

DSP

Date 27/11/2023

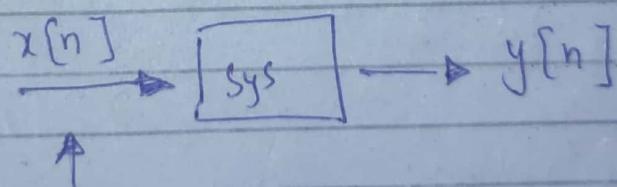
After Mids:-

Systems

- 1. Linear / Non-Linear System
- 2. Time Invariant / Time Variant //
- 3. Memory / Memoryless //
- 4. Stable / Unstable //
- 5. Static / Dynamic //

Type ① \rightarrow follows BIBO
Bounded i/p & Bounded o/p

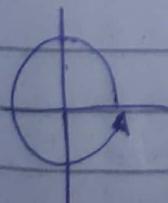
Type ②



e.g.

$$y = x(\sin t) \quad | \quad y = \sin t$$

$$y = \cos t \text{ (stable)}$$



$$-1 \leq y \leq 1$$

$$y = \tan(\theta) \quad (\text{unstable})$$

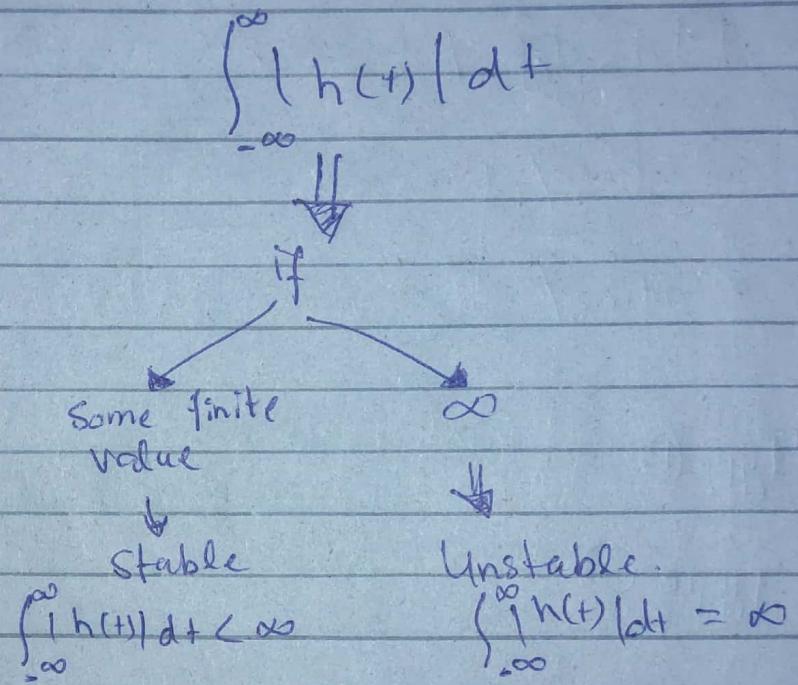
Type ②

Impulse Response

$$\begin{cases} h(t) = ? \\ \sum h(n) = ? \end{cases}$$

(i) Check whether it's cont/discrete.

For Cont.



For discrete,

$$\sum_{-\infty}^{\infty} |h(n)| < \infty \rightarrow \text{stable}$$

otherwise unstable.

Question :-

$$h(t) = e^{-t} \cdot \sin t \cdot u(t)$$

$$h(t) = e^{-at} \cdot \sin bt \cdot u(t).$$

$$h(t) = \frac{e^{-t}}{V} \cdot \frac{\sin t \cdot u(t)}{u}$$

$$= \int_{-\infty}^{\infty} e^{-t} \cdot \sin t \cdot u(t) dt$$

$$h(t) = \int_0^{\infty} e^{-t} \cdot \sin t \cdot dt$$

$\therefore =$

$$-e^{-t} \cdot \sin t + \int e^{-t} \cdot \cos t dt$$

IATE

$$-e^{-t} \cdot \sin t + \left[-e^{-t} \cdot \cos t \right]_0^{\infty}$$

$$= (0 - 0) + \frac{\cos t \cdot e^{-t}}{-1} \Big|_0^{\infty}$$

$$\int_0^{\infty} (-\sin t) \frac{e^{-t}}{-1} dt$$

$$\frac{1}{I} = -(\cos t \cdot e^{-t}) \Big|_0^{\infty} - \int_0^{\infty} e^{-t} \cdot \sin t dt$$

$$2I = -(0 - 1)$$

$$I = \frac{1}{2}$$

$$I = \int_{-\infty}^{\infty} |h(t)| = \frac{1}{2} < \infty$$

stable

If the sys is stable and then find the range of a and b
 if $h(t) = e^{-at} \cdot \sin(bt) \cdot u(t)$

$$I = \int_0^{\infty} e^{-at} \cdot \sin(bt) dt$$

$$I = \frac{e^{-at} \cdot \sin bt}{a} \Big|_0^{\infty} - \int_0^{\infty} \frac{e^{-at}}{a} \cdot \frac{\cos bt}{b} dt$$

