Chapter 9

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Introduction to AC Machines



Introduction

- r Primary AC motor drives
 - x Induction motors (asynchronous)
 - x Squirrel cage brushless
 - × Wound rotor brushed
 - x Synchronous Motors
 - x Permanent Magnet brushless
 - x Wound rotor brushed

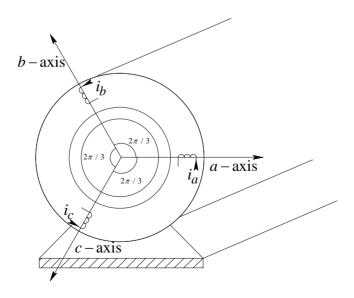
x These machines have similar stators but different rotor constructions.



Introduction

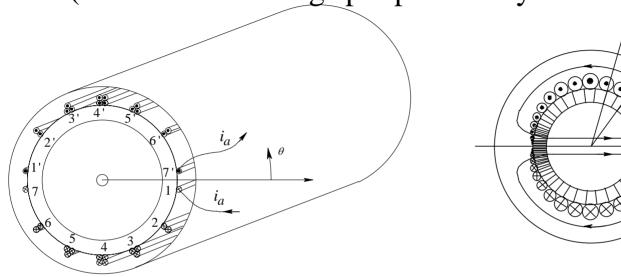
Stator windings produce a sinusoidal field distribution in the airgap.

These magnetic field distributions are displaced by 120 ° w.r.t. each other.



Sinusoidally-distributed Stator Windings

(number of windings per phase vary with angle)



Conductor density

Total

$$N_{s} = \int_{0}^{\pi} n_{s}(\theta) d\theta = \int_{0}^{\pi} \hat{n}_{s} \sin(\theta) d\theta = 2\hat{n}_{s}$$

$$\Rightarrow n_{s}(\theta) = \frac{N_{sp}}{2} \sin(\theta) \qquad 0 < \theta < \pi$$

 $N_{\rm sp}$ is the number of conductors/phase/pole.

I.M. – typically 4 pole; P.M. - multipole

 $d\theta$

 $\theta = 0$ magnetic axis of phase a





Air-gap Field Distribution

Apply Ampere's Law.

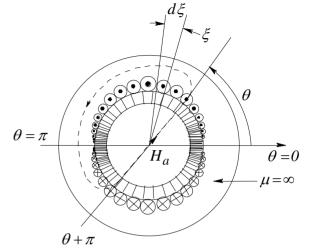
RHR gives direction. From symmetry

$$H_a(\theta) = -H_a(\theta + \pi)$$

(negative sign because line of integration points inwards at $\theta + \pi$)

$$\sum H.dl = H_m l_m + 2H_a(\theta) l_g = \int_0^{\pi} n_s(\theta + \xi) i_a d\xi$$

$$2H_a(\theta)\ell_g = \frac{N_s}{2}i_a \int_0^{\pi} \sin(\theta + \xi) d\xi = N_s i_a \cos(\theta)$$









 $\theta = 0$

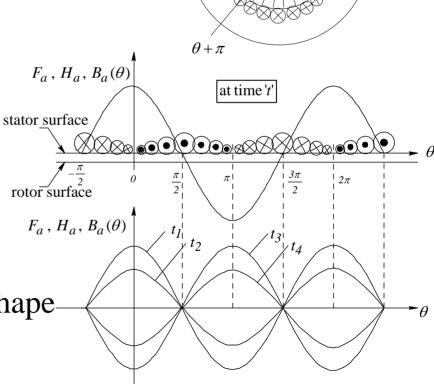
Air-gap Field Distribution

Radial magnetic field strength

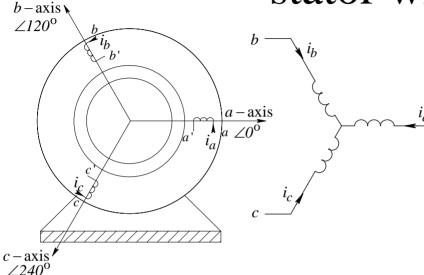
Radial magnetic flux density

Radial magnetomotive force

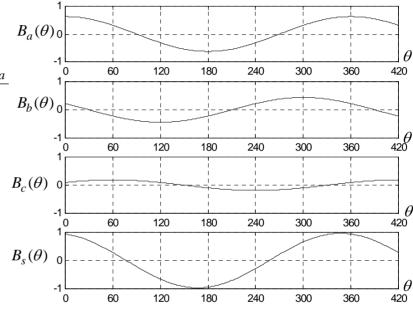
Field quantities have different magnitudes and units but same shape



