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analyzer.lib

This library contains a collection of tools to analyze signals.

It should be used using the an environment:

```
an = library("analyzer.lib");
process = an.functionCall;
```

Another option is to import stdfaust.lib which already contains the an environment:

```
import("stdfaust.lib");
process = an.functionCall;
```

Amplitude Tracking

amp_follower

Classic analog audio envelope follower with infinitely fast rise and exponential decay. The amplitude envelope instantaneously follows the absolute value going up, but then floats down exponentially.

Usage

```
_ : amp_follower(rel) : _
```

Where:

• rel: release time = amplitude-envelope time-constant (sec) going down

Reference

Musical Engineer's Handbook, Bernie Hutchins, Ithaca NY, 1975 Electronotes Newsletter, Bernie Hutchins

amp_follower_ud

Envelope follower with different up and down time-constants (also called a "peak detector").

```
_ : amp_follower_ud(att,rel) : _
```

Where:

- att: attack time = amplitude-envelope time constant (sec) going up
- rel: release time = amplitude-envelope time constant (sec) going down

Note

We assume rel >> att. Otherwise, consider rel \sim max(rel,att). For audio, att is normally faster (smaller) than rel (e.g., 0.001 and 0.01). Use amp_follower_ar below to remove this restriction.

Reference

• "Digital Dynamic Range Compressor Design — A Tutorial and Analysis", by Dimitrios Giannoulis, Michael Massberg, and Joshua D. Reiss http://www.eecs.qmul.ac.uk/~josh/documents/GiannoulisMassbergReiss-dynamicrangecompression-JAES2012.pdf

```
amp_follower_ar(att,rel)
```

Envelope follower with independent attack and release times. The release can be shorter than the attack (unlike in amp_follower_ud above).

Usage

```
_ : amp_follower_ar(att,rel) : _;
```

Spectrum-Analyzers

Spectrum-analyzers split the input signal into a bank of parallel signals, one for each spectral band. They are related to the Mth-Octave Filter-Banks in filter.lib. The documentation of this library contains more details about the implementation. The parameters are:

- M: number of band-slices per octave (>1)
- N: total number of bands (>2)
- ftop = upper bandlimit of the Mth-octave bands ($\langle SR/2 \rangle$)

In addition to the Mth-octave output signals, there is a highpass signal containing frequencies from ftop to SR/2, and a "dc band" lowpass signal containing frequencies from 0 (dc) up to the start of the Mth-octave bands. Thus, the N output signals are

```
\label{eq:highpass} \mbox{(ftop), MthOctaveBands(M,N-2,ftop), dcBand(ftop*2^(-M*(N-1)))}
```

A Spectrum-Analyzer is defined here as any band-split whose bands span the relevant spectrum, but whose band-signals do not necessarily sum to the original signal, either exactly or to within an allpass filtering. Spectrum analyzer outputs are normally at least nearly "power complementary", i.e., the power spectra of the individual bands sum to the original power spectrum (to within some negligible tolerance).

Increasing Channel Isolation

Go to higher filter orders - see Regalia et al. or Vaidyanathan (cited below) regarding the construction of more aggressive recursive filter-banks using elliptic or Chebyshev prototype filters.

References

- "Tree-structured complementary filter banks using all-pass sections", Regalia et al., IEEE Trans. Circuits & Systems, CAS-34:1470-1484, Dec. 1987
- "Multirate Systems and Filter Banks", P. Vaidyanathan, Prentice-Hall, 1993
- Elementary filter theory: https://ccrma.stanford.edu/~jos/filters/

mth_octave_analyzer[N]

Octave analyzer.

Usage

Where:

```
_: mth_octave_analyzer(0,M,ftop,N) : par(i,N,_); // Oth-order Butterworth
_: mth_octave_analyzer6e(M,ftop,N) : par(i,N,_); // 6th-order elliptic
Also for convenience:
_: mth_octave_analyzer3(M,ftop,N) : par(i,N,_); // 3d-order Butterworth
_: mth_octave_analyzer5(M,ftop,N) : par(i,N,_); // 5th-roder Butterworth
mth_octave_analyzer_default = mth_octave_analyzer6e;
```

• 0: order of filter used to split each frequency band into two

- M: number of band-slices per octave
- ftop: highest band-split crossover frequency (e.g., 20 kHz)
- N: total number of bands (including dc and Nyquist)

Mth-Octave Spectral Level

Spectral Level: Display (in bar graphs) the average signal level in each spectral band.

mth_octave_spectral_level6e

Spectral level display.

Usage:

```
_ : mth_octave_spectral_level6e(M,ftop,NBands,tau,dB_offset) : _;
```

Where:

- M: bands per octave
- ftop: lower edge frequency of top band
- NBands: number of passbands (including highpass and dc bands),
- tau: spectral display averaging-time (time constant) in seconds,
- dB offset: constant dB offset in all band level meters.

Also for convenience:

```
mth_octave_spectral_level_default = mth_octave_spectral_level6e;
spectral_level = mth_octave_spectral_level(2,10000,20);
```

[third|half]_octave_[analyzer|filterbank]

A bunch of special cases based on the different analyzer functions described above:

```
third_octave_analyzer(N) = mth_octave_analyzer_default(3,10000,N);
third_octave_filterbank(N) = mth_octave_filterbank_default(3,10000,N);
half_octave_analyzer(N) = mth_octave_analyzer_default(2,10000,N);
half_octave_filterbank(N) = mth_octave_filterbank_default(2,10000,N);
octave_filterbank(N) = mth_octave_filterbank_default(1,10000,N);
octave_analyzer(N) = mth_octave_analyzer_default(1,10000,N);
```

See mth_octave_spectral_level_demo.

Arbritary-Crossover Filter-Banks and Spectrum Analyzers

These are similar to the Mth-octave analyzers above, except that the band-split frequencies are passed explicitly as arguments.

analyzer

Analyzer.

Usage

```
_ : analyzer(0,freqs) : par(i,N,_); // No delay equalizer Where:
```

- 0: band-split filter order (ODD integer required for filterbank[i])
- freqs: (fc1,fc2,...,fcNs) [in numerically ascending order], where Ns=N-1 is the number of octave band-splits (total number of bands N=Ns+1).

If frequencies are listed explicitly as arguments, enclose them in parens:

```
_ : analyzer(3,(fc1,fc2)) : _,_,_
```

basic.lib

A library of basic elements for Faust organized in 5 sections:

- Conversion Tools
- Counters and Time/Tempo Tools
- Array Processing/Pattern Matching
- Selectors (Conditions)
- Other Tools (Misc)

It should be used using the ba environment:

```
ba = library("basic.lib");
process = ba.functionCall;
```

Another option is to import stdfaust.lib which already contains the ba environment:

```
import("stdfaust.lib");
process = ba.functionCall;
Conversion Tools
samp2sec(n)
Converts a number of samples to a duration in seconds.
Usage
samp2sec(n) : _
Where:
  • n: number of samples
sec2samp(d)
Converts a duration in seconds to a number of samples.
Usage
sec2samp(d) : _
Where:
  • d: duration in seconds
db2linear(1)
Converts a loudness in dB to a linear gain (0-1).
Usage
db2linear(1) : _
Where:
  • 1: loudness in dB
```

linear2db(g)
Converts a linear gain (0-1) to a loudness in dB.
Usage
<pre>linear2db(g) : _</pre>
Where:
• g: a linear gain
lin2LogGain
Converts a linear gain (0-1) to a log gain (0-1).
Usage
_ : lin2LogGain : _
log2LinGain
Converts a log gain (0-1) to a linear gain (0-1).
Usage
_ : log2LinGain : _
tau2pole(tau)
Returns a real pole giving exponential decay. Note that t60 (time to decay 60 dB) is ${\sim}6.91$ time constants.
Usage
_ : smooth(tau2pole(tau)) : _
Where:

• tau: time-constant in seconds

pole2tau(pole)
Returns the time-constant, in seconds, corresponding to the given real, positive pole in $(0,1)$.
Usage
pole2tau(pole) : _
Where:
• pole: the pole
midikey2hz(mk)
Converts a MIDI key number to a frequency in Hz (MIDI key $69 = A440$).
Usage
midikey2hz(mk) : _
Where:
• mk: the MIDI key number
pianokey2hz(pk)
Converts a piano key number to a frequency in Hz (piano key $49 = A440$).
$\mathbf{U}\mathbf{sage}$
pianokey2hz(pk) : _
Where:
• pk: the piano key number

hz2pianokey(f)

Converts a frequency in Hz to a piano key number (piano key 49 = A440).

Usage

hz2pianokey(f) : _
Where:

• f: frequency in Hz

Counters and Time/Tempo Tools

countdown(n,trig)

Starts counting down from n included to 0. While trig is 1 the output is n. The countdown starts with the transition of trig from 1 to 0. At the end of the countdown the output value will remain at 0 until the next trig.

Usage

countdown(n,trig) : _

Where:

- n: the starting point of the countdown
- trig: the trigger signal (1: start at n; 0: decrease until 0)

countup(n,trig)

Starts counting up from 0 to n included. While trig is 1 the output is 0. The countup starts with the transition of trig from 1 to 0. At the end of the countup the output value will remain at n until the next trig.

Usage

countup(n,trig) : _

Where:

- n: the starting point of the countup
- trig: the trigger signal (1: start at 0; 0: increase until n)

sweep(period,run)
Counts from 0 to 'period' samples repeatedly, while 'run' is 1. Outsputs zero while 'run' is 0.
Usage
<pre>sweep(period,run) : _</pre>
time()
A simple timer that counts every samples from the beginning of the process.
Usage
time : _
tempo(t)
Converts a tempo in BPM into a number of samples.
Usage
tempo(t) : _
Where:
• t: tempo in BPM
period(p)
Basic sawtooth wave of period p.

Usage
<pre>period(p) : _</pre>
Where:
• p: period as a number of samples
<pre>pulse(p)</pre>
Pulses (10000) generated at period p.
Usage
<pre>pulse(p) : _</pre>
Where:
• p: period as a number of samples
<pre>pulsen(n,p)</pre>
Pulses (11110000) of length ${\tt n}$ generated at period ${\tt p}$.
Usage
<pre>pulsen(n,p) : _</pre>
Where:
n: the length of the pulse as a number of samplesp: period as a number of samples
beat(t)

Pulses at tempo t.

beat(t) : _

Where:

• t: tempo in BPM

Array Processing/Pattern Matching

count(1)

Count the number of elements of list l.

Usage

count $((10,20,30,40)) \rightarrow 4$

Where:

• 1: list of elements

take(e,1)

Take an element from a list.

Usage

 $take(3,(10,20,30,40)) \rightarrow 30$

Where:

- p: position (starting at 1)
- 1: list of elements

subseq(1, p, n)

Extract a part of a list.

```
subseq((10,20,30,40,50,60), 1, 3) \rightarrow (20,30,40)
subseq((10,20,30,40,50,60), 4, 1) \rightarrow 50
```

Where:

- 1: list
- p: start point (0: begin of list)
- n: number of elements

Note:

Faust doesn't have proper lists. Lists are simulated with parallel compositions and there is no empty list

Selectors (Conditions)

```
if(c,t,e))
```

if-then-else implemented with a select 2.

Usage

• if(c, then, else) : _

Where:

- c: condition
- t: signal selected while c is true
- e: signal selected while c is false

selector(i,n)

Selects the ith input among n at compile time.

```
\_,\_,\_,\_: selector(2,4) : \_ // selects the 3rd input among 4 Where:
```

- i: input to select (int, numbered from 0, known at compile time)
- n: number of inputs (int, known at compile time, n > i)

selectn(N,i)

Selects the ith input among N at run time.

Usage

```
_,_,_ : selectn(4,2) : _ // selects the 3rd input among 4 Where:
```

- N: number of inputs (int, known at compile time, N > 0)
- i: input to select (int, numbered from 0)

Example test program

```
N=64;
process = par(n,N, (par(i,N,i) : selectn(N,n)));
```

select2stereo(bpc)

Select between 2 stereo signals.

Usage

```
_,_,_ : select2stereo(bpc) : _,_,_,
```

Where:

• bpc: the selector switch (0/1)

Other

latch(c)

Latch input on positive-going transition of "clock" ("sample-and-hold").

```
Usage
_ : latch(clocksig) : _
Where:
   • clocksig: hold trigger (0 for hold, 1 for bypass)
sAndH
Sample And Hold.
Usage
_ : sAndH(t) : _
Where:
   • t: hold trigger (0 for hold, 1 for bypass)
peakhold(mode,sig)
Outputs current max value above zero.
Usage
_ : peakhold(mode) : _;
Where:
{\tt mode} means: 0 - Pass through. A single sample 0 trigger will work as a reset. 1 -
Track and hold max value.
peakholder(holdtime)
Tracks abs peak and holds peak for 'holdtime' samples.
Usage
_ : peakholder(holdtime) : _;
```

impulsify

Turns the signal from a button into an impulse $(1,0,0,\ldots)$ when button turns on).

