1-D Rate Conversion

- Decimation
 - Reduce the sampling rate of a discrete-time signal.
 - Low sampling rate reduces storage and computation requirements.
- Interpolation
 - Increase the sampling rate of a discrete-time signal.
 - Higher sampling rate preserves fidelity.

1-D Periodic Subsampling

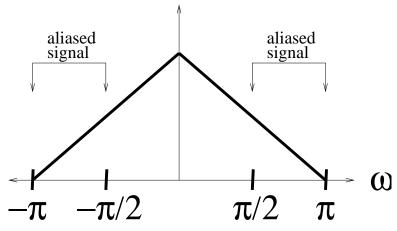
• Time domain subsampling of x(n) with period D

$$y(n) = x(Dn)$$
 $x(n) \longrightarrow y(n)$

• Frequency domain representation

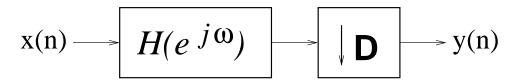
$$Y(e^{j\omega}) = \frac{1}{D} \sum_{k=0}^{D-1} X\left(e^{j(\omega - 2\pi k)/D}\right)$$

- Problem: Frequencies above π/D will alias.
- Example when D=2



• Solution: Remove frequencies above π/D .

Decimation System



- ullet Apply the filter $H(e^{j\omega})$ to remove high frequencies
 - For $|\omega| < \pi$

$$H(e^{j\omega}) = \operatorname{rect}\left(\frac{D\omega}{2\pi}\right)$$

– For all ω

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} \operatorname{rect}\left(D\frac{\omega - k2\pi}{2\pi}\right)$$

= $\operatorname{prect}_{2\pi/D}(\omega)$

- Impulse response

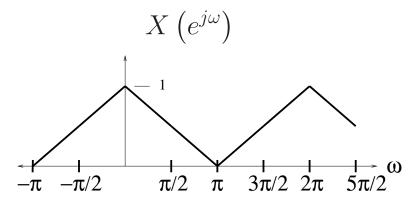
$$h(n) = \frac{1}{D}\mathrm{sinc}(n/D)$$

• Frequency domain representation

$$Y(e^{j\omega}) = \frac{1}{D} \sum_{k=0}^{D-1} H\left(e^{j(\omega - 2\pi k)/D}\right) X\left(e^{j(\omega - 2\pi k)/D}\right)$$

Graphical View of Decimation for D = 2

• Spectral content of signal



• Spectral content of filtered signal

$$H\left(e^{j\omega}\right)X\left(e^{j\omega}\right)$$

$$-\pi -\pi/2 \qquad \pi/2 \qquad \pi \qquad 3\pi/2 \quad 2\pi \qquad 5\pi/2$$

• Spectral content of decimated signal

$$Y(e^{j\omega}) = \frac{1}{D} \sum_{k=0}^{D-1} H\left(e^{j(\omega - 2\pi k)/D}\right) X\left(e^{j(\omega - 2\pi k)/D}\right)$$

$$-2\pi - \pi \qquad \pi \qquad 2\pi \quad 3\pi \qquad 4\pi \quad 5\pi$$

Decimation for Images

- Extension to decimation of images is direct
- Apply 2-D Filter

$$f(i,j) = h(i,j) * x(i,j)$$

• Subsample result

$$y(i,j) = f(Di, Dj)$$

• Ideal choice of filter is

$$h(m,n) = \frac{1}{D^2} \mathrm{sinc}(m/D) \mathrm{sinc}(n/D)$$

- Problems:
 - Filter has infinite extent.
 - Filter is not strictly positive.

Alternative Filters for Image Decimation

• Direct subsampling

$$h(m,n) = \delta(m,n)$$

- Advantages/Disadvantages:
 - * Low computation
 - * Excessive aliasing
- Block averaging

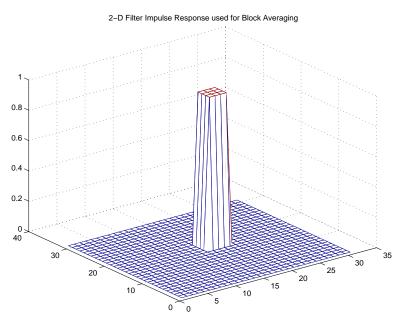
$$h(m,n) = \delta(m,n) + \delta(m+1,n) + \delta(m,n+1) + \delta(m+1,n+1)$$

- Advantages/Disadvantages:
 - * Low computation
 - * Some aliasing
- Sinc function

