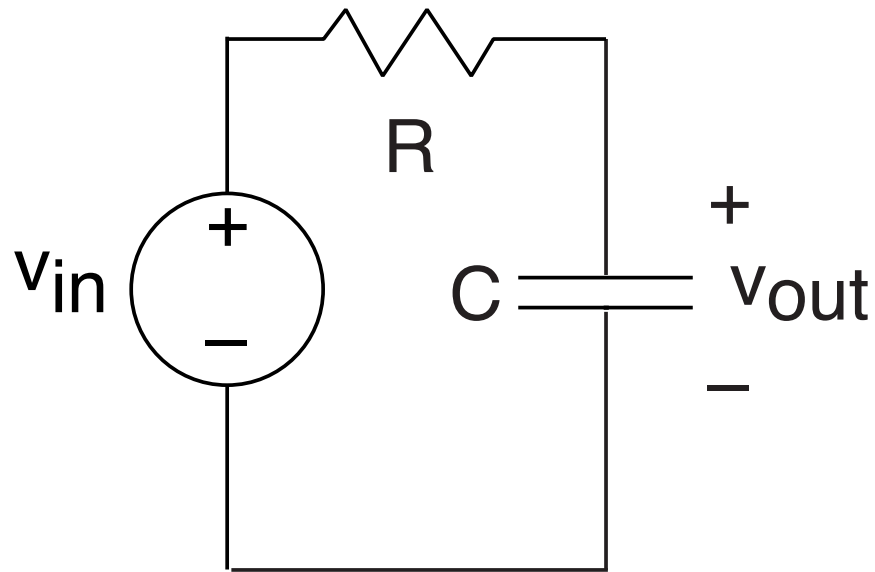


Fundamentals of Electrical Engineering

Circuits with Capacitors and Inductors

- Filters
- Thévenin and Mayer-Norton equivalents
- Complex power

What is the circuit doing?



