

23/4/09

Summer 2005

Q1 (b).

$$s = \frac{2}{T} \frac{z-1}{z+1}$$

$$C(s) = \frac{M(s)}{E(s)} = \frac{1}{s+a} = \frac{1}{\frac{2}{T} \frac{z-1}{z+1} + a} = \frac{1(z+1)}{2(z-1) + aT(z+1)}$$

$$= \frac{1(z+1)}{(aT+2)z + (aT-2)}$$

Matched pole zero

$$C(s) = \frac{1}{s+a}$$

 $\Rightarrow s = -a$  is a pole

$$\Rightarrow z = e^{-aT}$$

$$\frac{1(z+1)}{(aT+2)z + (aT-2)} = \frac{1}{z - e^{-aT}}$$

$$\frac{1}{z - e^{-aT}} = \frac{\frac{1(z+1)}{aT+2}}{z + \frac{aT-2}{aT+2}} = \frac{\frac{1(z+1)}{aT+2}}{z + \frac{\frac{aT}{2}-1}{\frac{aT}{2}+1}}$$

$$\Rightarrow -e^{-aT} = \frac{\frac{aT}{2}-1}{\frac{aT}{2}+1} \Rightarrow e^{-aT} = \frac{1 - \frac{Ta}{2}}{1 + \frac{Ta}{2}}$$

$$(c). Y(z) = h_0 + h_1 z^{-1} + h_2 z^{-2} + \dots$$

$$U(z) = \frac{1}{1-z^{-1}}$$

$$Y(z) = G(z)U(z) = (g_0 + g_1 z^{-1} + g_2 z^{-2} + \dots) \frac{1}{1-z^{-1}}$$

$$\Rightarrow (1-z^{-1})(h_0 + h_1 z^{-1} + h_2 z^{-2} + \dots) = (g_0 + g_1 z^{-1} + g_2 z^{-2} + \dots)$$

$$\Rightarrow h_0 + (h_1 - h_0)z^{-1} + (h_2 - h_1)z^{-2} + \dots = g_0 + g_1 z^{-1} + g_2 z^{-2} + \dots$$

$$\Rightarrow G(z) = h_0 + (h_1 - h_0)z^{-1} + (h_2 - h_1)z^{-2} + \dots$$

$$Y(z) = (h_0 + (h_1 - h_0)z^{-1} + (h_2 - h_1)z^{-2} + \dots) U(z)$$

$$y(k) = (h_1 - h_0)u(k-1) + (h_2 - h_1)u(k-2)$$

$$y(k+1) = (h_1 - h_0)u(k) + (h_2 - h_1)u(k-1) + (h_3 - h_2)u(k-2)$$

$$y(k+2) = (h_1 - h_0)u(k+1) + (h_2 - h_1)u(k) + (h_3 - h_2)u(k-1) + (h_4 - h_3)u(k-2)$$

$$\begin{bmatrix} y(k+1) \\ y(k+2) \end{bmatrix} = \begin{bmatrix} h_1 - h_0 & 0 \\ h_2 - h_1 & h_1 - h_0 \end{bmatrix} \begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} + \underline{y_f(k)}$$

$$\underline{y_f(k)} = \begin{bmatrix} h_2 - h_1 & h_3 - h_2 \\ h_3 - h_2 & h_4 - h_3 \end{bmatrix} \begin{bmatrix} u(k-1) \\ u(k-2) \end{bmatrix} = \begin{bmatrix} -0.2 & 0.5 \\ 0.5 & -0.2 \end{bmatrix} \begin{bmatrix} 1 \\ 0.5 \end{bmatrix} = \begin{bmatrix} 0.05 \\ 0.4 \end{bmatrix}$$

$$\begin{bmatrix} y(k+1) \\ y(k+2) \end{bmatrix} = \begin{bmatrix} 0 & 0.5 \end{bmatrix}^T = \begin{bmatrix} 0 \\ 0.5 \end{bmatrix}$$

$$\begin{bmatrix} 0 \\ 0.5 \end{bmatrix} = \begin{bmatrix} 0.5 & 0 \\ -0.2 & 0.5 \end{bmatrix} \begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} + \begin{bmatrix} 0.05 \\ 0.4 \end{bmatrix}$$

