# TSKS14 Multiple Antenna Communications

Lecture 2, 2020

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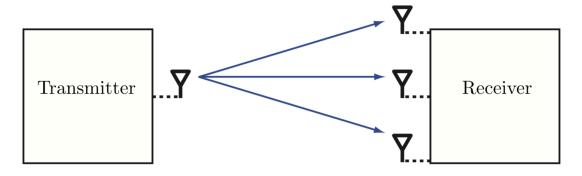
#### Taxonomy: Point-to-point channels

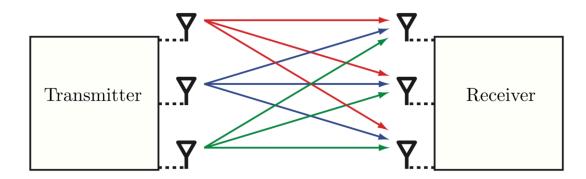


Receiver Transmitter Transmitter Receiver

(a) Point-to-point SISO channel.







(b) Point-to-point SIMO channel.

(d) Point-to-point MIMO channel.



Single-input single-output (SISO)
Single-input multiple-output (SIMO)

Multiple-input single-output (MISO)
Multiple-input multiple-output (MIMO)

#### Outline of this lecture

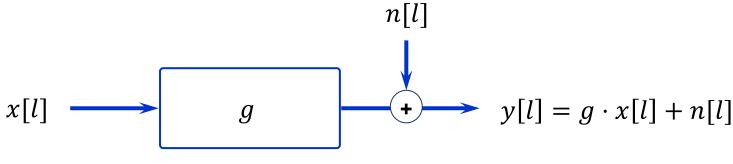
- Recap of SISO channel capacity
- Capacity of SIMO and MISO channels
- Channel models for line-of-sight channels
- Beamwidth with uniform linear arrays



#### Not all transmit power is received

- Wireless signals spread out as a sphere
  - Fraction of received signal at distance d:  $\frac{\lambda^2}{(4\pi d)^2}$
- Cellular communication: 70 dB to 130 dB is lost!
  - Channel gain  $|g|^2$  is around -70 dB to -130 dB

Also known as pathloss



Channel response



#### Capacity of memoryless SISO channel

• AWGN channel with a complex channel gain g:

$$y[l] = g \cdot x[l] + n[l]$$

- $x[l] \sim CN(0, q)$ , energy per sample: q = P/B
- $n[l] \sim CN(0, N_0)$

Memoryless channel: Drop dependence on  $\it l$ 

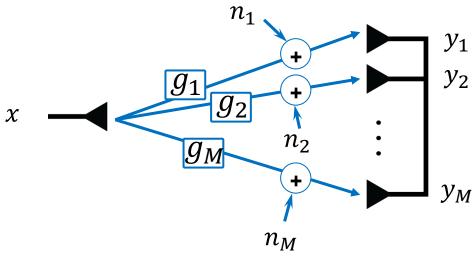
• Capacity computation: Treat  $g \cdot x \sim CN(0, q|g|^2)$  as the signal!

$$C = \log_2\left(1 + \frac{q|g|^2}{N_0}\right)$$
 bits per complex sample



How can we increase the SNR?

#### Single-input multiple-output (SIMO) channel



Noise:  $n_m \sim CN(0, N_0)$ 

• Received vector signal:

$$\begin{bmatrix} y_1 \\ \vdots \\ y_M \end{bmatrix} = \begin{bmatrix} g_1 \\ \vdots \\ g_M \end{bmatrix} x + \begin{bmatrix} n_1 \\ \vdots \\ n_M \end{bmatrix}$$

• Short form:

