

Exercise 1.a

Consider the system

$$\begin{aligned}\dot{x}_1(t) &= -x_1(t) + x_2(t) \\ \dot{x}_2(t) &= -x_1^3(t) - x_2^3(t)\end{aligned}$$

and the Lyapunov function $V(x) = bx_1^4 + ax_2^2(t)$. Select proper values of a, b such that $V(x)$ can be used to prove the stability of the origin.

$$Vdot=-4bx_1^4+4bx_1^3x_2-2ax_1^3x_2-2ax_2^4$$

- a. $a=1, b=1$ ($Vdot=-4x_1^4+2x_1^3x_2-2x_2^4$)
- b. $a=2, b=1$ ($Vdot=-4x_1^2-4x_2^4<0$)**
- c. $a=0, b=1$ (V is not a Lyapunov function)

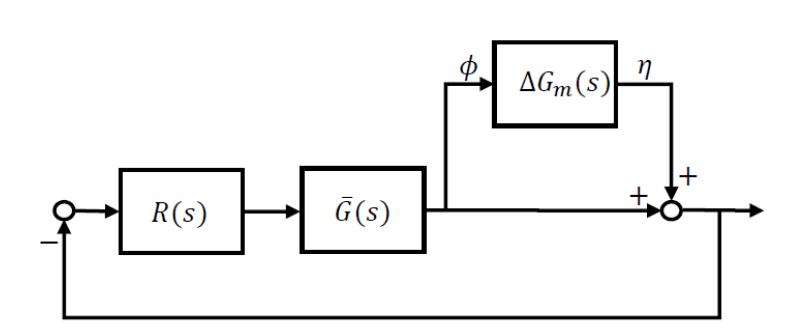
Exercise 1.b

In view of the previous answer, the origin is (**select the strongest property**)

- a. stable
- b. locally asymptotically stable
- c. globally asymptotically stable**

Exercise 2

Given the following feedback system, assume that in the nominal case ($\Delta G_m(s)=0$) the system is asymptotically stable and that $\Delta G_m(s)$ is asymptotically stable. Select the sufficient condition guaranteeing that in the perturbed case the asymptotic stability of the closed-loop system is maintained.



- a. $|T(jw)\Delta(jw)| < 1$, for any $w \geq 0$
- b. $|S(jw)\Delta G_m(jw)| < 1$, for any $w \geq 0$
- c. $|S(jw)R(jw)\Delta G_m(jw)| < 1$, for any $w \geq 0$

Exercise 3.a

Consider the system with the following transfer function $G(s)$

$$\begin{bmatrix} \frac{a}{(s+1)} & \frac{b}{(s+1)(s+2)} \\ \frac{c}{(s+2)} & \frac{d}{(s+2)} \end{bmatrix}$$

The set of poles (with the proper multiplicity) is

- a. -1,-1,-2,-2,-2
- b. -1,-2,-2**
- c. -1, -2

$$\det G(s) = \frac{ad(s+2)-bc}{(s+1)(s+2)^2}$$

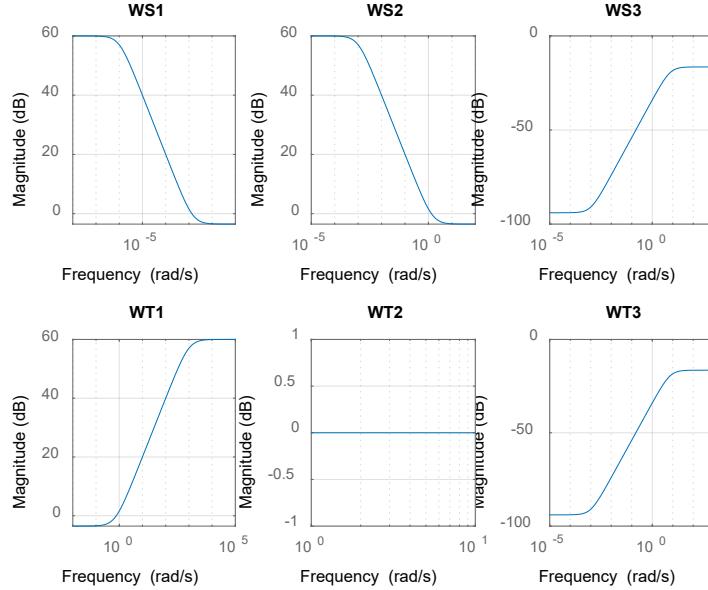
Exercise 3.b

The system must not have zeros at the origin. The zero is $s=(2ad-bc)/ad$. Therefore it must be $2ad$ not equal to bc