$$② h(t) = e^{at} u(t) + \bar{e}^{bt} u(t)$$

$$③ h(n) = \begin{cases} b^n & n < 0 \\ a^n & n \geq 0 \end{cases}$$

Piecewise

If system is stable determine the range of a and b for which sys will be stable,

$$\sum_{-\infty}^{\infty} |h(n)| < \infty$$

$$\sum_{-\infty}^{\infty} b^n + \sum_{0}^{\infty} a^n < \infty$$

$$\sum_{1}^{\infty} b^n + \sum_{0}^{\infty} a^n < \infty$$

$$\sum_{1}^{\infty} b^n + \sum_{0}^{\infty} a^n + 1 - 1 < \infty$$

$$\sum_{0}^{\infty} b^n + \left(\sum_{1}^{\infty} b^n - 1 \right) + \sum_{0}^{\infty} a^n - 1 < \infty$$

$$\sum_{0}^{\infty} b^n - 1 + 1 \rightarrow \sum_{0}^{\infty} |b^n| + \sum_{0}^{\infty} |a^n| + 1$$

$$\frac{1}{| -|b| } + \frac{1}{| -|a| } + 1 < \infty$$

Issue here

$b \neq \pm 1, a \neq \pm 1$

STATIC AND DYNAMIC SYSTEMS:-

↓
memoryless

↳ Memory

$$(a) \int (b) \sum (c) \frac{d}{dt} (d) t - td$$

(i) $y(t) = a \cdot x(t) \Rightarrow \underline{\text{Static}}$

$$\begin{array}{ll} \text{For } t=0 & y(0) = a \cdot x(0) \\ t=1 & y(1) = a \cdot x(1) \end{array}$$

(ii) $y(t) = t \cdot x(t^2) \Rightarrow \underline{\text{Dynamic}}$

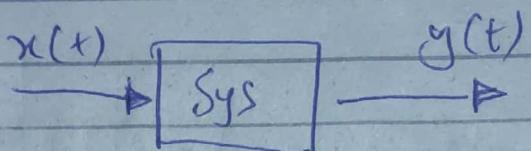
$$\begin{array}{ll} t=0 & y(0) = 0 \\ t=1 & y(1) = 1 \cdot x(1) \\ t=2 & y(2) = 2 \cdot x(4) \end{array}$$

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LTI Systems

→ Mostly used in books

→ Reason: Convolution.



$$y(t) = x(t) * h(t)$$

① Sys Must be linear

② Sys Must be Time Invariant

LTIV → Linear Time Variant

System

Invertible System

Time Check

① Delay (k units)

② Replace

$$n \rightarrow n-k$$

$$t \rightarrow t-k$$

Prob ①

$$y(t) = t \cdot x(t)$$

i) $x_1(t) \Rightarrow y_1(t) = t \cdot x_1(t)$

ii) $x_2(t) \Rightarrow y_2(t) = t \cdot x_2(t)$

$$y_1(t) + y_2(t) = t \cdot x_1(t) + t \cdot x_2(t)$$

$$y' = t(x_1(t) + x_2(t)) \quad \text{---} \textcircled{i}$$

(iii) $x_1(t) + x_2(t) \Rightarrow y'' = t(x_1(t) + x_2(t)) \quad \text{---} \textcircled{ii}$

Prob

$$y(t) = 10x(t^2) + 5$$

i) $x_1(t) \Rightarrow y_1(t) = 10x_1(t^2) + 5$

ii) $x_2(t) \Rightarrow y_2(t) = 10x_2(t^2) + 5$

$$y_1(t) + y_2(t) = 10x_1(t^2) + 10x_2(t^2) + 10$$

$$y' = 10(x_1(t^2) + x_2(t^2) + 1)$$

(iii) $x_1(t) + x_2(t) \Rightarrow y'' = 10[x_1(t^2) + x_2(t^2)] + 5$

$$\Rightarrow y'' = 10x_1(t^2) + 10x_2(t^2) + 5$$

LTV/LTI

- i) Delay $x(t)$ by k units
- ii) Replace t with $t-k$

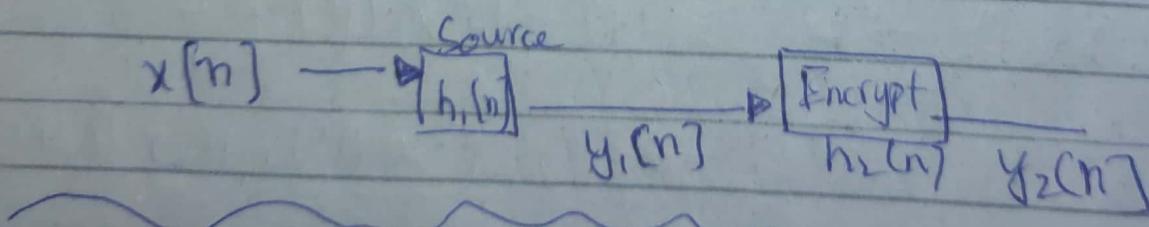
$$y(t) = x(t) - 3u(t)$$

(i) $x(t-y) \rightarrow y(t) = x(t-y) - 3u(t)$

(ii) $y(t-y) = x(t-y) - 3u(t-y)$

TV

Invertible & Non-Invertible Systems.



Date 4/12/2023

Convolution:-

{ = cont.

Σ = discrete.

For Converting to freq. domain.

