

EE4002 Control Engineering

B) State-Space Control

Chapter 1. The State Space Modelling Approach

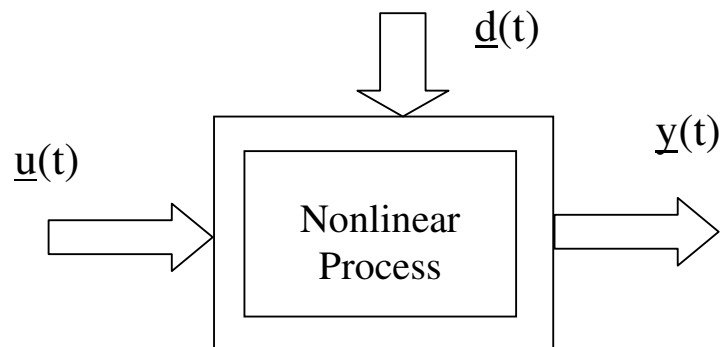
1.1 The State Space Model

Transformation of an N^{th} order multi-input-multi-output system to N first order differential equations:

Benefits

- Robust method of simulating high order differential equations
- Easy analysis of dynamics
- Allows for use of model reduction methods
- Can easily apply advanced control
- Used for estimation

Consider the N^{th} order nonlinear dynamical process:

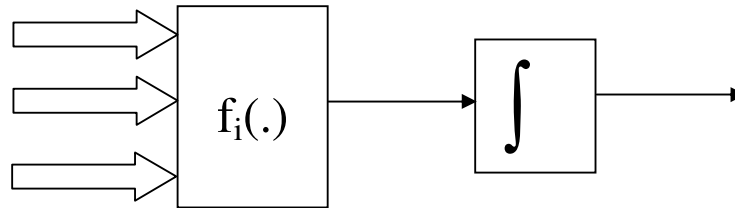


In general this process could be represented by a model consisting of :

It is possible to transform this set of coupled nonlinear differential equations to yield a set of N first order differential equations with states, $\{x_1(t), x_2(t), \dots, x_N(t)\}$.

The differential equation describing the dynamics of the i^{th} state $x_i(t)$ can be written as:

Which could be represented by the following simulation diagram:



The N first order differential equations could be written as:

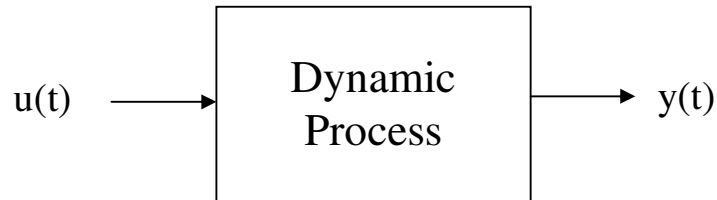
$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_N(t) \end{bmatrix} = \dot{\underline{x}}(t) = \begin{bmatrix} f_1(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \\ f_2(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \\ \vdots \\ f_N(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \end{bmatrix}$$

The output $\underline{y}(t)$ is then generated by:

$$\underline{y}(t) = \begin{bmatrix} y_1(t) \\ y_2(t) \\ \vdots \\ y_p(t) \end{bmatrix} = \begin{bmatrix} h_1(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \\ h_2(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \\ \vdots \\ h_p(\underline{x}(t), \underline{u}(t), \underline{d}(t)) \end{bmatrix}$$

1.1.1 Some Example State-Space Systems:

i) Third Order Linear Process



Modelled by the following differential equation:

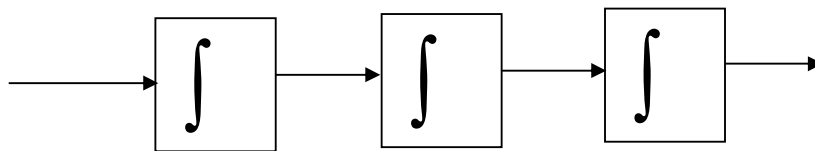
$$\frac{d^3}{dt^3} y(t) + 5 \frac{d^2}{dt^2} y(t) + 3 \frac{d}{dt} y(t) + y(t) = u(t)$$

Which could of course be represented by the transfer function model:



Rearrange to yield an expression for the highest derivative:

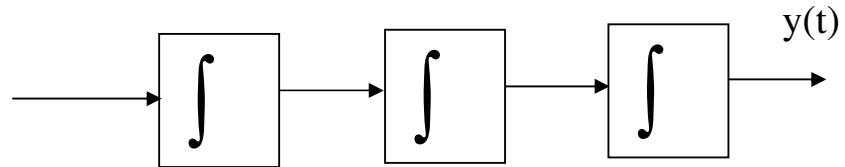
Form lower derivatives then by integration:



But we know that:

$$\ddot{y}(t) = u(t) - 5\dot{y}(t) - 3\dot{y}(t) - y(t)$$

This yields the following simulation diagram:



Assign the state variables as the outputs of integrators:

We can now specify the state vector for the process as:

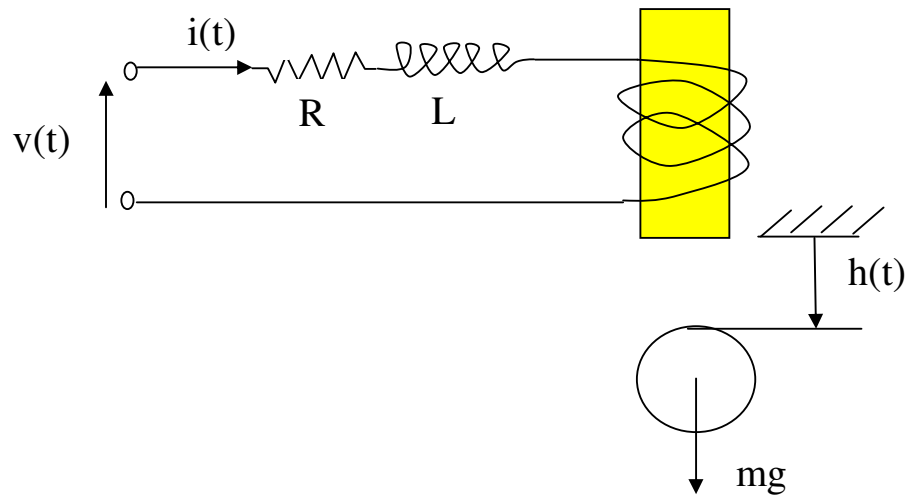
Now this could be written in matrix form as:

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix} = \begin{bmatrix} & & \\ & & \\ & & \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix} + \begin{bmatrix} \\ \\ \end{bmatrix} u(t)$$

$$y(t) = \begin{bmatrix} & & \\ & & \\ & & \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \end{bmatrix}$$

ii) Third Order Nonlinear Process

Consider the following magnetic suspension system:



Can be modelled by:

$$m \frac{d^2 h(t)}{dt^2} = mg - \frac{Ki^2(t)}{h^2(t)}$$

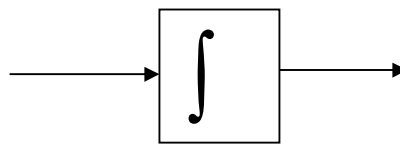
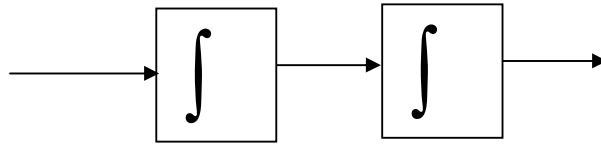
$$L \frac{di(t)}{dt} = v(t) - Ri(t)$$

Rewritten as:

$$\frac{d^2 h(t)}{dt^2} = g - \frac{Ki^2(t)}{mh^2(t)}$$

$$\frac{di(t)}{dt} = \frac{1}{L} (v(t) - Ri(t))$$

Could be represented by the following simulation diagram:



Note that three integrators are required – system is 3rd order

Now can arbitrarily assign the three states $\{x_1(t), x_2(t), x_3(t)\}$

And define the state vector:

The state-equations are now:

$$\dot{x}_1 = x_2(t)$$

$$\dot{x}_2 = g - \frac{Kx_3^2(t)}{mx_1^2(t)}$$

$$\dot{x}_3 = \frac{1}{L}(u(t) - Rx_3(t))$$

with output equation:

$$y(t) = x_1(t)$$

1.2 Derivation of the Linear State Space Model - by Linearisation

Consider first the linearisation of a multivariate function:

$$z = g(\underline{w}) \quad \text{where} \quad \underline{w} = \begin{bmatrix} w_1 \\ \vdots \\ w_N \end{bmatrix}$$

about the operating point vector:

$$\underline{w}_0 = \begin{bmatrix} w'_1 \\ \vdots \\ w'_N \end{bmatrix}$$

Use the first order Taylor's series approximation :

$$z \approx g(\underline{w}_0) + \left. \frac{\partial g}{\partial w_1} \right|_{\underline{w}=\underline{w}_0} (w_1 - w'_1) + \left. \frac{\partial g}{\partial w_2} \right|_{\underline{w}=\underline{w}_0} (w_2 - w'_2) + \dots + \left. \frac{\partial g}{\partial w_N} \right|_{\underline{w}=\underline{w}_0} (w_N - w'_N)$$

which could be written as:

$$(z - z_0) = \begin{bmatrix} \left. \frac{\partial g}{\partial w_1} \right|_{\underline{w}_0} & \left. \frac{\partial g}{\partial w_2} \right|_{\underline{w}_0} & \dots & \left. \frac{\partial g}{\partial w_N} \right|_{\underline{w}_0} \end{bmatrix} (\underline{w} - \underline{w}_0)$$

Now for simplicity first consider a linearisation of the i^{th} state equation:

$$\frac{dx_i(t)}{dt} = f_i(\underline{x}(t))$$

Could be linearised to yield:

$$(z - z_0) = \begin{bmatrix} \left. \frac{\partial f_i}{\partial x_1} \right|_{\underline{x}_0} & \left. \frac{\partial f_i}{\partial x_2} \right|_{\underline{x}_0} & \cdots & \left. \frac{\partial f_i}{\partial x_N} \right|_{\underline{x}_0} \end{bmatrix} (\underline{x} - \underline{x}_0)$$

But:

Hence the linearised equation becomes:

$$\frac{d}{dt} \Delta x_i(t) = \begin{bmatrix} \left. \frac{\partial f_i}{\partial x_1} \right|_{\underline{x}_0} & \left. \frac{\partial f_i}{\partial x_2} \right|_{\underline{x}_0} & \cdots & \left. \frac{\partial f_i}{\partial x_N} \right|_{\underline{x}_0} \end{bmatrix} \Delta \underline{x}(t)$$

Now let us expand to include input and disturbance vector:

$$\frac{dx_i(t)}{dt} = f_i(\underline{x}(t), \underline{u}(t), \underline{d}(t))$$

Which will have the linearisation about the operating point:

$$\begin{aligned} \frac{d}{dt} \Delta x_i(t) = & \begin{bmatrix} \left. \frac{\partial f_i}{\partial x_1} \right|_{op} & \left. \frac{\partial f_i}{\partial x_2} \right|_{op} & \cdots & \left. \frac{\partial f_i}{\partial x_N} \right|_{op} \end{bmatrix} \Delta \underline{x}(t) + \begin{bmatrix} \left. \frac{\partial f_i}{\partial u_1} \right|_{op} & \left. \frac{\partial f_i}{\partial u_2} \right|_{op} & \cdots & \left. \frac{\partial f_i}{\partial u_m} \right|_{op} \end{bmatrix} \Delta \underline{u}(t) \\ & + \begin{bmatrix} \left. \frac{\partial f_i}{\partial d_1} \right|_{op} & \left. \frac{\partial f_i}{\partial d_2} \right|_{op} & \cdots & \left. \frac{\partial f_i}{\partial d_s} \right|_{op} \end{bmatrix} \Delta \underline{d}(t) \end{aligned}$$

which could further be written as:

$$\frac{d}{dt} \Delta x_i(t) = [a_{i1} \quad a_{i2} \quad \cdots \quad a_{iN}] \Delta \underline{x}(t) + [b_{i1} \quad b_{i2} \quad \cdots \quad b_{im}] \Delta \underline{u}(t) + [e_{i1} \quad e_{i2} \quad \cdots \quad e_{is}] \Delta \underline{d}(t)$$