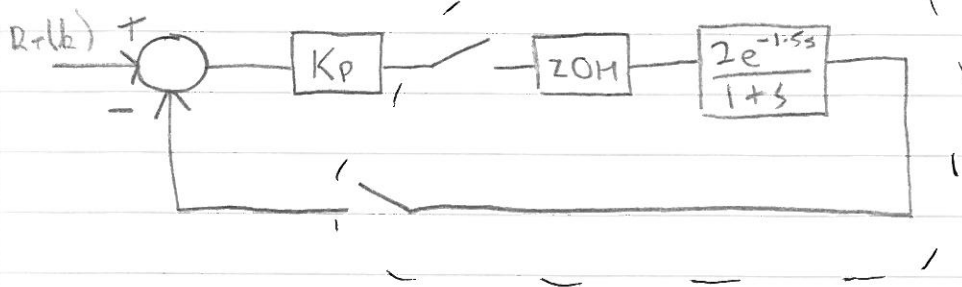
$$\begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} = \begin{bmatrix} 0.5 & 0 \\ -0.2 & 0.5 \end{bmatrix}^{-1} \begin{bmatrix} -0.05 \\ 0.1 \end{bmatrix}$$

$$\begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} = 4 \begin{bmatrix} 0.5 & 0 \\ 0.2 & 0.5 \end{bmatrix} \begin{bmatrix} -0.05 \\ 0.1 \end{bmatrix}$$

$$\begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} = 4 \begin{bmatrix} -0.025 \\ 0.04 \end{bmatrix}$$

$$\begin{bmatrix} u(k) \\ u(k+1) \end{bmatrix} = \begin{bmatrix} -0.1 \\ 0.16 \end{bmatrix}$$

Q2(c).



$$G(z) = \mathcal{Z} \left\{ \frac{1 - e^{-sT}}{s} \cdot \frac{2e^{-1.5s}}{1+s} \right\}$$

$$= 2(1 - z^{-1}) z^{-1} \mathcal{Z}_m \left\{ \frac{1}{s(s+1)} \right\} \Big|_{m=0.5}$$

$$\begin{aligned} m &= 1 - \frac{\theta}{T} \\ &= 1 - \frac{0.5}{1} \\ &= 0.5 \end{aligned}$$

Using modified  $\mathcal{Z}$  transform tables

$$\frac{1}{s(s+a)} \rightarrow \frac{z^{-1}}{a} \left( \frac{1}{1-z^{-1}} - \frac{e^{-aT}}{1-e^{-aT}z^{-1}} \right)$$

$$\frac{1}{s(s+1)} \rightarrow \frac{z^{-1}}{1} \left( \frac{1}{1-z^{-1}} - \frac{e^{-0.5}}{1-e^{-0.5}z^{-1}} \right)$$

$$= z^{-1} \left( \frac{1}{1-z^{-1}} - \frac{0.61}{1-0.37z^{-1}} \right)$$

$$= z^{-1} \left( \frac{1 - 0.37z^{-1} - 0.61 + 0.61z^{-1}}{(1-z^{-1})(1-0.37z^{-1})} \right)$$

$$= z^{-1} \left( \frac{0.39 + 0.24z^{-1}}{(1-z^{-1})(1-0.37z^{-1})} \right)$$

$$= G(z) = \frac{2z^{-2}(0.39 + 0.24z^{-1})}{1 - 0.37z^{-1}}$$

$$\Rightarrow K_P G(z) = \frac{2K_P (0.39 + 0.24z^{-1})}{z^2 - 0.37z}$$

$$z^2 - 0.37z = 0$$

$$z(z - 0.37) = 0$$

$$z = 0 \quad z = 0.37 \quad \text{poles}$$

$$0.39 + 0.24z^{-1} = 0$$

$$z = -0.615 \quad \text{zero}$$

$$\sum_{i=1}^n \frac{1}{\sigma - p_i} = \sum_{j=1}^m \frac{1}{\sigma - z_j}$$

$$\frac{1}{\sigma} + \frac{1}{\sigma - 0.37} = \frac{1}{\sigma + 0.615}$$

$$(\sigma - 0.37)(\sigma + 0.615) + \sigma(\sigma + 0.615) = \sigma(\sigma - 0.37)$$