(i) F.T

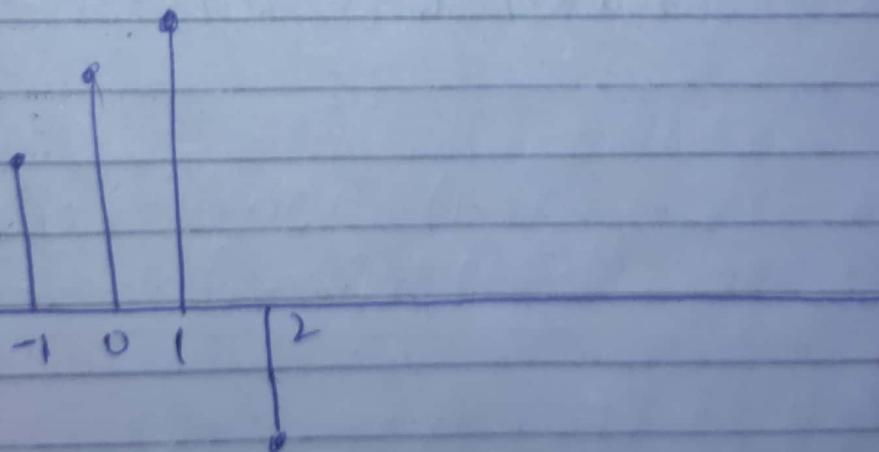
(ii) L.T

$$y[n] = \sum_{m=-\infty}^{\infty} x[m] \cdot h(n-m)$$

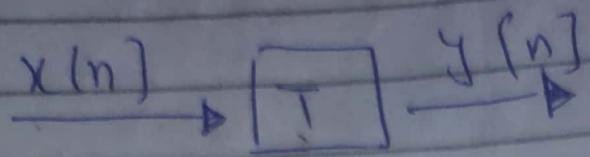
$$\text{Total o/p} = 2IR + 2SR$$

Any discrete signal is represented as:

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \cdot \delta(n-k)$$



$$x[n] = x(-1) \cdot \delta(n+1) + x(0) \cdot \delta(n) + x(1) \cdot \delta(n-1) \\ + x(2) \cdot \delta(n-2)$$



$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \cdot \delta(n-k)$$

$$y[n] = T[x[n]] = T \left(\sum_{k=-\infty}^{\infty} x[k] \cdot \delta(n-k) \right)$$

$$\text{II} = T[- \dots x(-1) \cdot \delta(n+1) + x(0) \cdot \delta(n) + \\ x(p) \cdot \delta(n+1) \dots - \dots]$$

$$\text{II} = [x(-1) \cdot T[\delta(n+1)] + x(0) \cdot T[\delta(n)] \\ + x(1) \cdot T[\delta(n+1)] \dots]$$

$$\text{II} = \dots x(-1) \cdot h(n+1) + x(0) \cdot h(n) + \\ x(1) \cdot h(n-1) \dots$$

$$\text{II} = \sum_{k=-\infty}^{\infty} x(k) \cdot h(n-k)$$

Important Properties of Convolution:-

i) Commutative Prop.

$$y[n] = x[n] * h[n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} x(k) \cdot h(n-k)$$

Let $n-k = m$ } change of
 $n-m = k$ } var

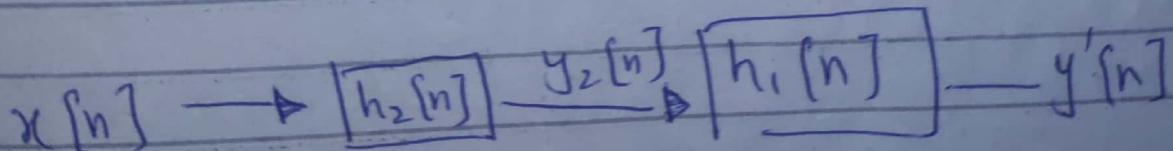
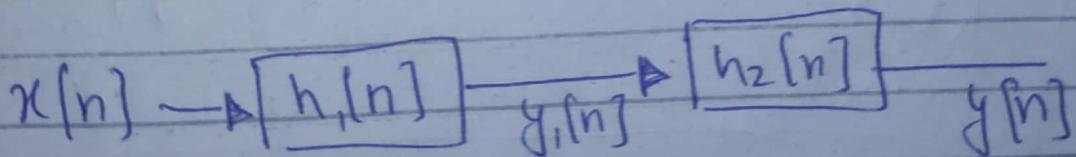
If $k \rightarrow -\infty$ then $m \rightarrow \infty$

$k \rightarrow \infty$ then $m \rightarrow -\infty$

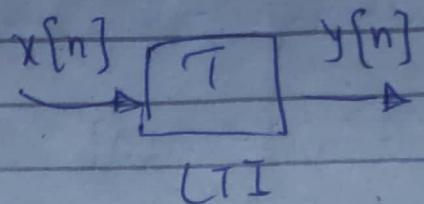
$$y[n] = \sum_{m=0}^{\infty} h[m] \cdot x[n-m]$$

$$y[n] = \sum_{m=-\infty}^{\infty} h[m] \cdot x[n-m] = h[n] * x[n]$$

~~Prop~~ (ii) Dist. Property.



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$$y[n] = \sum_{k=-\infty}^{\infty} x[k] \cdot h[n-k]$$

① Causal System (Realistic)

Present + past values
No future values

③ Non Causal
System (unreal)

Past + future
values

② Anti Causal
only future

So care must be taken if for
 $h[-1], h[-2], \dots$

Distr. Prop.