This could of course be repeated for all N state equations:

$$\frac{d}{dt} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \vdots \\ \Delta x_i \\ \vdots \\ \Delta x_N \end{bmatrix} = \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1N} \\ a_{21} & a_{22} & \cdots & a_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ a_{i1} & a_{i2} & \cdots & a_{iN} \\ \vdots & \vdots & \ddots & \vdots \\ a_{N1} & a_{N2} & \cdots & a_{NN} \end{bmatrix} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \vdots \\ \Delta x_i \\ \vdots \\ \Delta x_N \end{bmatrix} + \begin{bmatrix} b_{11} & b_{12} & \cdots & b_{1m} \\ b_{21} & b_{22} & \cdots & b_{2m} \\ \vdots & \vdots & \ddots & \vdots \\ b_{i1} & b_{i2} & \cdots & b_{im} \\ \vdots & \vdots & \ddots & \vdots \\ b_{N1} & b_{N2} & \cdots & b_{Nm} \end{bmatrix} \begin{bmatrix} \Delta u_1 \\ \Delta u_2 \\ \vdots \\ \Delta u_m \end{bmatrix} + \begin{bmatrix} e_{11} & e_{12} & \cdots & e_{1s} \\ e_{21} & e_{22} & \cdots & e_{2s} \\ \vdots & \vdots & \ddots & \vdots \\ e_{i1} & e_{i2} & \cdots & e_{is} \\ \vdots & \vdots & \ddots & \vdots \\ e_{N1} & e_{N2} & \cdots & e_{Ns} \end{bmatrix} \begin{bmatrix} \Delta d_1 \\ \Delta d_2 \\ \vdots \\ \Delta d_s \end{bmatrix}$$

Or in more compact form as:

Now consider the static output equation for the i^{th} output:

$$y_i(t) = h_i(\underline{x}(t), \underline{u}(t), \underline{d}(t))$$

The linearisation about the operating point $\{\underline{x}_0, \underline{u}_0, \underline{d}_0\}$ is:

$$\Delta y_i(t) = \begin{bmatrix} \left. \frac{\partial h_i}{\partial x_1} \right|_{op} & \left. \frac{\partial h_i}{\partial x_2} \right|_{op} & \cdots & \left. \frac{\partial h_i}{\partial x_N} \right|_{op} \end{bmatrix} \Delta \underline{x}(t) + \begin{bmatrix} \left. \frac{\partial h_i}{\partial u_1} \right|_{op} & \left. \frac{\partial h_i}{\partial u_2} \right|_{op} & \cdots & \left. \frac{\partial h_i}{\partial u_m} \right|_{op} \end{bmatrix} \Delta \underline{u}(t) + \begin{bmatrix} \left. \frac{\partial h_i}{\partial d_1} \right|_{op} & \left. \frac{\partial h_i}{\partial d_2} \right|_{op} & \cdots & \left. \frac{\partial h_i}{\partial d_s} \right|_{op} \end{bmatrix} \Delta \underline{d}(t)$$

which could be written as:

$$\Delta y_i(t) = [c_{i1} \quad c_{i2} \quad \cdots \quad c_{iN}] \Delta \underline{x}(t) + [d_{i1} \quad d_{i2} \quad \cdots \quad d_{im}] \Delta \underline{u}(t) + [f_{i1} \quad f_{i2} \quad \cdots \quad f_{is}] \Delta \underline{d}(t)$$

This could be repeated for all the P outputs:

$$\begin{bmatrix} \Delta y_1 \\ \Delta y_2 \\ \vdots \\ \Delta y_i \\ \vdots \\ \Delta y_P \end{bmatrix} = \begin{bmatrix} c_{11} & c_{12} & \cdots & c_{1N} \\ c_{21} & c_{22} & \cdots & c_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ c_{i1} & c_{i2} & \cdots & c_{iN} \\ \vdots & \vdots & \ddots & \vdots \\ c_{P1} & c_{P2} & \cdots & c_{PN} \end{bmatrix} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \vdots \\ \Delta x_i \\ \vdots \\ \Delta x_N \end{bmatrix} + \begin{bmatrix} d_{11} & d_{12} & \cdots & d_{1m} \\ d_{21} & d_{22} & \cdots & d_{2m} \\ \vdots & \vdots & \ddots & \vdots \\ d_{i1} & d_{i2} & \cdots & d_{im} \\ \vdots & \vdots & \ddots & \vdots \\ d_{P1} & d_{P2} & \cdots & d_{Pm} \end{bmatrix} \begin{bmatrix} \Delta u_1 \\ \Delta u_2 \\ \vdots \\ \Delta u_m \end{bmatrix} + \begin{bmatrix} f_{11} & f_{12} & \cdots & f_{1s} \\ f_{21} & f_{22} & \cdots & f_{2s} \\ \vdots & \vdots & \ddots & \vdots \\ f_{i1} & f_{i2} & \cdots & f_{is} \\ \vdots & \vdots & \ddots & \vdots \\ f_{P1} & f_{P2} & \cdots & f_{Ps} \end{bmatrix} \begin{bmatrix} \Delta d_1 \\ \Delta d_2 \\ \vdots \\ \Delta d_s \end{bmatrix}$$

which again could be written in more compact form as:

This yields the following linear state-space model for, a process about a particular operating point:

$$\begin{aligned} \frac{d}{dt} \underline{x}(t) &= A \underline{x}(t) + B \underline{u}(t) + E \underline{d}(t) \\ \underline{y}(t) &= C \underline{x}(t) + D \underline{u}(t) + F \underline{d}(t) \end{aligned}$$

1.2.1 Linearisation Examples:

i) The magnetic suspension system

$$m \frac{d^2 h(t)}{dt^2} = mg - \frac{Ki^2(t)}{h^2(t)}$$

$$L \frac{di(t)}{dt} = v(t) - Ri(t)$$

Find a linear model about the desired operating airgap $h=0.01\text{m}$

The process parameters are:

$$L=10\text{mH} \quad M=0.05 \text{ Kg} \quad g=10\text{ms}^{-2} \quad R=1 \text{ ohm} \quad K=2 \times 10^{-4} \text{Nm}^2/\text{A}^2$$

First find the operating point,

From the force equation:

$$m \frac{d^2 h(t)}{dt^2} = mg - \frac{Ki^2(t)}{h^2(t)} =$$

From the electrical equation:

$$L \frac{di(t)}{dt} = v(t) - Ri(t) =$$

The operating point vector will now be defined as:

$$\underline{x}_0 = \begin{bmatrix} h_0 \\ \dot{h}_0 \\ i_0 \end{bmatrix} = \begin{bmatrix} \quad \\ \quad \\ \quad \end{bmatrix}$$

Now linearise about this operating point vector:

The state equations can be written as:

$$\begin{aligned}\dot{x}_1 &= x_2(t) \\ \dot{x}_2 &= g - \frac{Kx_3^2(t)}{mx_1^2(t)} \\ \dot{x}_3 &= \frac{1}{L}(u(t) - Rx_3(t))\end{aligned}$$

Now define deviations of the states from their operating point values as:

Hence the linearised model is:

$$\begin{aligned}\frac{d}{dt}\Delta x_1(t) &= \left. \frac{\partial f_1}{\partial x_2} \right|_{OP} \Delta x_2(t) \\ \frac{d}{dt}\Delta x_2(t) &= \left. \frac{\partial f_2}{\partial x_1} \right|_{OP} \Delta x_1(t) + \left. \frac{\partial f_2}{\partial x_3} \right|_{OP} \Delta x_3(t) \\ \frac{d}{dt}\Delta x_3(t) &= \left. \frac{\partial f_3}{\partial x_3} \right|_{OP} \Delta x_3(t) + \left. \frac{\partial f_3}{\partial u} \right|_{OP} \Delta u(t)\end{aligned}$$

But we know that:

$$f_1(x_2) = x_2$$

$$f_2(x_1, x_3) = g - \frac{Kx_3^2}{Mx_1^2}$$

$$f_3(x_3, u) = \frac{u}{L} - \frac{R}{L}x_3$$

This yields the following linearised model about the operating point:

$$\begin{aligned}\frac{d}{dt}\Delta x_1(t) &= \Delta x_2(t) \\ \frac{d}{dt}\Delta x_2(t) &= \\ \frac{d}{dt}\Delta x_3(t) &= -100\Delta x_3(t) + 100\Delta u(t)\end{aligned}$$

