1

Digital Signal Processing

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Abstract—This manual provides a simple introduction to digital signal processing.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

https://github.com/AkshithaKola/ EE3900/blob/main/filter/code/ Sound%20Noise.way

2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find? Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the

- synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

Solution: The following is the python code to remove band noise

https://github.com/AkshithaKola/ EE3900/blob/main/filter/code /2.3.py

2.4 The output of the python script in Problem 2.3 is file the audio Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

The reduced noise audio file is

https://github.com/AkshithaKola/ EE3900/blob/main/filter/code/ Sound%20With%20 ReducedNoise.way

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

Solution:

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /3.2.py

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 3.2

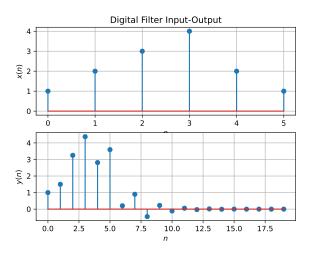


Fig. 3.2

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$Z\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)

resulting in (4.2). Similarly, it can be shown that

$$Z\{x(n-k)\} = z^{-k}X(z)$$
 (4.6)

4.2 Obtain X(z) for x(n) defined in problem 3.1.

Solution:

$$X(z) = \mathbb{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5}$$

$$(4.7)$$

$$= (4.7)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.9}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.10)

$$\implies \frac{Y(z)}{X(z)} = \frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{4.11}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.12)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.13)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.14}$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.15}$$

and from (4.13),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.16)

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.17}$$

using the formula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a| \tag{4.18}$$

Solution: Let $a^n u(n) = v(n)$

$$V(z) = \sum_{n=0}^{\infty} a^n u(n) z^{-n}$$
 (4.19)

$$=\sum_{n=0}^{\infty} a^n z^{-n}$$
 (4.20)

$$=\sum_{n=0}^{\infty} (az^{-1})^n \tag{4.21}$$

$$=\frac{1}{1-az^{-1}}\tag{4.22}$$

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.23)

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of x(n).

Solution: The following code plots Fig. 4.6.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /4.5.py

Using (4.11), we observe that $|H(e^{J\omega})|$ is given by

$$|H(e^{J\omega})| = \left| \frac{1 + e^{-2J\omega}}{1 + \frac{1}{2}e^{-J\omega}} \right|$$

$$= \sqrt{\frac{(1 + \cos 2\omega)^2 + (\sin 2\omega)^2}{\left(1 + \frac{1}{2}\cos \omega\right)^2 + \left(\frac{1}{2}\sin \omega\right)^2}}$$
(4.24)

$$=\sqrt{\frac{2(1+\cos 2\omega)}{\frac{5}{4}+\cos \omega}}\tag{4.26}$$

$$=\sqrt{\frac{2(2\cos^2\omega)}{\frac{5}{4}+\cos\omega}}\tag{4.27}$$

$$=\frac{4|\cos\omega|}{\sqrt{5+4\cos\omega}}\tag{4.28}$$

The period of numerator is π and the period of denominator is 2π

∴The period of $|H(e^{j\omega})|$ is LCM $(\pi,2\pi)$

$$\left| H\left(e^{J(\omega+2\pi)}\right) \right| = \frac{4\left|\cos\left(\omega+2\pi\right)\right|}{\sqrt{5+4\cos\left(\omega+2\pi\right)}} \quad (4.29)$$

$$= \frac{4|\cos\omega|}{\sqrt{5 + 4\cos\omega}} \tag{4.30}$$

$$\left| H\left(e^{J(\omega + 2\pi)} \right) \right| = \left| H\left(e^{J\omega} \right) \right| \tag{4.31}$$

and so its fundamental period is 2π .

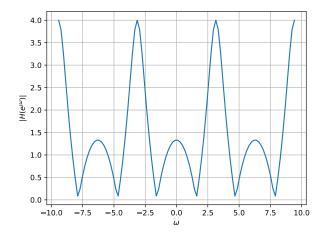


Fig. 4.6: $|H(e^{j\omega})|$

4.7 Express h(n) in terms of $H(e^{j\omega})$.

Solution: We have,

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k}$$
 (4.32)

on multiplying $e^{j\omega n}d\omega$ on both sides

$$\implies H(e^{j\omega})e^{j\omega n}d\omega = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k}e^{j\omega n}d\omega$$
(4.33)