Usage

```
button("gate") : impulsify ;
```

automat

Record and replay to the values the input signal in a loop.

Usage

```
hslider(...) : automat(bps, size, init) : _
```

Break Point Functions

bpf is an environment (a group of related definitions) that can be used to create break-point functions. It contains three functions :

- start(x,y) to start a break-point function
- end(x,y) to end a break-point function
- point(x,y) to add intermediate points to a break-point function

A minimal break-point function must contain at least a start and an end point :

```
f = bpf.start(x0,y0) : bpf.end(x1,y1);
```

A more involved break-point function can contains any number of intermediate points:

```
f = bpf.start(x0,y0) : bpf.point(x1,y1) : bpf.point(x2,y2) : bpf.end(x3,y3);
In any case the x_{i} must be in increasing order (for all i, x_{i} < x_{i+1}).
```

For example the following definition :

```
f = bpf.start(x0,y0) : ... : bpf.point(xi,yi) : ... : bpf.end(xn,yn);
```

implements a break-point function f such that :

```
• f(x) = y_{0} \text{ when } x < x_{0}
```

```
• f(x) = y_{n} \text{ when } x > x_{n}
```

```
• f(x) = y_{i} + (y_{i+1}-y_{i})*(x-x_{i})/(x_{i+1}-x_{i}) when x_{i} \le x and x \le x_{i+1}
```

bypass1(bpc,e)

Takes a mono input signal, route it to e and bypass it if bpc = 1.

Usage

```
_ : bypass1(bpc,e) : _
```

Where:

- bpc: bypass switch (0/1)
- e: a mono effect

bypass2(bpc,e)

Takes a stereo input signal, route it to e and bypass it if bpc = 1.

Usage

```
_,_ : bypass2(bpc,e) : _,_
```

Where:

- bpc: bypass switch (0/1)
- e: a stereo effect

 ${\bf compressor.lib}$

A library of compressor effects.

It should be used using the co environment:

```
co = library("compressor.lib");
process = co.functionCall;
```

Another option is to import stdfaust.lib which already contains the co environment:

```
import("stdfaust.lib");
process = co.functionCall;
```

Functions Reference

compressor_mono and compressor_stereo

Mono and stereo dynamic range compressors.

Usage

```
_ : compressor_mono(ratio,thresh,att,rel) : _
_,_ : compressor_stereo(ratio,thresh,att,rel) : _,_
```

Where:

- ratio: compression ratio (1 = no compression, > 1 means compression)
- thresh: dB level threshold above which compression kicks in (0 dB = max level)
- att: attack time = time constant (sec) when level & compression going up
- rel: release time = time constant (sec) coming out of compression

References

- http://en.wikipedia.org/wiki/Dynamic_range_compression
- https://ccrma.stanford.edu/~jos/filters/Nonlinear_Filter_Example_ Dynamic.html
- Albert Graef's "faust2pd"/examples/synth/compressor_.dsp
- More features: https://github.com/magnetophon/faustCompressors

limiter_*

A limiter guards against hard-clipping. It can be can be implemented as a compressor having a high threshold (near the clipping level), fast attack and release, and high ratio. Since the ratio is so high, some knee smoothing is desirable ("soft limiting"). This example is intended to get you started using compressor_* as a limiter, so all parameters are hardwired to nominal values here. Ratios: 4 (moderate compression), 8 (severe compression), 12 (mild limiting), or 20 to 1 (hard limiting) Att: 20-800 MICROseconds (Note: scaled by ratio in the 1176) Rel: 50-1100 ms (Note: scaled by ratio in the 1176) Mike Shipley likes 4:1 (Grammy-winning mixer for Queen, Tom Petty, etc.) Faster attack gives "more bite" (e.g. on vocals) He hears a bright, clear eq effect as well (not implemented here)

```
_ : limiter_1176_R4_mono : _;
_,_ : limiter_1176_R4_stereo : _,_;
```

Reference:

http://en.wikipedia.org/wiki/1176_Peak_Limiter

delay.lib

This library contains a collection of delay functions.

It should be used using the de environment:

```
de = library("delay.lib");
process = de.functionCall;
```

Another option is to import stdfaust.lib which already contains the de environment:

```
import("stdfaust.lib");
process = de.functionCall;
```

Basic Delay Functions

```
delay(n,d,x)
```

Simple d samples delay where n is the maximum delay length as a number of samples (it needs to be a power of 2). Unlike the Q delay operator, this function allows to preallocate memory which means that d can be changed dynamically at run time as long as it remains smaller than n.

Usage

```
_ : delay(n,d) : _
```

Where:

- ${\tt n}$: the max delay length as a power of 2
- d: the delay length as a number of samples (integer)

fdelay(n,d,x)

Simple d samples fractional delay based on 2 interpolated delay lines where ${\tt n}$ is the maximum delay length as a number of samples (it needs to be a power of 2 - see delay()).

Usage

```
_ : fdelay(n,d) : _
```

Where:

- n: the max delay length as a power of 2
- d: the delay length as a number of samples (float)

sdelay(N,it,dt)

s(mooth)delay: a mono delay that doesn't click and doesn't transpose when the delay time is changed.

Usage

```
_ : sdelay(N,it,dt) : _
```

Where:

- N: maximal delay in samples (must be a constant power of 2, for example 65536)
- it: interpolation time (in samples) for example 1024
- dt: delay time (in samples)

Lagrange Interpolation

fdelaylti and fdelayltv

Fractional delay line using Lagrange interpolation.

```
_ : fdelaylt[i|v](order, maxdelay, delay, inputsignal) : _
```

Where order=1,2,3,... is the order of the Lagrange interpolation polynomial.

fdelaylti is most efficient, but designed for constant/slowly-varying delay. fdelayltv is more expensive and more robust when the delay varies rapidly.

NOTE: The requested delay should not be less than (N-1)/2.

References

- https://ccrma.stanford.edu/~jos/pasp/Lagrange_Interpolation.html
- Timo I. Laakso et al., "Splitting the Unit Delay Tools for Fractional Delay Filter Design", IEEE Signal Processing Magazine, vol. 13, no. 1, pp. 30-60, Jan 1996.
- Philippe Depalle and Stephan Tassart, "Fractional Delay Lines using Lagrange Interpolators", ICMC Proceedings, pp. 341-343, 1996.

fdelay[n]

For convenience, fdelay1, fdelay2, fdelay3, fdelay4, fdelay5 are also available where n is the order of the interpolation.

Thiran Allpass Interpolation

Thiran Allpass Interpolation

Reference

https://ccrma.stanford.edu/~jos/pasp/Thiran Allpass Interpolators.html

fdelay[n]a

Delay lines interpolated using Thiran allpass interpolation.

• N=1,2,3, or 4 is the order of the Thiran interpolation filter, and the delay argument is at least N - 1/2.

Note

The interpolated delay should not be less than $\mathbb{N}-1/2$. (The allpass delay ranges from $\mathbb{N}-1/2$ to $\mathbb{N}+1/2$.) This constraint can be alleviated by altering the code, but be aware that allpass filters approach zero delay by means of pole-zero cancellations. The delay range $[\mathbb{N}-1/2,\mathbb{N}+1/2]$ is not optimal. What is?

Delay arguments too small will produce an UNSTABLE allpass!

Because allpass interpolation is recursive, it is not as robust as Lagrange interpolation under time-varying conditions. (You may hear clicks when changing the delay rapidly.)

First-order allpass interpolation, delay d in [0.5,1.5]

demo.lib

This library contains a set of demo functions based on examples located in the /examples folder.

It should be used using the dm environment:

```
dm = library("demo.lib");
process = dm.functionCall;
```

Another option is to import stdfaust.lib which already contains the dm environment:

```
import("stdfaust.lib");
process = dm.functionCall;
```

Analyzers

```
mth_octave_spectral_level_demo
```

Demonstrate mth_octave_spectral_level in a standalone GUI.

```
_ : mth_octave_spectral_level_demo(BandsPerOctave);
_ : spectral_level_demo : _; // 2/3 octave
```

Filters

parametric_eq_demo

A parametric equalizer application. Based on examples/filtering/parametricEqualizer.dsp

Usage:

```
_ : parametric_eq_demo : _ ;
```

spectral_tilt_demo

A spectral tilt application. Based on examples/filtering/parametricEqualizer.dsp

Usage

```
\_ : spectral_tilt_demo(N) : \_ ;
```

Where:

• N: filter order (integer)

All other parameters interactive

${\tt mth_octave_filterbank_demo}$ and ${\tt filterbank_demo}$

Graphic Equalizer: Each filter-bank output signal routes through a fader. Based on examples/filtering/filterBank.dsp

Usage

```
_ : mth_octave_filterbank_demo(M) : _
_ : filterbank_demo : _
```

Where:

• N: number of bands per octave

Effects

cubicnl_demo

Distortion demo application. Based on examples/dynamic/distortion.dsp

Usage:

```
_ : cubicnl_demo : _;
```

gate_demo

 ${\it Gate \ demo \ application.} \ {\it Based \ on \ examples/dynamic/noise} {\it Gate.dsp}$

Usage

```
_,_ : gate_demo : _,_;
```

compressor_demo

 $Compressor\ demo\ application.\ Based\ on\ examples/dynamic/compressor.dsp$

Usage

```
_,_ : compressor_demo : _,_;
```

exciter

 $Psychoacoustic\ harmonic\ exciter,\ with\ GUI.\ Based\ on\ examples/psychoacoustic/harmonic\ Exciter. dsp$

```
_ : exciter : _
```

References

- https://secure.aes.org/forum/pubs/ebriefs/?elib=16939

moog_vcf_demo

Illustrate and compare all three Moog VCF implementations above. Based on examples/filtering/moogVCF.dsp

Usage

```
_ : moog_vcf_demo : _;
```

$wah4_demo$

Wah pedal application. Based on examples/filtering/wahPedal.dsp

Usage

```
_ : wah4_demo : _;
```

crybaby_demo

Crybaby effect application. Based on examples/filtering/cryBaby.dsp

${\bf Usage}$

```
_ : crybaby_demo : _ ;
```

vocoder_demo

Use example of the vocoder function where an impulse train is used as excitation. Based on examples/filtering/vocoder.dsp

```
_ : vocoder_demo : _;
```

flanger_demo

Flanger effect application. Based on examples/phasing/flanger.dsp

Usage

```
_,_ : flanger_demo : _,_;
```

phaser2_demo

Phaser effect demo application. Based on examples/phasing/phaser.dsp

Usage

```
_,_ : phaser2_demo : _,_;
```

freeverb_demo

Freeverb demo application Based on examples/reverb/freeverb.dsp

Usage

```
_,_ : freeverb_demo : _,_;
```

stereo_reverb_tester

Handy test inputs for reverberator demos below. Based on examples/reverb
/reverbTester.dsp

```
_ : stereo_reverb_tester : _
```

fdnrev0_demo

A reverb application using fdnrev0. Based on examples/reverb/fdnrev.dsp

Usage

```
_,_ : fdnrev0_demo(N,NB,BBSO) : _,_
```

Where:

- n: Feedback Delay Network (FDN) order / number of delay lines used = order of feedback matrix / 2, 4, 8, or 16 [extend primes array below for 32, 64....]
- nb: Number of frequency bands / Number of (nearly) independent T60 controls / Integer 3 or greater
- bbso = Butterworth band-split order / order of lowpass/highpass bandsplit used at each crossover freq / odd positive integer

zita_rev_fdn_demo

Reverb demo application based on zita_rev_fdn. Based on examples/reverb/zitaRevFDN.dsp

Usage

```
si.bus(8) : zita_rev_fdn_demo : si.bus(8)
```

zita_rev1

Example GUI for zita_rev1_stereo (mostly following the Linux zita-rev1 GUI).

Only the dry/wet and output level parameters are "dezippered" here. If parameters are to be varied in real time, use ${\tt smooth(0.999)}$ or the like in the same way.

```
_,_ : zita_rev1 : _,_
```

Reference
http://www.kokkinizita.net/linuxaudio/zita-rev1-doc/quickguide.html
Generators
sawtooth_demo
An application demonstrating the different sawtooth oscillators of Faust. Based on examples/generator/sawtoothLab.dsp
Usage
sawtooth_demo : _
virtual_analog_oscillator_demo
$\label{thm:continuous} \mbox{Virtual analog oscillator demo application. Based on examples/generator/virtual Analog.dsp}$
Usage
<pre>virtual_analog_oscillator_demo : _</pre>
oscrs_demo
Simple application demoing filter based oscillators.
Usage
oscrs_demo : _

envelope.lib

This library contains a collection of envelope generators.

It should be used using the en environment:

```
en = library("envelope.lib");
process = en.functionCall;
Another option is to import stdfaust.lib which already contains the en environment:
import("stdfaust.lib");
process = en.functionCall;
```

Functions Reference

smoothEnvelope

An envelope with an exponential attack and release.