$$y = gx + n$$

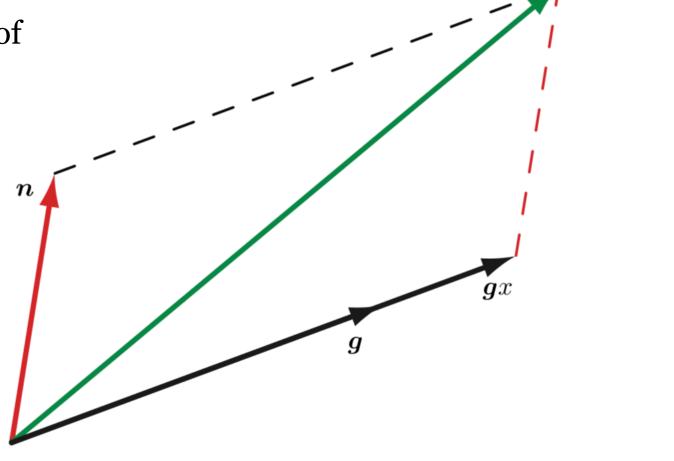


y = gx + n

#### Geometry of received signal

- Received vector **y** is summation of
  - Desired signal gx
  - Noise vector *n*
- Want to turn *y* into an estimate of *x*

Only part of *y* parallel to *g* contains desired signal

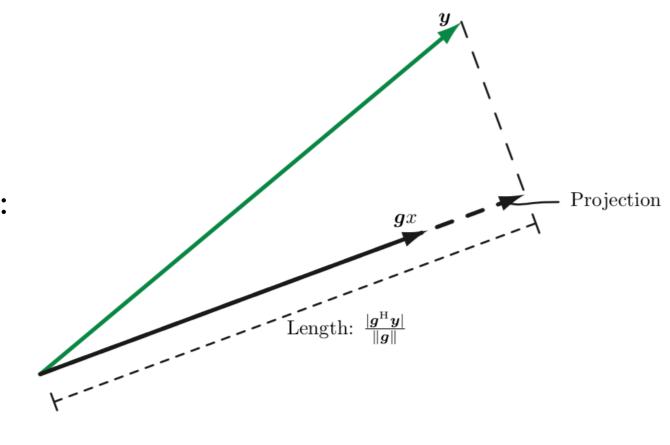




#### Capacity of SIMO channel

• Maximum ratio combining (MRC):

$$\frac{g^H}{\|g\|}y = \|g\|x + \frac{g^Hn}{\|g\|}$$
Combining vector Scalar



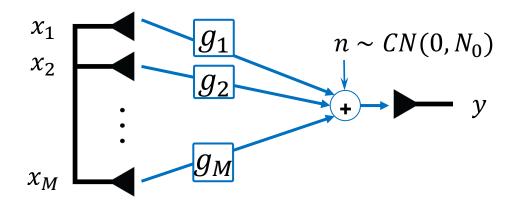
Like a SISO channel with  $g = \|\boldsymbol{g}\|$ 

• Capacity:

$$C = \log_2\left(1 + \frac{q||\boldsymbol{g}||^2}{N_0}\right) \text{ bits per complex sample}$$



#### Multiple-input single-output (MISO) channel



• Received vector signal:

$$y = [g_1 \quad \cdots \quad g_M] \begin{bmatrix} x_1 \\ \vdots \\ x_M \end{bmatrix} + n = \boldsymbol{g}^T \boldsymbol{x} + n$$

• Precoding:  $x = w\tilde{x}$ ,

Information signal:  $\tilde{x} \sim CN(0, q)$ 



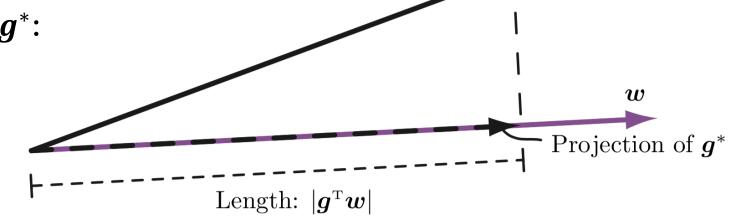
Unit-norm precoding vector

#### Geometry of received signal

• Channel projects signal onto  $g^*$ :

Maximum ratio transmission (MRT):

$$oldsymbol{w} = rac{oldsymbol{g}^*}{\|oldsymbol{g}\|}$$



$$y = \boldsymbol{g}^T \frac{\boldsymbol{g}^*}{\|\boldsymbol{g}\|} \tilde{x} + n = \|\boldsymbol{g}\| \tilde{x} + n$$

• Capacity:

$$C = \log_2\left(1 + \frac{q\|\boldsymbol{g}\|^2}{N_0}\right)$$
 bits per complex sample