Now integrating on both sides from $-\pi$ to π However,

$$\int_{-\pi}^{\pi} e^{j\omega(n-k)} d\omega = \begin{cases} 2\pi & n=k\\ 0 & \text{otherwise} \end{cases}$$
 (4.34)

and so,

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \tag{4.35}$$

$$=\frac{1}{2\pi}\sum_{k=-\infty}^{\infty}\int_{-\pi}^{\pi}h(k)e^{j\omega(n-k)}d\omega \qquad (4.36)$$

$$= \frac{1}{2\pi} 2\pi h(n) = h(n) \tag{4.37}$$

Thus,

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.38)$$

(4.39)

5 Impulse Response

5.1 Using long division, find h(n), n < 5 for H(z) in (4.11).

Solution: From (4.11), we have

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.1)

Substitute $z^{-1} = x$ to perform long division

$$\frac{2x-4}{\frac{1}{2}x+1} = \frac{2x-4}{x^2+1} = \frac{-x^2-2x}{-2x+1} = \frac{2x+4}{5}$$

From above division we can write,

$$1 + z^{-2} = (1 + \frac{1}{2}z^{-1})(2z^{-1} - 4) + 5$$
 (5.2)

$$\frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} = 2z^{-1} - 4 + \frac{5}{1+\frac{1}{2}z^{-1}}$$
 (5.3)

From (4.11), we can write

$$H(z) = -4 + 2z^{-1} + \frac{5}{1 + \frac{1}{2}z^{-1}}$$
 (5.4)

$$= -4 + 2z^{-1} + 5\sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n z^{-n}$$
 (5.5)

$$=1-\frac{1}{2}z^{-1}+5\sum_{n=2}^{\infty}\left(-\frac{1}{2}\right)^{n}z^{-n}$$
 (5.6)

$$= \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n z^{-n} + 4 \sum_{n=2}^{\infty} \left(-\frac{1}{2}\right)^n z^{-n} \quad (5.7)$$

$$=\sum_{n=-\infty}^{\infty}u(n)\left(-\frac{1}{2}\right)^{n}z^{-n}+$$

$$\sum_{n=-\infty}^{\infty} u(n-2) \left(-\frac{1}{2}\right)^{n-2} z^{-n}$$
 (5.8)

Therefore, from (4.1),

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.9)$$

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.10}$$

and there is a one to one relationship between

h(n) and H(z).

h(n) is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.11),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.11)

From (4.18),

$$\frac{1}{1 + \frac{1}{2}z^{-1}} \stackrel{\mathcal{Z}}{\rightleftharpoons} \left(-\frac{1}{2}\right)^n u(n) \quad |z| > \frac{1}{2} \tag{5.12}$$

$$\frac{z^{-2}}{1 + \frac{1}{2}z^{-1}} \stackrel{Z}{\rightleftharpoons} \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad |z| > \frac{1}{2}$$
(5.13)

$$\Rightarrow H(z) \stackrel{\mathcal{Z}}{\rightleftharpoons} \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad |z| > \frac{1}{2}$$
(5.14)
(5.15)

(Since Z-transform is a linear operator)

$$\therefore h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.16)

From (5.11), Consider the first part:

$$\frac{1}{1 + \frac{1}{2}z^{-1}} = \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n z^n \quad |z| > \frac{1}{2} \quad (5.17)$$

and

$$\frac{1}{1 + \frac{1}{2}z^{-1}} = \frac{2z}{1 + 2z}$$

$$= \sum_{n = -\infty}^{-1} (2)^{-n} (-1)^{-n+1} z^{-n} |z| < \frac{1}{2}$$
(5.18)

Therefore, ROC of H(z) will be

$$|z| \neq \frac{1}{2} \tag{5.20}$$

5.3 Sketch h(n). Is it bounded? Convergent? **Solution:** The following code plots Fig. 5.3.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /5.3.py

and run the code using the following command

python3 5.3.py

From the plot, we can conclude that it is convergent to $\boldsymbol{0}$

: It is bounded as well.