Usage

smoothEnvelope(ar,t) : _

- ar: attack and release duration (s)
- t: trigger signal (0-1)

ar

AR (Attack, Release) envelope generator (useful to create percussion envelopes).

Usage

asr

ASR (Attack, Sustain, Release) envelope generator.

```
asr(a,s,r,t) : _
```

Where:

- a, s, r: attack (sec), sustain (percentage of t), release (sec)
- t: trigger signal (>0 for attack, then release is when t back to 0)

adsr

ADSR (Attack, Decay, Sustain, Release) envelope generator.

Usage

```
adsr(a,d,s,r,t) : _
```

Where:

- a, d, s, r: attack (sec), decay (sec), sustain (percentage of t), release (sec)
- t: trigger signal (>0 for attack, then release is when t back to 0)

filter.lib

A library of filters and of more advanced filter-based sound processor organized in 18 sections:

- Basic Filters
- Comb Filters
- Direct-Form Digital Filter Sections
- Direct-Form Second-Order Biquad Sections
- Ladder/Lattice Digital Filters
- Useful Special Cases
- Ladder/Lattice Allpass Filters
- Digital Filter Sections Specified as Analog Filter Sections
- Simple Resonator Filters
- Butterworth Lowpass/Highpass Filters
- Special Filter-Bank Delay-Equalizing Allpass Filters
- Elliptic (Cauer) Lowpass Filters
- Elliptic Highpass Filters
- Butterworth Bandpass/Bandstop Filters
- Elliptic Bandpass Filters
- Parametric Equalizers (Shelf, Peaking)

- Mth-Octave Filter-Banks
- Arbritary-Crossover Filter-Banks and Spectrum Analyzers

It should be used using the fi environment:

```
fi = library("filter.lib");
process = fi.functionCall;
```

Another option is to import stdfaust.lib which already contains the fi environment:

```
import("stdfaust.lib");
process = fi.functionCall;
```

Basic Filters

zero(z)

One zero filter. Difference equation: y(n) = x(n) - z * x(n-1).

Usage

```
_ : zero(z) : _
```

Where:

• z: location of zero along real axis in z-plane

Reference

 $https://ccrma.stanford.edu/\sim jos/filters/One_Zero.html$

pole(p)

One pole filter. Could also be called a "leaky integrator". Difference equation: y(n) = x(n) + p * y(n-1).

Usage

```
_ : pole(z) : _
```

Where:

• p: pole location = feedback coefficient

Reference
https://ccrma.stanford.edu/~jos/filters/One_Pole.html
integrator
Same as pole(1) [implemented separately for block-diagram clarity].
dcblockerat(fb)
DC blocker with configurable break frequency. The amplitude response is substantially flat above fb, and sloped at about $+6$ dB/octave below fb. Derived from the analog transfer function $H(s) = s / (s + 2PIfb)$ by the low-frequency matching bilinear transform method (i.e., the standard frequency-scaling constant 2*SR).
Usage
_ : dcblockerat(fb) : _
Where:
• fb: "break frequency" in Hz, i.e., -3 dB gain frequency.
Reference
$https://ccrma.stanford.edu/{\sim}jos/pasp/Bilinear_Transformation.html$
deble sheet
dcblocker
DC blocker. Default dc blocker has -3dB point near 35 Hz (at $44.1~\mathrm{kHz}$) and high-frequency gain near 1.0025 (due to no scaling).
Usage
_ : dcblocker : _

Comb Filters

ff_comb and ff_fcomb

Feed-Forward Comb Filter. Note that ff_comb requires integer delays (uses delay() internally) while ff_fcomb takes floating-point delays (uses fdelay() internally).

Usage

```
_ : ff_comb(maxdel,intdel,b0,bM) : _
_ : ff_fcomb(maxdel,del,b0,bM) : _
```

Where:

- maxdel: maximum delay (a power of 2)
- intdel: current (integer) comb-filter delay between 0 and maxdel
- del: current (float) comb-filter delay between 0 and maxdel
- b0: gain applied to delay-line input
- bM: gain applied to delay-line output and then summed with input

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/Feedforward_Comb_Filters.html$

ffcombfilter

```
Typical special case of ff_{comb}() where: b0 = 1.
```

fb_comb and fb_fcomb

Feed-Back Comb Filter.

Usage

```
_ : fb_comb(maxdel,intdel,b0,aN) : _
_ : fb_fcomb(maxdel,del,b0,aN) : _
```

- maxdel: maximum delay (a power of 2)
- intdel: current (integer) comb-filter delay between 0 and maxdel
- del: current (float) comb-filter delay between 0 and maxdel

- b0: gain applied to delay-line input and forwarded to output
- aN: minus the gain applied to delay-line output before summing with the input and feeding to the delay line

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/Feedback_Comb_Filters.html$

.

rev1

Special case of fb_comb (rev1(maxdel,N,g)). The "rev1 section" dates back to the 1960s in computer-music reverberation. See the jcrev and brassrev in reverb.lib for usage examples.

fbcombfilter and ffbcombfilter

Other special cases of Feed-Back Comb Filter.

Usage

```
_ : fbcombfilter(maxdel,intdel,g) : _
_ : ffbcombfilter(maxdel,del,g) : _
```

Where:

- maxdel: maximum delay (a power of 2)
- intdel: current (integer) comb-filter delay between 0 and maxdel
- del: current (float) comb-filter delay between 0 and maxdel
- g: feedback gain

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/Feedback_Comb_Filters.html$

allpass_comb and allpass_fcomb

```
Schroeder Allpass Comb Filter. Note that 
allpass_comb(maxlen,len,aN) = ff_comb(maxlen,len,aN,1) : fb_comb(maxlen,len-1,1,aN);
```

which is a direct-form-1 implementation, requiring two delay lines. The implementation here is direct-form-2 requiring only one delay line.

Usage

```
_ : allpass_comb (maxdel,intdel,aN) : _
_ : allpass_fcomb(maxdel,del,aN) : _
```

Where:

- maxdel: maximum delay (a power of 2)
- intdel: current (integer) comb-filter delay between 0 and maxdel
- del: current (float) comb-filter delay between 0 and maxdel
- aN: minus the feedback gain

References

- https://ccrma.stanford.edu/~jos/pasp/Allpass_Two_Combs.html
- $https://ccrma.stanford.edu/\sim jos/pasp/Schroeder_Allpass_Sections.html$
- $https://ccrma.stanford.edu/\sim jos/filters/Four_Direct_Forms.html$

rev2

Special case of allpass_comb (rev2(maxlen,len,g)). The "rev2 section" dates back to the 1960s in computer-music reverberation. See the jcrev and brassrev in reverb.lib for usage examples.

allpass_fcomb5 and allpass_fcomb1a

Same as allpass_fcomb but use fdelay5 and fdelay1a internally (Interpolation helps - look at an fft of faust2octave on

```
`1-1' <: allpass_fcomb(1024,10.5,0.95), allpass_fcomb5(1024,10.5,0.95);`).
```

Direct-Form Digital Filter Sections

iir(tfN)

Nth-order Infinite-Impulse-Response (IIR) digital filter, implemented in terms of the Transfer-Function (TF) coefficients. Such filter structures are termed "direct form".

Usage

```
_ : iir(bcoeffs,acoeffs) : _
```

Where:

- order: filter order (int) = max(#poles, #zeros)
- bcoeffs: (b0,b1,...,b_order) = TF numerator coefficients
- acoeffs: $(a1,...,a_order) = TF$ denominator coeffs (a0=1)

Reference

 $https://ccrma.stanford.edu/{\sim}jos/filters/Four_Direct_Forms.html$

fir

FIR filter (convolution of FIR filter coefficients with a signal)

Usage

```
_ : fir(bv) : _
```

Where:

• bv = b0,b1,...,bn is a parallel bank of coefficient signals.

Note

by is processed using pattern-matching at compile time, so it must have this normal form (parallel signals).

Example

Smoothing white noise with a five-point moving average:

```
bv = .2,.2,.2,.2;
process = noise : fir(bv);
Equivalent (note double parens):
process = noise : fir((.2,.2,.2,.2));
```

conv and convN

Convolution of input signal with given coefficients.

```
Usage
```

```
_ : conv((k1,k2,k3,...,kN)) : _; // Argument = one signal bank
_ : convN(N,(k1,k2,k3,...)) : _; // Useful when N < count((k1,...))</pre>
```

tf1, tf2 and tf3

tfN = N'th-order direct-form digital filter.

Usage

```
_ : tf1(b0,b1,a1) : _
_ : tf2(b0,b1,b2,a1,a2) : _
_ : tf3(b0,b1,b2,b3,a1,a2,a3) : _
```

Where:

- a: the poles
- b: the zeros

Reference

https://ccrma.stanford.edu/~jos/fp/Direct_Form_I.html

notchw(width,freq)

Simple notch filter based on a biquad (tf2).

Usage:

```
_ : notchw(width,freq) : _
```

- width: "notch width" in Hz (approximate)
- $\bullet\,$ freq: "notch frequency" in Hz

Reference

 $https://ccrma.stanford.edu/\sim jos/pasp/Phasing_2nd_Order_Allpass_Filters. \\ html$

Direct-Form Second-Order Biquad Sections

Direct-Form Second-Order Biquad Sections

Reference

https://ccrma.stanford.edu/~jos/filters/Four_Direct_Forms.html

tf21, tf22, tf22t and tf21t

tfN = N'th-order direct-form digital filter where:

- tf21 is tf2, direct-form 1
- tf22 is tf2, direct-form 2
- tf22t is tf2, direct-form 2 transposed
- tf21t is tf2, direct-form 1 transposed

Usage

```
- : tf21(b0,b1,b2,a1,a2) : _
- : tf22(b0,b1,b2,a1,a2) : _
- : tf22t(b0,b1,b2,a1,a2) : _
- : tf21t(b0,b1,b2,a1,a2) : _
```

Where:

- a: the poles
- b: the zeros

Reference

 $https://ccrma.stanford.edu/{\sim}jos/fp/Direct_Form_I.html$

Ladder/Lattice Digital Filters

Ladder and lattice digital filters generally have superior numerical properties relative to direct-form digital filters. They can be derived from digital waveguide filters, which gives them a physical interpretation.

av2sv

Compute reflection coefficients sv from transfer-function denominator av.

Usage

```
sv = av2sv(av)
```

Where:

- av: parallel signal bank a1,...,aN
- sv: parallel signal bank s1,...,sN

where ro = ith reflection coefficient, and ai = coefficient of $z^(-i)$ in the filter transfer-function denominator A(z).

Reference

 $https://ccrma.stanford.edu/\sim jos/filters/Step_Down_Procedure.html~(where~reflection~coefficients~are~denoted~by~k~rather~than~s).$

bvav2nuv

Compute lattice tap coefficients from transfer-function coefficients.

Usage

```
nuv = bvav2nuv(bv,av)
```

Where:

- av: parallel signal bank a1,...,aN
- bv: parallel signal bank b0,b1,...,aN
- nuv: parallel signal bank nu1,...,nuN

where nui is the i'th tap coefficient, bi is the coefficient of $z^{(-i)}$ in the filter numerator, ai is the coefficient of $z^{(-i)}$ in the filter denominator

```
iir_lat2(bc,av)
```

Two-multiply latice IIR filter or arbitrary order.

Usage

```
_ : iir_lat2(bv,av) : _
```

Where:

- bv: zeros as a bank of parallel signals
- av: poles as a bank of parallel signals

allpassnt(n,sv)

Two-multiply lattice allpass (nested order-1 direct-form-ii allpasses).

Usage

```
_ : allpassnt(n,sv) : _
```

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

```
iir_kl(bc,av)
```

Kelly-Lochbaum ladder IIR filter or arbitrary order.

Usage

```
_ : iir_kl(bv,av) : _
```

Where:

- bv: zeros as a bank of parallel signals
- av: poles as a bank of parallel signals

allpassnklt(n,sv)

Kelly-Lochbaum ladder allpass.

```
_ : allpassklt(n,sv) : _
```

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

```
iir_lat1(bc,av)
```

One-multiply latice IIR filter or arbitrary order.

Usage

```
_ : iir_lat1(bv,av) : _
```

Where:

- $\bullet\,$ by: zeros as a bank of parallel signals
- av: poles as a bank of parallel signals

allpassn1mt(n,sv)

One-multiply lattice allpass with tap lines.

Usage

```
_ : allpassn1mt(n,sv) : _
```

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

iir_nl(bv,av)

Normalized ladder filter of arbitrary order.

```
_ : iir_nl(bv,av) : _
```

Where:

- bv: zeros as a bank of parallel signals
- av: poles as a bank of parallel signals

References

- J. D. Markel and A. H. Gray, Linear Prediction of Speech, New York: Springer Verlag, 1976.
- https://ccrma.stanford.edu/~jos/pasp/Normalized_Scattering_Junctions.html

allpassnnlt(n,sv)

Normalized ladder allpass filter of arbitrary order.