$$\cancel{\sigma^2} + 0.615\sigma - \cancel{0.37\sigma} - 0.228 + \sigma^2 + 0.615\sigma = \cancel{\sigma^2} - \cancel{0.37\sigma}$$

$$\sigma^2 + 1.23\sigma - 0.228 = 0$$

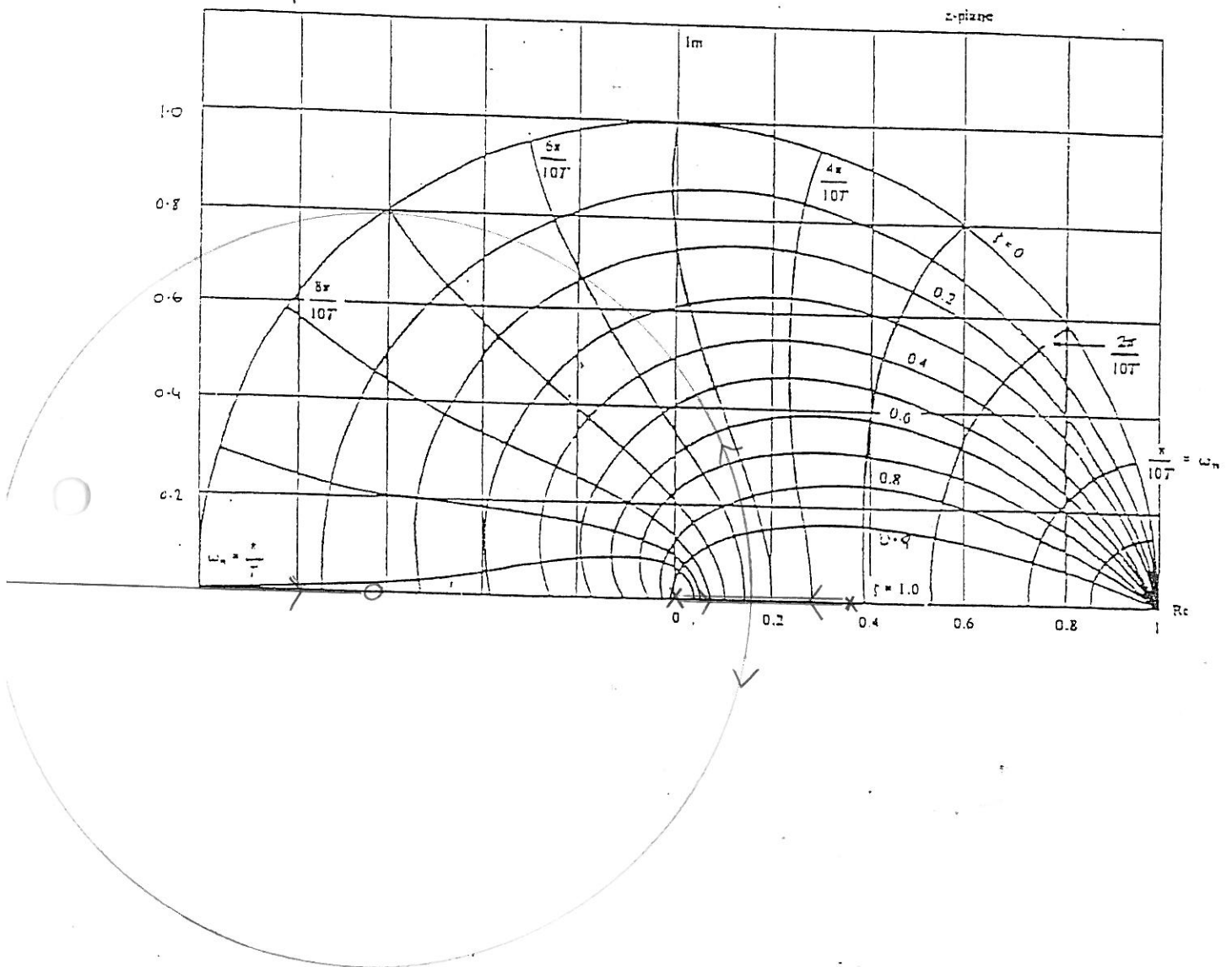
$$\sigma = \frac{-1.23 \pm \sqrt{1.23^2 + 4(0.228)}}{2}$$