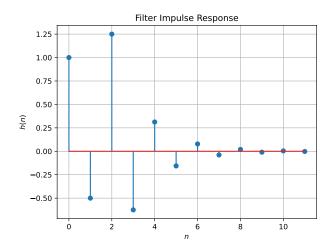


Fig. 5.3: h(n) as the inverse of H(z)

5.4 Is it convergent? Justify using the ratio test. **Solution:** Using the ratio test for convergence

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \to \infty} \left| \frac{\left(-\frac{1}{2}\right)^{n-1} \left(1 + \frac{1}{4}\right)}{\left(-\frac{1}{2}\right)^{n-2} \left(1 + \frac{1}{4}\right)} \right| \quad (5.21)$$

$$= \lim_{n \to \infty} \left| -\frac{1}{2} \right| \quad (5.22)$$

$$= \frac{1}{2} < 1 \quad (5.23)$$

 \therefore h(n) is Convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.24}$$

Is the system defined by (3.2) stable for the impulse response in (5.10)?

Solution: From 5.2,

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^n u(n) + \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$

$$= \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^n + \sum_{n=0}^{\infty} \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$

$$= \frac{1}{1 - \left(-\frac{1}{2}\right)} + \frac{1}{1 - \left(-\frac{1}{2}\right)}$$

$$= \frac{4}{3} < \infty$$
(5.28)

using the fomula for the sum of an infinite geometric progression

- ... The system is stable.
- 5.6 Verify the above result using a python code. **Solution:** The following code verifies whether the given system is stable or not

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /5.6.py

run the code using the following command

python3 5.6.py

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.29)$$

This is the definition of h(n).

Solution: The following code plots Fig. 5.7. Note that this is the same as Fig. 5.3.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /5.7.py

run the code using the following command

python3 5.7.py

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.30)

Comment. The operation in (5.30) is known as *convolution*.

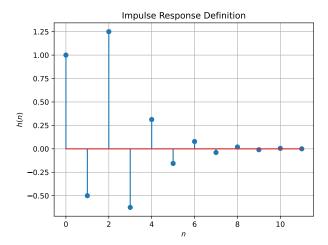


Fig. 5.7: h(n) from the definition

Solution:

$$x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.31)
=
$$\sum_{k=0}^{5} x(k)h(n-k)$$
 (5.32)

The following code plots Fig. 5.9.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /5.8.py

run the code using the following command

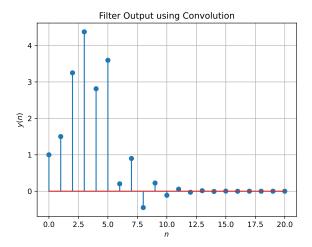


Fig. 5.8: y(n) from the definition of convolution

This plot is same as y(n) in Fig. ??

Therefore,

$$y(n) = x(n) * h(n)$$
 (5.33)

5.9 Express the above convolution using a Teoplitz matrix.

Solution: From (3.1), we can write

$$\mathbf{x} = \begin{pmatrix} 1 \\ 2 \\ 3 \\ 4 \\ 2 \\ 1 \end{pmatrix} \qquad \mathbf{h} = \begin{pmatrix} 1 \\ -0.5 \\ 1.25 \\ -0.62 \\ 0.31 \\ -0.16 \end{pmatrix} \tag{5.34}$$

Their convolution is given by the product of the following Toeplitz matrix T

$$\begin{pmatrix} 1 & 0 & 0 & 0 & 0 & 0 \\ -0.5 & 1 & 0 & 0 & 0 & 0 \\ 1.25 & -0.5 & 1 & 0 & 0 & 0 \\ -0.62 & 1.25 & -0.5 & 1 & 0 & 0 \\ 0.31 & -0.62 & 1.25 & -0.5 & 1 & 0 \\ -0.16 & 0.31 & -0.62 & 1.25 & -0.5 & 1 \\ 0 & -0.16 & 0.31 & -0.62 & 1.25 & -0.5 \\ 0 & 0 & -0.16 & 0.31 & -0.62 & 1.25 \\ 0 & 0 & 0 & -0.16 & 0.31 & -0.62 \\ 0 & 0 & 0 & 0 & -0.16 & 0.31 \\ 0 & 0 & 0 & 0 & 0 & -0.16 \\ 0 & 0 & 0 & 0 & 0 & -0.16 \end{pmatrix}$$

$$\mathbf{y} = \mathbf{x} \circledast \mathbf{h} = \mathbf{T}\mathbf{x} = \begin{pmatrix} 1\\ 1.5\\ 3.25\\ 4.38\\ 2.81\\ 3.59\\ 0.12\\ 0.78\\ -0.62\\ 0\\ -0.16 \end{pmatrix}$$
 (5.36)

