

CENG 384 - Signals and Systems for Computer Engineers
Spring 2018-2019
Written Assignment 1

Baykara, Azad
e2171320@ceng.metu.edu.tr

Kocaman, Alper
e2169589@ceng.metu.edu.tr

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1. (a)

$$\begin{aligned}y[n] &= 2x[n] + \frac{3}{4}y[n-1] - \frac{1}{8}y[n-2] \\y[n] - \frac{3}{4}y[n-1] + \frac{1}{8}y[n-2] &= 2x[n]\end{aligned}$$

(b) Take Fourier Transform of the difference equation.

$$\begin{aligned}\frac{1}{8}e^{-2jw}Y(e^{jw}) - \frac{3}{4}e^{-jw}Y(e^{jw}) + Y(e^{jw}) &= 2X(e^{jw}) \\Y(e^{jw}) \cdot \left[\frac{1}{8}e^{-2jw} - \frac{3}{4}e^{-jw} + 1\right] &= 2X(e^{jw}) \\H(e^{jw}) &= \frac{16}{e^{-2jw} - 6e^{-jw} + 8}\end{aligned}$$

(c)

$$\begin{aligned}H(e^{jw}) &= \frac{16}{e^{-2jw} - 6e^{-jw} + 8} = \frac{16}{(e^{-jw} - 4)(e^{-jw} - 2)} = \frac{A}{e^{-jw} - 4} + \frac{B}{e^{-jw} - 2} \\A = 8 \quad B = -8 \\H(e^{jw}) &= \frac{8}{e^{-jw} - 4} - \frac{8}{e^{-jw} - 2} \\&= \frac{4}{1 - \frac{1}{2}e^{-jw}} - \frac{2}{1 - \frac{1}{4}e^{-jw}} \\&\text{Take inverse FT} \\h[n] &= 4\left(\frac{1}{2}\right)^n u[n] - 2\left(\frac{1}{4}\right)^n u[n]\end{aligned}$$

(d)

Fourier transform of $x[n] = (\frac{1}{4})^n u[n]$ is $\frac{1}{1 - \frac{1}{4}e^{-jw}}$

$$y[n] = h[n] * x[n] \longleftrightarrow Y(e^{jw}) = H(e^{jw}) \cdot X(e^{jw})$$

$$\begin{aligned} Y(e^{jw}) &= \left(\frac{8}{e^{-jw} - 4} - \frac{8}{e^{-jw} - 2} \right) \cdot \frac{1}{1 - \frac{1}{4}e^{-jw}} \\ &= \left(\frac{8}{e^{-jw} - 4} - \frac{8}{e^{-jw} - 2} \right) \cdot \frac{4}{4 - e^{-jw}} \quad \text{let } z = e^{-jw} \\ &= \left(\frac{8}{z - 4} - \frac{8}{z - 2} \right) \cdot \frac{4}{4 - z} \\ &= \left(\frac{8}{z - 2} - \frac{8}{z - 4} \right) \cdot \frac{4}{z - 4} \\ &= \frac{32}{(z - 2)(z - 4)} - \frac{32}{(z - 4)^2} \\ &= \frac{16}{z - 4} - \frac{16}{z - 2} - \frac{32}{(z - 4)^2} \\ &= \frac{-4}{1 - \frac{1}{4}e^{-jw}} + \frac{8}{1 - \frac{1}{2}e^{-jw}} - \frac{2}{(1 - \frac{1}{4}e^{-jw})^2} \end{aligned}$$

Take inverse Fourier transform

$$y[n] = -4\left(\frac{1}{4}\right)^n u[n] + 8\left(\frac{1}{2}\right)^n u[n] - 2(n+1)\left(\frac{1}{4}\right)^n u[n]$$

2. If two LTI systems are connected parallel, then overall system's impulse response is

$$h[n] = h_1[n] + h_2[n]$$

and frequency response is

$$H(e^{jw}) = H_1(e^{jw}) + H_2(e^{jw})$$

In this question, overall system's frequency response $H(e^{jw})$ and impulse response of first system $h_1[n]$ are given. In order to find $h_2[n]$,

1-Frequency response of $h_1[n]$ should be found,

2-Frequency response of $h_2[n]$ which is $H_2(e^{jw})$ should be found using formula

$$H_2(e^{jw}) = H(e^{jw}) - H_1(e^{jw})$$

3-By applying inverse fourier transformation, find $h_2[n]$.

1-Frequency response of

$$h_1[n] = \left(\frac{1}{3}\right)^n u[n]$$

$$H_1(e^{jw}) = \frac{1}{1 - \frac{1}{3}e^{-jw}}$$

$$H_1(e^{jw}) = \frac{-3}{e^{-jw} - 3}$$

2-Frequency response of $h_2[n]$,

$$H_2(e^{jw}) = H(e^{jw}) - H_1(e^{jw})$$

$$H_2(e^{jw}) = \frac{5e^{-jw} - 12}{e^{-2jw} - 7e^{-jw} + 12} - \frac{-3}{e^{-jw} - 3}$$

$$H_2(e^{jw}) = \frac{5e^{-jw} - 12}{(e^{-jw} - 4)(e^{-jw} - 3)} - \frac{-3}{e^{-jw} - 3}$$