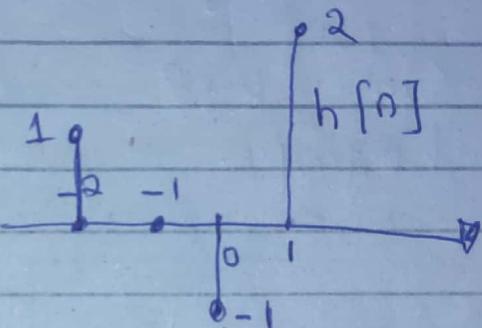
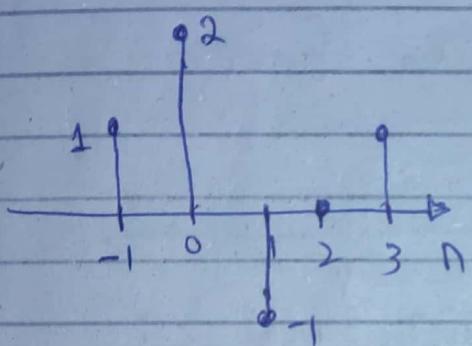
$$y[n] = \sum_{s=-\infty}^{\infty} h_2[s] \cdot g_1[n-s]$$

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Convolution (Graphically):

1. $x[n] \rightarrow \{1, 2, -1, 0, 1\}$

2. $h[n] \rightarrow \{1, 0, -1, 2\}$



3. $y[n] = ?$

starting index = $-1 - 2 = -3$

Ending index = 4

Length = $\text{length}(x[n]) + \text{length}(h[n]) - 1$
= $5 + 4 - 1 = 8$

4. $x(n) \rightarrow x(k)$
 $h(n) \rightarrow h(k)$

5. $y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k) \quad \text{or}$

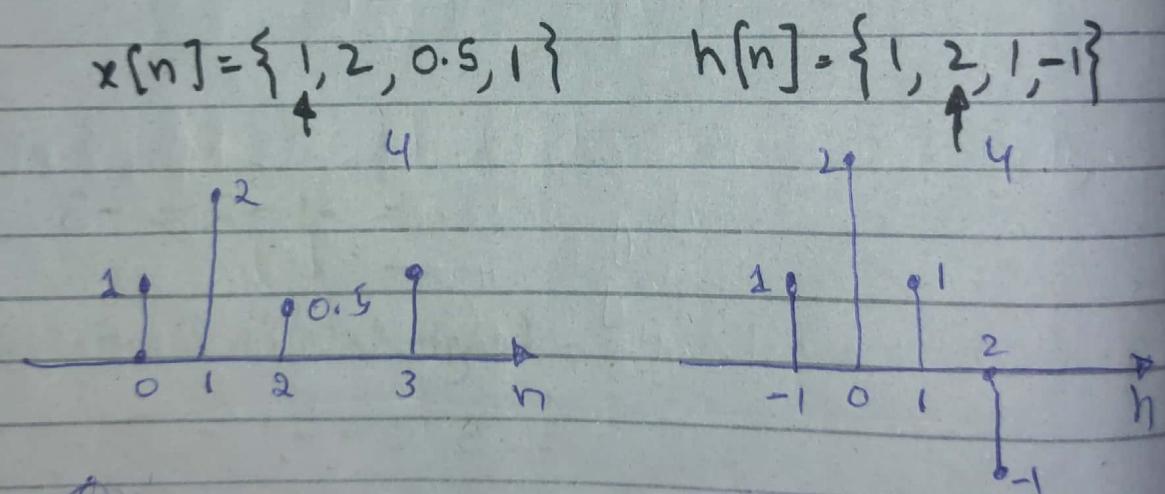
$$y(n) = \sum_{k=0}^{\infty} h(k) \cdot x(n-k)$$

6. $h(-k) \Rightarrow$ folding
Mujhe shifting nazar nahi
avahi.

