{-1}\dot{X}] + C(q, \dot{q})J^{-1}\dot{X} + BJ^{-1}\dot{X} + g(q) = J^T(q)F.$$

$$\Lambda(q) \ddot{X} + \Gamma(q, \dot{q}) \dot{X} + \eta(q) = F,$$

where

$$\Lambda(q) = J^{-T}(q)M(q)J^{-1}(q)$$

$$\Gamma(q, \dot{q}) = J^{-T}(q)C(q, \dot{q})J^{-1}(q) + J^{-T}(q)BJ^{-1}(q) - \Lambda(q)\dot{J}(q)J^{-1}(q)$$

$$\eta(q) = J^{-T}(q)g(q)$$

- Same controllers can be designed in task space using these dynamics.
- The end-effector position is seldom directly measured. So, even if we're working in task space we may have to use joint sensors and kinematics equations for X to be fed into control algorithm.