$$v_{in}(t) = A \cos(2\pi f_0 t) = \text{Re}[Ae^{j2\pi f_0 t}]$$

$$V_{out} = \frac{1}{j2\pi f_0 RC + 1} V_{in}$$

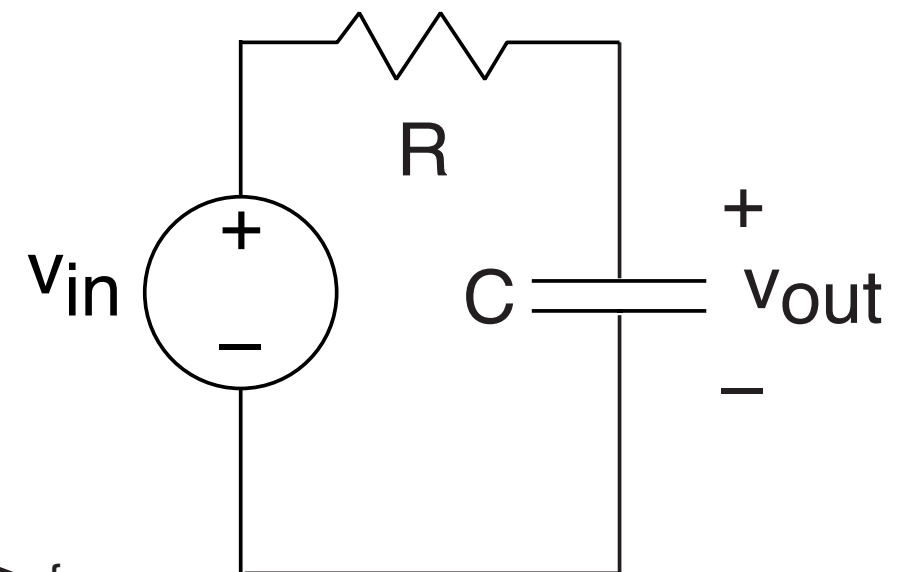
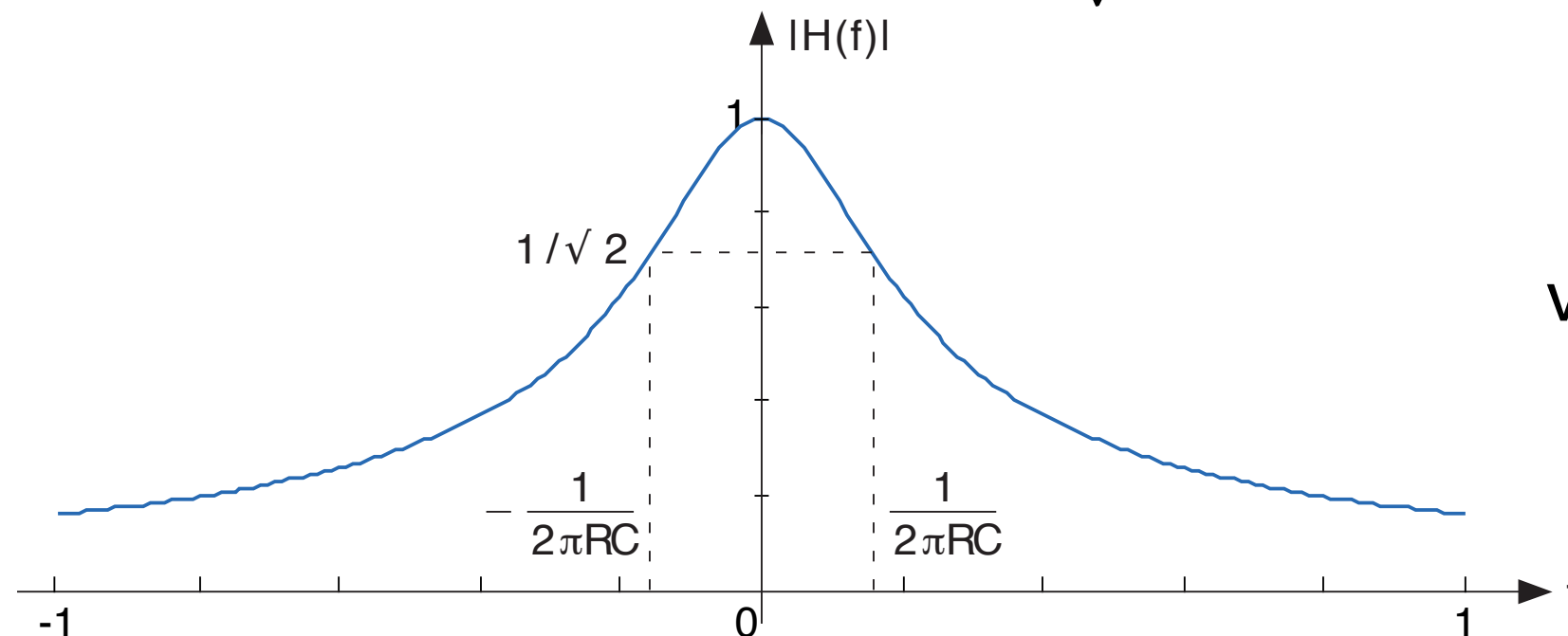
$$v_{out}(t) = \text{Re} \left[\frac{1}{j2\pi f_0 RC + 1} Ae^{j2\pi f_0 t} \right]$$