Usage:

```
_ : allpassnnlt(n,sv) : _
```

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1,1)

References

- J. D. Markel and A. H. Gray, Linear Prediction of Speech, New York: Springer Verlag, 1976.

Useful Special Cases

tf2np

Biquad based on a stable second-order Normalized Ladder Filter (more robust to modulation than tf2 and protected against instability).

```
_ : tf2np(b0,b1,b2,a1,a2) : _
```

Where:

- a: the poles
- b: the zeros

wgr(f,r,x)

Second-order transformer-normalized digital waveguide resonator.

Usage

```
_ : wgr(f,r) : _
```

Where:

- f: resonance frequency (Hz)
- r: loss factor for exponential decay (set to 1 to make a numerically stable oscillator)

References

- https://ccrma.stanford.edu/~jos/pasp/Digital_Waveguide_Oscillator. html

nlf2

Second order normalized digital waveguide resonator.

Usage

```
_ : nlf2(f,r) : _
```

- f: resonance frequency (Hz)
- \bullet r: loss factor for exponential decay (set to 1 to make a sinusoidal oscillator)

Reference

 $https://ccrma.stanford.edu/\sim jos/pasp/Power_Normalized_Waveguide_Filters.html$

apnl(a1,a2)

Passive Nonlinear Allpass based on Pierce switching springs idea. Switch between allpass coefficient a1 and a2 at signal zero crossings.

Usage

```
_ : apnl(a1,a2) : _
```

Where:

• a1 and a2: allpass coefficients

Reference

• "A Passive Nonlinear Digital Filter Design ..." by John R. Pierce and Scott A. Van Duyne, JASA, vol. 101, no. 2, pp. 1120-1126, 1997

Ladder/Lattice Allpass Filters

An allpass filter has gain 1 at every frequency, but variable phase. Ladder/lattice allpass filters are specified by reflection coefficients. They are defined here as nested allpass filters, hence the names allpassn*.

References

- https://ccrma.stanford.edu/~jos/pasp/Conventional_Ladder_Filters. html
- https://ccrma.stanford.edu/~jos/pasp/Nested_Allpass_Filters.html
- Linear Prediction of Speech, Markel and Gray, Springer Verlag, 1976

allpassn(n,sv)

Two-multiply lattice - each section is two multiply-adds.

```
_ : allpassn(n,sv) : _
```

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

References

• J. O. Smith and R. Michon, "Nonlinear Allpass Ladder Filters in FAUST", in Proceedings of the 14th International Conference on Digital Audio Effects (DAFx-11), Paris, France, September 19-23, 2011.

allpassnn(n,tv)

Normalized form - four multiplies and two adds per section, but coefficients can be time varying and nonlinear without "parametric amplification" (modulation of signal energy).

Usage:

```
_ : allpassnn(n,tv) : _
```

Where:

- n: the order of the filter
- tv: the reflexion coefficients (-PI PI)

allpasskl(n,sv)

Kelly-Lochbaum form - four multiplies and two adds per section, but all signals have an immediate physical interpretation as traveling pressure waves, etc.

Usage:

```
_ : allpassnkl(n,sv) : _
```

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

allpass1m(n,sv)

One-multiply form - one multiply and three adds per section. Normally the most efficient in special-purpose hardware.

Usage:

_ : allpassn1m(n,sv) : _

Where:

- n: the order of the filter
- sv: the reflexion coefficients (-1 1)

Digital Filter Sections Specified as Analog Filter Sections

tf2s and tf2snp

Second-order direct-form digital filter, specified by ANALOG transfer-function polynomials B(s)/A(s), and a frequency-scaling parameter. Digitization via the bilinear transform is built in.

Usage

_ : tf2s(b2,b1,b0,a1,a0,w1) : _

Where:

$$H(s) = \frac{b2 s^2 + b1 s + b0}{s^2 + a1 s + a0}$$

and w1 is the desired digital frequency (in radians/second) corresponding to analog frequency 1 rad/sec (i.e., s=j).

Example

A second-order ANALOG Butterworth lowpass filter, normalized to have cutoff frequency at 1 rad/sec, has transfer function

where a1 = sqrt(2). Therefore, a DIGITAL Butterworth lowpass cutting off at SR/4 is specified as tf2s(0,0,1,sqrt(2),1,PI*SR/2);

Method

Bilinear transform scaled for exact mapping of w1.

Reference

 $https://ccrma.stanford.edu/\sim jos/pasp/Bilinear_Transformation.html$

tf3s1f

Analogous to tf2s above, but third order, and using the typical low-frequency-matching bilinear-transform constant 2/T ("lf" series) instead of the specific-frequency-matching value used in tf2s and tf1s. Note the lack of a "w1" argument.

Usage

```
_ : tf3slf(b3,b2,b1,b0,a3,a2,a1,a0) : _
```

tf1s

First-order direct-form digital filter, specified by ANALOG transfer-function polynomials B(s)/A(s), and a frequency-scaling parameter.

Usage

```
tf1s(b1,b0,a0,w1)
```

Where:

$$b1 s + b0$$

$$H(s) = ----s + a0$$

and w1 is the desired digital frequency (in radians/second) corresponding to analog frequency 1 rad/sec (i.e., s = j).

Example

A first-order ANALOG Butterworth lowpass filter, normalized to have cutoff frequency at 1 rad/sec, has transfer function

 $\begin{array}{c} 1 \\ H(s) = ---- s + 1 \end{array}$

so b0 = a0 = 1 and b1 = 0. Therefore, a DIGITAL first-order Butterworth lowpass with gain -3dB at SR/4 is specified as

tf1s(0,1,1,PI*SR/2); // digital half-band order 1 Butterworth

Method

Bilinear transform scaled for exact mapping of w1.

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/Bilinear_Transformation.html$

tf2sb

Bandpass mapping of tf2s: In addition to a frequency-scaling parameter w1 (set to HALF the desired passband width in rad/sec), there is a desired center-frequency parameter wc (also in rad/s). Thus, tf2sb implements a fourth-order digital bandpass filter section specified by the coefficients of a second-order analog lowpass prototpe section. Such sections can be combined in series for higher orders. The order of mappings is (1) frequency scaling (to set lowpass cutoff w1), (2) bandpass mapping to wc, then (3) the bilinear transform, with the usual scale parameter 2*SR. Algebra carried out in maxima and pasted here.

Usage

```
_ : tf2sb(b2,b1,b0,a1,a0,w1,wc) : _
```

tf1sb

First-to-second-order lowpass-to-bandpass section mapping, analogous to tf2sb above.

```
_ : tf1sb(b1,b0,a0,w1,wc) : _
```

Simple Resonator Filters

resonlp, resonhp and resonbp

Simple resonant lowpass, highpass and bandpass filters based on tf2s.

Usage

```
_ : resonlp(fc,Q,gain) : _
_ : resonhp(fc,Q,gain) : _
_ : resonbp(fc,Q,gain) : _

Where:

• fc: center frequency (Hz)
• Q: q
• gain: gain (0-1)
```

Butterworth Lowpass/Highpass Filters

lowpass and highpass

Nth-order Butterworth lowpass or highpass filters.

Usage

```
_ : lowpass(N,fc) : _
_ : highpass(N,fc) : _
```

- N: filter order (number of poles) [nonnegative constant integer]
- fc: desired cut-off frequency (-3dB frequency) in Hz

T	•			
Ref	αr	Δn	\boldsymbol{c}	Δc
TCCI	CI	c_{11}	\cdot	

- https://ccrma.stanford.edu/ \sim jos/filters/Butterworth_Lowpass_Design.
- butter function in Octave ("[z,p,g] = butter(N,1,'s');")

lowpass0_highpass1(s,N,fc)

TODO

Special Filter-Bank Delay-Equalizing Allpass Filters

These special allpass filters are needed by filterbank et al. below. They are equivalent to (lowpass(N,fc) + |- highpass(N,fc))/2, but with canceling polezero pairs removed (which occurs for odd N).

lowpass_plus|minus_highpass

TODO

Elliptic (Cauer) Lowpass Filters

Elliptic (Cauer) Lowpass Filters

References

- http://en.wikipedia.org/wiki/Elliptic_filter
- functions neauer and ellip in Octave

lowpass3e(fc)

Third-order Elliptic (Cauer) lowpass filter.

```
_ : lowpass3e(fc) : _
```

Where:

• fc: -3dB frequency in Hz

Design

For spectral band-slice level display (see octave_analyzer3e):

```
[z,p,g] = ncauer(Rp,Rs,3); % analog zeros, poles, and gain, where Rp = 60 % dB ripple in stopband Rs = 0.2 % dB ripple in passband
```

lowpass6e(fc)

Sixth-order Elliptic/Cauer lowpass filter.

Usage

```
_ : lowpass6e(fc) : _
```

Where:

• fc: -3dB frequency in Hz

Design

For spectral band-slice level display (see octave_analyzer6e):

```
[z,p,g] = ncauer(Rp,Rs,6); % analog zeros, poles, and gain, where
Rp = 80 % dB ripple in stopband
Rs = 0.2 % dB ripple in passband
```

Elliptic Highpass Filters

highpass3e(fc)

Third-order Elliptic (Cauer) highpass filter. Inversion of lowpass3e wrt unit circle in s plane (s <- 1/s)

```
_ : highpass3e(fc) : _
```

Where:

• fc: -3dB frequency in Hz

highpass6e(fc)

Sixth-order Elliptic/Cauer highpass filter. Inversion of lowpass3e wrt unit circle in s plane (s <- 1/s)

Usage

```
_ : highpass6e(fc) : _
```

Where:

• fc: -3dB frequency in Hz

Butterworth Bandpass/Bandstop Filters

bandpass and bandstop

Order 2*Nh Butterworth bandpass filter made using the transformation s <- s + wc^2/s on lowpass(Nh), where wc is the desired bandpass center frequency. The lowpass(Nh) cutoff w1 is half the desired bandpass width. A notch-like "bandstop" filter is similarly made from highpass(Nh).

Usage

```
_ : bandpass(Nh,fl,fu) : _
_ : bandstop(Nh,fl,fu) : _
```

- Nh: HALF the desired bandpass/bandstop order (which is therefore even)
- fl: lower -3dB frequency in Hz
- fu: upper -3dB frequency in Hz Thus, the passband (stopband) width is fu-fl, and its center frequency is (fl+fu)/2.

Reference

 $\rm http://cnx.org/content/m16913/latest/$

Elliptic Bandpass Filters

bandpass6e

Order 12 elliptic bandpass filter analogous to bandpass(6).

bandpass12e

Order 24 elliptic bandpass filter analogous to bandpass (6).

Parametric Equalizers (Shelf, Peaking)

Parametric Equalizers (Shelf, Peaking)

References

- http://en.wikipedia.org/wiki/Equalization
- http://www.musicdsp.org/files/Audio-EQ-Cookbook.txt
- Digital Audio Signal Processing, Udo Zolzer, Wiley, 1999, p. 124
- https://ccrma.stanford.edu/~jos/filters/Low_High_Shelving_Filters.html>
- https://ccrma.stanford.edu/~jos/filters/Peaking Equalizers.html>
- maxmsp.lib in the Faust distribution
- bandfilter.dsp in the faust2pd distribution

low_shelf and lowshelf_other_freq

First-order "low shelf" filter (gain boost|cut between dc and some frequency)

Usage

```
_ : lowshelf(N,L0,fx) : _
_ : lowshelf_other_freq(N,L0,fx) : _
```

Where: * N: filter order 1, 3, 5, ... (odd only). * L0: desired level (dB) between dc and fx (boost L0>0 or cut L0<0) * fx: -3dB frequency of lowpass band (L0>0) or upper band (L0<0) (see "SHELF SHAPE" below).

The gain at SR/2 is constrained to be 1. The generalization to arbitrary odd orders is based on the well known fact that odd-order Butterworth band-splits are allpass-complementary (see filterbank documentation below for references).

Shelf Shape

The magnitude frequency response is approximately piecewise-linear on a log-log plot ("BODE PLOT"). The Bode "stick diagram" approximation L(lf) is easy to state in dB versus dB-frequency lf = dB(f):

- L0 > 0:
- L(lf) = L0, f between 0 and fx = 1st corner frequency;
- L(lf) = L0 N * (lf lfx), f between fx and f2 = 2nd corner frequency;
- L(lf) = 0, lf > lf2.
- lf2 = lfx + L0/N = dB-frequency at which level gets back to 0 dB.
- L0 < 0:
- L(lf) = L0, f between 0 and f1 = 1st corner frequency;
- L(lf) = -N * (lfx lf), f between f1 and lfx = 2nd corner frequency;
- L(lf) = 0, lf > lfx.
- lf1 = lfx + L0/N = dB-frequency at which level goes up from L0.

See lowshelf_other_freq.

high shelf and highshelf other freq

First-order "high shelf" filter (gain boost cut above some frequency).