#### Receive and transmit beamforming

Same capacity of SIMO and MISO channel

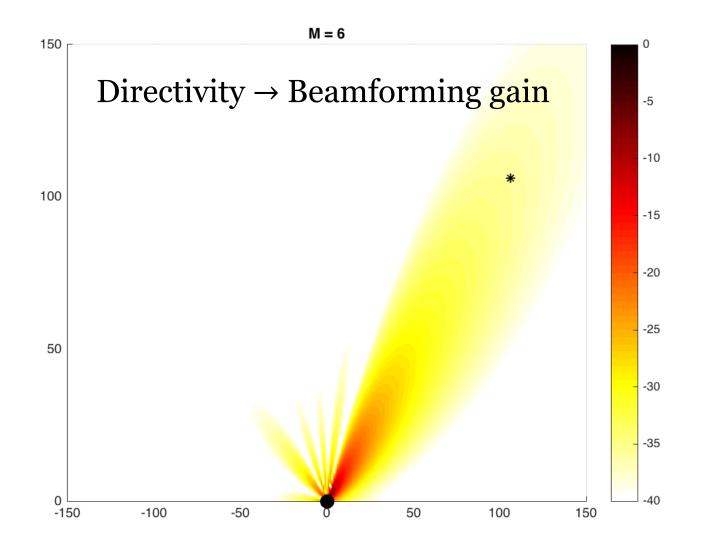
$$C = \log_2 \left( 1 + \frac{q \|\boldsymbol{g}\|^2}{N_0} \right)$$
 bits per complex sample

- Achieved by "beamforming" along channel vector g
- Beamforming gain of  $||g||^2$  (often proportional to M)
- Several different names
  - Maximum ratio combining/transmission (MRC, MRT)
  - Conjugate beamforming, matched filtering



#### Transmit beamforming

- Other names:
  - Array gain
  - Power gain

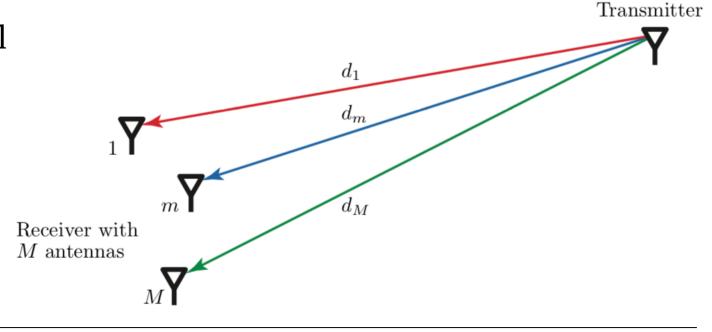




#### Free-space line-of-sight communication

- Direct path between transmitter and receiver
  - No other paths

• We will derive a channel model





#### Complex baseband representation

• Real-valued baseband signals (bandwidth B/2):

$$s_I(t)$$
 and  $s_Q(t)$ 

Complex baseband representation:

$$s_b(t) = s_I(t) + js_Q(t)$$

Real-valued passband signal

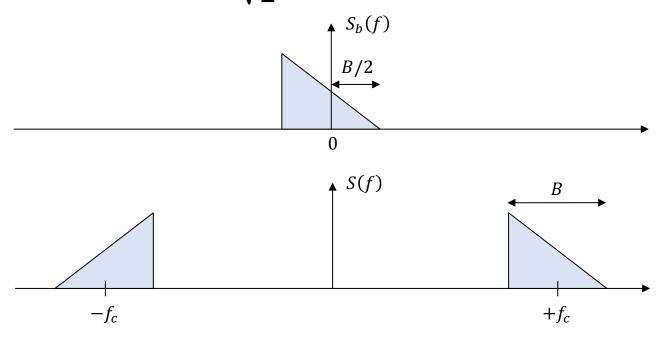
$$s(t) = s_I(t)\sqrt{2}\cos(2\pi f_c t) - s_Q(t)\sqrt{2}\sin(2\pi f_c t)$$
  