7. $h(n-k) \Rightarrow$ shifting

8. Multiply $x(k) \cdot h(n-k)$

9. Sum k & lo, Aap ka Kam
hogaya.



Step 1

Starting index of $x[n] = 0$

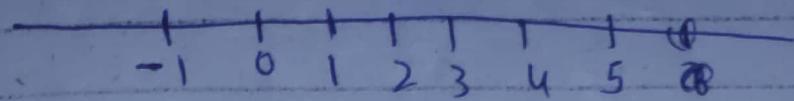
|| || || $h[n] = -1$

|| || || $y[n] = 0 + (-1) = -1$

Len of $x[n] = 4$

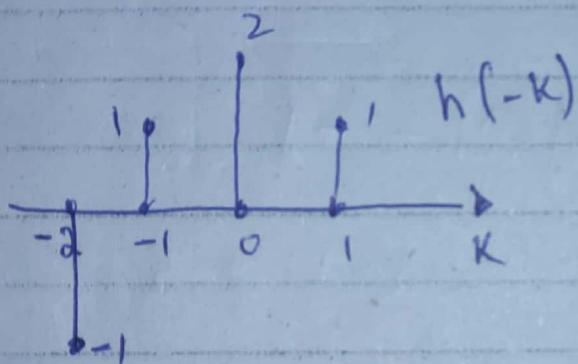
Len of $h[n] = 4$

|| || $y[n] = 7$

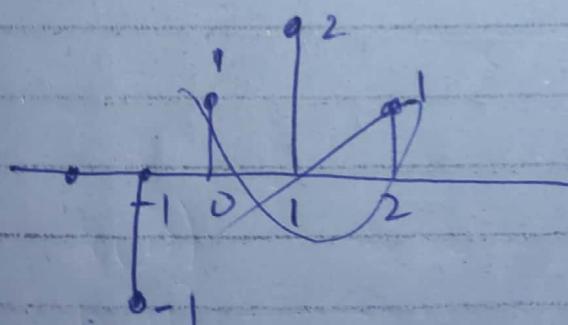


Hence, Ending index of $y[n] = 5$

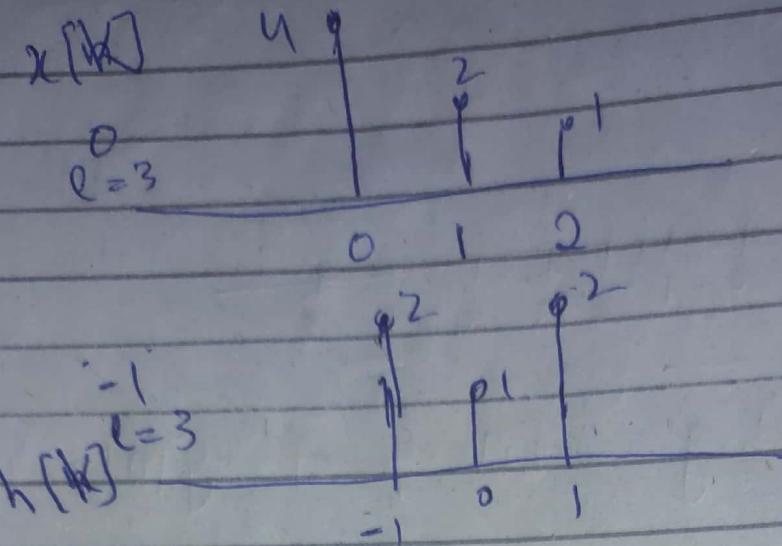
Replace n by k ,



$h(+1-k)$



Output sequence = $\{1, 4, 5.5, 3, 0.5, 0.5, -1\}$

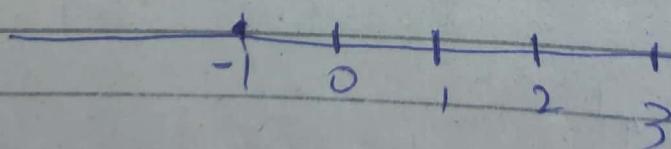


$y[-n]$

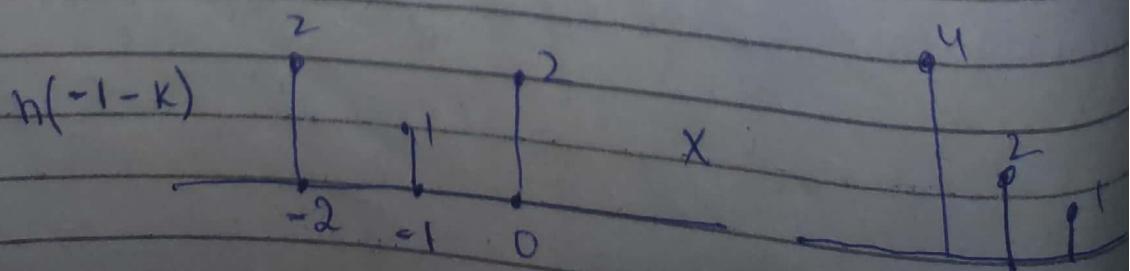
St. ind of $y[n] = -1$

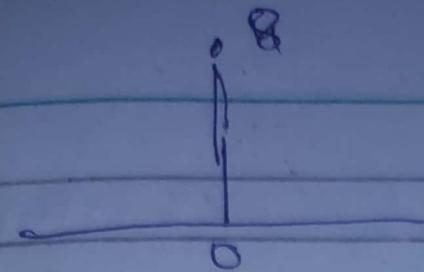
$$\text{len of } y[n] = 3 + 3 - 1 = 5$$

$y(n)$



For $\text{y}(n = -1) \therefore$

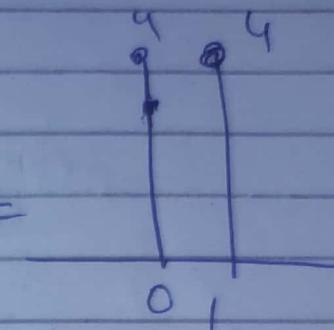
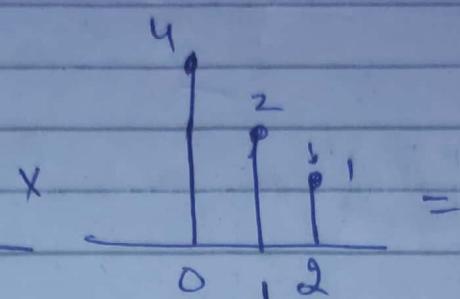
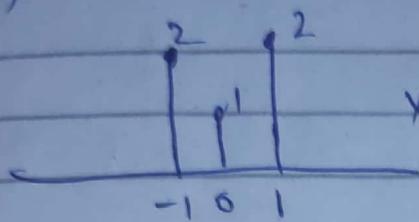