Usage

```
_ : highshelf(N,Lpi,fx) : _
_ : highshelf_other_freq(N,Lpi,fx) : _
```

Where:

- N: filter order 1, 3, 5, ... (odd only).
- Lpi: desired level (dB) between fx and SR/2 (boost Lpi>0 or cut Lpi<0)
- fx: -3dB frequency of highpass band (L0>0) or lower band (L0<0) (Use highshelf_other_freq() below to find the other one.)

The gain at dc is constrained to be 1. See lowshelf documentation above for more details on shelf shape.

peak_eq

Second order "peaking equalizer" section (gain boost or cut near some frequency) Also called a "parametric equalizer" section.

Usage

```
_ : peak_eq(Lfx,fx,B) : _;
```

Where:

- Lfx: level (dB) at fx (boost Lfx>0 or cut Lfx<0)
- fx: peak frequency (Hz)
- B: bandwidth (B) of peak in Hz

peak_eq_cq

Constant-Q second order peaking equalizer section.

Usage

```
_ : peak_eq_cq(Lfx,fx,Q) : _;
```

Where:

- Lfx: level (dB) at fx
- fx: boost or cut frequency (Hz)
- Q: "Quality factor" = fx/B where B = bandwidth of peak in Hz

peak_eq_rm

Regalia-Mitra second order peaking equalizer section

Usage

```
_ : peak_eq_rm(Lfx,fx,tanPiBT) : _;
```

- Lfx: level (dB) at fx
- fx: boost or cut frequency (Hz)
- tanPiBT: tan(PI*B/SR), where B = -3dB bandwidth (Hz) when $10^{(Lfx/20)} = 0 \sim PI*B/SR$ for narrow bandwidths B

Reference

P.A. Regalia, S.K. Mitra, and P.P. Vaidyanathan, "The Digital All-Pass Filter: A Versatile Signal Processing Building Block" Proceedings of the IEEE, 76(1):19-37, Jan. 1988. (See pp. 29-30.)

spectral_tilt

Spectral tilt filter, providing an arbitrary spectral rolloff factor alpha in (-1,1), where -1 corresponds to one pole (-6 dB per octave), and +1 corresponds to one zero (+6 dB per octave). In other words, alpha is the slope of the ln magnitude versus ln frequency. For a "pinking filter" (e.g., to generate 1/f noise from white noise), set alpha to -1/2.

Usage

```
_ : spectral_tilt(N,f0,bw,alpha) : _
```

Where:

- N: desired integer filter order (fixed at compile time)
- fo: lower frequency limit for desired roll-off band
- bw: bandwidth of desired roll-off band
- alpha: slope of roll-off desired in nepers per neper (ln mag / ln radian freq)

Examples

See spectral_tilt_demo.

Reference

Link to appear here when write up is done

levelfilter and levelfilterN

Dynamic level lowpass filter.

```
_ : levelfilter(L,freq) : _
_ : levelfilterN(N,freq,L) : _
```

Where:

- L: desired level (in dB) at Nyquist limit (SR/2), e.g., -60
- freq: corner frequency (-3dB point) usually set to fundamental freq
- N: Number of filters in series where L = L/N

Reference

 $https://ccrma.stanford.edu/realsimple/faust_strings/Dynamic_Level_Lowpass_Filter.html$

Mth-Octave Filter-Banks

Mth-octave filter-banks split the input signal into a bank of parallel signals, one for each spectral band. They are related to the Mth-Octave Spectrum-Analyzers in analysis.lib. The documentation of this library contains more details about the implementation. The parameters are:

- M: number of band-slices per octave (>1)
- N: total number of bands (>2)
- ftop: upper bandlimit of the Mth-octave bands (<SR/2)

In addition to the Mth-octave output signals, there is a highpass signal containing frequencies from ftop to SR/2, and a "dc band" lowpass signal containing frequencies from 0 (dc) up to the start of the Mth-octave bands. Thus, the N output signals are

```
highpass(ftop), MthOctaveBands(M,N-2,ftop), dcBand(ftop*2^(-M*(N-1)))
```

A Filter-Bank is defined here as a signal bandsplitter having the property that summing its output signals gives an allpass-filtered version of the filter-bank input signal. A more conventional term for this is an "allpass-complementary filter bank". If the allpass filter is a pure delay (and possible scaling), the filter bank is said to be a "perfect-reconstruction filter bank" (see Vaidyanathan-1993 cited below for details). A "graphic equalizer", in which band signals are scaled by gains and summed, should be based on a filter bank.

The filter-banks below are implemented as Butterworth or Elliptic spectrumanalyzers followed by delay equalizers that make them allpass-complementary.

Increasing Channel Isolation

Go to higher filter orders - see Regalia et al. or Vaidyanathan (cited below) regarding the construction of more aggressive recursive filter-banks using elliptic or Chebyshev prototype filters.

References

- "Tree-structured complementary filter banks using all-pass sections", Regalia et al., IEEE Trans. Circuits & Systems, CAS-34:1470-1484, Dec. 1987
- "Multirate Systems and Filter Banks", P. Vaidyanathan, Prentice-Hall, 1993
- Elementary filter theory: https://ccrma.stanford.edu/~jos/filters/

mth_octave_filterbank[n]

Allpass-complementary filter banks based on Butterworth band-splitting. For Butterworth band-splits, the needed delay equalizer is easily found.

Usage

```
_ : mth_octave_filterbank(0,M,ftop,N) : par(i,N,_); // Oth-order
_ : mth_octave_filterbank_alt(0,M,ftop,N) : par(i,N,_); // dc-inverted version
Also for convenience:
_ : mth_octave_filterbank3(M,ftop,N) : par(i,N,_); // 3d-order Butterworth
_ : mth_octave_filterbank5(M,ftop,N) : par(i,N,_); // 5th-roder Butterworth
mth_octave_filterbank_default = mth_octave_analyzer6e;
```

Where:

- 0: order of filter used to split each frequency band into two
- M: number of band-slices per octave
- ftop: highest band-split crossover frequency (e.g., 20 kHz)
- N: total number of bands (including dc and Nyquist)

Arbritary-Crossover Filter-Banks and Spectrum Analyzers

These are similar to the Mth-octave analyzers above, except that the band-split frequencies are passed explicitly as arguments.

filterbank

Filter bank.

Usage

```
_ : filterbank (0,freqs) : par(i,N,_); // Butterworth band-splits Where:
```

- 0: band-split filter order (ODD integer required for filterbank[i])
- freqs: (fc1,fc2,...,fcNs) [in numerically ascending order], where Ns=N-1 is the number of octave band-splits (total number of bands N=Ns+1).

If frequencies are listed explicitly as arguments, enclose them in parens:

```
_ : filterbank(3,(fc1,fc2)) : _,_,_
```

filterbanki

Inverted-dc filter bank.

Usage

```
_ : filterbanki(0,freqs) : par(i,N,_); // Inverted-dc version Where:
```

- 0: band-split filter order (ODD integer required for filterbank[i])
- freqs: (fc1,fc2,...,fcNs) [in numerically ascending order], where Ns=N-1 is the number of octave band-splits (total number of bands N=Ns+1).

If frequencies are listed explicitly as arguments, enclose them in parens:

```
_ : filterbanki(3,(fc1,fc2)) : _,_,_
```

hoa.lib

Faust library for high order ambisonic.

It should be used using the ho environment:

```
ho = library("ho.lib");
process = ho.functionCall;
```

Another option is to import stdfaust.lib which already contains the ho environment:

```
import("stdfaust.lib");
process = ho.functionCall;
```

encoder

Ambisonic encoder. Encodes a signal in the circular harmonics domain depending on an order of decomposition and an angle.

Usage

```
encoder(n, x, a) : _{-}
```

Where:

- n: the order
- x: the signal
- a: the angle

decoder

Decodes an ambisonics sound field for a circular array of loudspeakers.

Usage

```
_ : decoder(n, p) : _
```

Where:

- n: the order
- p: the number of speakers

Note

Number of loudspeakers must be greater or equal to 2n+1. It's preferable to use 2n+2 loudspeakers.

decoderStereo

Decodes an ambisonic sound field for stereophonic configuration. An "home made" ambisonic decoder for stereophonic restitution (30° - 330°): Sound field lose energy around 180°. You should use inPhase optimization with ponctual sources. #### Usage

_ :	decoderStereo(n)	:	_	
Wh	ere:			
•	n: the order			

Optimization Functions

Functions to weight the circular harmonics signals depending to the ambisonics optimization. It can be basic for no optimization, maxRe or inPhase.

optimBasic

The basic optimization has no effect and should be used for a perfect circle of loudspeakers with one listener at the perfect center loudspeakers array.

optimMaxRe

The maxRe optimization optimize energy vector. It should be used for an auditory confined in the center of the loudspeakers array.

${\tt optimInPhase}$

The inPhase Optimization optimize energy vector and put all loudspeakers signals n phase. It should be used for an auditory.

Usage

```
" optimInPhase(n) : _ "
here:
n: the order
```

wider

Can be used to wide the diffusion of a localized sound. The order depending signals are weighted and appear in a logarithmic way to have linear changes.

Usage

```
_ : wider(n,w) : _
Where:
      • n: the order
      • w: the width value between 0 - 1
```

map

It simulate the distance of the source by applying a gain on the signal and a wider processing on the soundfield.

Usage

```
map(n, x, r, a)
```

- \bullet n: the order
- x: the signal

```
• r: the radius
```

• a: the angle in radian

rotate

Rotates the sound field.

Usage

```
_ : rotate(n, a) : _ Where:
```

- n: the order
- a: the angle in radian

math.lib

Mathematic library for Faust. Some functions are implemented as Faust foreign functions of math.h functions that are not part of Faust's primitives. Defines also various constants and several utilities.

It should be used using the fi environment:

```
ma = library("math.lib");
process = ma.functionCall;
```

Another option is to import stdfaust.lib which already contains the ma environment:

```
import("stdfaust.lib");
process = ma.functionCall;
```

Functions Reference

SR

Current sampling rate (between 1Hz and 192000Hz). Constant during program execution.

Usage		
SR : _		
_		
BS		
Current block-size.	Can change during the execution.	
TI		
Usage		
BS : _		
-		
PI		
Constant PI in doub	le precisio.n	
Usage		
PI : _		
-		
FTZ		
zero. Usually not nee	samples under the "maximum subnoreded in C++ because the architecture favoring javascript for instance.	
Usage		
_ : ftz : _		
See: http://docs.ora	$ m acle.com/cd/E19957-01/806-3568/ncg_$	$_{ m math.html}$
-		
neg		
Invert the sign (-x) of	of a signal.	

Takes two input signals: x and n, and multiplies x by 2 to the power n.

Usage

```
_,_ : ldexp : _
```

scalb

Takes two input signals: x and n, and multiplies x by 2 to the power n.

Usage

```
_,_ : scalb : _
```

log1p(x)

Computes log(1 + x) without undue loss of accuracy when x is nearly zero.

Usage

```
_ : log1p : _
```

logb

Return exponent of the input signal as a floating-point number.

Usage

ilogb

Return exponent of the input signal as an integer number.

Usage
_ : ilogb : _
log2
Returns the base 2 logarithm of x.
Usage
_ : log2 : _
expm1
Return exponent of the input signal minus 1 with better precision.
Usage
_ : expm1 : _
acosh
Computes the principle value of the inverse hyperbolic cosine of the input signal
Usage
_ : acosh : _
asinh(x)
Computes the inverse hyperbolic sine of the input signal.
Usage
_ : asinh : _

Computes the inverse hyperbolic tangent of the input signal.
Usage
_ : atanh : _
sinh
Computes the hyperbolic sine of the input signal.
Usage
_ : sinh : _
cosh
Computes the hyperbolic cosine of the input signal.
Usage
_ : cosh : _
<u> </u>
tanh
Computes the hyperbolic tangent of the input signal.
Usage
_ : tanh : _
erf

Computes the error function of the input signal.

atanh(x)

Usage
_ : erf : _

erfc
Computes the complementary error function of the input signal.
Usage
_ : erfc : _
gamma
Computes the gamma function of the input signal.
Usage
_ : gamma : _
lgamma
Calculates the natural logorithm of the absolute value of the gamma function of the input signal.
Usage
_ : lgamma : _
Ј0
Computes the Bessel function of the first kind of order 0 of the input signal.
$\mathbf{U}\mathbf{sage}$
_ : JO : _

J	1

Computes the Bessel function of the first kind of order 1 of the input signal.