$$= \frac{-1.23 \pm 1.56}{2}$$

$$= -1.395 \quad 0.165$$

As  $K_P \uparrow \quad \xi \downarrow \quad \omega_n \uparrow$

Eventually closed loop system becomes unstable



*Z Plane Design Template*

*Please submit with your script*



24/4/09

Summer 2005

Q 3 (b).  $y(k+1) = a_1 y(k) + a_2 y(k-1) + b_1 u(k) + b_2 u(k-1)$   
 $z Y(z) = a_1 Y(z) + a_2 z^{-1} Y(z) + b_1 U(z) + b_2 z^{-1} U(z)$   
 $Y(z)(z - a_1 - a_2 z^{-1}) = U(z)(b_1 + b_2 z^{-1})$   
 $\Rightarrow \frac{Y(z)}{U(z)} = \frac{b_1 + b_2 z^{-1}}{z - a_1 - a_2 z^{-1}} = \frac{b_1 z + b_2}{z^2 - a_1 z - a_2} = \frac{B(z)}{A(z)}$

$$\hat{\Theta}(k) = [\hat{a}_1(k) \hat{a}_2(k) \hat{b}_1(k) \hat{b}_2(k)]^T = [2 \ -1 \ 0 \ 0.5]^T$$

$$a_1 = 2 \quad b_1 = 0$$

$$a_2 = -1 \quad b_2 = 0.5$$

$$n = 2 \Rightarrow n_q = n_s = n - 1 = 1$$

$$\Rightarrow Q(z) = z + q_1$$

$$S(z) = s_0 z + s_1$$

The Diophantine equation is:

$$A_d(z) = A(z)Q(z) + S(z)B(z)$$

$$A_d(z) = (z^2 - a_1 z - a_2)(z + q_1) + (s_0 z + s_1)(b_1 z + b_2)$$

$$A_d(z) = z^3 + q_1 z^2 - a_1 z^2 - a_1 q_1 z - a_2 z - a_2 q_1 + b_1 s_0 z^2 + b_2 s_0 z + b_1 s_1 z + b_2 s_1$$

$$= z^3 + (q_1 - a_1 + b_1 s_0) z^2 + (b_1 s_1 + b_2 s_0 - a_1 q_1 - a_2) z + b_2 s_1 - a_2 q_1$$

$$A_d(z) = z^3 + c_1 z^2 + c_2 z + c_3$$

$$q_1 - a_1 + b_1 s_0 = c_1 \Rightarrow q_1 + b_1 s_0 = c_1 + a_1$$

$$b_1 s_1 + b_2 s_0 - a_1 q_1 - a_2 = c_2 \Rightarrow b_1 s_1 + b_2 s_0 - a_1 q_1 = c_2 + a_2$$

$$b_2 s_1 - a_2 q_1 = c_3$$

$$\begin{bmatrix} 1 & b_1 & 0 \\ -a_1 & b_2 & b_1 \\ -a_2 & 0 & b_2 \end{bmatrix} \begin{bmatrix} q_1 \\ s_0 \\ s_1 \end{bmatrix} = \begin{bmatrix} c_1 + a_1 \\ c_2 + a_2 \\ c_3 \end{bmatrix}$$

$$\begin{bmatrix} 1 & 0 & 0 \\ -2 & 0.5 & 0 \\ 1 & 0 & 0.5 \end{bmatrix} \begin{bmatrix} q_1 \\ s_0 \\ s_1 \end{bmatrix} = \begin{bmatrix} c_1 + 2 \\ c_2 - 1 \\ c_3 \end{bmatrix}$$

Poles at  $z=0.8$  twice

$\Rightarrow$  place fast pole at  $0.8^5 = 0.33$

$$\begin{aligned} A_c(z) &= (z - 0.8)^2 (z - 0.33) \\ &= (z^2 - 1.6z + 0.64)(z - 0.33) \\ &= z^3 - 1.93z^2 + 1.168z - 0.2112 \end{aligned}$$

$$\Rightarrow c_1 = -1.93$$

$$c_2 = 1.168$$

$$c_3 = -0.2112$$

$$\begin{bmatrix} 1 & 0 & 0 \\ -2 & 0.5 & 0 \\ 1 & 0 & 0.5 \end{bmatrix} \begin{bmatrix} q_1 \\ s_0 \\ s_1 \end{bmatrix} = \begin{bmatrix} 0.07 \\ 0.168 \\ -0.2112 \end{bmatrix}$$