The following python code computes the convolution using Teoplitz matrix.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /5.9.py

run the code using the following command

python3 5.9.py

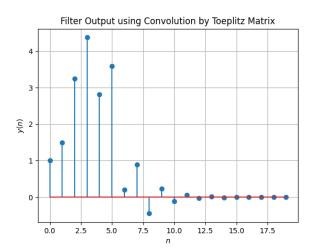


Fig. 5.9: y(n) from the definition of convolution using Teoplitz matrix

5.10 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.37)

Solution: From (5.30) we know that,

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.38)

Substitute k = n-i

$$\sum_{k=-\infty}^{\infty} x(k)h(n-k) = \sum_{n-i=-\infty}^{\infty} x(n-i)h(n-(n-i))$$
(5.39)

$$=\sum_{i=-\infty}^{-\infty}x(n-i)h(i)$$
 (5.40)

$$=\sum_{i=-\infty}^{\infty}x(n-i)h(i) \qquad (5.41)$$

Since, the order of limit doesn't matter in case of summation. Therefore, now we have

$$\sum_{k=-\infty}^{\infty} x(k)h(n-k) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.42)

from (5.30)

$$\therefore y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.43)

6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

Solution: The following code plots Fig. 6.1.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.1.py

run the above code using the command.

python3 6.1.py

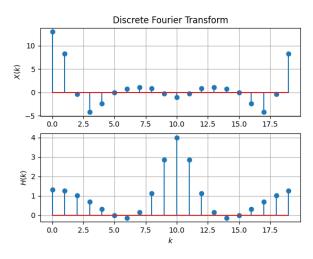


Fig. 6.1: Plots of the real parts of the DFT of x(n) and h(n)

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

Solution: Download and run the following code.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.2.py

run the above code using the command.

python3 6.2.py

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. 5.9. Note that this is the same as y(n) in Fig. 3.2.

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.3.py

run the above code using the command.

python3 6.3.py

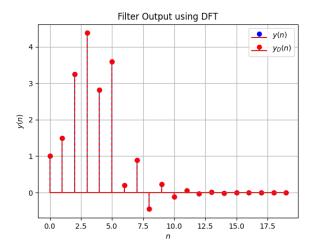


Fig. 6.3: y(n) from the DFT

6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT. **Solution:** Download the code from

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.4.py

and execute it using

\$ python3 6.4.py

Observe that Fig. (6.4) is the same as y(n) in Fig. (3.2).

6.5 Wherever possible, express all the above equations as matrix equations.

Solution: We use the DFT Matrix, where $\omega = e^{-\frac{j2k\pi}{N}}$, which is given by

$$\mathbf{W} = \begin{pmatrix} \omega^0 & \omega^0 & \dots & \omega^0 \\ \omega^0 & \omega^1 & \dots & \omega^{N-1} \\ \vdots & \vdots & \ddots & \vdots \\ \omega^0 & \omega^{N-1} & \dots & \omega^{(N-1)(N-1)} \end{pmatrix}$$
(6.4)

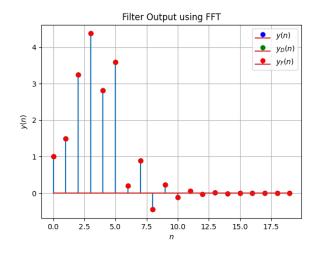


Fig. 6.4: y(n) using FFT and IFFT

i.e. $W_{jk} = \omega^{jk}$, $0 \le j, k < N$. Hence, we can write any DFT equation as

$$\mathbf{X} = \mathbf{W}\mathbf{x} = \mathbf{x}\mathbf{W} \tag{6.5}$$

where

$$\mathbf{x} = \begin{pmatrix} x(0) \\ x(1) \\ \vdots \\ x(n-1) \end{pmatrix}$$
 (6.6)

Using (6.3), the inverse Fourier Transform is given by

$$\mathbf{x} = \mathcal{F}^{-1}(\mathbf{X}) = \mathbf{W}^{-1}\mathbf{X} = \frac{1}{N}\mathbf{W}^{\mathbf{H}}\mathbf{X} = \frac{1}{N}\mathbf{X}\mathbf{W}^{\mathbf{H}}$$
(6.7)

$$\implies \mathbf{W}^{-1} = \frac{1}{N} \mathbf{W}^{\mathbf{H}} \tag{6.8}$$

where H denotes hermitian operator. We can rewrite (6.2) using the element-wise multiplication operator as

$$\mathbf{Y} = \mathbf{H} \cdot \mathbf{X} = (\mathbf{W}\mathbf{h}) \cdot (\mathbf{W}\mathbf{x}) \tag{6.9}$$