$$v_{out}(t) = \frac{A}{\sqrt{4\pi^2 f_0^2 R^2 C^2 + 1}} \cos(2\pi f_0 t - \tan^{-1} 2\pi f_0 RC)$$

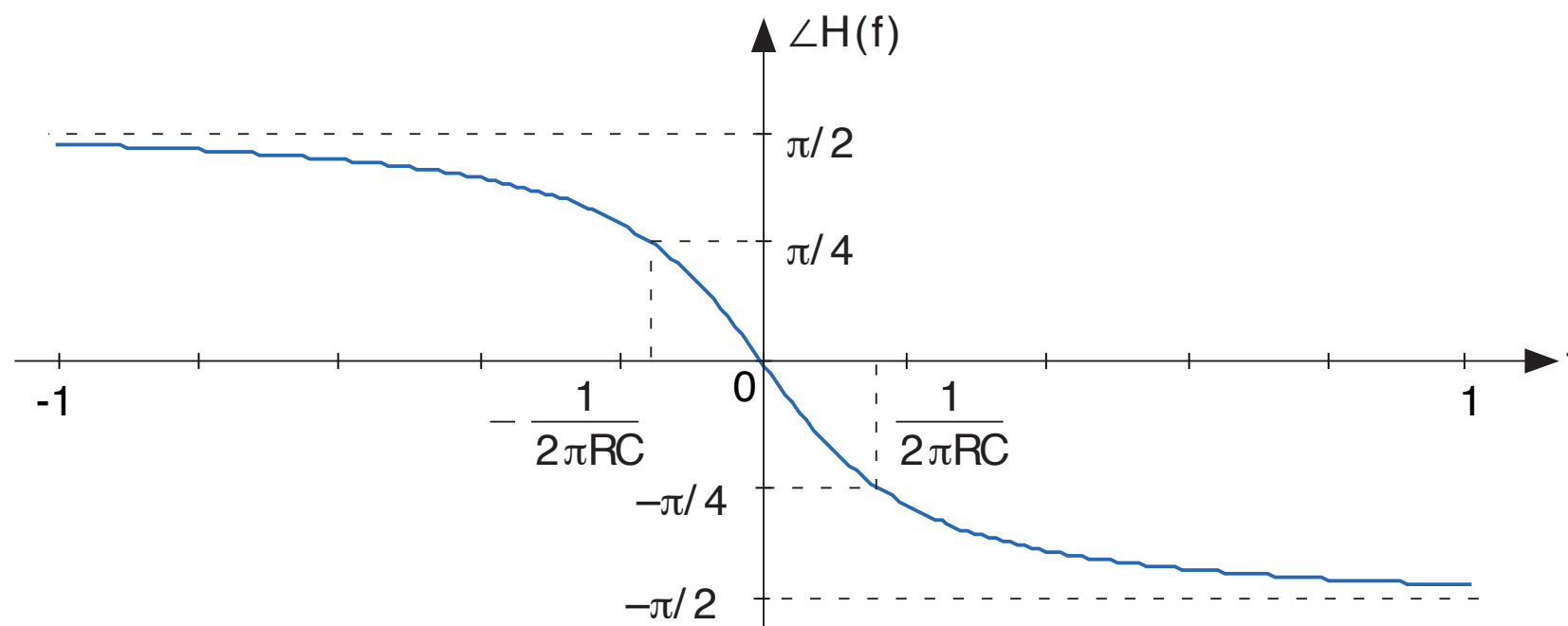
$$H(f) = |H(f)|e^{j\angle H(f)} = \frac{1}{\sqrt{4\pi^2 f^2 R^2 C^2 + 1}} e^{-j \tan^{-1} 2\pi f RC}$$