$$\Rightarrow \begin{bmatrix} q_1 \\ s_0 \\ s_1 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ -2 & 0.5 & 0 \\ 1 & 0 & 0.5 \end{bmatrix}^{-1} \begin{bmatrix} 0.07 \\ 0.168 \\ -0.2112 \end{bmatrix} = \begin{bmatrix} 0.07 \\ 0.616 \\ -0.5624 \end{bmatrix}$$

$$\Rightarrow Q(z) = z + 0.07$$

$$S(z) = 0.616 - 0.5624z$$

$$1(z) = t_0 A_0 = t_0 (z - 0.33)$$

$$t_0 = \frac{A_c(z)}{B(z)}$$

For unity DC gain

$$\lim_{z \rightarrow 1} \frac{t_0 B(z)}{A_c(z)} = 1$$

$$z \rightarrow 1$$

$$\Rightarrow t_0 = \frac{A_c(1)}{B(1)} = 0.02$$

21/4/09

Summer 05

Q 4 (a).  $\frac{d}{dt} \underline{x}(t) = A \underline{x}(t) + B \underline{u}(t)$

Taking Laplace transform yields

$$\underline{X}(s) = (sI - A)^{-1} (B \underline{U}(s) + \underline{x}(0))$$

Now solve for the state trajectory using the inverse Laplace transform

$$\underline{x}(t) = \mathcal{L}^{-1} \{ \underline{X}(s) \}$$

$$\mathcal{L}^{-1} \{ W(s) V(s) \} = w(t) \otimes v(t) = \int_0^t w(t-\tau) v(\tau) d\tau$$

Define  $\phi(s) = (sI - A)^{-1}$

Transition matrix

$$\underline{X}(s) = \phi(s) \underline{x}(0) + \underbrace{\phi(s) B}_{W(s)} \underbrace{\underline{U}(s)}_{V(s)}$$

Taking inverse Laplace transform yields

$$\underline{x}(t) = \phi(t) \underline{x}(0) + \mathcal{L}^{-1} \{ W(s) V(s) \}$$

$$\underline{x}(t) = \phi(t) \underline{x}(0) + w(t) \otimes v(t)$$

$$\underline{x}(t) = \phi(t) \underline{x}(0) + \int_0^t \underbrace{\phi(t-\tau)}_{w(t-\tau)} \underbrace{B \underline{u}(\tau)}_{v(\tau)} d\tau$$

Consider the zero-input response

$$\underline{x}(t) = \phi(t) \underline{x}(0)$$

which is the solution to  $\dot{\underline{x}} = A \underline{x}$

$$\dot{\underline{x}}(t) = \frac{d}{dt} (\phi(t) \underline{x}(0)) = \frac{d\phi}{dt} \underline{x}(0) \quad \dot{\underline{x}} = A \underline{x} = A \phi(t) \underline{x}(0)$$

$$\ddot{\underline{x}}(t) = \frac{d^2}{dt^2} (\phi(t) \underline{x}(0)) = \frac{d^2\phi}{dt^2} \underline{x}(0) \quad \ddot{\underline{x}} = A \dot{\underline{x}} = A^2 \phi(t) \underline{x}(0)$$

$$\ddot{\underline{x}}(t) = \frac{d^3}{dt^3} (\phi(t) \underline{x}(0)) = \frac{d^3\phi}{dt^3} \underline{x}(0) \quad \ddot{\underline{x}} = A \ddot{\underline{x}} = A^3 \phi(t) \underline{x}(0)$$

$$\frac{d\phi}{dt} = A \phi(t), \quad \frac{d^2\phi}{dt^2} = A^2 \phi(t), \quad \frac{d^3\phi}{dt^3} = A^3 \phi(t)$$

$$\Rightarrow \frac{d^i\phi}{dt^i} = A^i \phi(t)$$

This will be true if  $\phi(t) = I + \frac{At}{1!} + \frac{A^2 t^2}{2!} + \frac{A^3 t^3}{3!} + \dots$