Three-phase sinusoidally-distributed stator windings



r Example 9-3: 2-pole, $l_{\rm g}=1$ mm, $i_{\rm a}=$ 10 A, $i_b = -7$ A, $i_c = -3$ A, $N_{sp} = 50$.



$$B_a(\theta) = \frac{\mu_o N_s i_a}{2\ell_g} \cos \theta = 0.628 \cos \theta \ Wb/m^2$$

$$B_b(\theta) = -0.440 \times \cos(\theta - 120^0) \text{ Wb/m}^2$$
 $B_c(\theta) = -0.188 \times \cos(\theta - 240^0) \text{ Wb/m}^2$

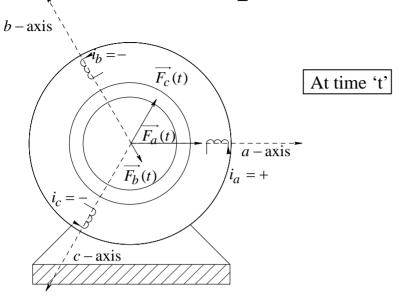
$$B_c(\theta) = -0.188 \times \cos(\theta - 240^0) \ Wb/m^2$$

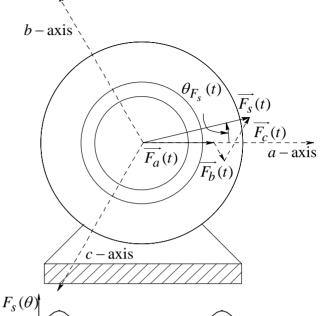
= combined stator-produced flux density EE4001, UCC





Space Vector to Represent Sinusoidal Distributions





r Complex number representation

$$F_a(\theta, t) = \frac{N_s}{2} i_a(t) \cos(\theta) \Leftrightarrow \overrightarrow{F_a}(t) = \frac{N_s}{2} i_a(t) \angle \theta^{o}$$

Similarly,
$$\overrightarrow{F_b}(t) = \frac{N_s}{2} i_b(t) \angle 120^{\circ}$$
; $\overrightarrow{F_c}(t) = \frac{N_s}{2} i_c(t) \angle 240^{\circ}$

And

= resultant stator space vector

Audio

for magnetomotive force

r Similar expressions for B and H



stator surface

 $\frac{5\pi}{2}$ rotor surface

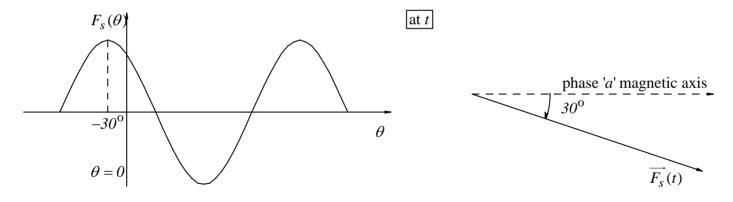


Example

Three-phase, sinusoidally-distributed stator with $\frac{N_s}{2}$ =50 turns At time t, $i_a = 10A$, $i_b = -10A$ and $i_c = 0A$

Find \vec{F}_s

$$\overrightarrow{F}_{s}(t) = \frac{N_{s}}{2} \left(i_{a} \angle 0^{0} + i_{b} \angle 120^{0} + i_{c} \angle 240^{0} \right)
= 50 \left\{ 10 + (-10) \left[\cos 120^{0} + j \sin 120^{0} \right] + (0) \left[\cos 240^{0} + j \sin 240^{0} \right] \right\}
\overrightarrow{F}_{s}(t) = 50 \times 17.32 \angle -30^{0} = 866 \angle -30^{0} \text{ A} \cdot \text{turns}$$



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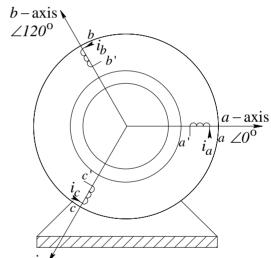