6.6 Verify the above equation by generating the DFT matrix in python.

Solution: Download the code from

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.6.py

and execute it using

\$ python3 6.6.py

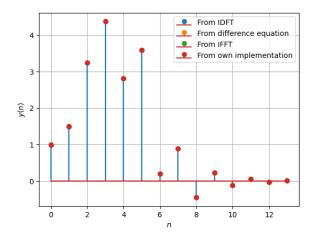


Fig. 6.6: y(n) using DFT matrix

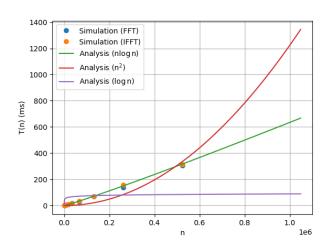


Fig. 6.7: Complexity of FFT/IFFT is O(n log n)

The above code plots (6.6)

6.7 Compute the 8-point FFT in C.

Solution: Download the following C code

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.7.c

run the above C code using the command

The above code generates the text files that are loaded in the following code

Download the following code

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.7.1.py https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /6.7.2.py

run the above code using the command

The above code plots the graphs 6.7 and 6.7

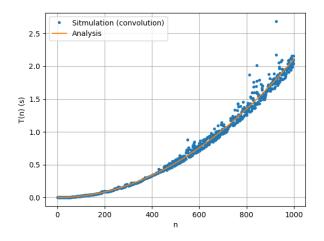


Fig. 6.7: Complexity of convolution is $O(n^2)$

7 FFT

7.1 Definitions

1. The DFT of x(n) is given by

$$X(k) \triangleq \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(7.1)

2. Let

$$W_N = e^{-j2\pi/N} \tag{7.2}$$

Then the N-point DFT matrix is defined as

$$\mathbf{F}_N = \begin{bmatrix} W_N^{mn} \end{bmatrix} \tag{7.3}$$

where W_N^{mn} are the elements of \mathbf{F}_N .

3. Let

$$\mathbf{I}_4 = \begin{pmatrix} \mathbf{e}_4^1 & \mathbf{e}_4^2 & \mathbf{e}_4^3 & \mathbf{e}_4^4 \end{pmatrix} \tag{7.4}$$

be the 4×4 identity matrix. Then the 4 point DFT permutation matrix is defined as

$$\mathbf{P}_4 = \begin{pmatrix} \mathbf{e}_4^1 & \mathbf{e}_4^3 & \mathbf{e}_4^2 & \mathbf{e}_4^4 \end{pmatrix} \tag{7.5}$$

4. The 4 point DFT diagonal matrix is defined as

$$\mathbf{D}_{4} = diag \begin{pmatrix} W_{N}^{0} & W_{N}^{1} & W_{N}^{2} & W_{N}^{3} \end{pmatrix}$$
 (7.6)

7.2 Problems

1. Show that

$$W_N^2 = W_{N/2} (7.7)$$

Solution: From defination

$$W_N = e^{-j2\pi/N} \tag{7.8}$$

$$W_N^2 = \left(e^{-j2\pi/N}\right)^2$$
 (7.9)
= $e^{-j2\pi/N/2}$ (7.10)

$$=e^{-j2\pi/N/2} (7.10)$$

$$=W_{N/2}$$
 (7.11)

2. Find P_6 .

Solution: $P_6 = \begin{pmatrix} e_4^1 & e_4^3 & e_4^5 & e_4^2 & e_4^4 & e_4^6 \end{pmatrix}$

$$\mathbf{P}_{6} = \begin{pmatrix} 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \end{pmatrix}$$
(7.12)

3. Find \mathbf{D}_3 .

Solution:

$$\mathbf{D}_3 = diag \left(W_3^0 \quad W_3^1 \quad W_3^2 \right) \tag{7.13}$$

$$= \begin{pmatrix} 1 & 0 & 0\\ 0 & e^{-j2\pi/3} & 0\\ 0 & 0 & (e^{-j2\pi/3})^2 \end{pmatrix}$$
(7.14)