Define the matrix exponential function as:

$$e^{At} = I + \frac{At}{1!} + \frac{A^2 t^2}{2!} + \frac{A^3 t^3}{3!} + \dots$$

$$\begin{aligned}\Rightarrow \phi(t) &= e^{At} \\ \phi(t-\tau) &= e^{A(t-\tau)} \\ &= e^{At} \cdot e^{-A\tau}\end{aligned}$$

$$\Rightarrow \underline{x}(t) = e^{At} \underline{x}(0) + \int_0^t e^{At} \cdot e^{-A\tau} B \underline{u}(\tau) d\tau$$

$$\underline{x}(t) = e^{At} \left( \underline{x}(0) + \int_0^t e^{-A\tau} B \underline{u}(\tau) d\tau \right)$$

(ii) The state trajectory is given by

$$\underline{x}(t) = \phi(t) \underline{x}(0) + \int_0^t \phi(t-\tau) B \underline{u}(\tau) d\tau$$

Consider that the initial time is  $t_0$  with initial state  $\underline{x}(t_0)$

$$\underline{x}(t) = \phi(t-t_0) \underline{x}(t_0) + \int_{t_0}^t \phi(t-\tau) B \underline{u}(\tau) d\tau$$

Consider what happens to the state vector over a time step  $T$   
 $t_0 = kT$ ;  $t = (k+1)T$

$$\underline{x}((k+1)T) = \phi((k+1)T - kT) \underline{x}(kT) + \int_{kT}^{(k+1)T} \phi((k+1)T - \tau) B \underline{u}(\tau) d\tau$$

$$\underline{x}((k+1)T) = \phi(T) \underline{x}(kT) + \int_{kT}^{(k+1)T} \phi((k+1)T - \tau) B \underline{u}(\tau) d\tau$$

Assume a zero order hold is utilized

$\Rightarrow \underline{u}(t) = \underline{u}(kT)$  for  $kT \leq t \leq (k+1)T$

$$\underline{x}((k+1)T) = \phi(T) \underline{x}(kT) + \int_{kT}^{(k+1)T} \phi((k+1)T - \tau) B d\tau \underbrace{\underline{u}(kT)}_{\text{CONSTANT OVER INTEGRAL}}$$

Now make the following substitution

$$\eta = (k+1)T - \tau$$

$$d\eta = -d\tau$$

$$\underline{x}((k+1)T) = \phi(T) \underline{x}(kT) - \int_T^0 \phi(\eta) B d\eta \underline{u}(kT)$$

$$(k+1)T \rightarrow (k+1)$$

$$kT \rightarrow k$$

$$\underline{x}(k+1) = \phi(T)\underline{x}(k) + \int_0^T \phi(\eta) B d\eta \underline{u}(k)$$

$$\underline{x}(k+1) = A_d \underline{x}(k) + B_d \underline{u}(k)$$

$$A_d = \phi(T) = e^{AT} = I + \frac{AT}{1!} + \frac{A^2 T^2}{2!} + \dots$$

$T$  is small

$$\Rightarrow A_d = I + AT$$

$$B_d = \int_0^T \phi(\eta) B d\eta = \int_0^T e^{A\eta} B d\eta = \frac{1}{A} e^{A\eta} B \Big|_0^T = \frac{1}{A} (e^{AT} - I) B$$

$$= A^{-1} (A_d - I) B$$

$$= A^{-1} (I + AT - I) B$$

$$= A^{-1} AT B$$

$$= T B$$

$$\underline{x}(k+1) = (I + AT) \underline{x}(k) + T B \underline{u}(k)$$

$$\underline{x}(k+1) - \underline{x}(k) = \underline{x}(k) + AT \underline{x}(k) + T B \underline{u}(k) - \underline{x}(k)$$

$$\Rightarrow \frac{\Delta \underline{x}(k+1)}{T} = A \underline{x}(k) + B \underline{u}(k)$$

$$(ii) \quad \frac{d}{dt} \underline{x}(t) = \begin{bmatrix} a & 0 \\ 0 & b \end{bmatrix} \underline{x}(t)$$