Space Vectors Representation of Combined Phase Currents and Voltages

r Mathematical concept

At time t

= stator current space vector



$$\vec{v}_{s}(t) = v_{a}(t) \angle 0^{0} + v_{b}(t) \angle 120^{0} + v_{c}(t) \angle 240^{0} \angle 240^{0}$$

$$= \hat{V}_{s}(t) \angle \theta_{v_{s}}(t)$$

$$= stator\ voltage\ space\ vector$$





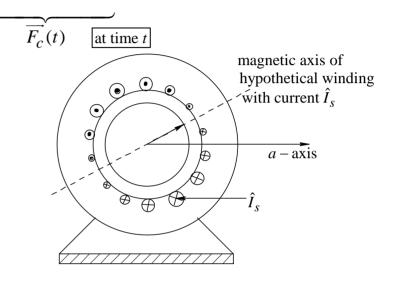
Physical interpretation of $i_s(t)$

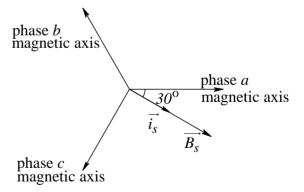
$$\frac{\overrightarrow{N_s}}{2}\overrightarrow{i_s}(t) = \underbrace{\frac{N_s}{2}}{i_a(t)} \underbrace{i_a(t)} \angle 0^0 + \underbrace{\frac{N_s}{2}}{i_b(t)} \underbrace{i_b(t)} \angle 120^0 + \underbrace{\frac{N_s}{2}}{i_c(t)} \underbrace{i_c(t)} \angle 240^0 = \overrightarrow{F_s}(t)$$

$$\overrightarrow{i_s}(t) = \frac{\overrightarrow{F_s}(t)}{N_s/2} \implies \widehat{I_s}(t) = \frac{\widehat{F_s}(t)}{N_s/2}$$
and $\theta_{i_s}(t) = \theta_{F_s}(t)$

 $\vec{F_s}(t)$ and $\vec{i_s}(t)$ are collinear

r Magnetic field is produced by combined effect of i_a , i_b and i_c but could equivalently be produced by hypothetical winding current $\vec{i_s}(t)$ at θ_{i_s} r helps in obtaining expression for torque





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Space Vector Components:

Finding Phase Currents from Current Space Vector

EE4001. UCC

$$\operatorname{Re}\left[\vec{i}_{s}\angle 0^{0}\right] = i_{a} + \operatorname{Re}\left[i_{b}\angle 120^{0}\right] + \operatorname{Re}\left[i_{b}\angle 240^{0}\right] = \frac{3}{2}i_{a}$$

$$\Rightarrow i_{a}(t) = \frac{2}{3}\operatorname{Re}(\vec{i}_{s}\angle 0^{\circ}) = \frac{2}{3}\hat{I}_{s}\cos\theta_{i_{s}}$$

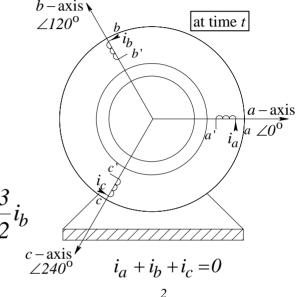
$$\operatorname{Re}\left[\vec{i}_{s}\angle -120^{0}\right] = \operatorname{Re}\left[i_{a}\angle -120^{0}\right] + i_{b} + \operatorname{Re}\left[i_{c}\angle 120^{0}\right] = \frac{3}{2}i_{b}$$

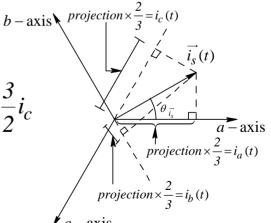
$$\frac{1}{2}i_{a} \qquad \frac{1}{2}i_{c} \qquad \frac{1}{2}i_{c}$$

$$\Rightarrow i_{b}(t) = \frac{2}{3}\operatorname{Re}(\vec{i}_{s}\angle -120^{\circ}) = \frac{2}{3}\hat{I}_{s}\cos\left(\theta_{i_{s}} -120^{\circ}\right)$$

$$\operatorname{Re}\left[\vec{i}_{s}\angle -240^{0}\right] = \operatorname{Re}\left[i_{a}\angle -240^{0}\right] + \operatorname{Re}\left[i_{b}\angle -240^{0}\right] + i_{c} = \frac{3}{2}i_{c}$$

$$\Rightarrow i_{c}(t) = \frac{2}{3}\operatorname{Re}(\vec{i}_{s}\angle -240^{\circ}) = \frac{2}{3}\hat{I}_{s}\cos\left(\theta_{i_{s}} -240^{\circ}\right)$$