4. Show that

$$\mathbf{F}_4 = \begin{bmatrix} \mathbf{I}_2 & \mathbf{D}_2 \\ \mathbf{I}_2 & -\mathbf{D}_2 \end{bmatrix} \begin{bmatrix} \mathbf{F}_2 & 0 \\ 0 & \mathbf{F}_2 \end{bmatrix} \mathbf{P}_4 \tag{7.15}$$

Solution:

$$\mathbf{F}_4 \mathbf{P}_4 = \begin{bmatrix} \mathbf{I}_2 & \mathbf{D}_2 \\ \mathbf{I}_2 & -\mathbf{D}_2 \end{bmatrix} \begin{bmatrix} \mathbf{F}_2 & 0 \\ 0 & \mathbf{F}_2 \end{bmatrix}$$
 (7.16)

$$= \begin{bmatrix} \mathbf{I}_2 \mathbf{F}_2 & \mathbf{D}_2 \mathbf{F}_2 \\ \mathbf{I}_2 \mathbf{F}_2 & -\mathbf{D}_2 \mathbf{F}_2 \end{bmatrix}$$
 (7.17)

$$= \begin{bmatrix} \mathbf{F}_2 & \mathbf{D}_2 \mathbf{F}_2 \\ \mathbf{F}_2 & -\mathbf{D}_2 \mathbf{F}_2 \end{bmatrix} \tag{7.18}$$

$$= \begin{bmatrix} W_2^0 & W_2^0 & W_2^0 & W_2^0 \\ W_2^0 & W_2^1 & W_2^1 & W_2^2 \\ W_2^0 & W_2^0 & -W_2^0 & -W_2^0 \\ W_2^0 & W_2^1 & -W_2^1 & -W_2^2 \end{bmatrix}$$
(7.19)

$$= \begin{bmatrix} 1 & 1 & 1 & 1\\ 1 & W_2^1 & W_2^1 & W_2^2\\ 1 & W_2^0 & -W_2^0 & -W_2^0\\ 1 & W_1^1 & -W_1^1 & -W_2^2 \end{bmatrix}$$
(7.20)

from eqn(7.7) we get $W_2 = W_4^2$

$$\mathbf{F}_{4}\mathbf{P}_{4} = \begin{bmatrix} W_{4}^{0} & W_{4}^{0} & W_{4}^{0} & W_{4}^{0} \\ W_{4}^{0} & W_{4}^{2} & W_{4}^{1} & W_{4}^{3} \\ W_{4}^{0} & W_{4}^{4} & W_{4}^{2} & W_{4}^{6} \\ W_{0}^{0} & W_{0}^{6} & W_{3}^{3} & W_{9}^{9} \end{bmatrix}$$
(7.21)

5. Show that

$$\mathbf{F}_{N} = \begin{bmatrix} \mathbf{I}_{N/2} & \mathbf{D}_{N/2} \\ \mathbf{I}_{N/2} & -\mathbf{D}_{N/2} \end{bmatrix} \begin{bmatrix} \mathbf{F}_{N/2} & 0 \\ 0 & \mathbf{F}_{N/2} \end{bmatrix} \mathbf{P}_{N} \quad (7.22)$$

Solution: Observe that for even N and letting \mathbf{f}_N^i denote the i^{th} column of \mathbf{F}_N ,

$$\begin{pmatrix} \mathbf{D}_{N/2} \mathbf{F}_{N/2} \\ -\mathbf{D}_{N/2} \mathbf{F}_{N/2} \end{pmatrix} = \begin{pmatrix} \mathbf{f}_N^2 & \mathbf{f}_N^4 & \dots & \mathbf{f}_N^N \end{pmatrix}$$
(7.23)

and

$$\begin{pmatrix} \mathbf{I}_{N/2} \mathbf{F}_{N/2} \\ \mathbf{I}_{N/2} \mathbf{F}_{N/2} \end{pmatrix} = \begin{pmatrix} \mathbf{f}_N^1 & \mathbf{f}_N^3 & \dots & \mathbf{f}_N^{N-1} \end{pmatrix}$$
(7.24)

Thus.