What is the circuit doing?

$$H(f) = |H(f)|e^{j\angle H(f)} = \frac{1}{\sqrt{4\pi^2 f^2 R^2 C^2 + 1}} e^{-j \tan^{-1} 2\pi f RC}$$



Lowpass filter



Filtering

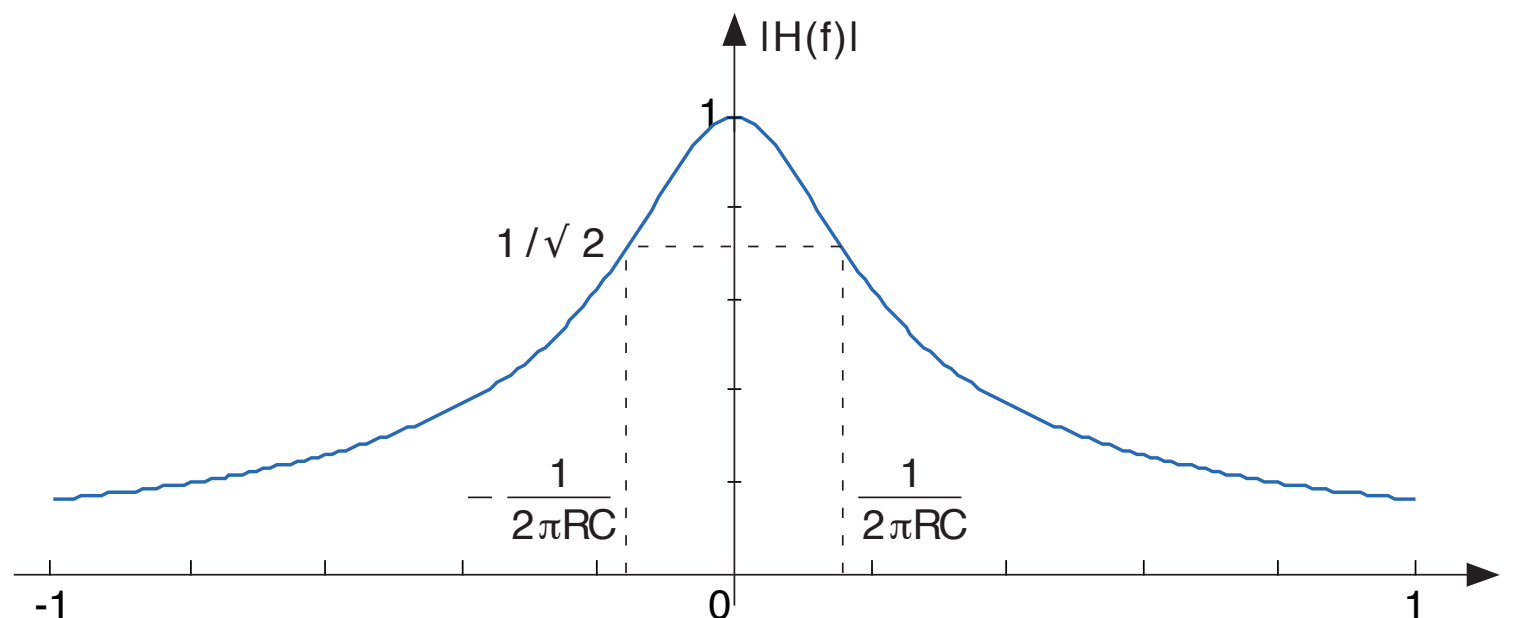
$$v_{\text{in}}(t) = A_1 \cos(2\pi f_1 t) + A_2 \cos(2\pi f_2 t)$$

$$v_{\text{in}}(t) = A_1 \cos(2\pi f_1 t) \implies v_{\text{out}}(t) = |H(f_1)| A_1 \cos(2\pi f_1 t + \angle H(f_1))$$