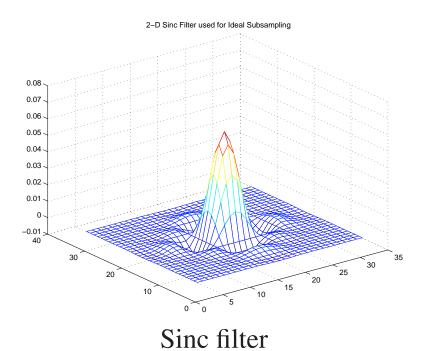
$$h(m,n) = \frac{1}{D^2} \operatorname{sinc}(m/D, n/D)$$

- Advantages/Disadvantages:
 - * Optimal, if signal is band limited...
 - * High computation

Decimation Filters



Block averaging filter



Original Image



• Full resolution

Image Decimation by 4 using Subsampling



• Severe aliasing

Image Decimation by 8 using Subsampling



• More severe aliasing

Image Decimation by 4 using Block Averaging



• Sharp, but with some aliasing

Image Decimation by 4 using Sinc Filter



• Theoretically optimal, but not necessarily the best visual quality

1-D Up-Sampling

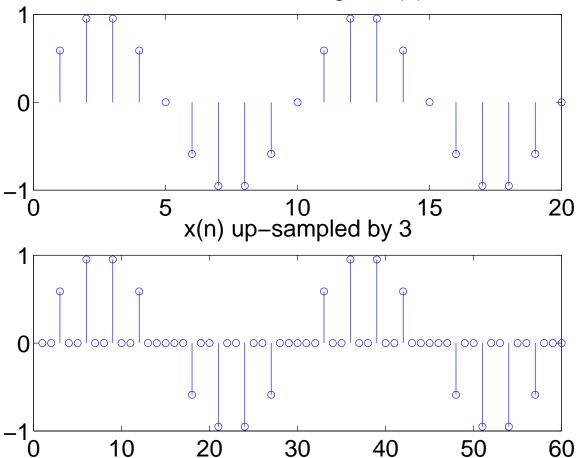
 \bullet Up-sampling by L

$$y(n) = \begin{cases} x(n/L) & \text{if } n = KL \text{ for some } K \\ 0 & \text{otherwise} \end{cases}$$

or the alternative form

$$y(n) = \sum_{m=-\infty}^{\infty} x(m)\delta(n - mL)$$





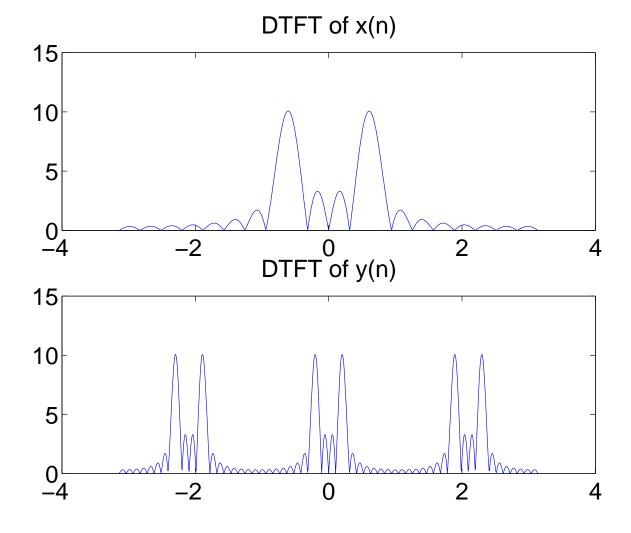
Up-Sampling in the Frequency Domain

 \bullet Up-sampling by L

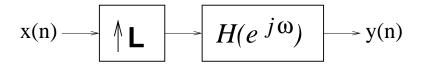
$$y(n) = \sum_{m=-\infty}^{\infty} x(m)\delta(n - mL)$$