$$y(-1) = 8$$

For $n = 0$

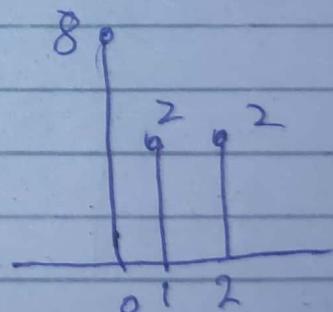
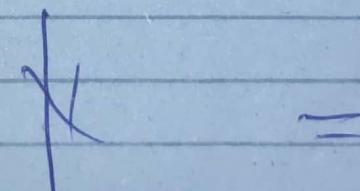
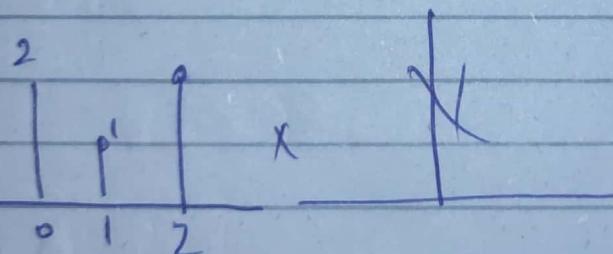
$h(-k)$



$$y(0) = 8$$

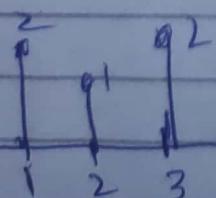
For $n = 1$;

$h(1-k)$



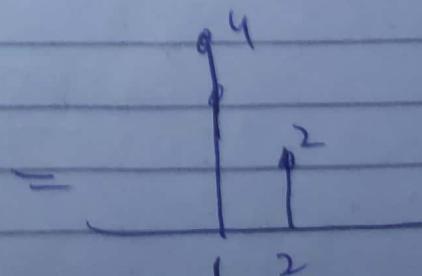
$$y(1) = 12$$

For $n = 2$



\times

$X/$



$$y(2) = 6$$

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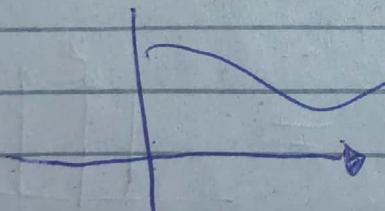
Time Domain

Frequency Domain

Time - Frequency Domain

DFT
DFS

Higher resolution
Correlation
3D Graph



$$DTFT(x(e^{j\omega}))$$

$$\begin{aligned} n &= 1 \dots N \\ n &= n-l \end{aligned}$$

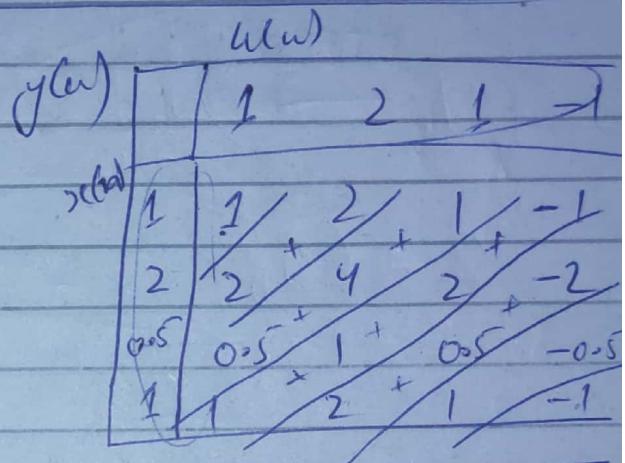
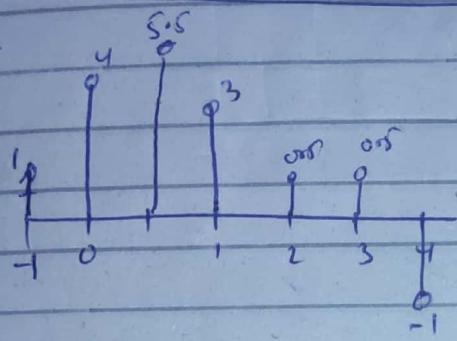
$$\begin{aligned} DFT(x(k)) \\ \xrightarrow{\text{FFT}} \end{aligned}$$