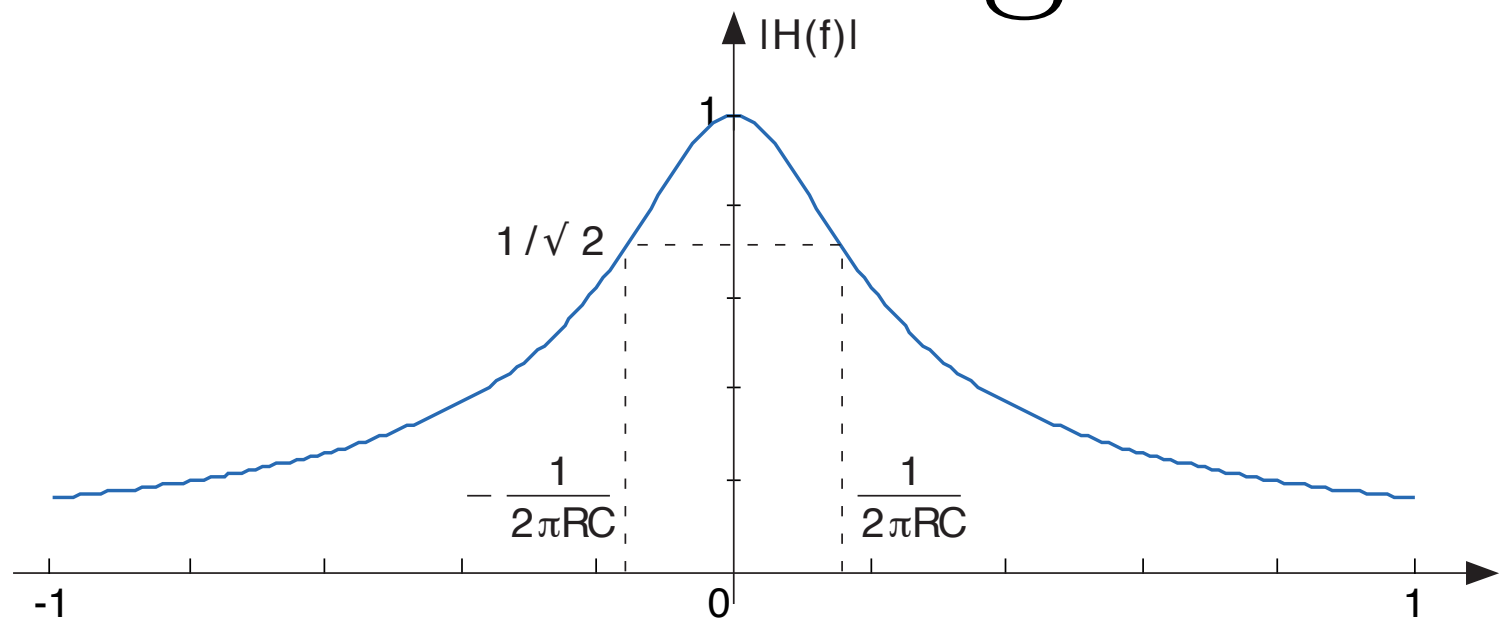
$$v_{\text{in}}(t) = A_2 \cos(2\pi f_2 t) \implies v_{\text{out}}(t) = |H(f_2)| A_2 \cos(2\pi f_2 t + \angle H(f_2))$$

Because the circuit is *linear*, **superposition applies**

$$v_{\text{out}}(t) = |H(f_1)| A_1 \cos(2\pi f_1 t + \angle H(f_1)) \\ + |H(f_2)| A_2 \cos(2\pi f_2 t + \angle H(f_2))$$