=  $Re\{\sqrt{2}s_b(t)e^{j2\pi f_c t}\}$ 

- Carrier frequency:  $f_c \gg B$
- Wavelength:  $\lambda = c/f_c$



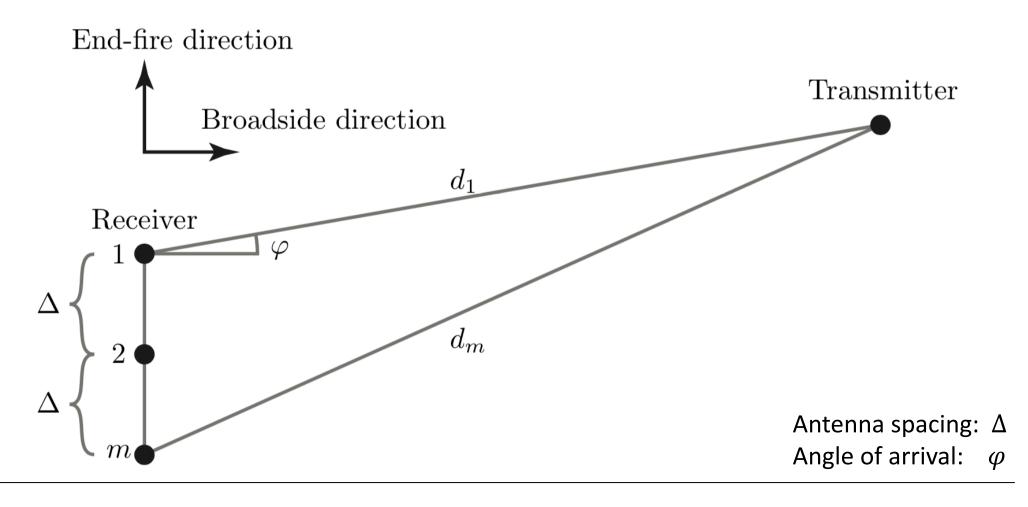
#### Complex baseband representation (2)

Fourier transforms:  $S_b(f) = \mathcal{F}\{s_b(t)\}$  $S(f) = \frac{1}{\sqrt{2}} \left( S_b(f - f_c) + S_b^*(-f - f_c) \right)$ 





#### Uniform linear array (ULA)





#### Impact of channel in complex baseband

• Transmitted signal:

$$s(t) = Re\{\sqrt{2}s_b(t)e^{j2\pi f_c t}\}$$

• Received signal at distance  $d_m$ :

$$\mu_m(t) = \frac{\lambda}{4\pi d_m} s\left(t - \frac{d_m}{c}\right) = \frac{\lambda}{4\pi d_m} Re\left\{\sqrt{2}s_b\left(t - \frac{d_m}{c}\right)e^{j2\pi f_c\left(t - \frac{d_m}{c}\right)}\right\}$$

- Receiver can compensate for time delay by selecting sampling time
  - Different  $d_m$ : Only one antenna has perfect sampling



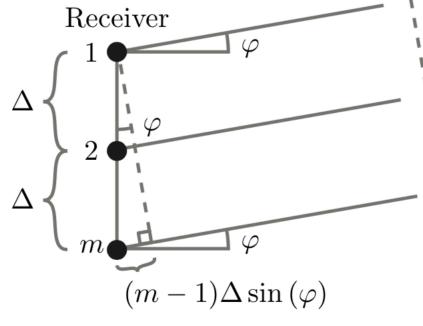
#### Far-field approximation

• Distances  $d_1, \dots, d_M$  approximately the same

Transmitter

Approximately same channel gain:  $\lambda^2$   $\lambda^2$ 

$$\frac{\lambda}{(4\pi d_m)^2} \approx \frac{\lambda}{(4\pi d)^2}$$



Spherical wavefronts

Planar wavefronts

$$Re\left\{\sqrt{2}s_{b}\left(t-\frac{d_{m}}{c}\right)e^{j2\pi f_{c}\left(t-\frac{d_{m}}{c}\right)}\right\}$$

Sample a



#### **ULA** channel model

- *M* antennas in receiver array
  - Angle of arrival  $\varphi$
  - Wavelength  $\lambda$
  - Distance Δ between antennas
  - Distance *d* to first antenna
- Channel vector:

$$\mathbf{g} = \frac{\lambda}{4\pi d} \begin{bmatrix} e^{-j2\pi \frac{\Delta \sin(\varphi)}{\lambda}} \\ \vdots \\ e^{-j2\pi \frac{(M-1)\Delta \sin(\varphi)}{\lambda}} \end{bmatrix}$$