The eigenvalues of  $A$  are roots of:

$$\det(\lambda I - A) = \begin{vmatrix} \lambda - a & 0 \\ 0 & \lambda - b \end{vmatrix} = 0$$

$$(\lambda - a)(\lambda - b) = 0$$

$$\lambda_1 = a \quad \lambda_2 = b$$

Since  $N=2$  we can write:

$$e^{At} = \alpha_0(t) I + \alpha_1(t) A$$

$$e^{at} = \alpha_0(t) I + \alpha_1(t) a$$

$$e^{bt} = \alpha_0(t) I + \alpha_1(t) b$$

$$\Rightarrow e^{at} - e^{bt} = \alpha_1(t)(a - b)$$

$$\Rightarrow \alpha_1(t) = \frac{e^{at} - e^{bt}}{a - b}$$

$$e^{at} = \lambda_0(t)I + \frac{a(e^{at} - e^{bt})}{a-b}$$

$$\Rightarrow \lambda_0(t)I = e^{at} - \frac{a(e^{at} - e^{bt})}{a-b}$$

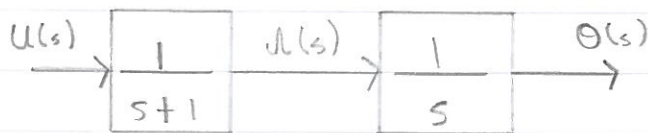
$$\Rightarrow e^{At} = e^{at} - \frac{a(e^{at} - e^{bt})}{a-b} + \frac{e^{at} - e^{bt}}{a-b} \begin{bmatrix} a & 0 \\ 0 & b \end{bmatrix}$$



25/4/09

Summer 2005

Q 4(c).



$$N(s) = \frac{1}{s+1} U(s)$$

$$sN(s) + N(s) = U(s)$$

$$\frac{dw}{dt} + w(t) = u(t)$$

$$\frac{dw}{dt} = -w + u$$

$$\Theta(s) = \frac{N(s)}{s}$$

$$N(s) = s\Theta(s)$$

$$w = \frac{d\Theta}{dt}$$

$$\text{or } \frac{d\Theta}{dt} = w$$

$$\frac{d}{dt} \begin{bmatrix} \Theta \\ w \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} \Theta \\ w \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

$$U(s) = C(s)E(s)$$

$$E(s) = \Theta_{des}(s) - \Theta(s)$$

$$\Rightarrow U(s) = C(s)(\Theta_{des}(s) - \Theta(s))$$

$$= (K_s + K_z)(\Theta_{des}(s) - \Theta(s))$$

$$= K_s \Theta_{des}(s) - K_s \Theta(s) + K_z \Theta_{des}(s) - K_z \Theta(s)$$

$$\Rightarrow u(t) = K \frac{d\Theta_{des}(t)}{dt} - K \frac{d\Theta}{dt} + K_z \Theta_{des}(t) - K_z \Theta(t)$$

Setpoint constant

$$\Rightarrow \frac{d\Theta_{des}(t)}{dt} = 0$$

$$\Rightarrow u(t) = K_z \Theta_{des}(t) - K \frac{d\Theta}{dt} - K_z \Theta(t)$$

$$= K_1 \Theta_{des}(t) - K_2 w(t) - K_1 \Theta(t)$$

$$= K_1 \Theta_{des}(t) - [K_1 \ K_2] \begin{bmatrix} \Theta \\ w \end{bmatrix}$$

$$C_{des}(s) = \det(sI - A + BK)$$

$$= \det \left( \begin{pmatrix} s & 0 \\ 0 & s \end{pmatrix} - \begin{pmatrix} 0 & 1 \\ 0 & -1 \end{pmatrix} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} (K_1 \ K_2) \right)$$

$$= \det \left( \begin{pmatrix} s & -1 \\ 0 & s+1 \end{pmatrix} + \begin{pmatrix} 0 & 0 \\ K_1 & K_2 \end{pmatrix} \right)$$



$$= \det \begin{pmatrix} s & -1 \\ K_1 & s+1+K_2 \end{pmatrix}$$

$$= s(s+1+K_2) + K_1$$

$$= s^2 + (K_2+1)s + K_1$$

$$\begin{aligned} C_{des}(s) &= (s + (2+2j))(s + (2-2j)) \\ &= s^2 + 2s - 2js + 2s + 2js + 4 - 4j + 4j - 4j^2 \\ &= s^2 + 4s + 8 \end{aligned}$$

$$\begin{aligned} K_1 &= 8 & K_2 + 1 &= 4 \\ & & \Rightarrow K_2 &= 3 \end{aligned}$$

$$\begin{aligned} K &= K_2 = 3 & K_z &= K_1 \\ & & 3z &= 8 \\ & & z &= 2.67 \end{aligned}$$

$$\Rightarrow C(s) = 3(s + 2.67)$$

30/4/09

Summer 2005

Q 5(a).  $G(s) = \frac{Y(s)}{U(s)} = \frac{f_0 + f_1 s + \dots + f_{N-1} s^{N-1}}{s^N + e_{N-1} s^{N-1} + \dots + e_0}$

$$\frac{Y(s)}{U(s)} = \frac{b_0/s^N + b_1/s^{N-1} + \dots + b_{N-1}/s}{1 + e_{N-1}/s + e_{N-2}/s^2 + \dots + e_0/s^N}$$