Filtering



Cutoff frequency: frequency f_c at which

$$|H(f_c)| = \frac{1}{\sqrt{2}} \max_f |H(f)|$$

For the RC lowpass,

$$|H(f)| = \frac{1}{\sqrt{4\pi^2 f^2 R^2 C^2 + 1}}$$

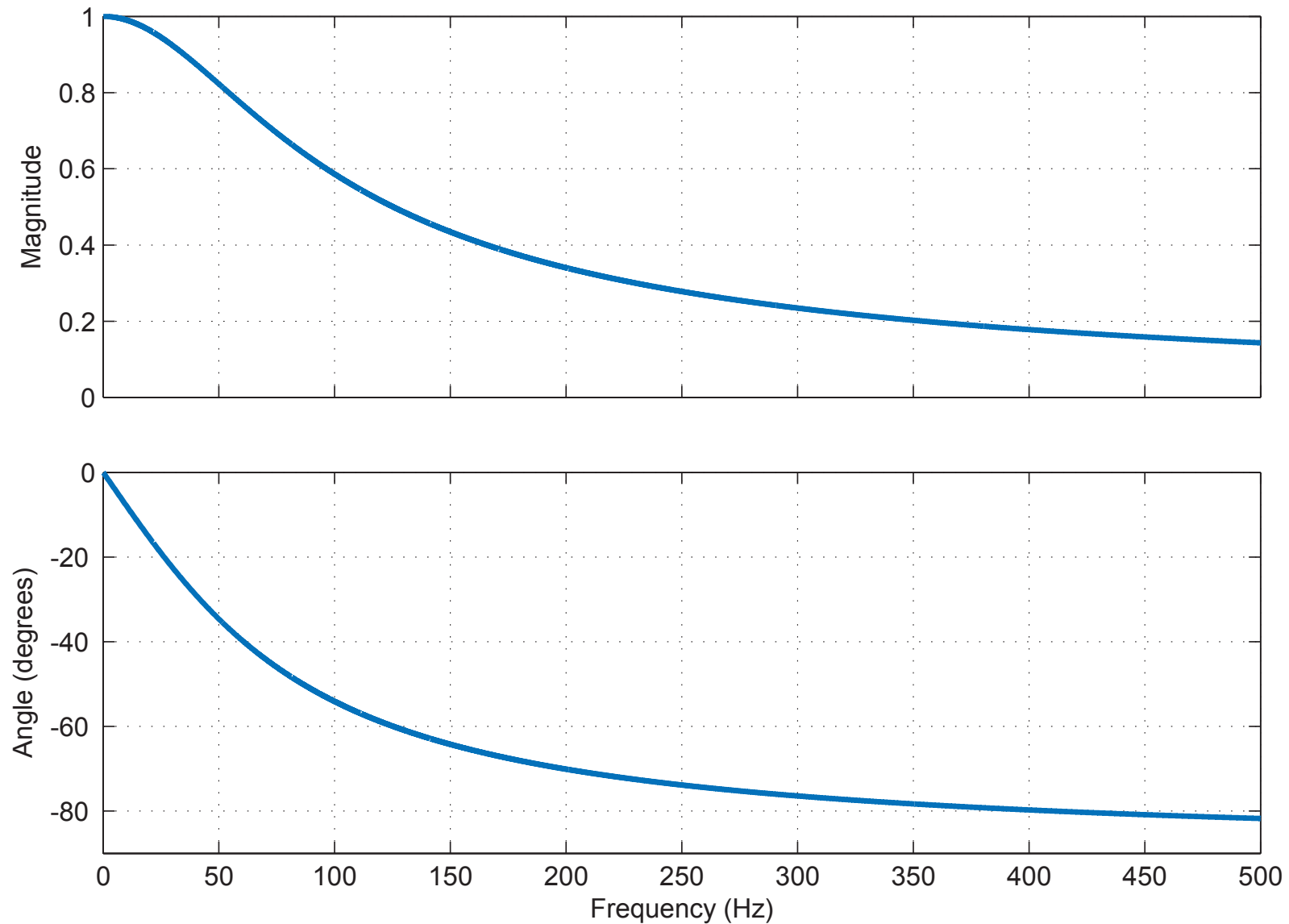
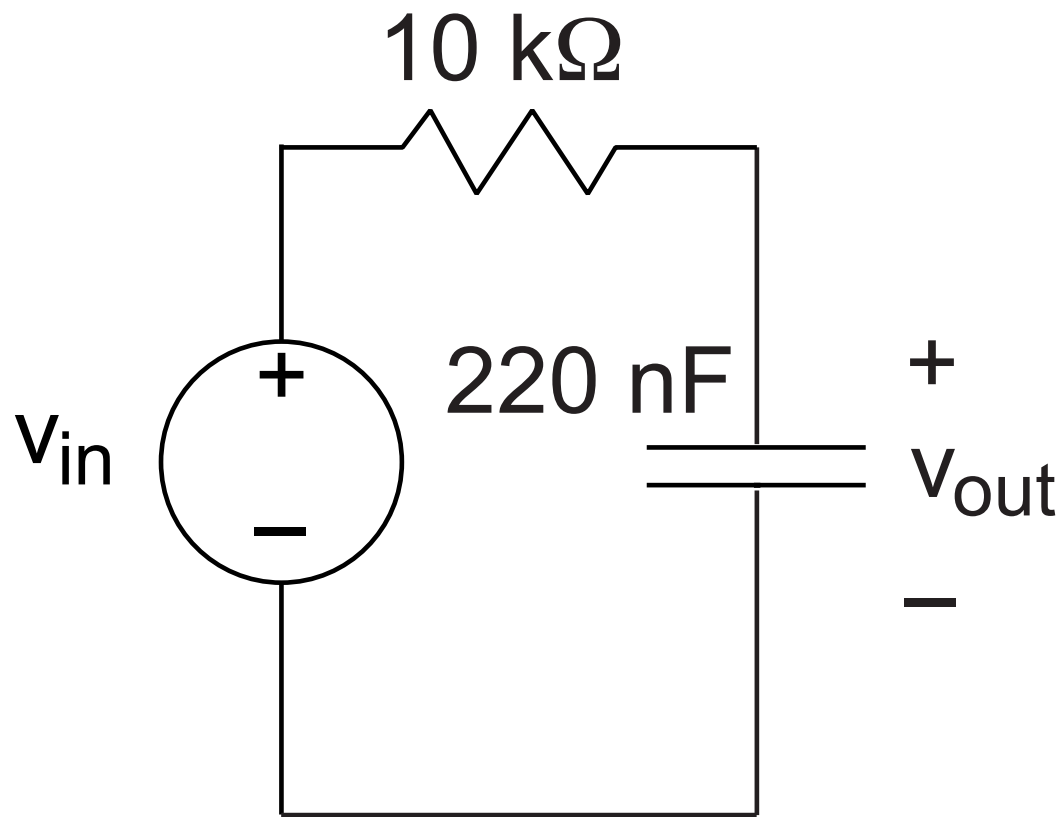
$$f_c = \frac{1}{2\pi RC}$$