And of course:

$$\Delta y(t) = h(t) - h_0 = \Delta x_1(t)$$

This then could be written in matrix form as:

$$\begin{aligned}\frac{d}{dt}\begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \Delta x_3 \end{bmatrix} &= \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & -100 \end{bmatrix} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \Delta x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 100 \end{bmatrix} \Delta u(t) \\ \Delta y(t) &= \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} \Delta x_1 \\ \Delta x_2 \\ \Delta x_3 \end{bmatrix}\end{aligned}$$

ii) Permanent magnet DC motor

Could be modelled by the following coupled equations:

$$\begin{aligned}\frac{di}{dt} &= \frac{1}{L}(v(t) - Ri(t) - K_m\omega(t)) \\ \frac{d\omega}{dt} &= \frac{1}{J}(K_M i(t) - B\omega(t) - T_L(t))\end{aligned}$$

As you can see these are linear differential equations:

If a tachometer of gain K_T V/rads⁻¹ is used to measure the speed, then the output equation could be written as:

Consider now that the motor is driving a nonlinear fan load:

$$T_L(t) = K_F \omega^2(t)$$

The process then would be modelled as:

$$\frac{di}{dt} = \frac{1}{L} (v(t) - Ri(t) - K_m \omega(t))$$

$$\frac{d\omega}{dt} = \frac{1}{J} (K_M i(t) - B\omega(t) - K_F \omega^2(t))$$

Generate a linear state-space model which describes the dynamics of this process close to the operating speed:

First find the operating point:

$$\frac{d\omega}{dt} = \frac{1}{J} (K_M i(t) - B\omega(t) - K_F \omega^2(t)) =$$

Now from the electrical equation:

$$\frac{di}{dt} = \frac{1}{L}(v(t) - Ri(t) - K_m \omega(t)) =$$

Assign the states: $\underline{x}(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} i(t) \\ \omega(t) \end{bmatrix}$

and input and output: $u(t) = v(t)$ and $y(t) = v_T(t) = K_T \omega(t)$

hence:

$$\dot{x}_1(t) = \frac{1}{L}(u(t) - Rx_1(t) - K_m x_2(t)) = f_1(x_1, x_2, u)$$

$$\dot{x}_2(t) = \frac{1}{J}(K_M x_1(t) - Bx_2(t) - K_F x_2^2(t)) = f_2(x_1, x_2)$$

$$y(t) = K_T x_2(t) = h(x_2)$$

The linearised model about the operating point is:

$$\begin{aligned} \frac{d}{dt} \Delta x_1(t) &= \left. \frac{\partial f_1}{\partial x_1} \right|_{OP} \Delta x_1(t) + \left. \frac{\partial f_1}{\partial x_2} \right|_{OP} \Delta x_2(t) + \left. \frac{\partial f_1}{\partial u} \right|_{OP} \Delta u(t) \\ \frac{d}{dt} \Delta x_2(t) &= \left. \frac{\partial f_2}{\partial x_1} \right|_{OP} \Delta x_1(t) + \left. \frac{\partial f_2}{\partial x_2} \right|_{OP} \Delta x_2(t) \\ \Delta y(t) &= \left. \frac{\partial h}{\partial x_2} \right|_{OP} \Delta x_2(t) \end{aligned}$$

Obviously the electrical equation is linear:

$$\dot{x}_1(t) = \frac{1}{L}(u(t) - Rx_1(t) - K_m x_2(t)) = f_1(x_1, x_2, u)$$

Now concentrating on the mechanical equation:

$$\dot{x}_2(t) = \frac{1}{J} (K_M x_1(t) - Bx_2(t) - K_F x_2^2(t)) = f_2(x_1, x_2)$$

And simply for the output equation:

$$y(t) = K_T x_2(t) = h(x_2)$$

This yields the following linear state-space model:

$$\begin{aligned} \frac{d}{dt} \Delta \underline{x} &= \begin{bmatrix} & \end{bmatrix} \Delta \underline{x}(t) + \begin{bmatrix} \end{bmatrix} \Delta u(t) \\ \Delta y(t) &= \begin{bmatrix} \end{bmatrix} \Delta \underline{x}(t) \end{aligned}$$