1	1 0
Usage	
_ : J1 : _	
Jn	
Computes the Bessel function of the first kind of order n (f the second input signal.	irst input signal) of
Usage	
, : Jn : _	
чо	
Computes the linearly independent Bessel function of the sec of the input signal.	cond kind of order 0
Usage	
_ : YO : _	
Y1	
Computes the linearly independent Bessel function of the sec of the input signal.	cond kind of order 1
Usage	
_ : YO : _	

Yn

Computes the linearly independent Bessel function of the second kind of order n (first input signal) of the second input signal.

```
Usage
_,_ : Yn : _
fabs, fmax, fmin
Just for compatibility. . .
fabs = abs
fmax = max
fmin = min
np2(n)
Gives the next power of 2 of x.
Usage
np2(n) : _
Where:
   • n: an integer
frac(n)
Gives the fractional part of n.
Usage
frac(n) : _
Where:
   • n: a decimal number
```

isnan(x)

Return non-zero if and only if x is a NaN.

Usage

```
_ : isnan : _
```

Where:

 $\bullet\,$ x: signal to analyse

chebychev

Chebychev transformation of order n.

Usage

```
_ : chebychev(n) : _
```

Where:

• n: the order of the polynomial

Semantics

```
T[0](x) = 1,

T[1](x) = x,

T[n](x) = 2x*T[n-1](x) - T[n-2](x)
```

Reference

http://en.wikipedia.org/wiki/Chebyshev_polynomial

chebychevpoly

Linear combination of the first Chebyshev polynomials.

```
_ : chebychevpoly((c0,c1,...,cn)) : _
```

Where:

• cn: the different Chebychevs polynomials such that: chebychevpoly((c0,c1,...,cn)) = Sum of chebychev(i)*ci

Reference

http://www.csounds.com/manual/html/chebyshevpoly.html

diffn(x)

Negated first-roder difference.

Usage

```
_ : diffn : _
```

misceffect.lib

This library contains a collection of audio effects.

It should be used using the ef environment:

```
ef = library("misceffect.lib");
process = ef.functionCall;
```

Another option is to import stdfaust.lib which already contains the ef environment:

```
import("stdfaust.lib");
process = ef.functionCall;
```

Dynamic

```
cubicnl(drive,offset)
```

Cubic nonlinearity distortion.

```
_ : cubicnl(drive,offset) : _
_ : cubicnl_nodc(drive,offset) : _
```

Where:

- drive: distortion amount, between 0 and 1
- offset: constant added before nonlinearity to give even harmonics. Note: offset can introduce a nonzero mean feed cubicnl output to dcblocker to remove this.

References:

- https://ccrma.stanford.edu/~jos/pasp/Cubic_Soft_Clipper.html
- $\bullet \ \ https://ccrma.stanford.edu/\sim jos/pasp/Nonlinear_Distortion.html$

gate_mono and gate_stereo

Mono and stereo signal gates.

Usage

```
_ : gate_mono(thresh,att,hold,rel) : _
or
_,_ : gate_stereo(thresh,att,hold,rel) : _,_
```

Where:

- thresh: dB level threshold above which gate opens (e.g., -60 dB)
- att: attack time = time constant (sec) for gate to open (e.g., 0.0001 s = 0.1 ms)
- hold: hold time = time (sec) gate stays open after signal level < thresh (e.g., 0.1 s)
- rel: release time = time constant (sec) for gate to close (e.g., 0.020 s = 20 ms)

References

- http://en.wikipedia.org/wiki/Noise gate
- http://www.soundonsound.com/sos/apr01/articles/advanced.asp
- http://en.wikipedia.org/wiki/Gating_(sound_engineering)

Filtering

speakerbp(f1,f2)

Dirt-simple speaker simulator (overall bandpass eq with observed roll-offs above and below the passband).

Low-frequency speaker model = +12 dB/octave slope breaking to flat near f1. Implemented using two dc blockers in series.

High-frequency model = -24 dB/octave slope implemented using a fourth-order Butterworth lowpass.

Example based on measured Celestion G12 (12" speaker): speakerbp(130,5000);

Usage

```
_ : speakerbp(130,5000) : _
```

piano_dispersion_filter(M,B,f0)

Piano dispersion allpass filter in closed form.

Usage

```
_ : piano_dispersion_filter(1,B,f0) : +(totalDelay),_ : fdelay(maxDelay) : _ Where:
```

- M: number of first-order allpass sections (compile-time only) Keep below 20. 8 is typical for medium-sized piano strings.
- B: string inharmonicity coefficient (0.0001 is typical)
- \bullet f0: fundamental frequency in Hz

Outputs

- MINUS the estimated delay at f0 of all pass chain in samples, provided in negative form to facilitate subtraction from delay-line length.
- Output signal from all pass chain

stereo_width(w)

Stereo Width effect using the Blumlein Shuffler technique.

```
\_,\_ : stereo_width(w) : \_,\_
```

Where:

 \bullet w: stereo width between 0 and 1

At w=0, the output signal is mono ((left+right)/2 in both channels). At w=1, there is no effect (original stereo image). Thus, w between 0 and 1 varies stereo width from 0 to "original".

Reference

• "Applications of Blumlein Shuffling to Stereo Microphone Techniques" Michael A. Gerzon, JAES vol. 42, no. 6, June 1994

Time Based

echo(maxDuration,duration,feedback)

A simple echo effect.

Usage

```
_ : echo(maxDuration,duration,feedback) : _
```

Where:

- maxDuration: the max echo duration in seconds
- duration: the echo duration in seconds
- feedback: the feedback coefficient

Pitch Shifting

transpose

A simple pitch shifter based on 2 delay lines.

```
\_ : transpose(w, x, s) : \_
```

Where:

- w: the window length (samples)
- x: crossfade duration duration (samples)
- s: shift (semitones)

Meshes

mesh_square

Square Rectangular Digital Waveguide Mesh.

Usage

```
\verb"bus"(4*N) : \verb"mesh_square"(N) : \verb"bus"(4*N)";
```

Where:

• N: number of nodes along each edge - a power of two (1,2,4,8,...)

Reference

https://ccrma.stanford.edu/~jos/pasp/Digital_Waveguide_Mesh.html

Signal Order In and Out

The mesh is constructed recursively using 2x2 embeddings. Thus, the top level of mesh_square(M) is a block 2x2 mesh, where each block is a mesh(M/2). Let these blocks be numbered 1,2,3,4 in the geometry NW,NE,SW,SE, i.e., as 1 2 3 4 Each block has four vector inputs and four vector outputs, where the length of each vector is M/2. Label the input vectors as Ni,Ei,Wi,Si, i.e., as the inputs from the North, East South, and West, and similarly for the outputs. Then, for example, the upper left input block of M/2 signals is labeled 1Ni. Most of the connections are internal, such as 1Eo -> 2Wi. The 8*(M/2) input signals are grouped in the order 1Ni 2Ni 3Si 4Si 1Wi 3Wi 2Ei 4Ei and the output signals are 1No 1Wo 2No 2Eo 3So 3Wo 4So 4Eo or

In: 1No 1Wo 2No 2Eo 3So 3Wo 4So 4Eo

Out: 1Ni 2Ni 3Si 4Si 1Wi 3Wi 2Ei 4Ei

Thus, the inputs are grouped by direction N,S,W,E, while the outputs are grouped by block number 1,2,3,4, which can also be interpreted as directions

NW, NE, SW, SE. A simple program illustrating these orderings is process = mesh_square(2);.

Example

Reflectively terminated mesh impulsed at one corner:

```
\label{eq:mesh_square_test(N,x) = mesh_square(N)~(busi(4*N,x)) // input to corner with { busi(N,x) = bus(N) : par(i,N,*(-1)) : par(i,N-1,_), +(x); }; process = 1-1' : mesh_square_test(4); // all modes excited forever}
```

In this simple example, the mesh edges are connected as follows:

```
1No -> 1Ni, 1Wo -> 2Ni, 2No -> 3Si, 2Eo -> 4Si, 3So -> 1Wi, 3Wo -> 3Wi, 4So -> 2Ei, 4Eo -> 4Ei
```

A routing matrix can be used to obtain other connection geometries.

miscoscillator.lib

This library contains a collection of sound generators.

It should be used using the os environment:

```
os = library("miscoscillator.lib");
process = os.functionCall;
```

Another option is to import stdfaust.lib which already contains the os environment:

```
import("stdfaust.lib");
process = os.functionCall;
```

Wave-Table-Based Oscillators

```
sinwaveform(tablesize)
```

Sine waveform ready to use with a rdtable.

Usage

```
{\tt sinwaveform(tablesize):\_}
```

Where:

• tablesize: the table size

coswaveform(tablesize)
Cosine waveform ready to use with a rdtable.
Usage
coswaveform(tablesize) : _
Where:
• tablesize: the table size
<pre>phasor(tablesize,freq)</pre>
A simple phasor to be used with a rdtable.
Usage
<pre>phasor(tablesize,freq) : _</pre>
Where:
• tablesize: the table size
• freq: the frequency of the wave (Hz)
. (6.
oscsin(freq)
Sine wave oscillator.
Usage
oscsin(freq) : _
Where:
• freq: the frequency of the wave (Hz)

osc(freq) Default sine wave oscillator (same as oscrs). Usage osc(freq) : _ Where: ullet frequency of the wave (Hz) oscos(freq) Cosine wave oscillator. Usage osccos(freq) : _ Where: • freq: the frequency of the wave (Hz) oscp(freq,p) A sine wave generator with controllable phase. Usage oscp(freq,p) : _ Where: • freq: the frequency of the wave (Hz) • p: the phase in radian osci(freq)

Interpolated phase sine wave oscillator.

```
osci(freq) : _{-}
```

Where:

• freq: the frequency of the wave (Hz)

Virtual Analog Oscillators

Mostly elements from old "oscillator.lib".

Virtual analog oscillators and filter-based oscillators.

Low-frequency oscillators have prefix lf_ (no aliasing suppression, signal-means not necessarily zero)

Low Frequency Impulse and Pulse Trains, Square and Triangle Waves

Low Frequency Impulse and Pulse Trains, Square and Triangle Waves

```
\label{lem:lem:lemon} \begin{split} &\texttt{lf\_imptrain}, & &\texttt{lf\_pulsetrainpos}, & &\texttt{lf\_squarewavepos}, & &\texttt{lf\_squarewave}, \\ &\texttt{lf\_trianglepos} \end{split}
```

Usage

```
lf_imptrain(freq) : _
lf_pulsetrainpos(freq,duty) : _
lf_squarewavepos(freq) : _
lf_squarewave(freq) : _
lf_trianglepos(freq) : _
```

Where:

- freq: frequency in Hz
- duty: duty cycle between 0 and 1

Notes

- Suffix 'pos' means the function is nonnegative, otherwise ~ zero mean
- All impulse and pulse trains jump to 1 at time 0

Low Frequency Sawtooths

Low Frequency Sawtooths

```
lf_rawsaw, lf_sawpos, lf_sawpos_phase
```

Sawtooth waveform oscillators for virtual analog synthesis et al. The 'simple' versions (lf_rawsaw, lf_sawpos and saw1), are mere samplings of the ideal continuous-time ("analog") waveforms. While simple, the aliasing due to sampling is quite audible. The differentiated polynomial waveform family (saw2, sawN, and derived functions) do some extra processing to suppress aliasing (not audible for very low fundamental frequencies). According to Lehtonen et al. (JASA 2012), the aliasing of saw2 should be inaudible at fundamental frequencies below 2 kHz or so, for a 44.1 kHz sampling rate and 60 dB SPL presentation level; fundamentals 415 and below required no aliasing suppression (i.e., saw1 is ok).

Usage

```
lf_rawsaw(periodsamps) : _
lf_sawpos(freq) : _
lf_sawpos_phase(phase,freq) : _
saw1(freq) : _
```

Bandlimited Sawtooth

Bandlimited Sawtooth

```
sawN(N,freq), sawNp, saw2dpw(freq), saw2(freq), saw3(freq), saw4(freq),
saw5(freq), saw6(freq), sawtooth(freq), saw2f2(freq) saw2f4(freq)
```

Method 1 (saw2)

Polynomial Transition Regions (PTR) (for aliasing suppression)

Reference

- Kleimola, J.; Valimaki, V., "Reducing Aliasing from Synthetic Audio Signals Using Polynomial Transition Regions," in Signal Processing Letters, IEEE, vol.19, no.2, pp.67-70, Feb. 2012
- https://aaltodoc.aalto.fi/bitstream/handle/123456789/7747/publication6.pdf?sequence=9
- http://research.spa.aalto.fi/publications/papers/spl-ptr/

Method 2 (sawN)

Differentiated Polynomial Waves (DPW) (for aliasing suppression)

Reference

"Alias-Suppressed Oscillators based on Differentiated Polynomial Waveforms", Vesa Valimaki, Juhan Nam, Julius Smith, and Jonathan Abel, IEEE Tr. Acoustics, Speech, and Language Processing (IEEE-ASLP), Vol. 18, no. 5, May 2010.

Other Cases

Correction-filtered versions of saw2: saw2f2, saw2f4 The correction filter compensates "droop" near half the sampling rate. See reference for sawN.