Applies to both SIMO and MISO channels



#### Channel capacity with line-of-sight channel

• Recall:

$$C = \log_2 \left( 1 + \frac{q \|\boldsymbol{g}\|^2}{N_0} \right)$$
 bits per complex sample

• We have  $\|\boldsymbol{g}\|^2 = \beta M$  where  $\beta = \frac{\lambda^2}{(4\pi d)^2}$ :

where 
$$\beta = \frac{\lambda^2}{(4\pi d)^2}$$
:

$$C = \log_2\left(1 + \frac{q\beta M}{N_0}\right)$$

Beamforming gain: M

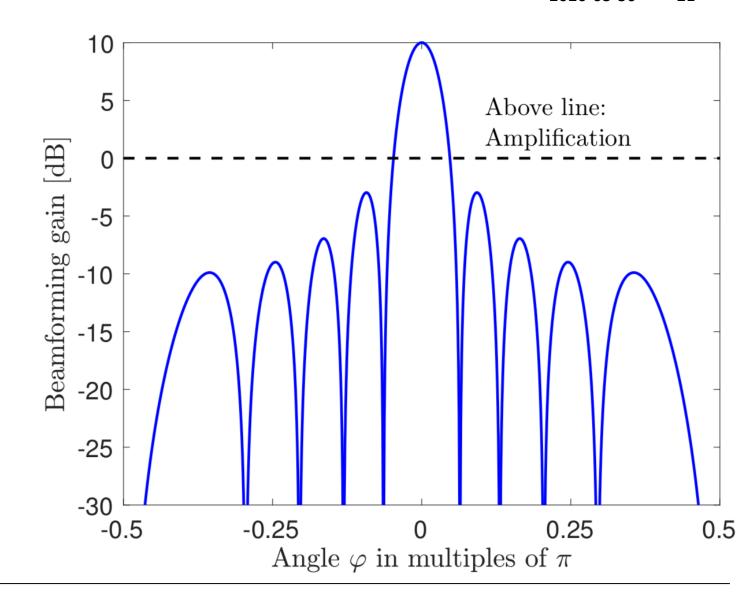


#### Beamwidth

- Beamforming gain in more than one direction
- Example: M = 10,  $\Delta = \lambda/2$
- First-null beamwidth:

$$\approx \frac{4}{M}$$

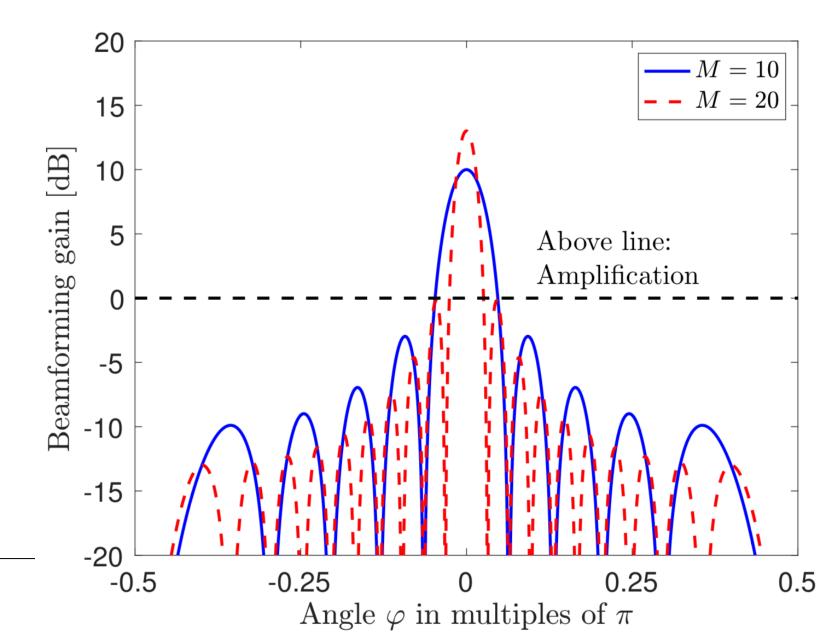
More antennas: Narrower beams





### Two benefits of beamforming

- Beamforming gain
  - Stronger signal at desired location
- Narrower beam
  - Less interference at undesired locations





#### End of Lecture 2

## TSKS14 Multiple Antenna Communications