$$n-l \dots N$$

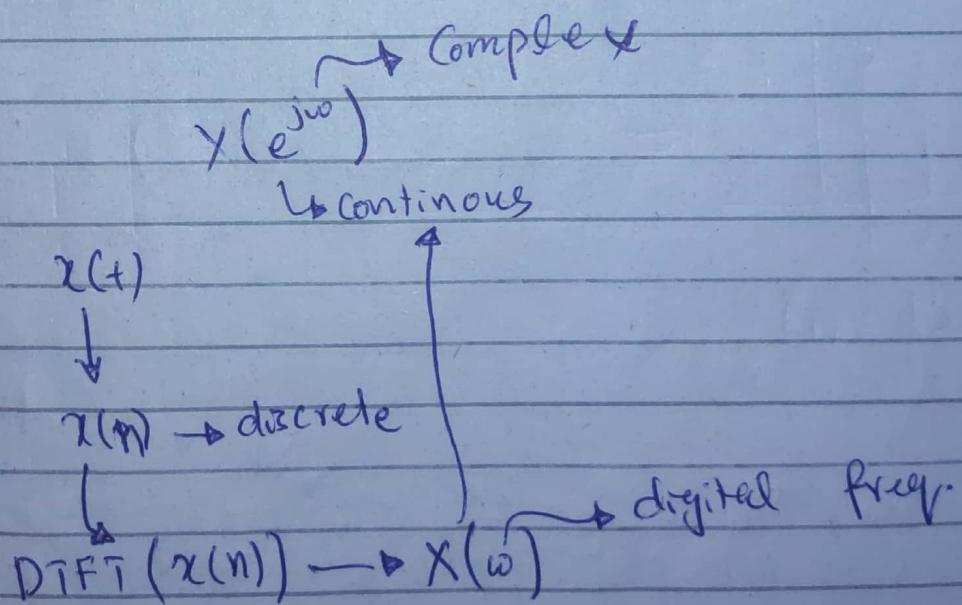
For $n=3$

$$y(n) = 8, 8, 12$$

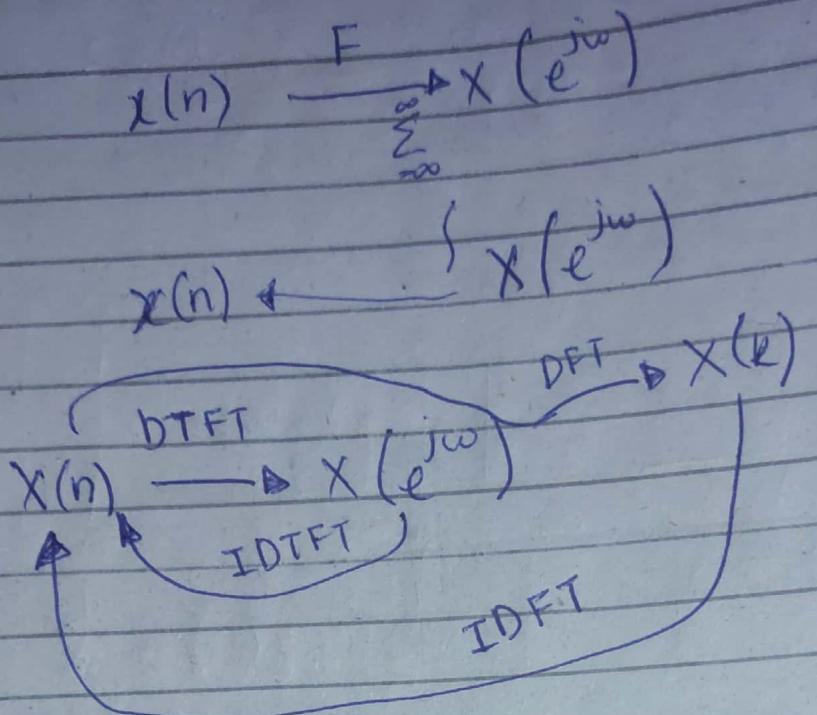
$$y(3) = 2$$



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DFT \rightarrow output is discrete.
 $\downarrow x(k)$
 \downarrow discrete.



$$DTFT\{x(n)\} = \sum_{n=0}^{\infty} x(n) \cdot e^{-j\omega n}$$

$$X(\omega) = \sum_{n=0}^{\infty} x(n) \cdot e^{-j\omega n}$$

$$x(n) \leftrightarrow X(e^{j\omega})$$

No. of samples

$$\begin{aligned} \omega &= \frac{\pi}{4} = \frac{1}{8} \cdot \frac{2\pi}{\cancel{8}} = \frac{\pi}{4} \\ &= 2 \cdot \frac{2\pi}{8} = \frac{\pi}{2} \\ 3 \cdot \frac{2\pi}{8} &= 3\pi/4 \\ 7 \cdot \frac{2\pi}{8} &= 7\pi/4 \end{aligned}$$

generally we can write as:

$$\omega = k \cdot \frac{2\pi}{N}$$

We Sampled our frequency.

Now DFT is:

$$X(k) = \sum_{n=0}^{N-1} x(n) \cdot e^{-j \frac{2\pi k}{N} n}$$

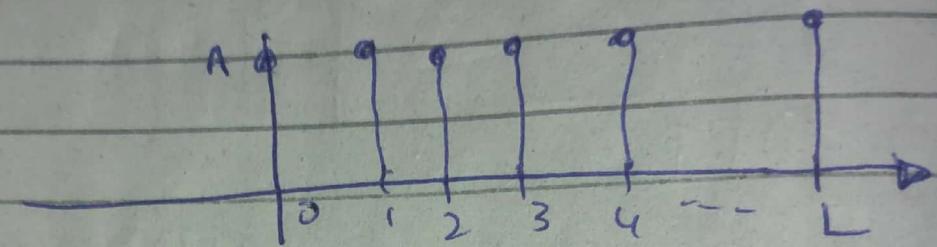
$$x(n) \xrightarrow{\text{DTFT}} X(\omega)$$
$$x(n) \xleftarrow{\text{IDTFT}} X(\omega)$$
$$\sum_{n=0}^{\infty} x(n) \cdot e^{-j\omega n}$$
$$\frac{1}{2\pi} \int_0^{2\pi} X(\omega) \cdot e^{j\omega n} d\omega$$

$$x(n) \xrightarrow{\text{DFT}} X(k)$$
$$\sum_{n=0}^{N-1} x(n) \cdot e^{-j \frac{2\pi n k}{N}}$$

$$X(k) \xrightarrow{\text{IDFT}} x(n)$$
$$\sum_{k=0}^{N-1} X(k) \cdot e^{j \frac{2\pi n k}{N}}$$