Usage

```
sawN(N,freq) : _
sawNp(N,freq,phase) : _
saw2dpw(freq) : _
saw2(freq) : _
saw3(freq) : _ // based on sawN
saw4(freq) : _ // based on sawN
saw5(freq) : _ // based on sawN
saw6(freq) : _ // based on sawN
saw6(freq) : _ // based on sawN
sawtooth(freq) : _ // = saw2
saw2f2(freq) : _
saw2f4(freq) : _
```

Where:

- N: polynomial order
- freq: frequency in Hz
- phase: phase

Bandlimited Pulse, Square, and Impulse Trains

Bandlimited Pulse, Square, and Impulse Trains

 $\verb"pulsetrain", \verb"pulsetrain", \verb"square", \verb"imptrain", imptrain", \verb"triangle", triangle", triangle"$

All are zero-mean and meant to oscillate in the audio frequency range. Use simpler sample-rounded lf_* versions above for LFOs.

```
pulsetrainN(N,freq,duty) : _
pulsetrain(freq, duty) : _ // = pulsetrainN(2)
squareN(N, freq) : _
square : _ // = squareN(2)
imptrainN(N,freq) : _
imptrain : _ // = imptrainN(2)
triangleN(N,freq) : _
triangle : _ // = triangleN(2)
```

Where:

- N: polynomial order
- freq: frequency in Hz

Filter-Based Oscillators

Filter-Based Oscillators

Usage

osc[b|r|rs|rc|s|w](f), where f = frequency in Hz.

References

- $\bullet \ \, http://lac.linuxaudio.org/2012/download/lac12-slides-jos.pdf$
- https://ccrma.stanford.edu/~jos/pdf/lac12-paper-jos.pdf

oscb

Sinusoidal oscillator based on the biquad

oscr, oscrs and oscs

Sinusoidal oscillator based on 2D vector rotation, = undamped "coupled-form" resonator = lossless 2nd-order normalized ladder filter.

oscr = oscrs, oscrs generates a sine wave and oscs a cosine.

Reference:

 $https://ccrma.stanford.edu/\sim jos/pasp/Normalized_Scattering_Junctions. \\ html$

oscs

Sinusoidal oscillator based on the state variable filter = undamped "modified-coupled-form" resonator = "magic circle" algorithm used in graphics

oscw, oscwq, oscwc and oscws

Sinusoidal oscillator based on the waveguide resonator wgr.

```
oscwc - unit-amplitude cosine oscillator
```

oscws - unit-amplitude sine oscillator

oscq - unit-amplitude cosine and sine (quadrature) oscillator

oscw - default = oscwc for maximum speed

Reference

 $https://ccrma.stanford.edu/\sim jos/pasp/Digital_Waveguide_Oscillator.html$

noise.lib

A library of noise generators.

It should be used using the no environment:

```
no = library("noise.lib");
process = no.functionCall;
```

Another option is to import stdfaust.lib which already contains the no environment:

```
import("stdfaust.lib");
process = no.functionCall;
```

Functions Reference

noise()
White noise generator (outputs random number between -1 and 1).
$\mathbf{U}\mathbf{sage}$
noise : _
multirandom(n)
Generates multiple decorrelated random numbers in parallel.
Usage
multirandom(n) : _
Where:
• n: the number of decorrelated random numbers in parallel
multinoise(n)
Generates multiple decorrelated noises in parallel.
$\mathbf{U}\mathbf{sage}$
multinoise(n) : _
Where:
• n: the number of decorrelated random numbers in parallel
noises(n,i)
TODO.

pink_noise

Pink noise (1/f noise) generator (third-order approximation)

Usage

```
pink_noise : _;
```

Reference:

 $https://ccrma.stanford.edu/\sim jos/sasp/Example_Synthesis_1_F_Noise.html$

pink_noise_vm

Multi pink noise generator.

Usage

```
pink_noise_vm(N) : _;
```

Where:

• N: number of latched white-noise processes to sum, not to exceed sizeof(int) in C++ (typically 32).

References

- $\bullet \ \ http://www.dsprelated.com/showarticle/908.php$
- $\bullet \ \, \rm http://www.firstpr.com.au/dsp/pink-noise/\#Voss-McCartney$

lfnoise, lfnoise0 and lfnoiseN

Low-frequency noise generators (Butterworth-filtered downsampled white noise)

Usage

Example

phafla.lib

```
A library of compressor effects.
```

It should be used using the pf environment:

```
pf = library("phafla.lib");
process = pf.functionCall;
```

Another option is to import stdfaust.lib which already contains the pf environment:

```
import("stdfaust.lib");
process = pf.functionCall;
```

Functions Reference

flanger_mono and flanger_stereo

Flanging effect.

Usage:

```
_ : flanger_mono(dmax,curdel,depth,fb,invert) : _;
_,_ : flanger_stereo(dmax,curdel1,curdel2,depth,fb,invert) : _,_;
_,_ : flanger_demo : _,_;
```

Where:

- dmax: maximum delay-line length (power of 2) 10 ms typical
- curdel: current dynamic delay (not to exceed dmax)
- depth: effect strength between 0 and 1 (1 typical)
- fb: feedback gain between 0 and 1 (0 typical)
- invert: 0 for normal, 1 to invert sign of flanging sum

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/Flanging.html$

phaser2_mono and phaser2_stereo

Phasing effect.

Phaser

```
_: phaser2_mono(Notches,phase,width,frqmin,fratio,frqmax,speed,depth,fb,invert) : _;
_,_ : phaser2_stereo(") : _,_;
_,_ : phaser2_demo : _,_;
```

Where:

- Notches: number of spectral notches (MACRO ARGUMENT not a signal)
- phase: phase of the oscillator (0-1)
- width: approximate width of spectral notches in Hz
- frqmin: approximate minimum frequency of first spectral notch in Hz
- fratio: ratio of adjacent notch frequencies
- frqmax: approximate maximum frequency of first spectral notch in Hz
- speed: LFO frequency in Hz (rate of periodic notch sweep cycles)
- depth: effect strength between 0 and 1 (1 typical) (aka "intensity") when depth=2, "vibrato mode" is obtained (pure allpass chain)
- fb: feedback gain between -1 and 1 (0 typical)
- invert: 0 for normal, 1 to invert sign of flanging sum

Reference:

- https://ccrma.stanford.edu/~jos/pasp/Phasing.html
- http://www.geofex.com/Article_Folders/phasers/phase.html
- 'An Allpass Approach to Digital Phasing and Flanging', Julius O. Smith III, Proc. Int. Computer Music Conf. (ICMC-84), pp. 103-109, Paris, 1984.
- CCRMA Tech. Report STAN-M-21: https://ccrma.stanford.edu/STANM/stanms/stanm21/

pm.lib

Faust physical modeling library.

It should be used using the fi environment:

```
pm = library("pm.lib");
process = pm.functionCall;
```

Another option is to import **stdfaust.lib** which already contains the **pm** environment:

```
import("stdfaust.lib");
process = pm.functionCall;
```

chain(A:B:...)

Creates a chain of bidirectional blocks. Blocks must have 3 inputs and outputs. The first input/output correspond to the left going signal, the second input/output correspond to the right going signal and the third input/output is the mix of the main signal output. The implied one sample delay created by the ~ operator is generalized to the left and right going waves. Thus, n blocks in chain() will add an n samples delay to both the left and right going waves. ### Usage

 $\label{lem:chain} rightGoingWaves, leftGoingWaves, mixedOutput : chain(A:B) : rightGoingWaves, leftGoingWaves, mixedOutput : with{}$

```
A = _,_,_;
B = _,_,_;
};
```

Requires

```
filter.lib (crossnn)
```

input(x)

Adds a waveguide input anywhere between 2 blocks in a chain of blocks (see chain()). ### Usage

```
string(x) = chain(A:input(x):B)
```

Where x is the input signal to be added to the chain.

output()

Adds a waveguide output anywhere between 2 blocks in a chain of blocks and sends it to the mix output channel (see chain()). ### Usage

```
chain(A:output:B)
```

terminations(a,b,c)

Creates terminations on both sides of a chain() without closing the inputs and outputs of the bidirectional signals chain. As for chain(), this function adds a 1 sample delay to the bidirectional signal both ways. ### Usage

```
rightGoingWaves,leftGoingWaves,mixedOutput : terminations(a,b,c) : rightGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoingWaves,leftGoin
```

Requires

```
filter.lib (crossnn)
```

fullTerminations(a,b,c)

Same as terminations() but closes the inputs and outputs of the bidirectional chain (only the mixed output remains). ### Usage

```
terminations(a,b,c) : _
with{
    a = *(-1); // left termination
    b = chain(D:E:F); // bidirectional chain of blocks (D, E, F, etc.)
    c = *(-1); // right termination
};
```

Requires

```
filter.lib (crossnn)
```

leftTermination(a,b)

Creates a termination on the left side of a chain() without closing the inputs and outputs of the bidirectional signals chain. This function adds a 1 sample delay near the termination. ### Usage

```
rightGoingWaves,leftGoingWaves,mixedOutput : terminations(a,b) : rightGoingWaves,leftGoingWaves
with{
    a = *(-1); // left termination
    b = chain(D:E:F); // bidirectional chain of blocks (D, E, F, etc.)
};
```

Requires

```
filter.lib (crossnn)
```

rightTermination(b,c)

Creates a termination on the right side of a chain() without closing the inputs and outputs of the bidirectional signals chain. This function adds a 1 sample delay near the termination. ### Usage

```
with{
    b = chain(D:E:F); // bidirectional chain of blocks (D, E, F, etc.)
    c = *(-1); // right termination
};
```

rightGoingWaves,leftGoingWaves,mixedOutput : terminations(b,c) : rightGoingWaves,leftGoingWaves,leftGoingWaves

Requires

```
filter.lib (crossnn)
```

waveguide(nMax,n)

A simple waveguide block based on a 4th order fractional delay. ### Usage

```
\label{lem:condition} rightGoingWaves, leftGoingWaves, mixedOutput : waveguide (nMax,n) : rightGoingWaves, leftGoingWaves, l
```

of the waveguide in samples. ### Requires filter.lib (fdelay4)

idealString(length,reflexion,xPosition,x)

An ideal string with rigid terminations and where the plucking position and the pick-up position are the same. ### Usage

```
1-1' : idealString(length,reflexion,xPosition,x)
```

With: * length: the length of the string in meters * reflexion: the coefficient of reflexion (0-0.9999999) * pluckPosition: the plucking position (0.001-0.999) * x: the input signal for the excitation ### Requires filter.lib (fdelay4,crossnn)

reverb.lib

A library of reverb effects.

It should be used using the re environment:

```
re = library("reverb.lib");
process = re.functionCall;
```

Another option is to import stdfaust.lib which already contains the re environment:

```
import("stdfaust.lib");
process = re.functionCall;
```

Functions Reference

jcrev and satrev

These artificial reverberators take a mono signal and output stereo (satrev) and quad (jcrev). They were implemented by John Chowning in the MUS10 computer-music language (descended from Music V by Max Mathews). They are Schroeder Reverberators, well tuned for their size. Nowadays, the more expensive freeverb is more commonly used (see the Faust examples directory).

jcrev reverb below was made from a listing of "RV", dated April 14, 1972, which was recovered from an old SAIL DART backup tape. John Chowning thinks this might be the one that became the well known and often copied JCREV.

satrev was made from a listing of "SATREV", dated May 15, 1971, which was recovered from an old SAIL DART backup tape. John Chowning thinks this might be the one used on his often-heard brass canon sound examples, one of which can be found at https://ccrma.stanford.edu/~jos/wav/FM_BrassCanon2.wav

```
_ : jcrev : _,_,_
_ : satrev : _,_
```

mono_freeverb and stereo_freeverb

A simple Schroeder reverberator primarily developed by "Jezar at Dreampoint" that is extensively used in the free-software world. It uses four Schroeder allpasses in series and eight parallel Schroeder-Moorer filtered-feedback comb-filters for each audio channel, and is said to be especially well tuned.

Usage

```
_ : mono_freeverb(fb1, fb2, damp, spread) : _;
_,_ : stereo_freeverb(fb1, fb2, damp, spread) : _,_;
```

Where:

- fb1: coefficient of the lowpass comb filters (0-1)
- fb2: coefficient of the allpass comb filters (0-1)
- damp: damping of the lowpass comb filter (0-1)
- spread: spatial spread in number of samples (for stereo)

fdnrev0

Pure Feedback Delay Network Reverberator (generalized for easy scaling).