Filtering

Several filter types are common.

- Lowpass
- Highpass
- Bandpass

RC Lowpass in Action

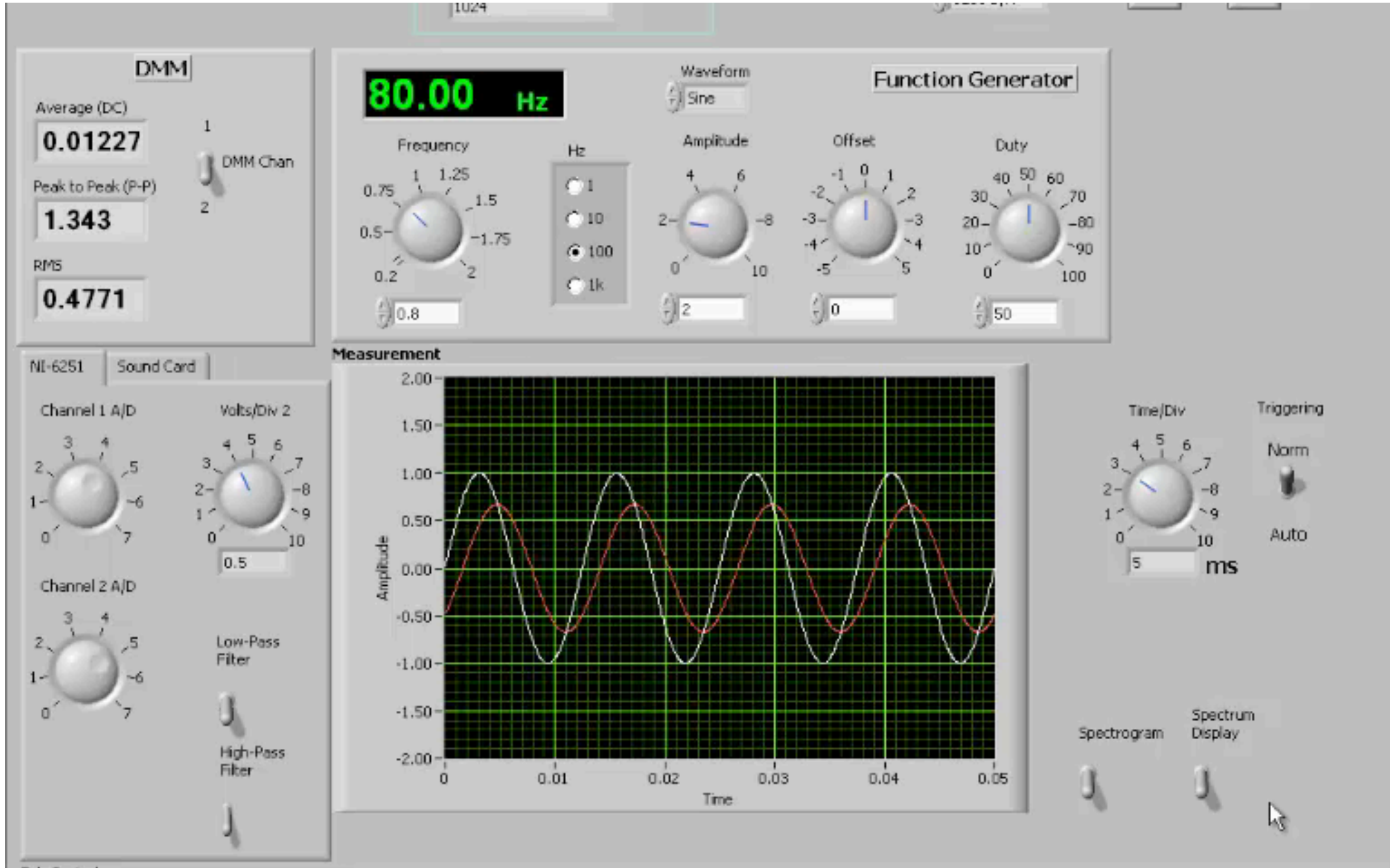


$$v_{\text{in}}(t) = \sin(2\pi ft) \quad v_{\text{out}}(t) = |H(f)| \sin(2\pi ft + \angle H(f))$$

$$H(f) = |H(f)|e^{j\angle H(f)} = \frac{1}{\sqrt{4\pi^2 f^2 R^2 C^2 + 1}} e^{-j \tan^{-1} 2\pi f RC}$$

$$f_c = \frac{1}{2\pi RC} = \frac{1}{2\pi \times 10^4 \times 220 \times 10^{-9}} = 72.34 \text{ Hz}$$

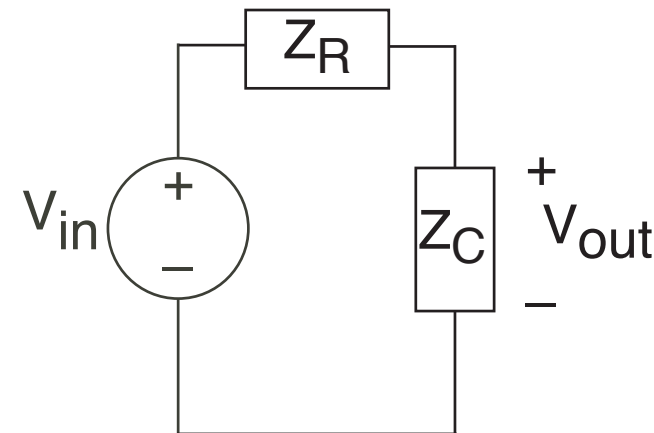
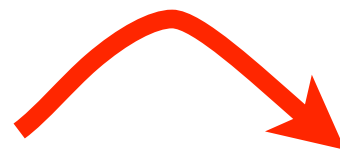
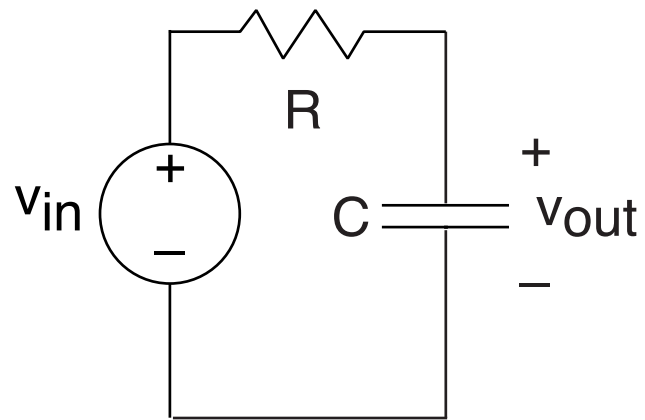
RC Lowpass in Action