• In the frequency domain

$$Y(e^{j\omega}) = X(e^{j\omega L})$$

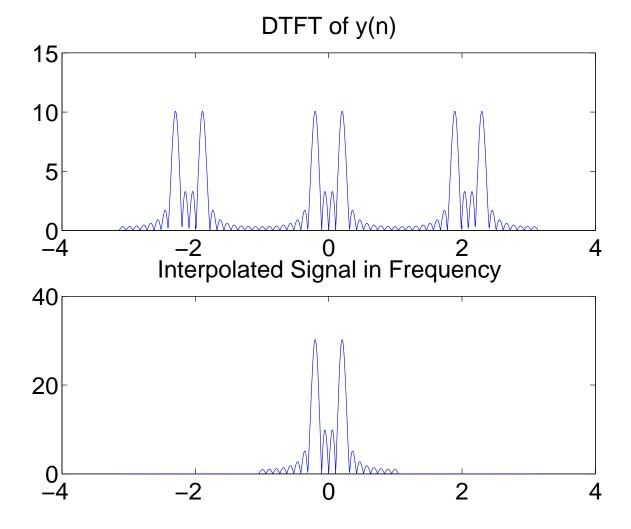


Interpolation in the Frequency Domain

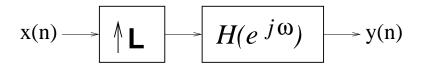


• Interpolating filter has the form

$$Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega L})$$



Interpolating Filter



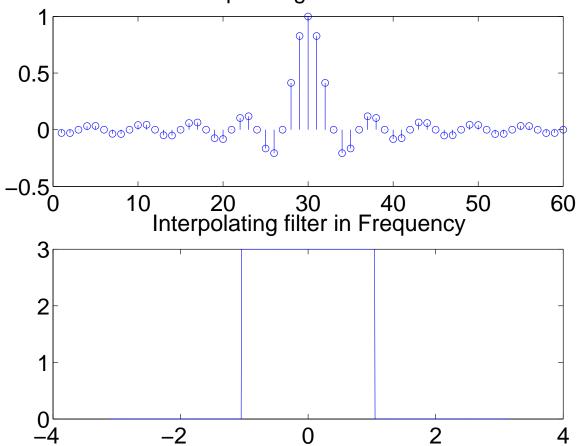
• In the frequency domain

$$H(e^{j\omega}) = L\mathrm{prect}_{2\pi/L}(\omega)$$

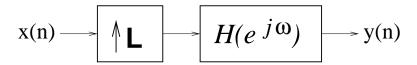
• In the time domain

$$h(n) = \operatorname{sinc}\left(\frac{n}{L}\right)$$

Interpolating filter in Time

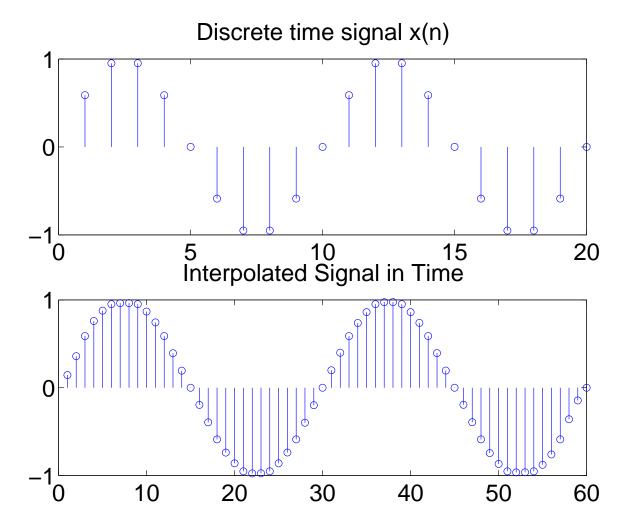


Interpolation in the Time Domain



• In the frequency domain

$$Y(e^{j\omega}) = X(e^{j\omega L})$$



2-D Interpolation

$$x(n) \longrightarrow \bigwedge(L,L) \longrightarrow H(e^{jv},e^{j\mu}) \longrightarrow y(n)$$

Up-sampling

$$y(m,n) = \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} x(k,l) \delta(m-kL,n-lL)$$

• In the frequency domain

$$H(e^{j\mu},e^{j\nu}) = \operatorname{prect}_{2\pi/L}(\mu)\operatorname{prect}_{2\pi/L}(\nu)$$

$$h(m,n) = \operatorname{sinc}\left(m/L,n/L\right)$$

- Problems: sinc function impulse response
 - Infinite support; infinite computation
 - Negative sidelobes; ringing artifacts at edges

Pixel Replication

$$x(n) \longrightarrow H(e^{jv}, e^{j\mu}) \longrightarrow y(n)$$

• Impulse response of filter

$$h(m,n) = \left\{ \begin{array}{l} 1 \ \ \text{for} \ 0 \leq m \leq L-1 \ \text{and} \ 0 \leq n \leq L-1 \\ 0 \ \ \text{otherwise} \end{array} \right.$$

ullet Replicates each pixel L^2 times

Bilinear Interpolation

$$x(n) \longrightarrow H(e^{jv}, e^{j\mu}) \longrightarrow y(n)$$

• Impulse response of filter

$$h(m,n) = \Lambda(m/L) \Lambda(n/L)$$

• Results in linear interpolation of intermediate pixels