Usage

```
<1,2,4,...,N signals> <: fdnrev0(MAXDELAY,delays,BBSO,freqs,durs,loopgainmax,nonl) :> <1,2,4,...,N signals>
```

Where:

- N: 2, 4, 8, ... (power of 2)
- MAXDELAY: power of 2 at least as large as longest delay-line length
- delays: N delay lines, N a power of 2, lengths perferably coprime
- BBSO: odd positive integer = order of bandsplit desired at freqs
- freqs: NB-1 crossover frequencies separating desired frequency bands
- durs: NB decay times (t60) desired for the various bands
- loopgainmax: scalar gain between 0 and 1 used to "squelch" the reverb

• nonl: nonlinearity (0 to 0.999..., 0 being linear)

Reference

 $https://ccrma.stanford.edu/{\sim}jos/pasp/FDN_Reverberation.html$

zita_rev_fdn

Internal 8x8 late-reverberation FDN used in the FOSS Linux reverb zita-rev1 by Fons Adriaensen fons@linuxaudio.org. This is an FDN reverb with allpass comb filters in each feedback delay in addition to the damping filters.

Usage

bus(8) : zita_rev_fdn(f1,f2,t60dc,t60m,fsmax) : bus(8)

Where:

- f1: crossover frequency (Hz) separating dc and midrange frequencies
- f2: frequency (Hz) above f1 where T60 = t60 m/2 (see below)
- t60dc: desired decay time (t60) at frequency 0 (sec)
- t60m: desired decay time (t60) at midrange frequencies (sec)
- fsmax: maximum sampling rate to be used (Hz)

Reference

- http://www.kokkinizita.net/linuxaudio/zita-rev1-doc/quickguide.html
- https://ccrma.stanford.edu/~jos/pasp/Zita_Rev1.html

zita_rev1_stereo

Extend zita_rev_fdn to include zita_rev1 input/output mapping in stereo mode.

Usage

```
_,_ : zita_rev1_stereo(rdel,f1,f2,t60dc,t60m,fsmax) : _,_
```

Where:

rdel = delay (in ms) before reverberation begins (e.g., 0 to ~100 ms) (remaining args and refs as for zita_rev_fdn above)

zita_rev1_ambi

Extend zita_rev_fdn to include zita_rev1 input/output mapping in "ambisonics mode", as provided in the Linux C++ version.

Usage

```
_,_ : zita_rev1_ambi(rgxyz,rdel,f1,f2,t60dc,t60m,fsmax) : _,_,_,_
Where:
```

 $rgxyz = relative gain of lanes 1,4,2 to lane 0 in output (e.g., -9 to 9) (remaining args and references as for zita_rev1_stereo above)$

route.lib

A library of basic elements to handle signal routing in Faust.

It should be used using the si environment:

```
ro = library("route.lib");
process = ro.functionCall;
```

Another option is to import stdfaust.lib which already contains the si environment:

```
import("stdfaust.lib");
process = ro.functionCall;
```

Functions Reference

cross(n)

```
Cross n signals: (x1,x2,...,xn) \rightarrow (xn,...,x2,x1).
```

Usage

```
_,_, : cross(3) : _,_,_
```

Where:

• n: number of signals (int, must be known at compile time)

```
Note
Special case: cross2:
cross2 = _, cross(2),_;
crossnn(n)
Cross two bus(n)s.
Usage
_,_,... : crossmm(n) : _,_,...
   • n: the number of signals in the bus
crossn1(n)
Cross bus(n) and bus(1).
Usage
_,_,... : crossn1(n) : _,_,...
Where:
   • n: the number of signals in the first bus
interleave(row,col)
Interleave row col cables from column order to row order. input: x(0), x(1), x(2)
..., x(row\text{col-1}) output: x(0+0row), x(0+1row), x(0+2row), ..., x(1+\theta row),
x(1+1row), x(1+2row), \dots
Usage
_,_,_,_: interleave(3,2) : _,_,_,_,
Where:
```

- row: the number of row (int, known at compile time)
- column: the number of column (int, known at compile time)

butterfly(n)

Addition (first half) then substraction (second half) of interleaved signals.

Usage

```
_,_,_ : butterfly(4) : _,_,_
```

Where:

• n: size of the butterfly (n is int, even and known at compile time)

hadamard(n)

Hadamard matrix function of size $n = 2^k$.

Usage

```
_,_,_ : hadamard(4) : _,_,_,
```

Where:

• n: 2^k, size of the matrix (int, must be known at compile time)

Note:

Implementation contributed by Remy Muller.

recursivize(p,q)

Create a recursion from two arbitrary processors p and q.

```
_,_ : recursivize(p,q) : _,_
```

Where:

- $\bullet\,$ p: the forward arbitrary processor
- $\bullet\,$ q: the feedback arbitrary processor

signal.lib

A library of basic elements to handle signals in Faust.

It should be used using the si environment:

```
si = library("signal.lib");
process = si.functionCall;
```

Another option is to import stdfaust.lib which already contains the si environment:

```
import("stdfaust.lib");
process = si.functionCall;
```

Functions Reference

bus(n)

n parallel cables

Usage

```
bus(4) : _,_,_,
```

Where:

• n: is an integer known at compile time that indicates the number of parallel cables.

block(n)

Block - terminate n signals.

Usage

```
_,_ : interpolate(i) : _
```

Where:

- i: interpolation control between 0 and 1 (0: first input; 1: second input)

smooth(s)

Exponential smoothing by a unity-dc-gain one-pole lowpass.

Usage:

```
_ : smooth(tau2pole(tau)) : _
```

Where:

 $\bullet\,$ tau: desired smoothing time constant in seconds, or

```
hslider(...) : smooth(s) : _
```

Where:

• s: smoothness between 0 and 1. s=0 for no smoothing, s=0.999 is "very smooth", s>1 is unstable, and s=1 yields the zero signal for all inputs. The exponential time-constant is approximately 1/(1-s) samples, when s is close to (but less than) 1.

Reference:

 $https://ccrma.stanford.edu/\sim jos/mdft/Convolution_Example_2_ADSR.html$

smoo

Smoothing function based on smooth ideal to smooth UI signals (sliders, etc.) down.

Usage

```
hslider(...) : smoo;
```

polySmooth

A smoothing function based on **smooth** that doesn't smooth when a trigger signal is given. This is very useful when making polyphonic synthesizer to make sure that the value of the parameter is the right one when the note is started.

Usage

```
hslider(...) : polysmooth(g,s,d) : _
```

Where:

- g: the gate/trigger signal used when making polyphonic synths
- s: the smoothness (see smooth)
- d: the number of samples to wait before the signal start being smoothed after ${\tt g}$ switched to 1

bsmooth

Block smooth linear interpolation during a block of samples.

Usage

```
hslider(...) : bsmooth : _
```

```
lag_ud(up, dn, signal)
```

Lag filter with separate times for up and down.

```
Usage
_ : lag_ud(up, dn, signal) : _;
dot(n)
Dot product for two vectors of size n.
Usage
_,_,_,_ : dot(3) : _
Where:
   • n: size of the vectors (int, must be known at compile time)
spat.lib
This library contains a collection of tools for sound spatialization.
It should be used using the sp environment:
sp = library("spat.lib");
process = sp.functionCall;
Another option is to import stdfaust.lib which already contains the sp envi-
ronment:
import("stdfaust.lib");
process = sp.functionCall;
panner(g)
A simple linear gain panner.
Usage
_ : panner(g) : _,_
```

Where:

• g: the panning (0-1)

spat(n,a,d)

GMEM SPAT: n-outputs spatializer

Usage

```
\_ : spat(n,r,d) : \_,\_,...
```

Where:

- n: number of outputs
- r: rotation (between 0 et 1)
- d: distance of the source (between 0 et 1)

stereoize(p)

Transform an arbitrary processor ${\sf p}$ into a stereo processor with 2 inputs and 2 outputs.

Usage

```
_,_ : stereoize(p) : _,_
```

Where:

• p: the arbitrary processor

synth.lib

This library contains a collection of envelope generators.

It should be used using the sy environment:

```
sy = library("synth.lib");
process = sy.functionCall;
```

Another option is to import stdfaust.lib which already contains the sy environment:

```
import("stdfaust.lib");
process = sy.functionCall;
```

popFilterPerc

A simple percussion instrument based on a "poped" resonant bandpass filter.

Usage

```
popFilterDrum(freq,q,gate) : _;
```

Where:

- freq: the resonance frequency of the instrument
- q: the q of the res filter (typically, 5 is a good value)
- gate: the trigger signal (0 or 1)

dubDub

A simple synth based on a sawtooth wave filtered by a resonant lowpass.

Usage

```
dubDub(freq,ctFreq,q,gate) : _;
```

Where:

- freq: frequency of the sawtooth
- ctFreq: cutoff frequency of the filter
- q: Q of the filter
- gate: the trigger signal (0 or 1)

sawTrombone

A simple trombone based on a lowpassed sawtooth wave.

Usage

```
sawTrombone(att,freq,gain,gate) : _
```

Where:

- att: exponential attack duration in s (typically 0.01)
- freq: the frequency
- gain: the gain (0-1)
- gate: the gate (0 or 1)

combString

Simplest string physical model ever based on a comb filter.

Usage

```
combString(freq,res,gate) : _;
```

Where:

- freq: the frequency of the string
- res: string T60 (resonance time) in second
- gate: trigger signal (0 or 1)

additiveDrum

A simple drum using additive synthesis.

Usage

```
additiveDrum(freq,freqRatio,gain,harmDec,att,rel,gate) : _
```

Where:

- freq: the resonance frequency of the drum
- freqRatio: a list of ratio to choose the frequency of the mode in function of freq e.g.(1 1.2 1.5 ...). The first element should always be one (fundamental).
- gain: the gain of each mode as a list (1 0.9 0.8 ...). The first element is the gain of the fundamental.
- harmDec: harmonic decay ratio (0-1): configure the speed at which higher modes decay compare to lower modes.
- att: attack duration in second
- rel: release duration in second
- gate: trigger signal (0 or 1)

additiveDrum

An FM synthesizer with an arbitrary number of modulators connected as a sequence.

```
freqs = (300,400,...);
indices = (20,...);
fm(freqs,indices) : _
```

Where:

- freqs: a list of frequencies where the first one is the frequency of the carrier and the others, the frequency of the modulator(s)
- indices: the indices of modulation (Nfreqs-1)

vaeffect.lib

A library of virtual analog filter effects.

It should be used using the ve environment:

```
ve = library("vaeffect.lib");
process = ve.functionCall;
```

Another option is to import stdfaust.lib which already contains the ve environment:

```
import("stdfaust.lib");
process = ve.functionCall;
```

Functions Reference

```
moog_vcf(res,fr)
```

Moog "Voltage Controlled Filter" (VCF) in "analog" form. Moog VCF implemented using the same logical block diagram as the classic analog circuit. As such, it neglects the one-sample delay associated with the feedback path around the four one-poles. This extra delay alters the response, especially at high frequencies (see reference [1] for details). See moog_vcf_2b below for a more accurate implementation.

Usage

```
moog_vcf(res,fr)
```

Where:

• fr: corner-resonance frequency in Hz (less than SR/6.3 or so)

• res: Normalized amount of corner-resonance between 0 and 1 (0 is no resonance, 1 is maximum)

References

- https://ccrma.stanford.edu/~stilti/papers/moogvcf.pdf
- https://ccrma.stanford.edu/~jos/pasp/vegf.html

moog_vcf_2b[n]

Moog "Voltage Controlled Filter" (VCF) as two biquads. Implementation of the ideal Moog VCF transfer function factored into second-order sections. As a result, it is more accurate than moog_vcf above, but its coefficient formulas are more complex when one or both parameters are varied. Here, res is the fourth root of that in moog_vcf, so, as the sampling rate approaches infinity, moog_vcf(res,fr) becomes equivalent to moog_vcf_2bn (when res and fr are constant). moog_vcf_2b uses two direct-form biquads (tf2). moog_vcf_2bn uses two protected normalized-ladder biquads (tf2np).

Usage

```
moog_vcf_2b(res,fr)
moog_vcf_2bn(res,fr)
```

Where:

- fr: corner-resonance frequency in Hz
- res: Normalized amount of corner-resonance between 0 and 1 (0 is min resonance, 1 is maximum)

wah4(fr)

Wah effect, 4th order.

Usage

```
_ : wah4(fr) : _
```

Where:

 $\bullet\,$ fr: resonance frequency in Hz

Reference
$https://ccrma.stanford.edu/{\sim}jos/pasp/vegf.html$
autowah(level)
Auto-wah effect.
Usage
_ : autowah(level) : _;
Where:
• level: amount of effect desired (0 to 1).
crybaby(wah)
Digitized CryBaby wah pedal.
Usage
_ : crybaby(wah) : _
Where:
• wah: "pedal angle" from 0 to 1
Reference
$https://ccrma.stanford.edu/{\sim}jos/pasp/vegf.html$

vocoder

A very simple vocoder where the spectrum of the modulation signal is analyzed using a filter bank.

${\bf Usage}$

_ : vocoder(nBands,att,rel,BWRatio,source,excitation) : _;

Where:

- nBands: Number of vocoder bands
- att: Attack time in seconds
- rel: Release time in seconds
- BWRatio: Coefficient to adjust the bandwidth of each band (0.1 2)
- source: Modulation signal
- \bullet excitation: Excitation/Carrier signal