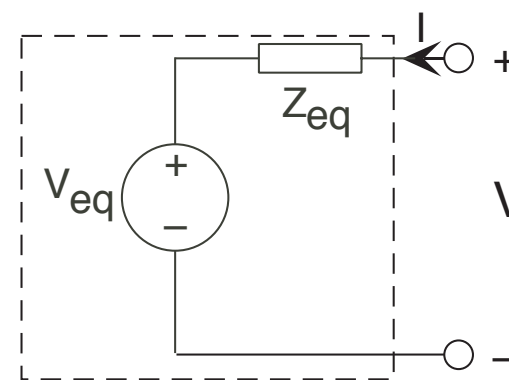
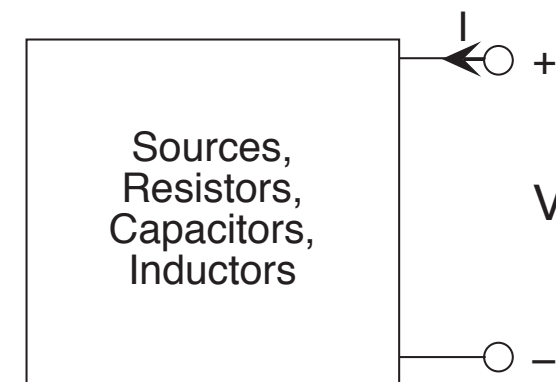
The Big Picture

$$v = V e^{j2\pi f t}, \quad i = I e^{j2\pi f t}$$

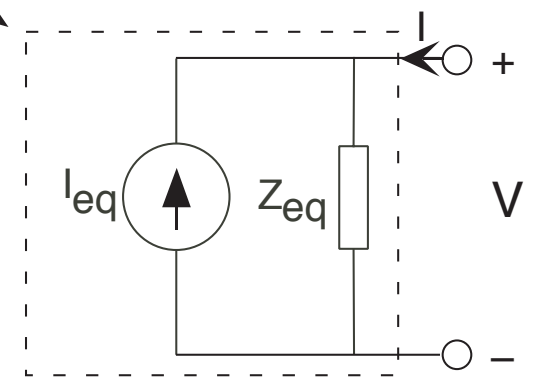
$$Z \equiv \frac{V}{I}$$



- Series/parallel “shortcuts” can be used
- Thévenin/Mayer-Norton equivalents have impedance “versions”

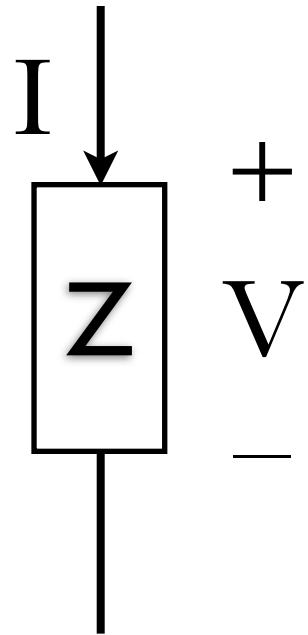


Thévenin Equivalent



Mayer-Norton Equivalent

Complex Power



$$Z \equiv \frac{V}{I}$$

$$v = V e^{j2\pi f t}, \quad i = I e^{j2\pi f t}$$

$$V = |V| e^{j\phi}, \quad I = |I| e^{j\theta}$$

$$v(t) = |V| \cos(2\pi f t + \phi) \quad i(t) = |I| \cos(2\pi f t + \theta)$$

$$p(t) = v(t) \cdot i(t)$$

Complex Power

$$V = |V|e^{j\phi}, \quad I = |I|e^{j\theta}$$

$$v(t) = \frac{1}{2} (V e^{j2\pi ft} + V^* e^{-j2\pi ft})$$

$$i(t) = \frac{1}{2} (I e^{j2\pi ft} + I^* e^{-j2\pi ft})$$

$$p(t) = v(t) \cdot i(t)$$

$$p(t) = \frac{1}{4} (V I^* + V^* I + V I e^{j4\pi ft} + V^* I^* e^{-j4\pi ft})$$

$$= \frac{1}{2} \operatorname{Re} [V I^*] + \frac{1}{2} \operatorname{Re} [V I e^{j4\pi ft}]$$

$$= \frac{1}{2} \operatorname{Re} [V I^*] + \frac{1}{2} |V| |I| \cos (4\pi ft + \phi + \theta)$$

$$P_{\text{ave}} = \frac{1}{2} \operatorname{Re} [V I^*]$$

Complex Power

$$P = \frac{1}{2}VI^* \quad P_{\text{ave}} = \frac{1}{2}\text{Re}[VI^*]$$

$$V = ZI$$

$$VI^* = Z|I|^2 = \frac{1}{Z^*}|V|^2$$

element	P	P_{ave}
R	$\frac{1}{2}R I ^2$	$\frac{1}{2}R I ^2$
C	$\frac{1}{2} \cdot \frac{1}{j2\pi fC} I ^2$	0
L	$\frac{1}{2} \cdot j2\pi fL I ^2$	0

Using Impedances

- By “thinking” of each element as a complex-valued resistor—as an impedance—a general picture emerges of what circuits do and how they behave
 - * Series/parallel rules
 - * Thévenin and Mayer-Norton equivalents
 - * Transfer functions and filtering
 - * Complex power
- *Only* applies when the source is a sinusoid