Ex 2

$$x(n) = \begin{cases} A & 0 \leq n \leq L \\ 0 & \text{otherwise} \end{cases}$$



DTFT $\rightarrow X(\omega) = ?$

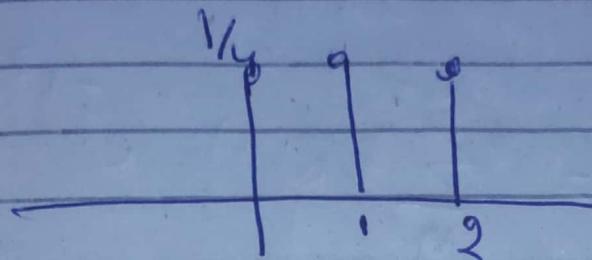
$$X(\omega) = \sum_{n=0}^{L-1} A \cdot e^{-j\omega n}$$

$$= A \cdot \sum_{n=0}^{L-1} e^{-j\omega n}$$

=

Ex ③

$$x(n) = \begin{cases} k_4 & 0 \leq n \leq 2 \\ 0 & \text{otherwise} \end{cases}$$



$$N=3$$

$$k=0, 1, 2$$

DFT $x(k) = ?$

$$x(n) = a^n \quad 0 \leq n \leq N-1$$

$$x(k) = \sum_{n=0}^{N-1} a^n \cdot e^{-j \frac{2\pi}{N} kn}$$

$$= \sum_{n=0}^{N-1} \left(a \cdot e^{-j \frac{2\pi k}{N}} \right)^n$$

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Compute the N-Points DFT

$$x(n) = \delta(n)$$

$$x(k) = \sum_{n=0}^{N-1} \delta(n) \cdot e^{-j \frac{2\pi k}{N} n}$$

$$x(k) = \delta(0) \cdot e^{-j \frac{2\pi k}{N} 0}$$

$$x(k) = 1 \begin{cases} \text{Mag} \\ + 0j \\ \text{phase} \end{cases}$$

$$\text{mag} = \sqrt{1^2 + 0^2} = 1$$

$$\text{phase} = \tan^{-1}\left(\frac{0}{1}\right) = \tan^{-1}\left(\frac{0}{1}\right) = 0$$

$$x(n) = \underline{\delta(n-n_0)} \quad 0 < n_0 < N$$

$$x(k) = \sum_{n=0}^{N-1} \delta(n-n_0) e^{-j \frac{2\pi k}{N} n}$$

$$x(k) = 1 \quad e^{-j \frac{2\pi k n_0}{N}}$$

at

$$\delta(n-n_0) = \begin{cases} 1 & ; n=n_0 \\ 0 & ; n \neq n_0 \end{cases}$$

$$x(k) = e^{-j\frac{2\pi k}{N}n_0}$$

delay in time, produces phase shift
in frequency domain.

$$(iii) x(n) = a^n \quad 0 \leq n \leq N-1$$

$$x(k) = \sum_{n=0}^{N-1} a^n \cdot e^{-j\frac{2\pi k}{N}n}$$

$$x(k) = \sum_{n=0}^{N-1} \left(a \cdot e^{-j\frac{2\pi k}{N}} \right)^n$$

$$x(k) = \frac{\left(a \cdot e^{-j\frac{2\pi k}{N}} \right)^{(N)}}{a \cdot e^{-j\frac{2\pi k}{N}} - 1}$$

$$x(k) = \frac{a e^{-j\frac{2\pi k N}{N}} - 1}{a \cdot e^{-j\frac{2\pi k}{N}} - 1}$$

$$\sum_{n=1}^{N_2} a^n = \frac{a^{N_1} - a^{N_2-1}}{1-a}$$

$$e^{-j2\pi k} = 1$$

$$x(n) = \begin{Bmatrix} 0 & 1 & 2 & 3 \\ 1 & 1 & 1 & 0 \end{Bmatrix}$$

For N=4

$$x(k) = \sum_{n=0}^{N-1} x(n) \cdot e^{-j \frac{2\pi k n}{N}} ; \quad k =$$

$$x(k) = x(0) + x(1) \cdot e^{-j \frac{2\pi k}{N}} +$$

Inverse DFT :-

0	1	2	3	4
1	1	1	1	1

$$x(k) = \{6, -2+2j, -2, -2, -2j\}$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) e^{-j \frac{2\pi k n}{N}}$$

Properties of DFT :-

Linearity :-

$$\text{DFT}\{a_1 x_1(t) + a_2 x_2(t)\} = a_1$$

$$= \sum_{n=0}^{N-1} a_1 x_1(n) e^{-j \frac{2\pi k n}{N}} + \sum_{n=0}^{N-1} a_2 x_2(n) e^{-j \frac{2\pi k n}{N}}$$

$$= a_1 \sum_{n=0}^{N-1} x_1(n) e^{-j \frac{2\pi k n}{N}} + a_2 \sum_{n=0}^{N-1} x_2(n) e^{-j \frac{2\pi k n}{N}}$$

$$= a_1 X_1(k) + a_2 X_2(k)$$

Periodicity :-

$$x(kn + N) = x(n)$$

$$x(k+N) = x(k)$$

Proof :-

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) \cdot e^{\frac{j2\pi kn}{N}}$$

$$x(n+N) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) e^{\frac{j2\pi k(n+N)}{N}}$$

$$x(n+N) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) \cdot e^{\frac{j2\pi kn}{N}} \cdot e^{j2\pi k}$$

$$x(n+N) = \frac{1}{N} \sum_{k=0}^{N-1} x(k) \cdot e^{\frac{j2\pi kn}{N}}$$