- a. $a=b=c=d=1$**
- b. $a=b=d=1, c=2$
- c. for any value of a,b,c,d

Exercise 4

Consider the design of a H_2/H_{∞} controller with shaping functions for a SISO system. Which one of the following pairs (W_{Si}, W_{Tj}) of the functions shown in the figure is reasonable and coherent with the goals of the control design procedure?



- a. WS3-WT2 WT2 does not make sense
- b. WS2- WT1 acceptable**
- c. WS1 -WT3 WT3 is not acceptable, small gain at high frequency

Exercise 5

Consider the system

$$\begin{aligned}\dot{x}_1(t) &= bu(t) \\ \dot{x}_2(t) &= x_1(t) + u(t)\end{aligned}$$

and assume that you want to design an infinite horizon LQ control with $Q=\text{diag}(q_1, q_2)$, $R=1$.

- A. Compute the necessary conditions guaranteeing that the solution of the infinite horizon LQ control is stabilizing.
 - a. any b , $q_1>0$, $q_2>0$ no, lack of reachability if $b=0$
 - b. $b\neq 0$, $q_1>0$, $q_2\geq 0$ no lack of observability if $q_2=0$
 - c. $b\neq 0$, any $q_1\geq 0$, $q_2>0$ yes**

B. With $Q=I$, $b=1$, the solution of the steady-state Riccati equation is $P=I$. What are the closed-loop eigenvalues? Recall the steady state Riccati equation of LQinf control: $A'P + PA + Q - PBR^{-1}B'P = 0$

- a. -1,-2
- b. -2,-2
- c. -1,-1

C. Assume now to implement the feedback control law $u(t) = -\rho Kx(t)$ (K is again the solution of the LQinf problem), specify the set of values of ρ guaranteed by LQ control so that the closed-loop system remains asymptotically stable.

- a. ρ in $(0, \infty)$
- b. ρ in $(0.5, \infty)$
- c. ρ in $(-0.5, 0.5)$

A.

$$A = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} b \\ 1 \end{bmatrix} \rightarrow \text{reachability matrix } M_r = [B \ AB] = \begin{bmatrix} b & 0 \\ 1 & b \end{bmatrix} \rightarrow \text{condition } b \neq 0$$

$$Q = \begin{bmatrix} q_1 & 0 \\ 0 & q_2 \end{bmatrix} \rightarrow Q^{1/2} = \begin{bmatrix} \sqrt{q_1} & 0 \\ 0 & \sqrt{q_2} \end{bmatrix} \rightarrow \text{observability matrix } M_o = \begin{bmatrix} Q^{1/2} \\ Q^{1/2}A \end{bmatrix} = \begin{bmatrix} \sqrt{q_1} & 0 \\ 0 & \sqrt{q_2} \\ 0 & 0 \\ \sqrt{q_2} & 0 \end{bmatrix} \rightarrow q_2 > 0$$

B.

$$K = R^{-1}B'P = \begin{bmatrix} 1 & 1 \end{bmatrix} \rightarrow A - BK = \begin{bmatrix} -1 & -1 \\ 0 & -1 \end{bmatrix} \rightarrow \text{eigenvalues } -1, -1$$

C.

The loop transfer function is $K(sI-A)^{-1}B = (2s+1)/s^2$. In order to compute the gain margin, consider the characteristic equation $s^2 + 2ps + p = 0$ with stable roots for any $p > 0$. The gain margin is $(0, \infty)$.

Exercise 6

EKF: see the notes and the slides

Exercise 7

MPC: see the notes and the slides