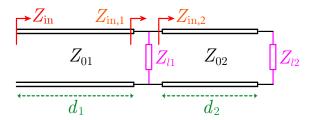
2020-2021, EE303 – Recitation 11

Question 1

Consider a combination of two loads and two lossless transmission lines.



The transmission lines are described as follows:

•
$$Z_{01} = 100 \ \Omega$$
, $d_1 = \lambda/4$, • $Z_{02} = 200 \ \Omega$, $d_2 = 5\lambda/8$ (1)

Given the load impedances $Z_{l1} = (100 - j200) \Omega$ and $Z_{l2} = (100 + j100) \Omega$, do the following.

- In the *second* transmission line, find the positions (measured from the load) of the first maximum and the first minimum of the voltage pattern.
- Find VSWR in both transmission lines.
- Find the input impedance of the overall combination, i.e., Z_{in} .
- * Extra question: Is it possible to make VSWR in the first line *unity* by selecting a suitable Z_{l1} without changing any other item?

Solution

The reflection coefficient at the second load:

$$\Gamma_{l2} = \frac{Z_{l2} - Z_{02}}{Z_{l2} + Z_{02}} = \frac{100 + j100 - 200}{100 + j100 + 200} = \frac{-1 + j}{3 + j} = -\frac{1}{5} + j\frac{2}{5}$$
 (2)

$$|\Gamma_{l2}| = \sqrt{\left(-\frac{1}{5}\right)^2 + \left(\frac{2}{5}\right)^2} = \frac{1}{\sqrt{5}}$$
 (3)

$$\varphi_{l2} = \pi - \tan^{-1}\left[(2/5)/(1/5)\right] = \pi/2 + \tan^{-1}\left[(1/5)/(2/5)\right]$$
 (4)

$$\Gamma_{l2} = |\Gamma_{l2}| \exp(j\varphi_{l2}) = \frac{1}{\sqrt{5}} \exp[j\pi/2 + j \tan^{-1}(1/2)]$$
 (5)

First maximum/minimum in the second line (in this case, the first maximum occurs earlier than the first minimum):

$$z'_{\text{max,first}} = \frac{\lambda \varphi_{l2}}{4\pi} = \frac{\lambda(\pi/2)}{4\pi} + \frac{\lambda \tan^{-1}(1/2)}{4\pi} = \frac{\lambda}{8} + \frac{\lambda}{4\pi} \tan^{-1}(1/2)$$
 (6)

$$z'_{\text{min,first}} = z'_{\text{max,first}} \pm \frac{\lambda}{4} = \frac{3\lambda}{8} + \frac{\lambda}{4\pi} \tan^{-1}(1/2)$$
 (7)

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VSWR for the second transmission line:

$$VSWR_2 = \frac{1 + |\Gamma_{l2}|}{1 - |\Gamma_{l2}|} = \frac{1 + 1/\sqrt{5}}{1 - 1/\sqrt{5}} = \frac{\sqrt{5} + 1}{\sqrt{5} - 1} \approx 2.62$$
 (8)

$$Z_{in,2} = Z_{02} \frac{Z_{l2} + jZ_{02} \tan(\beta_2 d_2)}{Z_{02} + jZ_{l2} \tan(\beta_2 d_2)} = 200 \frac{100 + j100 + j200 \tan(5\pi/4)}{200 + j(100 + j100) \tan(5\pi/4)}$$
(9)

$$=200\frac{100+j100+j200}{200+j100-100} = 200\frac{1+j3}{1+j} = (400+j200)\ \Omega \tag{10}$$

$$Z_{in,1} = Z_{l1} \parallel Z_{in,2} = (100 - j200) \parallel (400 + j200) = (160 - j120) \Omega$$
 (11)

Reflection coefficient at the effective load of the first transmission line:

$$\Gamma_{l1} = \frac{Z_{in,1} - Z_{01}}{Z_{in,1} + Z_{01}} = \frac{160 - j120 - 100}{160 - j120 + 100} = \frac{3 - j6}{13 - j6} \approx 0.37 - j0.29$$
(12)

VSWR for the first transmission line:

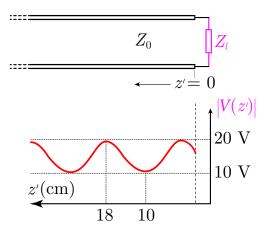
$$VSWR_2 = \frac{1 + |\Gamma_{l1}|}{1 - |\Gamma_{l1}|} \approx 2.77 \tag{13}$$

Input impedance of the overall combination:

$$Z_{in} = \frac{Z_{01}^2}{Z_{in,1}} = \frac{100^2}{(160 - j120)} = (40 + j30) \Omega$$
 (14)

Question 2

Consider a lossless transmission line terminated by a load Z_l with a given voltage pattern. Assume that the phase velocity is 3×10^8 m/s along the line.