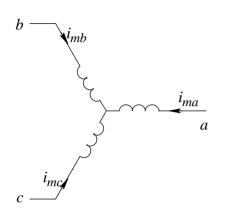


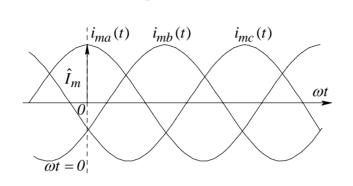


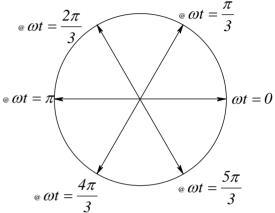


Balanced Sinusoidal Steady-State Excitation

(Rotor Open-Circuited – neglect stator winding resistance and leakage inductance)







$$\Rightarrow \overrightarrow{i_{ms}}(t) = \hat{I}_{ms} \angle \omega t \quad \text{where} \quad \hat{I}_{ms} = \frac{3}{2}\hat{I}_{m}$$

$$r \text{ Rotating MMF } \overrightarrow{F_{ms}}(t) = \frac{N_s}{2}\overrightarrow{i_s}(t) = \hat{F}_{ms} \angle \omega t \quad \text{where} \quad \hat{F}_{ms} = \frac{3}{2}\frac{N_s}{2}\hat{I}_{m} = \frac{N_s}{2}\hat{I}_{ms}$$
& Flux density $\overrightarrow{B_{ms}}(t) = \left(\frac{\mu_o}{\ell_g}\right)\frac{N_s}{2}\overrightarrow{i_{ms}}(t)$

r Constant amplitude





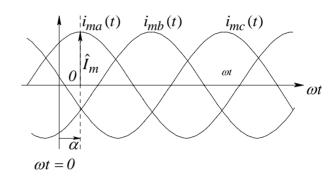
Relation Between Space Vectors and Phasors

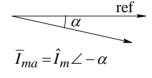
u Time domain

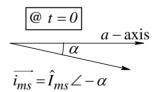
u Phasor

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u Space Vector





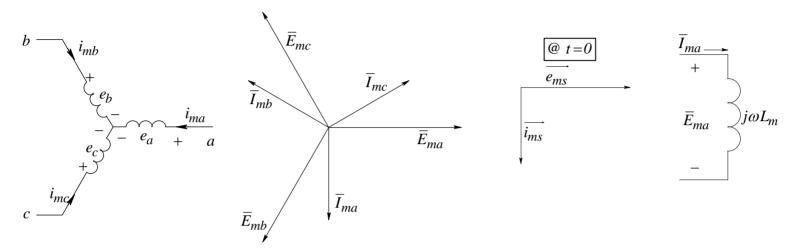


u Space Vector ⇔ phasor





Voltages in the stator windings



Where the three phase magnetizing inductance (2 pole), $L_m = \frac{3}{2} \frac{\pi \mu_o r l}{l_g} \left(\frac{N_s}{2}\right)^2$





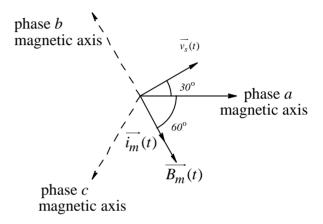
Example

$$v_a(t) = 120\sqrt{2}\cos\omega t$$

$$v_b(t) = 120\sqrt{2}\cos(\omega t - 120^\circ)$$

$$v_c(t) = 120\sqrt{2}\cos(\omega t - 240^\circ)$$

$$\vec{v}_s = \frac{3}{2} \times 120\sqrt{2} \angle 30^0 = 254.56 \angle 30^0 V$$



$$\overrightarrow{i_{ms}} = \frac{\overrightarrow{v}_s}{i\omega L_m} = \frac{254.56 \angle (30^0 - 90^0)}{2\pi \times 60 \times 0.777} = 0.869 \angle -60^0 A$$

$$\overrightarrow{B_{ms}} = \frac{\mu_o N_s \vec{i}_{ms}}{2\ell_g} = \frac{4\pi \times 10^{-7} \times 50 \times 0.869 \angle -60^0}{10^{-3}} = 0.055 \angle -60^0 \text{ Wb/m}^2$$