$$\begin{pmatrix} \mathbf{I}_{2}\mathbf{F}_{2} & \mathbf{D}_{2}\mathbf{F}_{2} \\ \mathbf{I}_{2}\mathbf{F}_{2} & -\mathbf{D}_{2}\mathbf{F}_{2} \end{pmatrix} = \begin{pmatrix} \mathbf{I}_{N/2} & \mathbf{D}_{N/2} \\ \mathbf{I}_{N/2} & -\mathbf{D}_{N/2} \end{pmatrix} \begin{pmatrix} \mathbf{F}_{N/2} & 0 \\ 0 & \mathbf{F}_{N/2} \end{pmatrix}$$
$$= \begin{pmatrix} \mathbf{f}_{N}^{1} & \dots & \mathbf{f}_{N}^{N-1} & \mathbf{f}_{N}^{2} & \dots & \mathbf{f}_{N}^{N} \end{pmatrix}$$
(7.25)

and so,

$$\begin{pmatrix} \mathbf{I}_{N/2} & \mathbf{D}_{N/2} \\ \mathbf{I}_{N/2} & -\mathbf{D}_{N/2} \end{pmatrix} \begin{pmatrix} \mathbf{F}_{N/2} & 0 \\ 0 & \mathbf{F}_{N/2} \end{pmatrix} \mathbf{P}_{N}$$
$$= \begin{pmatrix} \mathbf{f}_{N}^{1} & \mathbf{f}_{N}^{2} & \dots & \mathbf{f}_{N}^{N} \end{pmatrix} = \mathbf{F}_{N}$$
(7.26)

6. Find

$$\mathbf{P}_4\mathbf{x} \tag{7.27}$$

Solution: We have,

$$\mathbf{P}_{4}\mathbf{x} = \begin{pmatrix} \mathbf{e}_{4}^{1} & \mathbf{e}_{4}^{3} & \mathbf{e}_{4}^{2} & \mathbf{e}_{4}^{4} \end{pmatrix} \begin{pmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \end{pmatrix} = \begin{pmatrix} x(0) \\ x(2) \\ x(1) \\ x(3) \end{pmatrix} (7.28)$$

7. Show that

$$\mathbf{X} = \mathbf{F}_N \mathbf{x} \tag{7.29}$$

where \mathbf{x}, \mathbf{X} are the vector representations of x(n), X(k) respectively.

Solution: Writing the terms of X,

$$X(0) = x(0) + x(1) + \dots + x(N-1)$$
(7.30)

$$X(1) = x(0) + x(1)e^{-\frac{\sqrt{2}\pi}{N}} + \dots + x(N-1)e^{-\frac{\sqrt{2}(N-1)\pi}{N}}$$
(7.31)

$$X(N-1) = x(0) + x(1)e^{-\frac{12(N-1)\pi}{N}} + \dots + x(N-1)e^{-\frac{12(N-1)(N-1)\pi}{N}}$$
(7.32)

Clearly, the term in the m^{th} row and n^{th} column is given by $(0 \le m \le N - 1 \text{ and } 0 \le n \le N - 1)$

$$T_{mn} = x(n)e^{-\frac{j2mn\pi}{N}}$$
 (7.33)

and so, we can represent each of these terms as a matrix product

$$\mathbf{X} = \mathbf{F}_N \mathbf{x} \tag{7.34}$$

where $\mathbf{F}_N = \left[e^{-\frac{-j2mn\pi}{N}}\right]_{mn}$ for $0 \le m \le N-1$ and $0 \le n \le N-1$.

8. Derive the following Step-by-step visualisation of 8-point FFTs into 4-point FFTs and so on

$$\begin{bmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \end{bmatrix} = \begin{bmatrix} X_1(0) \\ X_1(1) \\ X_1(2) \\ X_1(3) \end{bmatrix} + \begin{bmatrix} W_8^0 & 0 & 0 & 0 \\ 0 & W_8^1 & 0 & 0 \\ 0 & 0 & W_8^2 & 0 \\ 0 & 0 & 0 & W_8^3 \end{bmatrix} \begin{bmatrix} X_2(0) \\ X_2(1) \\ X_2(2) \\ X_2(3) \end{bmatrix}$$

$$(7.35)$$

$$\begin{bmatrix} X(4) \\ X(5) \\ X(6) \\ X(7) \end{bmatrix} = \begin{bmatrix} X_1(0) \\ X_1(1) \\ X_1(2) \\ X_1(3) \end{bmatrix} - \begin{bmatrix} W_8^0 & 0 & 0 & 0 \\ 0 & W_8^1 & 0 & 0 \\ 0 & 0 & W_8^2 & 0 \\ 0 & 0 & 0 & W_8^3 \end{bmatrix} \begin{bmatrix} X_2(0) \\ X_2(1) \\ X_2(2) \\ X_2(3) \end{bmatrix}$$