- Find the wavelength and VSWR in the line.
- Find the magnitudes of the incident and reflected voltage waves.
- Find the reflection coefficient at the load (Γ_l) .

- Find the load impedance if the characteristic impedance of the transmission line is $Z_0 = 200 \ \Omega$.
- If the load involves a single inductor or capacitor (in addition to a resistor), find the value of the inductance or capacitance.
- Find the impedance value measured at z' = 8 cm.

Solution

$$\lambda = 0.32 \text{ m}$$
 (from peak-to-peak distance) (15)

VSWR =
$$20/10 = 2 = \frac{1 + |\Gamma_l|}{1 - |\Gamma_l|} \longrightarrow |\Gamma_l| = 1/3$$
 (16)

$$z'_{\text{max,first}} = \lambda \varphi_l / (4\pi) = 2 \text{ cm} \longrightarrow \varphi_l = 0.08\pi / \lambda = \pi/4 \text{ rad}$$
 (17)

$$\Gamma_l = |\Gamma_l| \exp(j\varphi_l) = (1/3) \exp(j\pi/4) = \sqrt{2}(1+j)/6$$
 (18)

From the maximum voltage:

$$|V_0^+|(1+|\Gamma_l|) = 20 \longrightarrow |V_0^+| = 15 \text{ V} \quad \text{and} \quad |V_0^-| = |\Gamma_l||V_0^+| = 5 \text{ V}$$
 (19)

Load impedance:

$$\Gamma_l = \frac{Z_l - Z_0}{Z_l + Z_0} \longrightarrow Z_l = Z_0 \frac{1 + \Gamma_l}{1 - \Gamma_l} \tag{20}$$

$$Z_l = 200 \frac{1 + \sqrt{2}/6 + j\sqrt{2}/6}{1 - \sqrt{2}/6 - j\sqrt{2}/6} \approx (277.9 + j147.4) \Omega$$
 (21)

Note: $f = u_p/\lambda = (3/0.32) \times 10^8$ Hz and $\omega \approx 5.9 \times 10^9$ rad/s

Load: A resistor with 277.9 Ω resistance and an inductor with an inductance of

$$L \approx 147.4/\omega \approx 25 \text{ nH}$$
 (22)

Impedance at z' = 8 cm (capacitive, even though the load is inductive):

$$Z(z' = 8 \text{ cm}) = Z_0^2 / Z_l \approx (112.3 - 59.6) \Omega$$
 (23)

Question 3

A series combination of a 30 Ω resistor and a $1/(80\pi)$ nF capacitor is connected to a lossless transmission line as a load at 1 GHz. The characteristic impedance of the transmission line is 50 Ω , while the phase velocity is 3×10^8 m/s. Find the impedance values measured at z' = 7.5 cm and z' = 15 cm. Also find how the values change if the frequency is increased to 2 GHz.

Solution

At 1 GHz, the load impedance can be found as

$$Z_l(1 \text{ GHz}) = 30 + \frac{1}{j2\pi \times 10^9 \times 10^{-9}/(80\pi)} = (30 - j40) \Omega$$
 (24)

$$z_1 = 0.075 \text{ m} = \lambda/4 \text{ [inverter] at 1 GHz}$$
 (25)

$$z_2 = 0.15 \text{ m} = \lambda/2 \text{ [repeater] at 1 GHz}$$
 (26)

$$Z(z_1, 1 \text{ GHz}) = \frac{50^2}{(30 - j40)} = (30 + j40) \Omega$$
 (27)

$$Z(z_2, 1 \text{ GHz}) = (30 - j40) \Omega$$
 (28)

$$Z_l(2 \text{ GHz}) = 30 + \frac{1}{j2\pi \times 2 \times 10^9 \times 10^{-9}/(80\pi)} = (30 - j20) \Omega$$
 (29)

$$z_1 = 0.075 \text{ m} = \lambda/2 \text{ [repeater] at 2 GHz}$$
 (30)

$$z_2 = 0.15 \text{ m} = \lambda \text{ [repeater] at 2 GHz}$$
 (31)

$$Z(z_1, 2 \text{ GHz}) = Z(z_2, 2 \text{ GHz}) = (30 - j20) \Omega$$
 (32)