$$H_2(e^{jw}) = \frac{-2}{1 - \frac{1}{4}e^{-jw}}$$

3-Impulse response of $H_2(e^{jw})$ by using inverse FT,

$$h_2[n] = -2\left(\frac{1}{4}\right)^n u[n]$$

3. (a) Input signal $x(t)$ consists of a periodic cosine function and an aperiodic square wave which can be separated as

$$x(t) = x_1(t) + x_2(t)$$

where

$$x_1(t) = \frac{\sin 2\pi t}{\pi t}$$

$$x_2(t) = \cos 3\pi t$$

By taking fourier transformation,

$$F\{x_1(t)\} = X_1(j\omega) = \begin{cases} 1, & \text{if } |\omega| < 2\pi \\ 0, & \text{otherwise} \end{cases}$$

$$X_2(j\omega) = \pi[\delta(\omega - 3\pi) + \delta(\omega + 3\pi)]$$

Thus, fourier transform of $x(t)$,

$$X(j\omega) = X_1(j\omega) + X_2(j\omega)$$

And the graph of $X(j\omega)$,

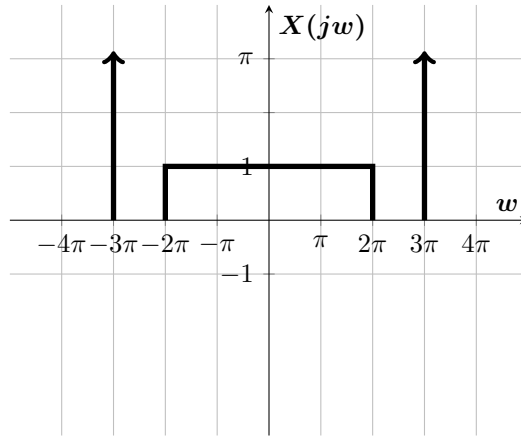


Figure 1: w vs. $X(jw)$.

- (b) From part a), Nyquist frequency is 3π .
Since

$$w_s = 2 \times w_m$$

$$w_s = 6\pi$$

Nyquist rate is 6π .

Sampling rate is

$$T = \frac{2\pi}{6\pi} = \frac{1}{3}$$

- (c) By using formula,

$$X_p(j\omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X(j(\omega - k\omega_s))$$

$$X_p(j\omega) = 3 \sum_{k=-\infty}^{\infty} X(j(\omega - 6\pi k))$$

(Graph of this question is below, top of the next page)

4. (a)

$$N = \frac{2\pi}{w_s} = 2$$

$$X_p(j\omega) = \frac{1}{T} \sum_{\forall k} X(j(\omega - kw_s))$$

$$X_d(e^{j\omega}) = X_p(j\frac{\omega}{T})$$

$$X_d(e^{j\omega}) = \begin{cases} \frac{2}{\pi} \omega & \text{if } |\omega| \leq \frac{\pi}{2} \\ 0 & \text{otherwise} \end{cases}, X_d(e^{j\omega}) = X_d(e^{j(\omega+N)})$$

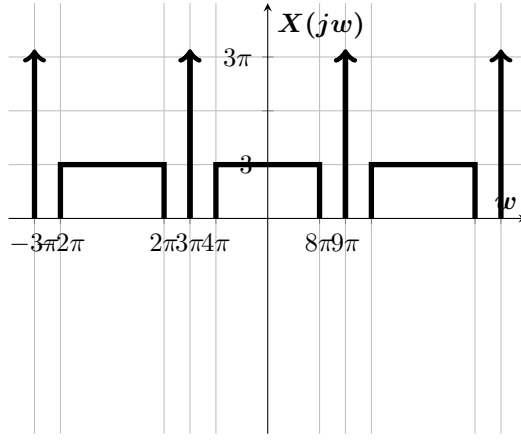


Figure 2: w vs. $X(j(w - 6\pi k))$.

(b) From discrete time Fourier transform table, we know that

$$e^{jw_0 n} \longleftrightarrow 2\pi \sum_{\forall k} \delta(w - w_0 - 2\pi k)$$

$$h[n] = \cos \pi n = \frac{1}{2}(e^{j\pi n} + e^{-j\pi n})$$

$$\begin{aligned} \text{Then, } H(e^{jw}) &= \frac{1}{2}(2\pi \sum_{\forall k} \delta(w - \pi - 2\pi k) + 2\pi \sum_{\forall k} \delta(w + \pi - 2\pi k)) \\ &= \pi(\sum_{\forall k} \delta(w - \pi - 2\pi k) + \delta(w + \pi - 2\pi k)) \end{aligned}$$

(c)

$$y_d[n] = x_d[n] \cdot h[n] \longleftrightarrow Y_d(e^{jw}) = \frac{1}{2\pi} X_d(e^{jw}) * H(e^{jw})$$

Convolution over 1 period ($-\pi$ to π)

$$Y_d(e^{jw}) = \frac{1}{2\pi} \cdot \pi(\sum_{\forall k} \delta(w - \pi) + \delta(w + \pi)) * X_d(e^{jw})$$

Shift $X_d(e^{jw})$ to the left and right by π . Hence,

$$Y_d(e^{jw}) = \begin{cases} \frac{1}{\pi} w & , \frac{\pi}{2} \leq |w| \leq \frac{3\pi}{2} \\ 0 & \text{otherwise} \end{cases}, Y_d(e^{jw}) = X_d(e^{j(w+2\pi)})$$