$$(7.36)$$

4-point FFTs into 2-point FFTs

$$\begin{bmatrix} X_1(0) \\ X_1(1) \end{bmatrix} = \begin{bmatrix} X_3(0) \\ X_3(1) \end{bmatrix} + \begin{bmatrix} W_4^0 & 0 \\ 0 & W_4^1 \end{bmatrix} \begin{bmatrix} X_4(0) \\ X_4(1) \end{bmatrix}$$
(7.37)

$$\begin{bmatrix} X_1(2) \\ X_1(3) \end{bmatrix} = \begin{bmatrix} X_3(0) \\ X_3(1) \end{bmatrix} - \begin{bmatrix} W_4^0 & 0 \\ 0 & W_4^1 \end{bmatrix} \begin{bmatrix} X_4(0) \\ X_4(1) \end{bmatrix}$$
 (7.38)

$$\begin{bmatrix} X_2(0) \\ X_2(1) \end{bmatrix} = \begin{bmatrix} X_5(0) \\ X_5(1) \end{bmatrix} + \begin{bmatrix} W_4^0 & 0 \\ 0 & W_4^1 \end{bmatrix} \begin{bmatrix} X_6(0) \\ X_6(1) \end{bmatrix}$$
 (7.39)

$$\begin{bmatrix} X_2(2) \\ X_2(3) \end{bmatrix} = \begin{bmatrix} X_5(0) \\ X_5(1) \end{bmatrix} - \begin{bmatrix} W_4^0 & 0 \\ 0 & W_4^1 \end{bmatrix} \begin{bmatrix} X_6(0) \\ X_6(1) \end{bmatrix}$$
 (7.40)

$$P_{8} \begin{bmatrix} x(0) \\ x(1) \\ x(2) \\ x(3) \\ x(4) \\ x(5) \\ x(6) \\ x(7) \end{bmatrix} = \begin{bmatrix} x(0) \\ x(2) \\ x(4) \\ x(6) \\ x(1) \\ x(3) \\ x(5) \\ x(7) \end{bmatrix}$$
 (7.41)

$$P_{4} \begin{bmatrix} x(0) \\ x(2) \\ x(4) \\ x(6) \end{bmatrix} = \begin{bmatrix} x(0) \\ x(4) \\ x(2) \\ x(6) \end{bmatrix}$$
 (7.42)

$$P_{4} \begin{bmatrix} x(1) \\ x(3) \\ x(5) \\ x(7) \end{bmatrix} = \begin{bmatrix} x(1) \\ x(5) \\ x(3) \\ x(7) \end{bmatrix}$$
 (7.43)

Therefore,

$$\begin{bmatrix} X_3(0) \\ X_3(1) \end{bmatrix} = F_2 \begin{bmatrix} x(0) \\ x(4) \end{bmatrix}$$
 (7.44)

$$\begin{bmatrix} X_4(0) \\ X_4(1) \end{bmatrix} = F_2 \begin{bmatrix} x(2) \\ x(6) \end{bmatrix}$$
 (7.45)

$$\begin{bmatrix} X_5(0) \\ X_5(1) \end{bmatrix} = F_2 \begin{bmatrix} x(1) \\ x(5) \end{bmatrix}$$
 (7.46)

$$\begin{bmatrix} X_6(0) \\ X_6(1) \end{bmatrix} = F_2 \begin{bmatrix} x(3) \\ x(7) \end{bmatrix}$$
 (7.47)

9. For

$$\mathbf{x} = \begin{pmatrix} 1\\2\\3\\4\\2\\1 \end{pmatrix} \tag{7.48}$$

compte the DFT using (7.29) **Solution:** Download the code from

https://github.com/AkshithaKola/ EE3900/blob/main/filter/code /7.9.py

and execute it using

\$ python3 7.9.py

10. Repeat the above exercise using the FFT after zero padding **x**.

Solution: Download the code from

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /7.10.py

and execute it using

\$ python3 7.10.py

11. Write a C program to compute the 8-point FFT. **Solution:** Download the code from

https://github.com/AkshithaKola/ EE3900/blob/main/**filter**/code /7.11.c

compile it using

\$ gcc 7.11.c -lm

and execute it using

\$./a.out