Cross multiplying yields:

$$(1 + e_{N-1}/s + e_{N-2}/s^2 + \dots + e_0/s^N) Y(s) = (b_0/s^N + b_1/s^{N-1} + b_2/s^{N-2} + \dots + b_{N-1}/s) U(s)$$

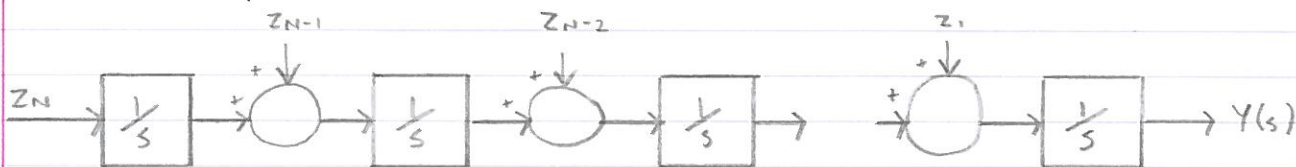
Solving for  $Y(s)$

$$Y(s) = \frac{1}{s} (f_{N-1} U - e_{N-1} Y) + \frac{1}{s^2} (f_{N-2} U - e_{N-2} Y) + \dots + \frac{1}{s^N} (f_0 U - e_0 Y)$$

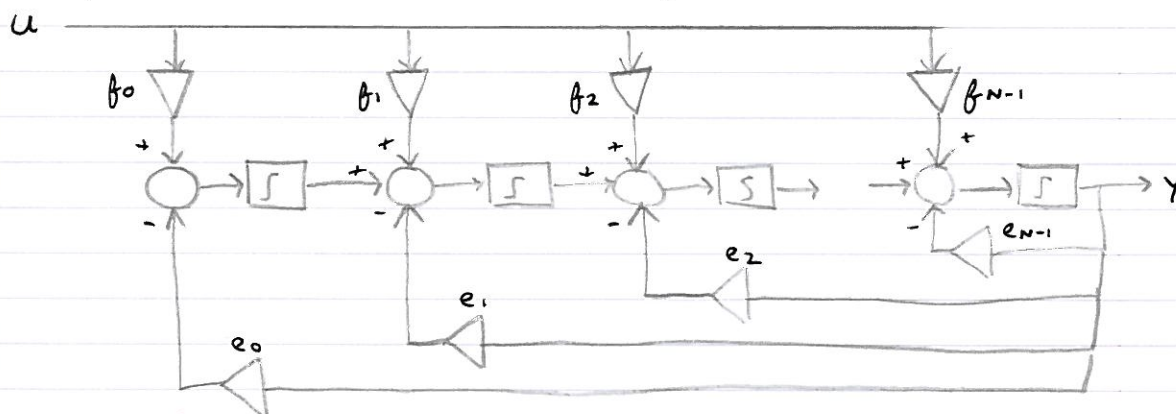
which could be written as  $z_i = f_{N-i} U - e_{N-i} Y$

$$Y(s) = \frac{1}{s} (z_1 + \frac{1}{s} (z_2 + \frac{1}{s} (z_3 + \frac{1}{s} (\dots z_{N-1} + \frac{1}{s} (z_N))))))$$

Could be represented as



This yields the observer canonical form



$$\frac{Y(s)}{U(s)} = \frac{K_a + Ks}{s^2 + (b+c)s + bc} = \frac{Ks + K_a}{s^2 + (b+c)s + bc}$$

$\uparrow$   $\uparrow$   
 $e_1$   $e_0$

$$Y(s) = \frac{1}{s} (K U - (b+c) Y) + \frac{1}{s^2} (K_a U - b c Y)$$

